

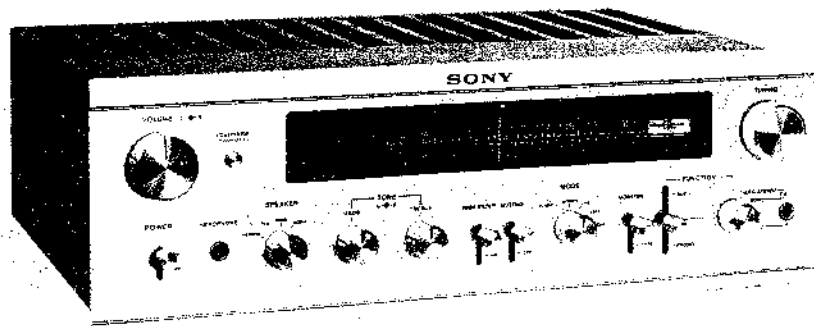


Set using ISD screws

This manual provides service information for all modifications on STR-6055 model.

STR-6055

USA Model
Canada Model
EP Model
UK Model
AEP Model



FM STEREO / FM-AM RECEIVER

SPECIFICATIONS

Fm Tuner Section

Tuning range: 87.5 MHz to 108 MHz
Usable sensitivity: 2.6 μ V (IHF)
2.2 μ V (S/N = 30 dB)
Signal-to-noise ratio: 70 dB
Capture ratio: 1.5 dB
Frequency response: 20 Hz to 15 kHz \pm 1 dB
Stereo separation: 38 dB at 400 Hz

A-m Tuner Section

Tuning range: 530 kHz to 1,605 kHz
Sensitivity: 48 dB/m, built-in bar antenna (S/N = 20 dB)
20 μ V, external antenna
Signal-to-noise ratio: 50 dB

Audio Amplifier Section

Dynamic output power: 145 watts (4 ohms), both channels operating (IHF constant power supply method)
100 watts (8 ohms), both channels operating
Continuous output power: 50 watts (4 ohms) per channel, both channels operating
40 watts (8 ohms) per channel, both channels operating
60 watts (4 ohms), each channel
43 watts (8 ohms), each channel

Harmonic distortion: less than 0.2% at 1 kHz at rated output
less than 0.1% at 1 watt output
Frequency response: PHONO: RIAA curve \pm 0.5 dB
TAPE: 10 Hz to 60 kHz
REC/PB: 10 Hz to 60 kHz

General

Power consumption: 160 watts (USA Model)
200 watts (UK, AEP and EP Model)
160 VA (Canada Model)
Power requirement: 117 volts ac (USA and Canada Model)
100, 117, 220, 240 volts ac (EP, UK and AEP Model)
Dimensions: 440(w) x 148(h) x 345(d) mm
17⁷/₁₆(w) x 5¹¹/₁₆(h) x 13⁷/₁₆(d) inches
Net weight: 12 kg (26 lb 7 oz)

SONY

SERVICE MANUAL

⊗ 134

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SECTION 1 TECHNICAL DESCRIPTION

1-1. SPECIFICATIONS

Fm Tuner Section	
Antenna:	300 ohms balanced
Tuning range:	87.5 MHz to 108 MHz
Sensitivity:	2.6 μ V (IHF usable sensitivity) 2.2 μ V (S/N=30 dB) 1.8 μ V (S/N=20 dB)
Signal-to-noise ratio:	70 dB
Capture ratio:	1.5 dB
Selectivity:	80 dB
Image rejection:	70 dB
I-f rejection:	90 dB
Spurious rejection:	100 dB
A-m suppression:	65 dB
Frequency response:	20 Hz to 15 kHz \pm 1 dB
Separation:	38 dB at 400 Hz
Harmonic distortion:	Mono: 0.2%, IHF (400 Hz 100% Mod) Stereo: 0.5%, IHF (400 Hz 100% Mod)
19 kHz, 38 kHz suppression:	60 dB
SCA suppression:	55 dB
Muting level:	less than 5 μ V

A-m Tuner Section

Antenna:	Built-in ferrite bar antenna with external antenna terminal
Tuning range:	530 kHz to 1,605 kHz
Sensitivity:	48 dB/m, built-in bar antenna (S/N=20dB) 20 μ V, external antenna
Signal-to-noise ratio:	50 dB
I-f rejection:	46 dB at 1,000 kHz
Harmonic distortion:	0.8%

Audio Amplifier Section

Dynamic output power: (IHF)	145 watts (4 ohms), both channels operating (constant power supply method) 100 watts (8 ohms), both channels operating (constant power supply method)
Continuous output power: (rated output)	50 watts (4 ohms) per chan- nel, both channels operating 40 watts (8 ohms) per chan- nel, both channels operating 60 watts (4 ohms), each chan- nel 43 watts (8 ohms), each chan- nel
20Hz~20kHz output power:	30 watts (8 ohms) both chan- nels operating
Power bandwidth:	15 Hz to 30 kHz, IHF
Harmonic distortion:	less than 0.2% at 1kHz at rated output less than 0.1% at 1 watt out- put
Frequency response:	PHONO : RIAA curve \pm 0.5 dB TAPE : 10 Hz to 60 kHz REC/PB : 10 Hz to 60 kHz
Input sensitivity and impedance:	PHONO : 1.8 mV 47k AUX : 140 mV 100k TAPE : 140 mV 100k REC/PB : 140 mV 100k
Signal output and output impedance:	REC OUT : 250 mV 10k REC/PB : 30 mV 80k
Signal-to-noise ratio:	PHONO : greater than 70 dB (weighting network "B") TAPE : greater than 90 dB (weighting network "A") REC/PB : greater than 90 dB (weighting network "A")
Tone controls:	BASS : \pm 10dB at 100 Hz TREBLE : \pm 10dB at 10kHz

	<u>General</u>
Power consumption:	160 watts (USA Model) 200 watts (EP, UK and AEP Model) 160 VA (Canada Model)
Power requirement:	117 volts ac (USA and Canada Model) 100, 117, 220, 240 volts ac (EP, UK and AEP Model)
Dimensions:	440(w) x 148(h) x 345(d) mm 17 ⁵ / ₁₆ (w) x 5 ¹³ / ₁₆ (h) x 13 ⁹ / ₁₆ (d) inches.
Net weight:	12 kg (26 lb 7 oz)
Shipping weight:	16 kg (35 lb 5 oz)

Stage/Control

Function

AFC circuit
D101, D102
C120

L104.

