

O P E R A T I N G A N D S E R V I C E M A N U A L

VHF OSCILLATOR

3200B



HEWLETT  PACKARD

Table 1-1. Specifications

Frequency range: 10 to 500 MHz in six bands: 10 to 18.8 MHz; 18.5 to 35 MHz; 35 to 68 MHz; 68 to 130 MHz; 130 to 260 MHz; 260 to 500 MHz.

Frequency accuracy: within $\pm 2\%$ after $\frac{1}{2}$ hour warmup.

Frequency calibration: increments of less than 4%.

Frequency stability (after 4-hour warmup under 0.2 mw load): short term (5 minutes) $\pm 0.002\%$; long term (1 hour) $\pm 0.02\%$; line voltage (5-volt change) $\pm 0.001\%$.

RF output:

Maximum power (across 50-ohm external load): >200 mw (10 to 130 MHz); >150 mw (130 to 260 MHz); >25 mw (260 to 500 MHz).

Range: 0 to >120 db attenuation from maximum output.

Load impedance: 50 ohms nominal.

RF leakage: sufficiently low to permit measurements at $1 \mu\text{V}$. RFI: meets requirements of MIL-I-6181D.

Amplitude modulation: externally modulated.

Range: 0 to 30%.

Distortion: <1% at 30% AM.

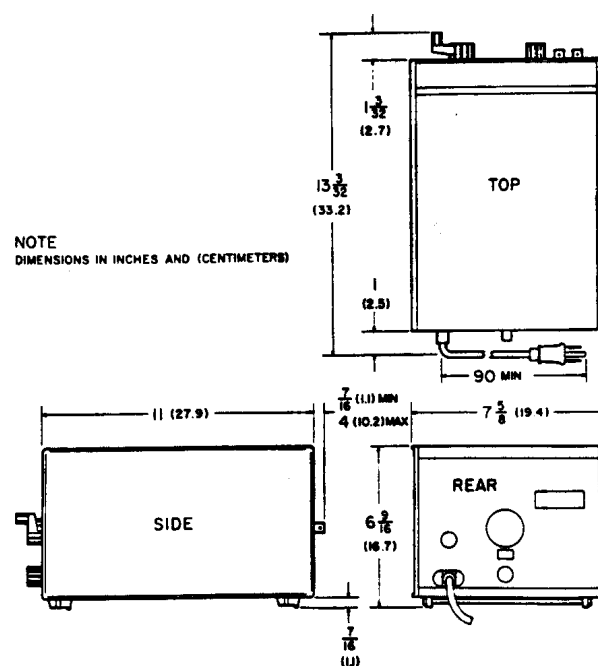
External requirements: approximately 20 volts rms into 600 ohms for 30% AM, 200 Hz to 100 kHz.

Pulse modulation: externally modulated.

External requirements: 2.5 volt negative pulse into 2000 ohms.

Power: 105 to 125 v or 210 to 250 v, 50 to 1000 Hz, 30 w.

Dimensions:



Weight: net 15 lbs. (6.8 kg), shipping 19 lbs. (8.6 kg).

Accessories available: 13515A Frequency Doubler
Probe: 00501B, 00514B, 00517B Output Cables;
00502B, 00506B Patching Cables.

SECTION IV PRINCIPLES OF OPERATION

4-1. GENERAL

4-2. The 3200B VHF Oscillator, shown in the block diagram (Figure 4-1), covers the frequency range from 10 to 500 MHz with six overlapping bands. This instrument was designed primarily as a high output source of CW. Circuitry has been incorporated to allow both amplitude and pulse modulation. A variable, piston-type, waveguide-below cutoff attenuator is used to control the output.

4-3. RF OSCILLATOR

4-4. The oscillator circuit, shown in Figure 4-2, generates all of the radio frequencies in the 10 to 500 MHz range of the instrument. The circuit employs two 6DZ4 triodes with push-pull connected plates and grounded cathodes. Simple, fixed feedback is accomplished by means of capacitive dividers consisting of C102 and C104, together with the grid-to-cathode capacitance of the triodes. Chokes L109 and L110, along with good cathode grounding and short grid connections, serve to maintain high grid circuit impedance, preventing phase shift in the feedback voltage divider and consequential oscillation in the circuit at parasitic frequencies.

4-5. Tuning of the oscillator is accomplished by means of a tank circuit which includes coils L101 to L106 and capacitors C103A and C103B. There are six different tank coils; one for each frequency range. The coils are wrapped on a specially formed turret which is detented into one of six frequency range positions by the RANGE switch on the front panel. Wiping contacts are used to connect the selected coil into the tank circuit. The center of each tank coil is neutral and is used as a feed point for plate power, which is connected through a switch operated by the turret drive and a common supply ring on the turret. One hundred ohm resistors (R101 to R106) are connected between the coils and the B+ feed point to break up undesirable RF paths, without introducing appreciable plate voltage or RF loss.