An automatic frequency control circuit is incorporated in the oscillator circuit to eliminate frequency drift and precise tuning difficulty. The principle of afc operation is as follows: When the tuner is correctly tuned, the intermediate frequency is 10.7 MHz and no dc correction voltage is produced by the ratio detector as shown in the "S" curve response of Fig. 1-1. Thus the voltage applied to diode D101 is determined solely by the positive fixed reverse bias voltage supplied by zener diode D102. Now assume that the local oscillator frequency changes by +Δf. This means that the new intermediate frequency is 10.7-MHz +Δf. See Fig. 1-1. As the result a positive dc component is fed back to the anode of D101, decreasing the reverse voltage to it, and making D101's barrier capacitance increase. This decreases the local oscillator's frequency, since the series circuit composed of C120 and D101 is connected in parallel to the tank circuit of the local oscillator. Conversely, if the local oscillator frequency decreases a negative dc voltage is fed back to D101 increasing the local oscillator frequency.

1.2. DETAILED CIRCUIT ANALYSIS

The following describes the function or operation of all stages and controls. The text sequence follows signal paths. Stages are listed by transistor reference designation at the left margin; major components are also listed in a similar manner. Refer to the block diagram on page 12 to 14 and the schematic diagram on page 51 to 56.

Stage/Control

Function

Fm Front End

Balun B1 This transformer matches 300-ohm twin lead to the fm front-end's input stage, thereby coupling the receiver signal to the front-end.

Passive rf circuit A triple-tuned circuit is employed between the antenna and mixer transistor. This passive coupling circuit contains no active amplifiers, so it is perfectly linear and cannot produce distortion and overload components. Thus, the factors that contribute to spurious responses are eliminated ahead of the mixer.

Local oscillator L104, Q102 Supplies heterodyning voltage to the mixer via L104. The circuit is a modified Hartley type with feedback applied to the emitter of Q102 from the tap on

C120

Mixer Q101

IFT101

RF signals and local oscillator voltage are heterodyned in the gate-source junction of mixer Q101 to produce 10.7 MHz i-f output signal.

Transformer IFT101 and capacitor C106 and C107 form a 10.7 MHz "high-C" tuned circuit. This type of circuit has the advantage of reducing the higher order harmonics of 10.7 MHz which cause cross-modulation or spurious interference.

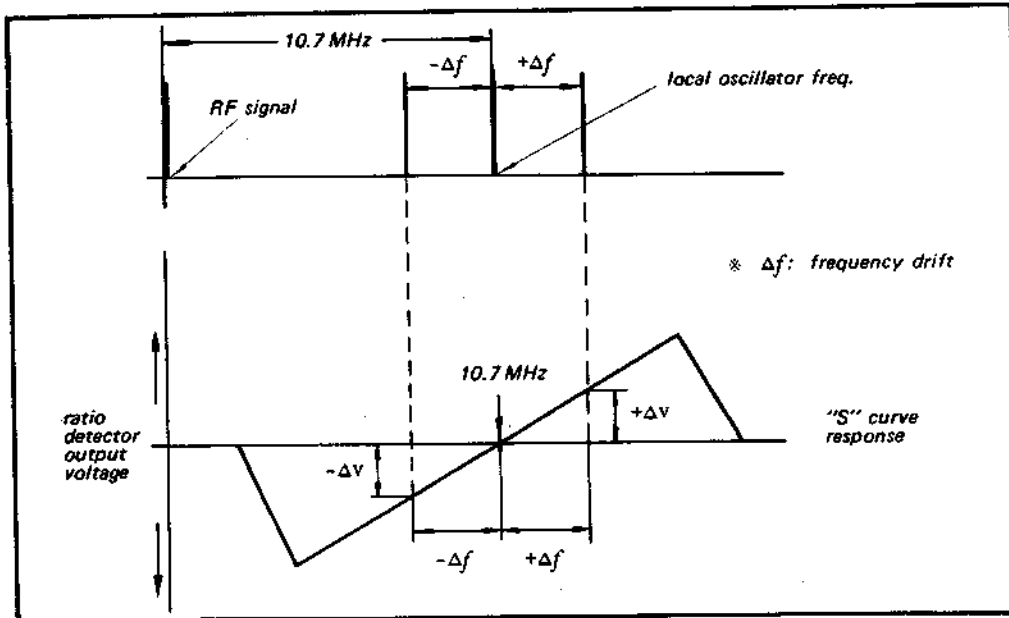


Fig. 1-1. Local oscillator frequency drift and afc voltage relationship

Stage/Control

Function

Stage/Control

Function

I-f preamplifier
Q103

The i-f signal coupled to the base of i-f preamplifier Q103 by the secondary winding of IFT101 is amplified to achieve a favorable signal-to-noise ratio before application to the filters in the i-f strip.

Diode limiters
D201 to D206
D209, D210

filters. Q206 provides power to drive the ratio detector.

Limiting is accomplished by diode pairs, connected in parallel and poled in opposite directions. The diodes conduct when the signal across them exceeds the barrier potential of about 0.6 volts in the forward direction. Thus the signal is limited in both directions to 1.2 volts peak-to-peak. The diodes provide symmetrical limiting.

I-f Amplifier

I-f amplifiers
Q201 to Q205
CF201 to CF206

These i-f stages are basically RC coupled amplifiers (except Q205) that provide essentially flat response. The selectivity of this section is determined by three pairs of filters CF201, CF202, CF203, CF204, CF205 and CF206 in the interstage coupling path. Each of these filters is a two-section ceramic filter that operates in the "trapped-energy" mode. The filters provide extremely-sharp skirt selectivity and flat response inside the passband. Thus, these filters largely determine overall tuner selectivity.

Ratio detector
D207, D208

T201 and diodes D207 and D208 form a balanced ratio detector that transforms the frequency-modulated signal into an audio signal. Output appears across C216.

Muting circuit
Q207, Q208,
Q209, Q210
D211 to D213

The i-f signal is extracted from limiter diodes D203 and D204 to drive the muting circuit. The extracted i-f signal is amplified by Q208 (FET) enough to drive voltage doubler D212 and D211 through tuned transformer T202. D213 provides positive fixed bias for Q209 through D212 and D211. T202 determines the bandwidth necessary to con-

I-f output
Q206

Signal at the base of Q206 has had all amplitude variations removed by the preceding limiters, and only selected signals have been passed by ceramic

Stage/Control

Function

trol the muting circuit without generating interstation or detuning noise. The output of the voltage doubler is a positive dc voltage proportional to the carrier levels of weak rf signals. Q209 and Q210 form a switching circuit which is driven by the voltage doubler. Q209 is normally cut off, thus forcing Q210 into conduction.

The collector of Q210 is connected to the gate of FET Q207 through MUTING switch S7. FET Q207 acts as an electronic switch which is inserted between the ratio detector and MPX decoder, and is controlled by the gate voltage applied.

With the MUTING switch ON, fm signals of average strength keep Q209 saturated, thus cutting off Q210. This causes Q207 to conduct and maintain normal operation.

Weak stations and interstation noise can not produce sufficient dc voltage at the base of Q209 to keep it conducting. As a result, Q209 cuts off.

This saturates Q210 and cuts off Q207, accordingly, the audio output is muted. With the MUTING switch OFF, Q207 is kept conducting regardless of the strength of the fm signal by a positive bias voltage on its gate. RV201 adjusts the muting level.

Stereo-mono automatic-switching circuit Q210, D409

The collector of Q210 is also connected to the output terminal of the MPX decoder's frequency doubler through diode D409.

This prevents noisy stereo reception by automatically switching the MPX decoder's operation into the monaural mode. This is needed because in fm stereo broadcasting, the S/N ratio of a demodulated stereo signal degrades much more rapidly than that of a mono signal when

Stage/Control

Function

the input signal strength decreases. As Q210 is forced into conduction by weak stations, the frequency doubler's output is effectively grounded, stopping the operation of the stereo demodulator. Thus, automatic switching of stereo to mono level is achieved according to the input rf signal.

TUNING meter M901

In fm mode, center-zero meter assures correct tuning by utilizing the ratio detector's characteristic.

R263

As indicated in Fig. 1-1, no dc voltage is produced across R263 when the tuner is correctly tuned. Deflection on the meter indicates the amount of deviation from the carrier frequency received.

Note that the meter will indicate zero-reading when the tuner is not receiving any off-the-air signal.

SCA trap L203, C220

The composite signal containing monaural information from 0 to 15 kHz, the 19 kHz pilot carrier, and the fm stereo signal at 38 kHz is fed to Q207 through trap L203-C220. This trap removes the unwanted SCA signal to the base of Q401 (the 19 kHz amplifier) through Q207.

MPX Decoder

19 kHz amplifier Q401

This stage serves two functions. It extracts the 19 kHz pilot signal by means of a tuned circuit at its drain, and provides a low-impedance source of composite stereo signal (without pilot carrier) at its source. By using an FET, harmonics of the 19 kHz and 38 kHz components are reduced to a low level thereby causing less carrier leak or beat interference.

Frequency doubler D401, D402

Signals developed at the collector of Q401 are transformer coupled to a fullwave rectifier consisting of D401 and D402.

<u>Stage/Control</u>	<u>Function</u>
38 kHz amplifier Q402	The output of this rectifier is not filtered, resulting in two positive pulses for each input cycle. Thus, the 19 kHz pilot-carrier frequency is effectively doubled by D401 and D402. However, the waveform is not sinusoidal at the base of Q402.
STEREO lamp circuit Q403	The 38 kHz pulses produced by D401 and D402 are amplified by Q402. The tank circuit at the collector of Q402 is tuned to 38 kHz to restore these pulses to a sinusoidal waveform. This signal is transformer coupled to the bridge-type demodulator to supply sampling drive for the demodulator.
Multiplex demodulator D405, D406, D407, D408	The STEREO indicator lights when the FUNCTION switch is set to the FM-AUTO STEREO position and an fm stereo signal is received. The emitter of Q402 is connected to the base of Q403 (which is normally cut off). The circuit operates as follows: When a composite stereo signal is applied to the multiplex decoder, the 38 kHz pulses produced at the output of the frequency doubler yield a higher average current flow through Q402. This forces Q403 into conduction, lighting STEREO indicator lamp PL904.
	The demodulator circuit employs four diodes in a balanced-bridge arrangement. This system has the advantage of cancelling residual rf components (38 kHz signal, some 19 kHz signal, and higher-order harmonics of these frequencies.) "L" and "R" components are developed at each side of the bridge as the result of demodulation, when the receiver is operated in the stereo mode. In the monaural mode, diodes D405 and D408 are forward biased by supply voltage through