4-6. ATTENUATOR

4-7. The output from the oscillator is coupled to a pickup loop (L107) mounted at the end of a piston in the waveguide-below-cutoff attenuator. The distance between the pickup coil and the output coils on the turret is a fixed amount with the probe fully inserted. With this dimension fixed on each fre-

quency range, the coupling is different. On the high range, with the attenuator probe at maximum penetration, the pickup coil becomes overcoupled to the output coil. In some instances, this overcoupling loads the tuned output circuit to an extent which causes the oscillator to stop. Overcoupling will also cause frequency "pulling" creating dial errors. With this type of attenuator setup, the maximum penetration does not mean maximum output.

4-8. MODULATION

4-9. Amplitude modulation of the 3200B is accomplished by capacitance coupling a modulating signal to the oscillator plate supply. This is shown in block diagram form (Figure 4-3). With this type of modulation, up to 30% AM is possible with less than 1% distortion at input frequencies from 200 Hz to 100 kHz.

4-10. Pulse modulation of the 3200B is explained by the circuitry of Figure 4-4. With no signal input, a -32 volt bias is applied to the oscillator tube grids through R401, junction A, L109, L110, R107, and R108. This bias will stop tube conduction, thereby stopping oscillation. Thus, with no input signal, there will be no output from the 3200B.

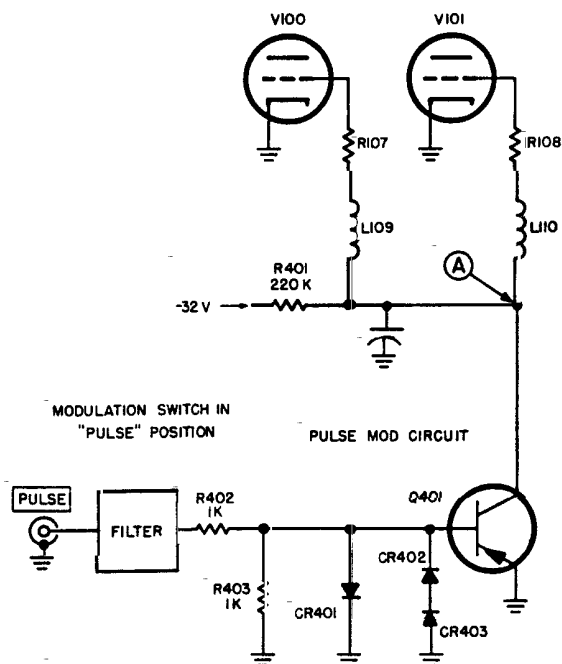


Figure 4-4. Pulse Modulation 3200B

4-11. A 2.5 volt negative-going pulse applied to the base of transistor Q401 will cause the transistor to conduct as the pulse approaches the negative peak. The impedance between the emitter and collector will decrease sufficiently to dissipate the -32 volt bias across R401. This puts junction A at a very low impedance with respect to ground, allowing the tubes to oscillate. As the pulse starts in the positive direction returning to zero potential, the transistor stops conduction, restoring the -32 volt bias to the oscillator tube grids which stops the oscillator. Essentially, the transistor becomes a fast action switch controlling the -32 volt bias to the oscillator grids.

4-12. The circuitry to the base of the transistor has resistors to limit the input impedance of the PULSE input connector and diodes to limit voltage, which could cause damage to the transistor.

4-13. Positive-going pulses to the pulse input connector, with Mod Input Switch in "PULSE" position, will result in no output from the instrument.

4-14. EXTERNAL B+ OPERATION

4-15. In the external B+ mode of operation, the internal plate voltage supply is disconnected. The plate supply lead is connected to the AM/B+ connector. A controlled external power supply connected to the AM/B+ connector will produce CW operation of the instrument. During external B+ operation a -6 volt bias is connected internally to the grid switch transistor, Q401, to keep it in the "ON" condition removing the -32 V bias from the oscillator grids. This circuitry is shown in the block diagram (Figure 4-5).

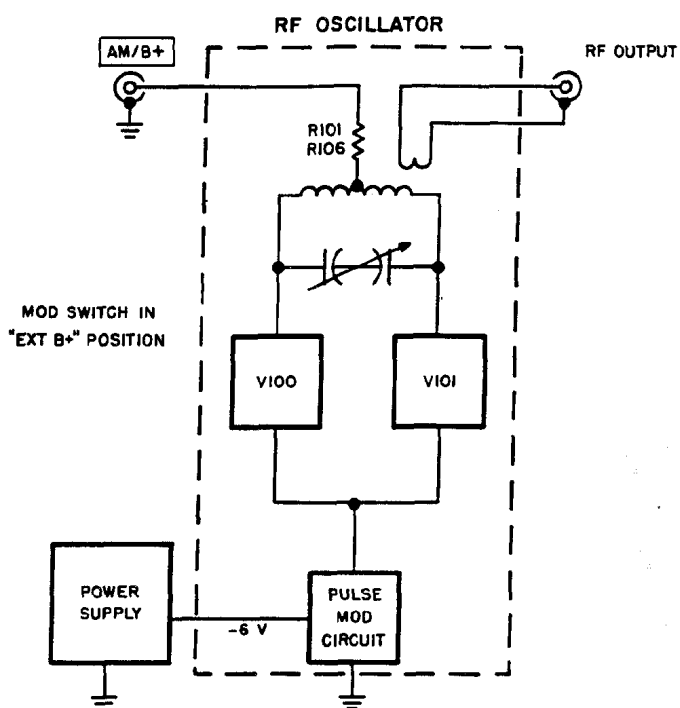


Figure 4-5. External B+ Operation

4-16. The Power Supply consists of two regulated supplies having a common power transformer. The outputs are 150 vdc, 40 mA for the oscillator tube plate supply, and 6.6 vdc, 450 mA for the oscillator tube heater supply. The regulated supplies are also a source of bias for operating the pulse modulation circuit (Z400). A -6 V bias is provided to operate the pulse modulation circuit, and -32 V is provided to turn off the oscillator. A block diagram of the complete supply is shown in Figure 4-6.