<u>Stage/Control</u>	<u>Function</u>
De-emphasis capacitors C413, C414, C422, C423	R405, the stereo indicator lamp, R412, R414, and R413, so these diodes merely act as small resistances. Under this condition, the monaural signal is applied to both "L" and "R" audio amplifiers.
Audio preamplifier Q404, Q405 Q406, Q407	These capacitors provide the roll off at high audio frequencies necessary to compensate for pre-emphasis at the transmitter. S10 should be set to the proper time constant. Specified de-emphasis time constant is 75 micro-seconds in USA and CAN-ADA, 50 micro-seconds in Europe.
Separation control RV401	Demodulated L and R signals are amplified by these stages to the level required at the input of the following low pass filter.
LPF401	The network that connects the emitters of Q404 and Q405 provides a form of negative feedback between left and right channels. Any residual "L" signal in the "R" channel (which is about 180° out of phase) is cancelled out by the "L" signal from the "L" channel. The same is true of residual "R" signal in the "L" channel. RV401 is therefore set for maximum separation.
A-m Tuner	Filters out the unwanted higher-order harmonics of 19 kHz and 38 kHz leakage to obtain clear audio.
Local oscillator Q305	Note: As two kinds of a-m circuit exist due to production changes, we describe the each circuit operation corresponding to respective serial numbers. USA Model Serial No. up to 810,400 Canada Model Serial No. up to 700,300 EP Model Serial No. up to 553,000
	This stage supplies injection voltage to the mixer via L303. In

Stage/Control

Function

this oscillator circuit feedback is applied to the emitter of Q305 from low-impedance winding on oscillator coil L304.

Mixer Q301 Incoming rf signals and local-oscillator voltage are heterodyned in the base-emitter junction of Q301 to produce the 455-kHz output. FET Q302 couples the output of the mixer to IFT-301. This stage functions as the gain control element of the AGC system as will be explained later. IFT301 is a transformer tuned to 455 kHz. It develops the i-f signal, and provides a path to ground through bypass capacitor C307 for the other heterodyne products. The low-impedance output winding of IFT301 provides link coupling to i-f amplifier Q303.

I-f amplifier Q303 This stage is basically an RC-coupled amplifier. The selectivity of the stage is determined by a double ceramic filter (CF301) in the interstage-coupling path. This filter provides extremely sharp skirt selectivity inside the passband.

CF301

I-f amplifier Q304 This circuit provides the power to drive diode detector D302.

Detector D302 The i-f signal from the secondary of IFT302 is rectified by diode D302. The i-f components of the output signal are filtered by C317, C318, and R320. The output appearing across R322, R323 is therefore clean audio.

TUNING meter M901 In a-m mode, deflection on this meter indicates the relative signal strength of the selected a-m station. The a-m i-f signal is extracted from secondary winding of IFT302 to drive TUNING METER M901. The extracted i-f signal is rectified by D303. The negative dc component at

Stage/Control

Function

the anode of D303 is filtered by C313, R336, and fed to the TUNING METER.

Agc circuit The negative dc component at the anode of diode D302 is filtered by R325, C336, C335, R335, C322, R326, C323 and fed back to the gate of Q302 to control its gain. The time constants of the agc filter components remove audio variations from the agc voltage.

USA Model..... Serial No. 810,401 and later
 Canada Model..... Serial No. 700,301 and later
 EP Model..... Serial No. 553,001 and later
 UK Model..... Serial No. 600,001 and later
 AEP Model..... Serial No. 900,001 and later

Antenna circuit A-m signals are received by the antenna tank circuit formed by L904, C302, L902, CV901, CT301, C305 and C304. C302 is selected not for its effect upon tuning, but to reduce spurious radiation by the local oscillator.

Low-pass filter L301, C302 The low pass filter (L301 - C302) reduces the spurious radiation caused by local oscillator which may interfere another receiver or communication system through the external antenna.

Local oscillator Q305 This stage supplies the injection voltage necessary to receive a-m signal. In this modified Hartley oscillator circuit, feedback is applied to the emitter of Q305 from a low-impedance winding on oscillator coil T301.

Mixer Q301 Incoming rf signal is fed to the base of Q301, while the local oscillator voltage is injected to the emitter circuit of Q301. These two signals are heterodyned in the base-emitter junction of Q301 to produce the 455-kHz output. This stage functions as the gain control element of the agc system due to Q302 in the emitter circuit, as will be explained later.

CFT301 CFT301 is a combination unit which contains a double-tuned circuit and one ceramic filter

<u>Stage/Control</u>	<u>Function</u>
	tuned to 455 kHz. It develops the i-f signal, and determines the selectivity inside the pass-band. It also provides link coupling to i-f amplifier Q303.
I-f amplifier Q303	This stage is basically an RC-coupled amplifier and amplifies the i-f signal to the proper level required by the following stages.
I-f amplifier Q304	Q304 and IFT301 form a tuned amplifier circuit which provides power to drive diode detector D302.
Detector D302	The i-f signal from the secondary side of IFT301 is rectified by diode D302. The i-f components of the output signal are filtered by C318, R320 and C319 and then cleaned audio signal is fed to the audio pre-amplifier through FUNCTION switch S1.
TUNING Meter M901	The detector (D302) output is also fed to TUNING meter M901 as the dc component in the rectified a-m signal is roughly proportional to the input signal level (not exactly for strong signals due to agc action).
AGC circuit	There are two feedback loops which provide proper agc operation. One is the minor loop applying AGC to the i-f amplifier Q304 base circuit. The other is the major feedback loop applying dc from the emitter circuit of Q304 to the emitter circuit of Q301 through Q302. The minor feedback loop consists of D301, R317, C326, R326, C325 and R314. The a-m i-f signal is extracted from the collector circuit of Q304 through C314 and rectified by diode D301. The output of the diode D301 is a positive dc voltage roughly proportional (not exactly due to agc action) to the carrier levels of input signal and fed to the base of Q304 through a filter circuit. Thus the out-

<u>Stage/Control</u>	<u>Function</u>
	put of diode D301 controls the current flow in Q304 and its emitter voltage as well. Major feedback is produced by the emitter circuit of Q304, R315, C322, C321, R325 and Q302. The emitter voltage of Q304 is applied to the base of Q302 through the filter circuit, determining the positive bias on Q302. As the Q302 shunts the emitter resistor of mixer Q301, it controls the operation of Q301 as a forward agc element. When the strong signal is received, Q302 is forced into conduction, shorting Q301 emitter to ground through R305. As a result, current flow in the Q301 (mixer) increase, reducing its current gain and allowing stable operation in a strong field-strength area.