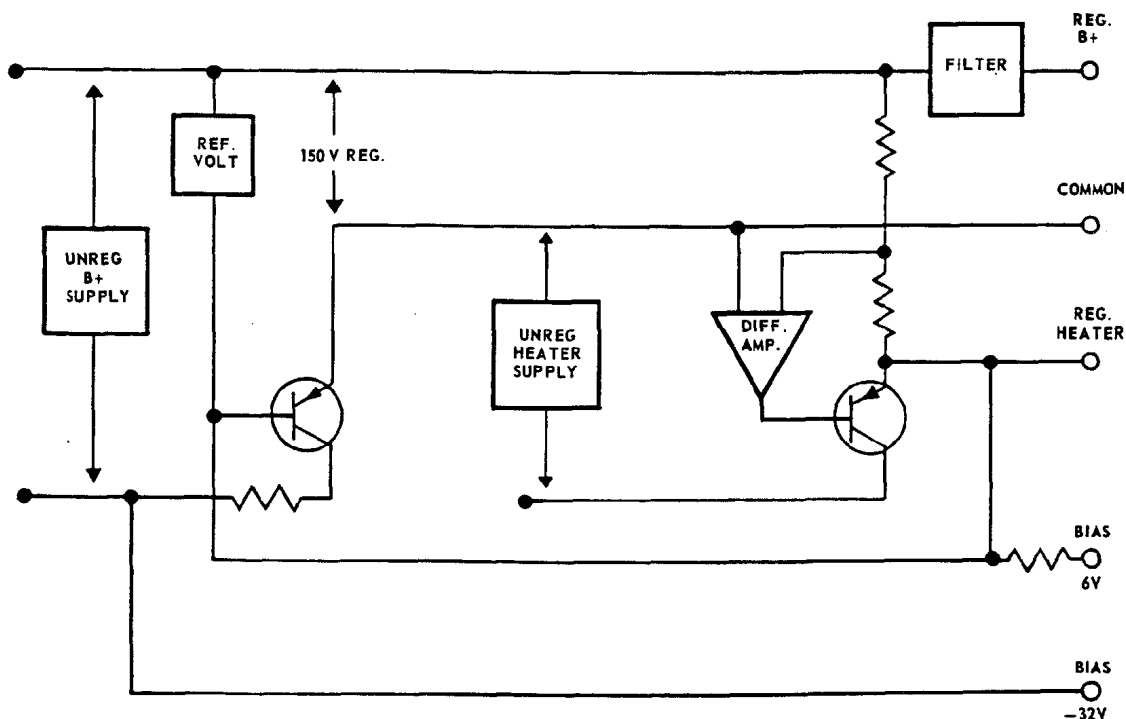


Figure 4-6. Power Supply 3200B

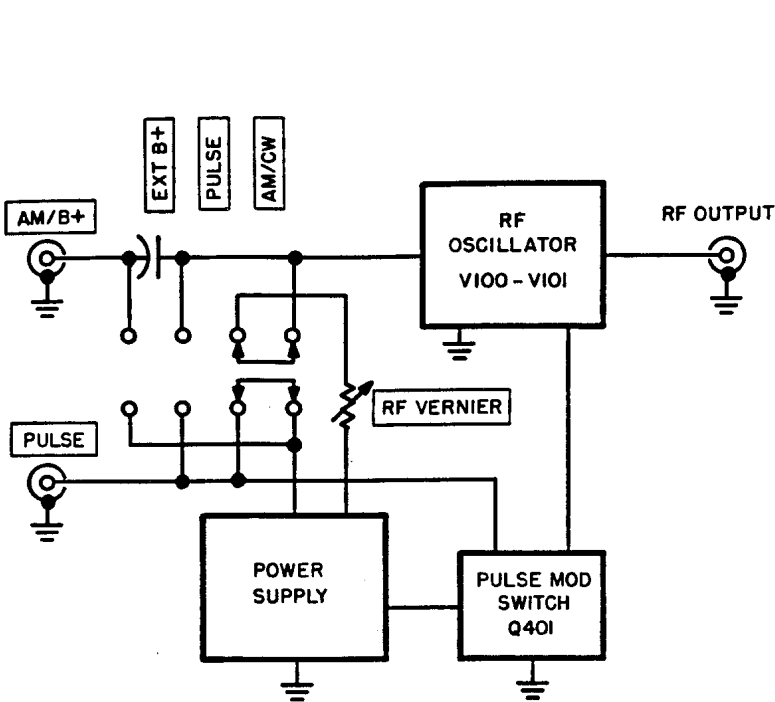


Figure 4-1. Block Diagram 3200B

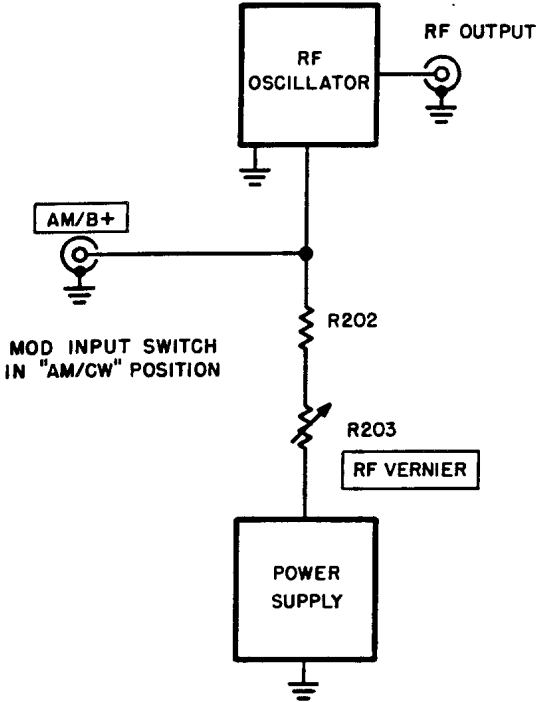


Figure 4-3. Amplitude Modulation
3200B

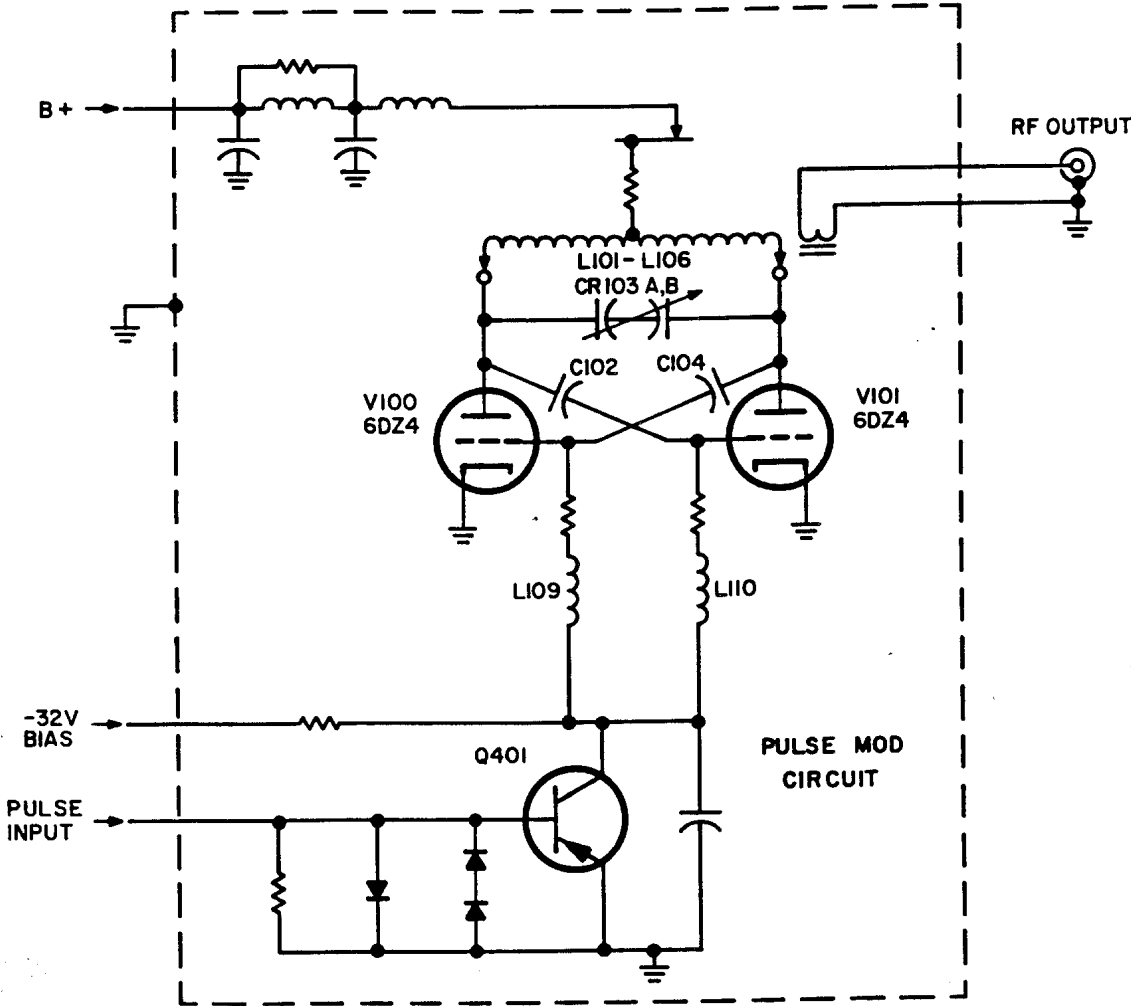


Figure 4-2. RF Oscillator Circuits

SECTION VII
SCHEMATIC DIAGRAMS

7-1. INTRODUCTION

7-2. This section includes manual backdating changes, the schematic diagram and component identification illustrations. Table 7-1 contains the notational information for the schematic diagram of Figure 7-1. The DC voltages drawn on the sche-

matic are typical voltages and are not to be construed as specifications. They were measured with an -hp- 410B Voltmeter.