Preamplifier Section

Equalizer amplifier Q501, Q502	These direct-coupled stages amplify the small signal provided by the phono cartridge to the level required at the input of the following tone-control amplifier.
Bias circuit R503, R508	Dc bias voltage for Q501 is extracted from R508 in the emitter circuit of Q502 and fed back to the base of Q501 through R502 and R503. This dc negative feedback technique provides stable operation during temperature changes.
Equalization circuit R509, R510, R511, R505 C505, C506	RIAA equalization is achieved by the negative-feedback loop containing R509, R510, R511, R505, C505 and C506. Be sure to use replacement components with the exact same values.
Equalization circuit	R513 (R563) in the output circuit prevents interaction between left and right channel-equalization when the MODE switch is set to L+R.

Stage/ControlFunction

MODE switch
S4

In the STEREO position of S4, left and right input signals are routed to their respective amplifiers. In the L+R position, the left and right signals are added and the sum is then fed to both amplifier channels. A rotary switch having two sections is used to obtain L+R signal even if the MONITOR switch is set to the TAPE position.

VOLUME control
RV601 (RV651)

The equalized phono signals and signals applied to the other input terminals are fed to the VOLUME control through the MONITOR and MODE switches. The level of the signal applied to the following tone-control amplifier is determined by the setting of RV601.

LOUDNESS switch
S5

This switch and R601, R602, C601, C602 compensate for the characteristics of the human ear which vary according to the loudness of the sound being heard. When this switch is set to ON and the VOLUME control is set for 30 dB attenuation, the overall frequency response is increased 10 dB at 50 Hz and 4 dB at 10 kHz with reference to the level at 1 kHz.

Tone-control
amplifier
Q601, Q602
(Q651, Q652)

This direct-coupled two-stage amplifier has basically flat response, but it operates as a negative-feedback type tone-control circuit. The output generated at the collector circuit of Q602 is fed back to the emitter circuit of Q601 through the treble and bass tone-control network.

TREBLE control
RV603 (RV653)

Increases or decreases the amount of negative feedback voltage determined by the setting of RV603. It has a range of ± 10 dB at 10 kHz.

Stage/ControlFunction

BASS control
RV604 (RV654)

Similar to the treble control except for filter components and frequency characteristics, however in this circuit the negative feedback voltage is determined by the setting of RV604. This has a range of ± 10 dB at 100 Hz.

HIGH FILTER
S6

The high-cutoff filter (R616 and C613) eliminates unwanted high-frequency components (5 kHz and higher) from the input signal when this switch is ON.

Power Amplifier Section

Preamplifier
Q701, Q702

Q701 and Q702 form a para-phase amplifier but signal output is extracted from the collector circuit of Q701. This circuit has a various advantages in direct coupling system. One is high stability despite temperature variations and another is high input impedance without reducing the amplifier's gain. The ac output appears across load resistor R705 (R755) in the collector circuit. An emitter decoupling circuit is formed by the emitter-base resistance of Q702, C702 and R708 in the base circuit of Q702.

This circuit forms a frequency-selective ac bypass circuit to reduce the amplifier's gain at very low frequencies. Common emitter-resistor R706 keeps the dc current flow constant in the Q701 and Q702, thus increasing dc stability.

Bias power supply
D701, D751

These diodes are forward biased by positive and negative power supply voltage through RV701 and RV751. They provide a stabilized voltage to bias transistor Q701 that is used to make the output terminal balance at zero dc through RV701.

Dc balance adj.
RV701 (RV751)

<u>Stage/Control</u>	<u>Function</u>
Thermal compensation and noise suppressor D711	As all the stages are directly coupled, dc stability is required. The negative temperature coefficient of D711 provides thermal compensation for the following driver stage. It also acts as a noise suppressor to reduce the popping noise due to unbalanced current flow in the following stages when the power switch is turned off.
Driver Q703	Though this stage is a conventional flat amplifier, it determines the output voltage swings because the following stages are basically in the emitter-follower configuration. The ac load resistor for this stage is R712.
Dc bias adj. (idling current) Q704, RV702	Q704 is forced to conduct and operates as a small resistance providing the necessary forward bias on the two cascaded emitter-followers. RV702 controls the base bias of Q704, determining the impedance between the emitter and collector of Q704, and thereby controls the dc bias voltage for the following complementary circuit.
Thermal compensator for dc bias D702	The negative temperature coefficient of D702 provides thermal compensation for the complementary and power transistor circuits. D702 is attached to the power transistor's heat sink to detect temperature change in the power transistors.
Complementary circuit Q707, Q708	These transistors operate as emitter-followers to provide the current swings demanded of the output stages and also provide the necessary phase inversion. Phase inversion is performed by using PNP and NPN type transistors.
Power transistor Q709, Q710	The output transistors (Q709 and Q710) are connected directly to a power supply of

<u>Stage/Control</u>	<u>Function</u>
	about $\pm 40V$. Q709 supplies power to the load during the positive half cycle and Q710 operates during the negative half cycle. As all the stages are directly coupled and designed to obtain zero potential at the output terminal, the large coupling capacitor at the output (which may cause power loss or distortion at low frequencies) is eliminated.
Protection circuit	To protect overloaded power transistors from destruction, a new protection circuit is employed. If a short circuit at the output terminals occurs, the protection circuit holds the current in the power transistor low enough not to make it overheat and also limit the input drive signals. Fig. 1-2 shows a partial schematic diagram detailing the protection circuit. With reference to this diagram, the protection circuit operates as follows: Since the protection circuit is identical for positive going half cycles and negative going half cycles, only the positive going half cycle operation is described here. Q705 limits the positive-going half cycle of the drive voltage applied to the base of Q707 when power consumption at the Q709 collector exceeds the safety margin. Since power dissipation at the collector can be considered a function of collector voltage and current, the trigger signal for Q705 is taken from the collector and emitter. Base voltage is partly determined by the ratio of resistance of R719 and series resistance of R726 and RL (load). Base voltage is also determined by the current flow in the R733 and the collector voltage of Q709. Under normal operation, Q705 is cut off.

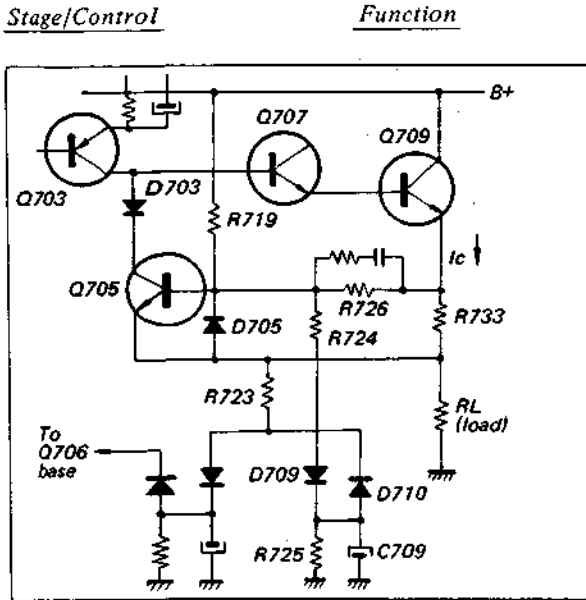


Fig. 1-2. Simplified protection circuit

When excessive current flows in the power transistor or power dissipation at the collector of power transistor exceeds the specified value, Q705 turns on and limits the input drive voltage to the power transistor. Limiting operation is also actuated by the condition of the load. The base voltage of Q705 is determined by the resistances R733, R726, R724, R725, and RL (load). D709 is employed to stop reverse voltage from applied being during the negative going-half cycle. Q705 turns limiting the input drive voltage to the power transistor when the load resistance decreases to some extent. Under reactive load conditions in class B amplifiers maximum current will flow when the voltage across the power transistor is maximum and this is the worst case for secondary breakdown. See Fig. 1-3. As all speakers have reactive properties, a protection circuit which covers the reactive region is required.

Fig. 1-3 shows the operating load lines for one half of a class B output stage under con-

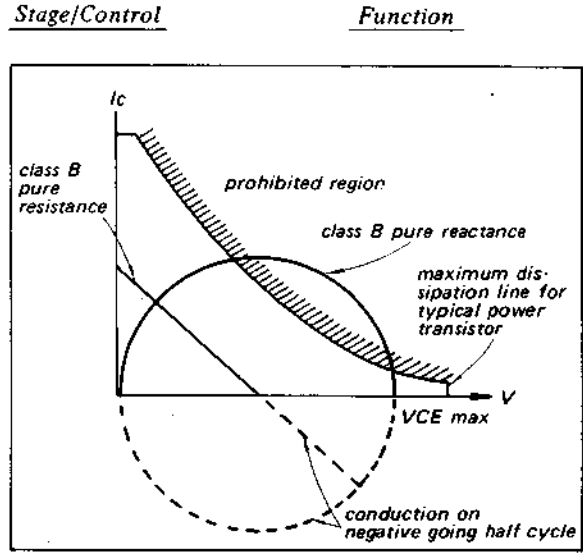


Fig. 1-3. Resistive and reactive load lines for class B output stage showing breakdown risk in purely resistive load

ditions of equal load impedance; in one case the load is purely resistive and in the other case purely reactive. It is apparent that the reactive load case could result in transistor failure. D710, C709 and R723 form a circuit charging the base voltage according to the reactive voltage induced in the load to obtain proper protection operation. C709 and R725 form a discharging circuit to detect reactive dc voltage. D705 protects Q705 from breakdown between base and emitter due to detected reactive voltage across C709. D703 protects Q705 from the breakdown between collector and emitter during the negative-going half cycle.

Power supply
rectifier
D801

A full-wave bridge rectifier provides a positive and a negative dc power supply for the power amplifier.

Rectifier
D802

A half-wave rectifier (D802) and ripple filter (C809, R801, R802 and C810) supply well-filtered dc power to the preamplifier section.

Stage/ControlFunction

Ripple filter
Q711
R741, R740
C714, C713
Q761, R791
R790,
C764, C763

These components reduce the ripple voltages in the dc power supply for the preamplifier and driver stages of the power amplifier section to an extremely-low value. Q711 and Q761 serve as an electronic filter to supply well filtered dc voltages of about $\pm 37V$ to each stage.