7-3. For location and identification of mechanical components, refer to Figures 6-1 through 6-5.

TABLE 7-1. SCHEMATIC DIAGRAM NOTES







Resistance is in ohms, capacitance in picofarads and inductance in microhenries unless otherwise specified.

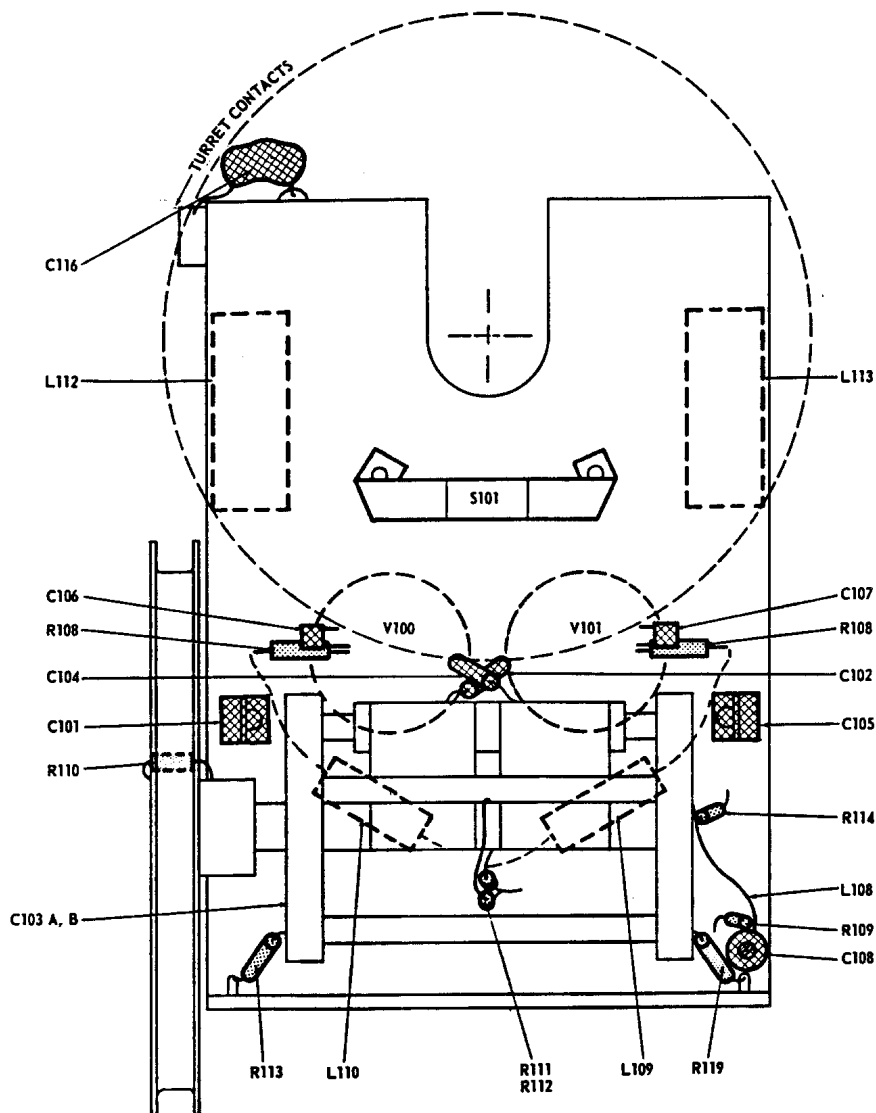
- 2 S101 is operated by cam on turret; contacts close on 10 - 18.8 MHz and 18.5 - 35 MHz ranges only.
- 3 Grid voltage varies with frequency -9 to -25 Vdc.
- 4 S200 is part of R203.
- 5 Use 1/2 amp. SLO BLO fuse for 115 V operation and 1/4 amp. SLO BLO fuse for 230 V operation.
- 6 Due to the interlock feature of V300 (OA2), the Power Supply will not operate with this tube removed from its socket.
- 7 The following capacitors are special and the values are not critical:

| Ref. Sym. | Description |
|-----------|--|
| C117 | Capacitance between ring on pulley and base plate |
| C119 | Capacitance between resistor leads and Rotor Shaft |
| C120 | Capacitance across Rotor thrust bearing |
| C121 | Capacitance between C103 tie rod and C103 Rotor |

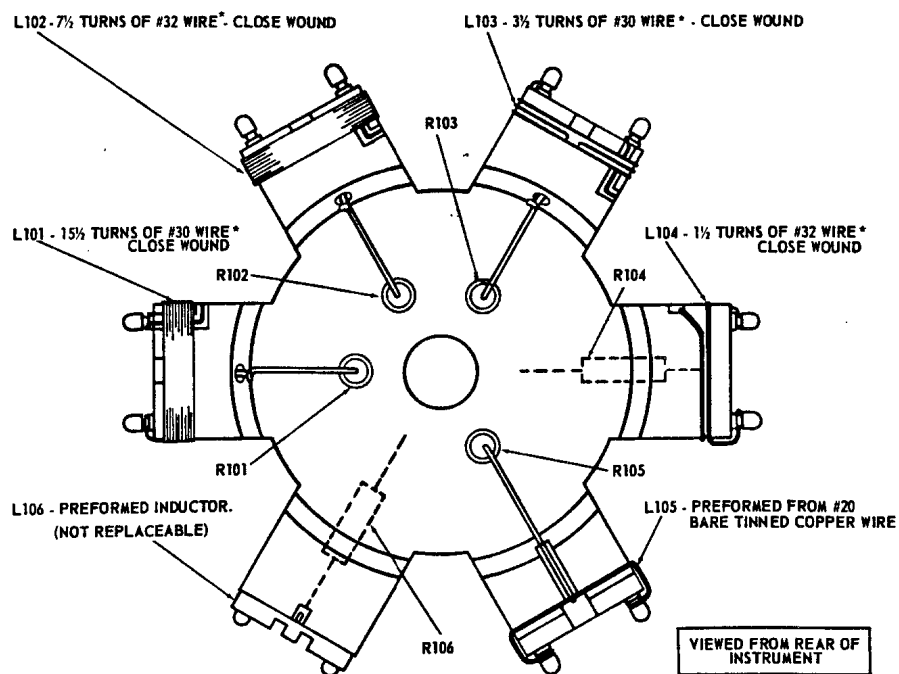
Z100 Inductor vs. Range:

| | |
|------|---------------|
| L101 | 10 - 18.8 MHz |
| L102 | 18.5 - 35 MHz |
| L103 | 35 - 68 MHz |
| L104 | 68 - 130 MHz |
| L105 | 130 - 260 MHz |
| L106 | 260 - 500 MHz |

- 9  Color code of wires using RETMA standard colors vs. numbers, e.g., 6-0-2: Blue-Black-Red
- 10  Front Panel Designation
- 11  Rear Panel Designation
- 12  Knob Control
- 13  Screwdriver Adjustment
- 14  Wiper moves toward CW when control is rotated clockwise



RF TUBE SHELF



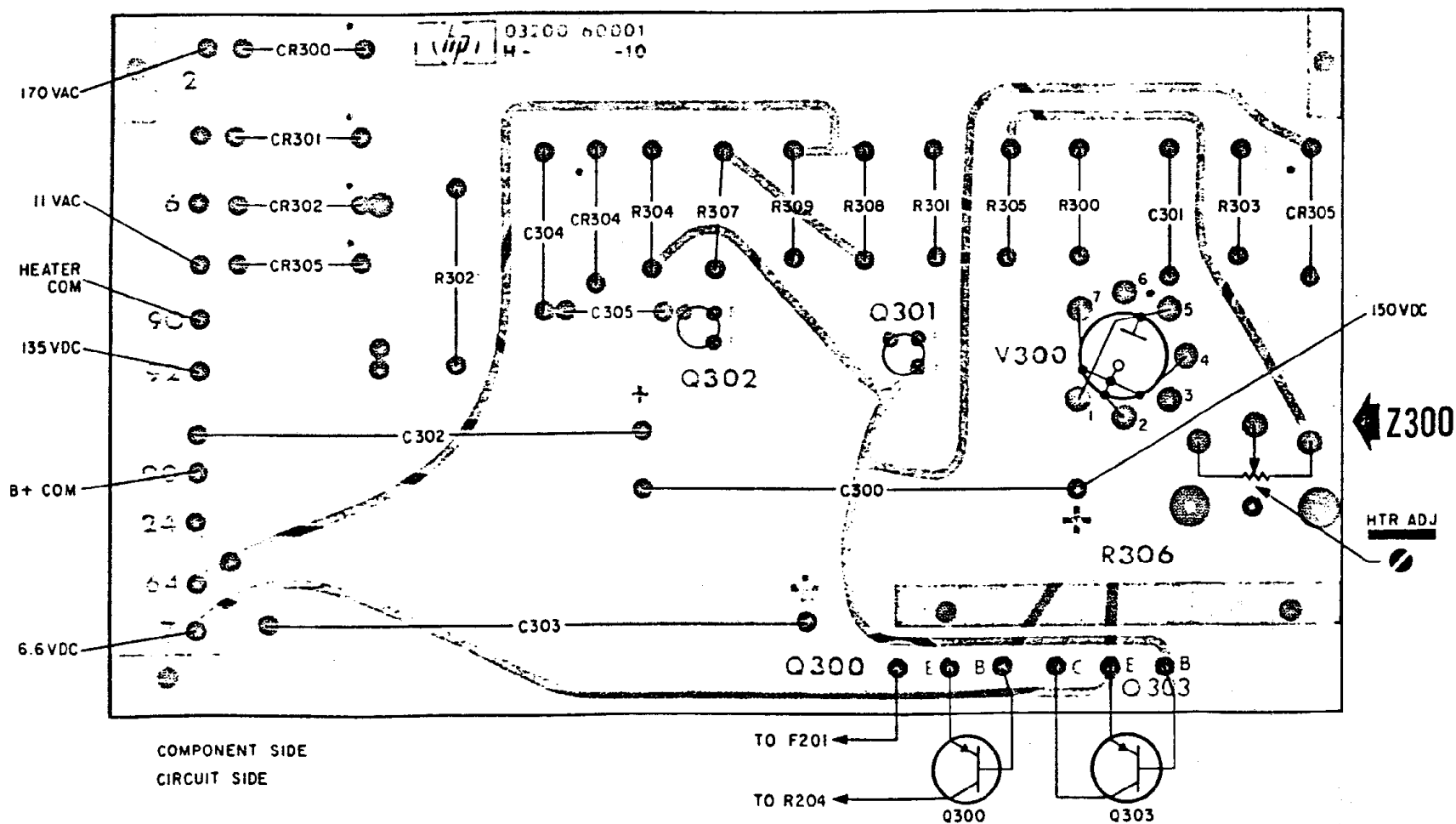
Z100

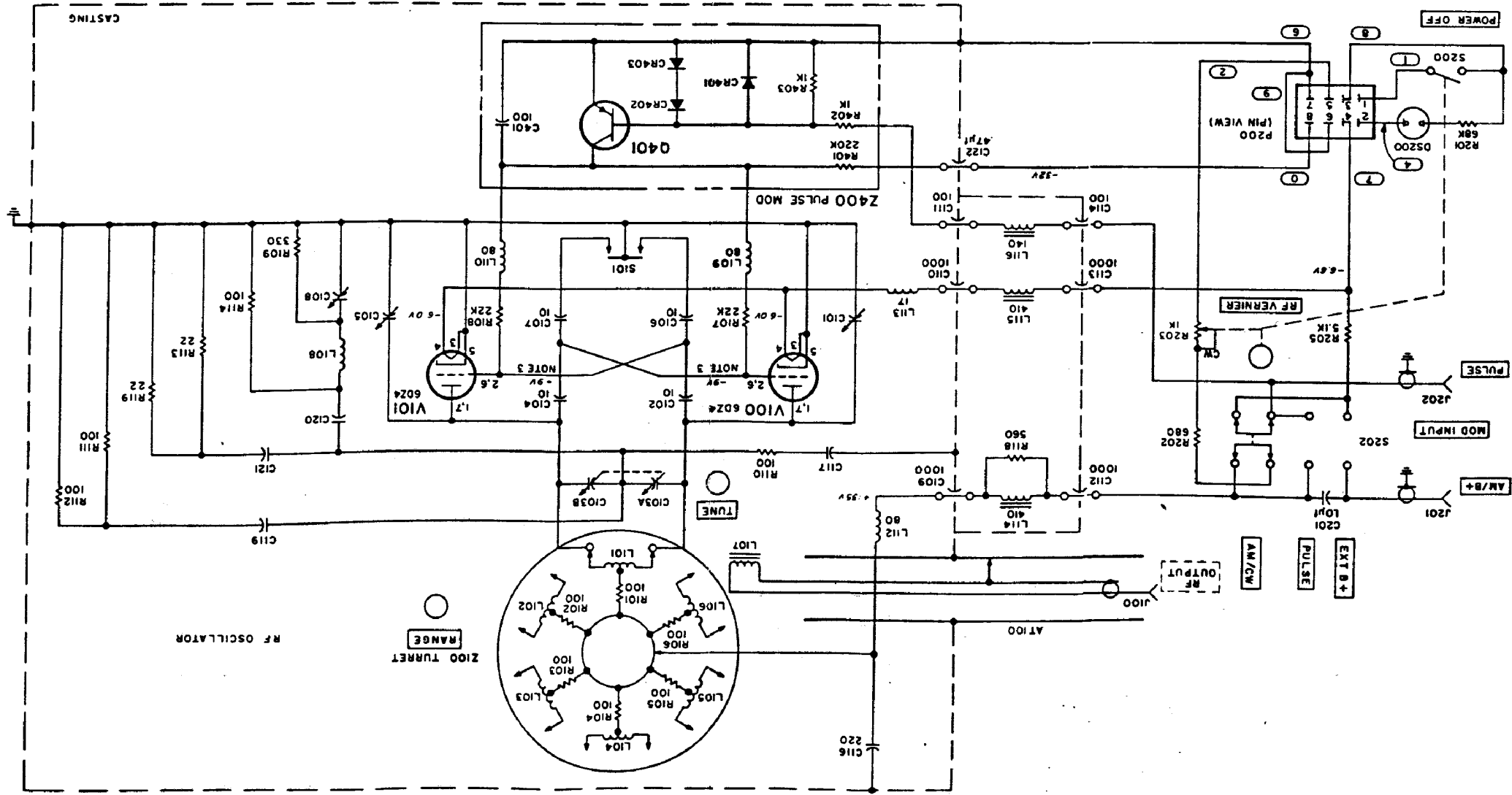
* SINGLE POLYURETHANE COVERED

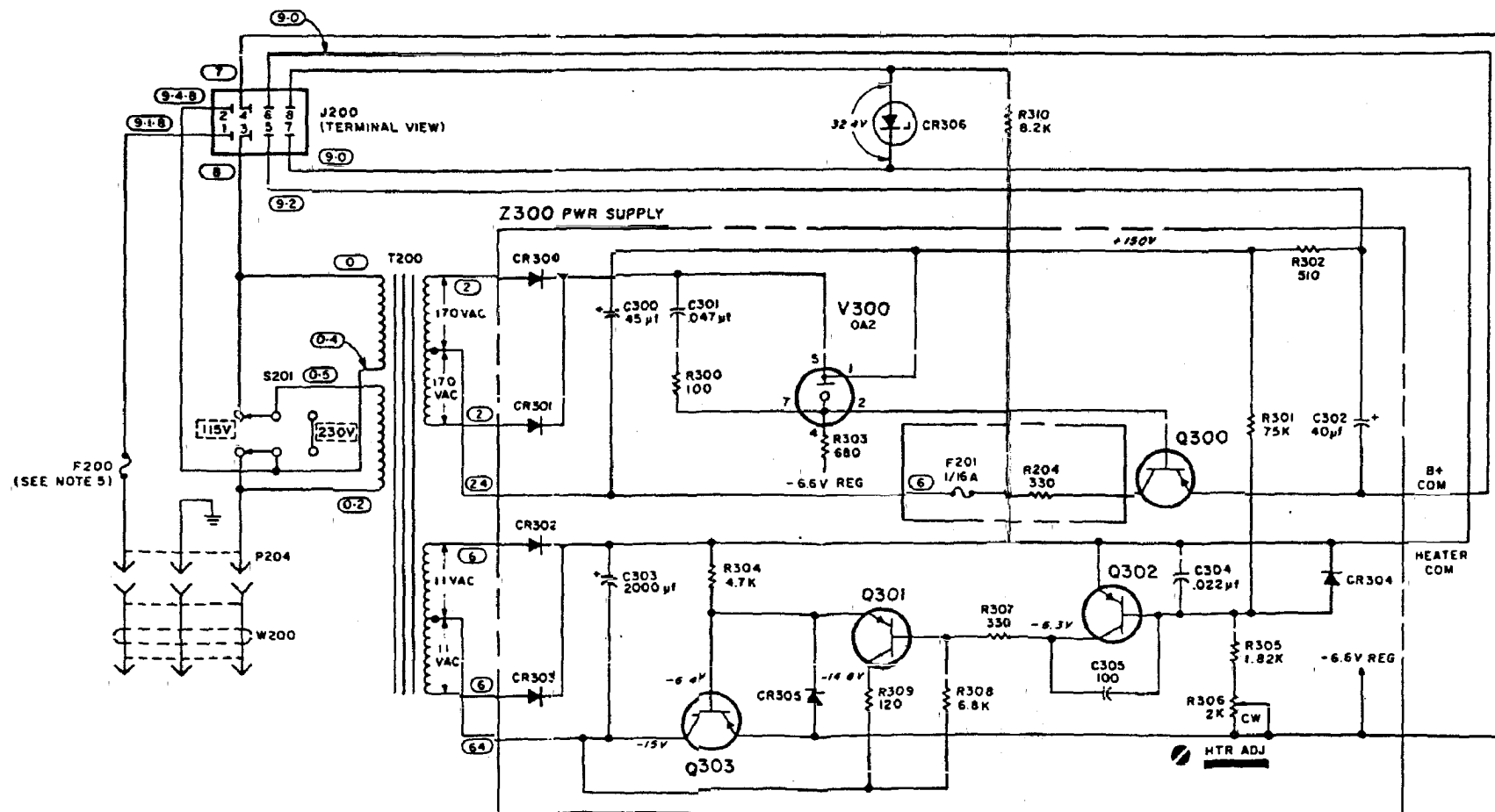
RF TURRET ASSEMBLY



| | | |
|----------------|-----------------|---------|
| PULSE INPUT | PULSE OUTPUT | - 32.4V |
|----------------|-----------------|---------|







REFERENCE DESIGNATORS

| RF OSC | CHASSIS | Z300 | Z400 |
|----------------------------------|-----------------------------|-----------------------------------|---------------------------|
| C101-114, 116-117, 119-122 | C201 DS 200 F200, 201 | C300-305 CR300-305 Q300-303 | C401 CR401-403 Q401 |
| J100 | J200-202 | R300-309 | R401-403 |
| L101-110, 112-116 | P200, 204 R201-205 | V300 | |
| R101-115, 118, 119 | S200-202 T200 | | |
| S101 | W200 | | |
| V100, 101 | | | |
| NOT USED C115, 118 L111 | NOT USED P201-203 | | |
| R115-117 | | | |

MODEL 3200B
VHF OSCILLATOR
SER PREFIXED 735

Figure 7-1
SCHEMATIC DIAGRAM