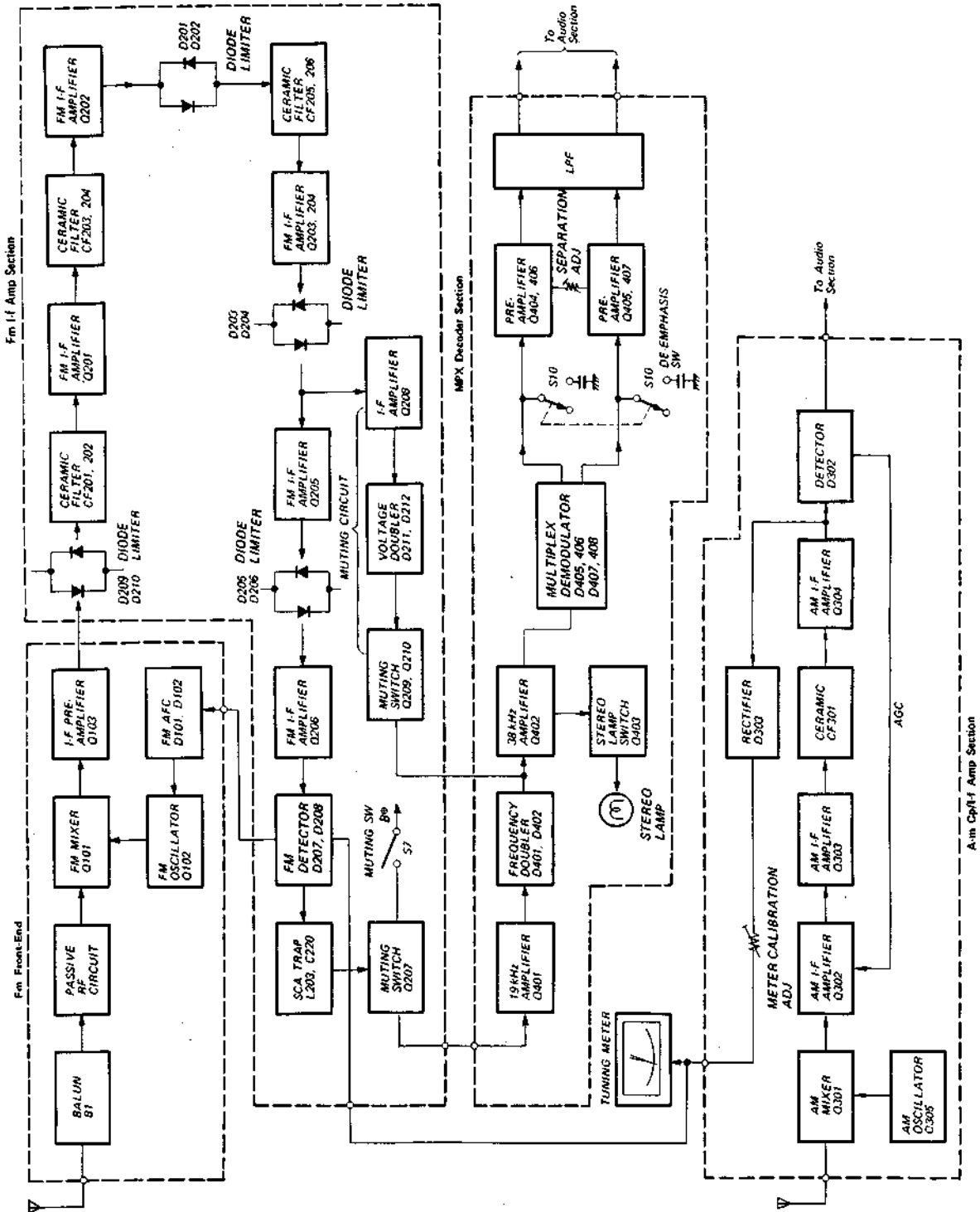
Stage/ControlFunction

Voltage regulator
Q801, D803, D804

Dc output from the rectifier is filtered by C807 and applied to series regulator Q801. Since the voltage at the base of Q801 is kept constant by means of zener diodes D803 and D804, the emitter voltage remains constant regardless of load or line-voltage variations. The regulated and well filtered output of 15V is supplied to the tuner section.

1-3. BLOCK DIAGRAM - Tuner Section -

Note: Applicable Serial Numbers:
 USA Model Up to 810,400
 Canada Model Up to 700,300
 EP Model Up to 553,000



- Tuner Section -

Note: Applicable Serial Numbers:

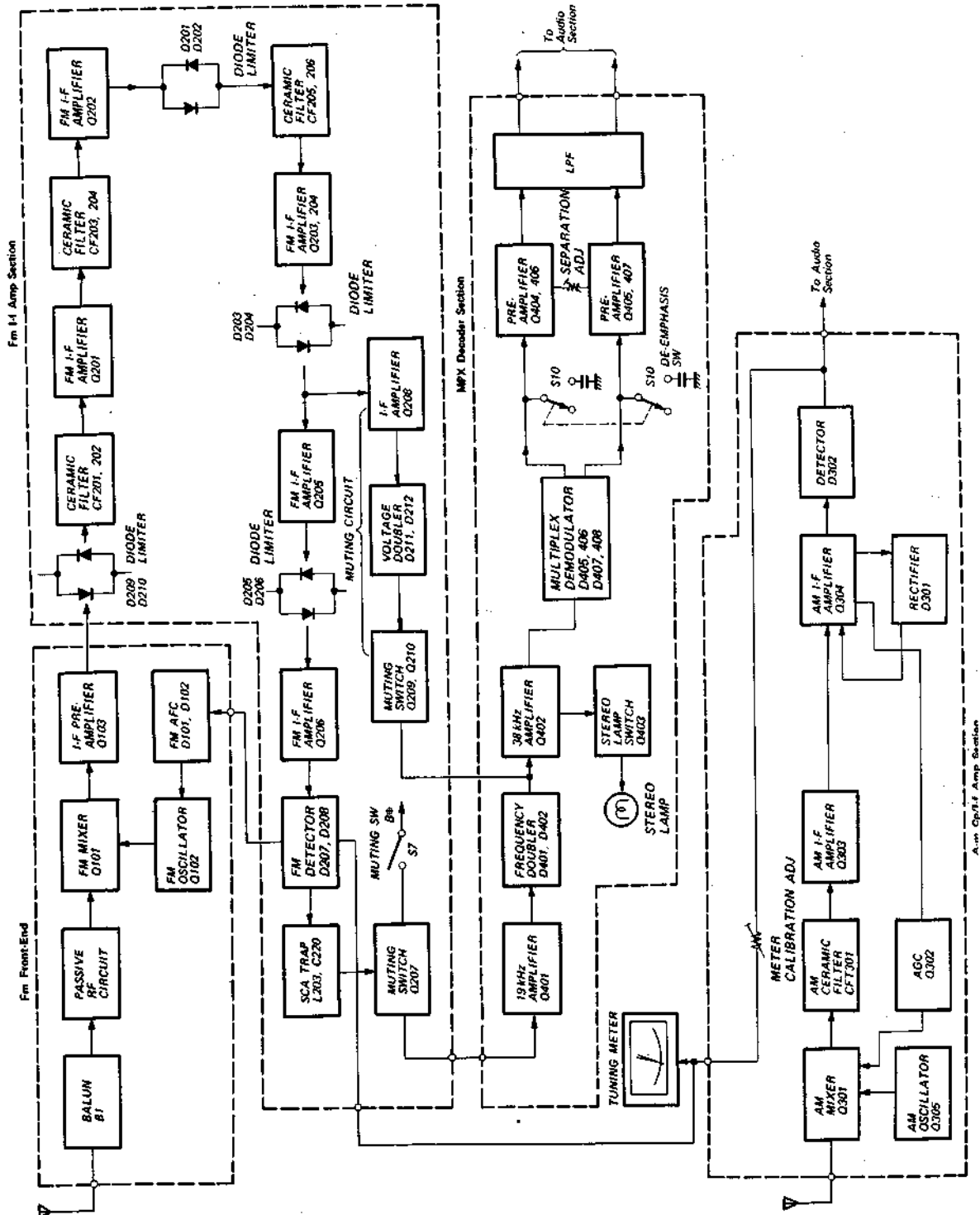
USA Model.....810,401 and later

Canada Model700,301 and later

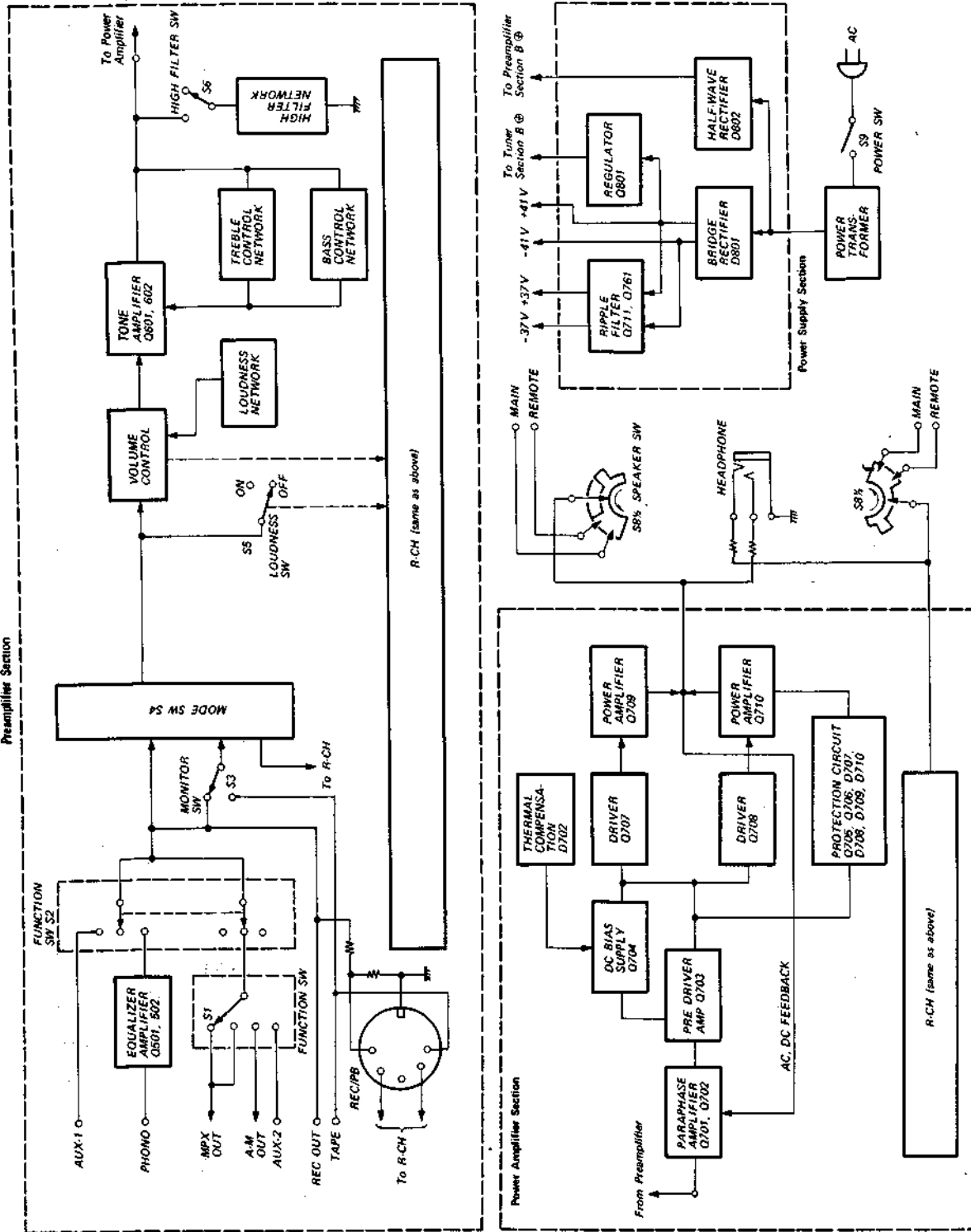
EP Model563,001 and later

UK Model600,001 and later

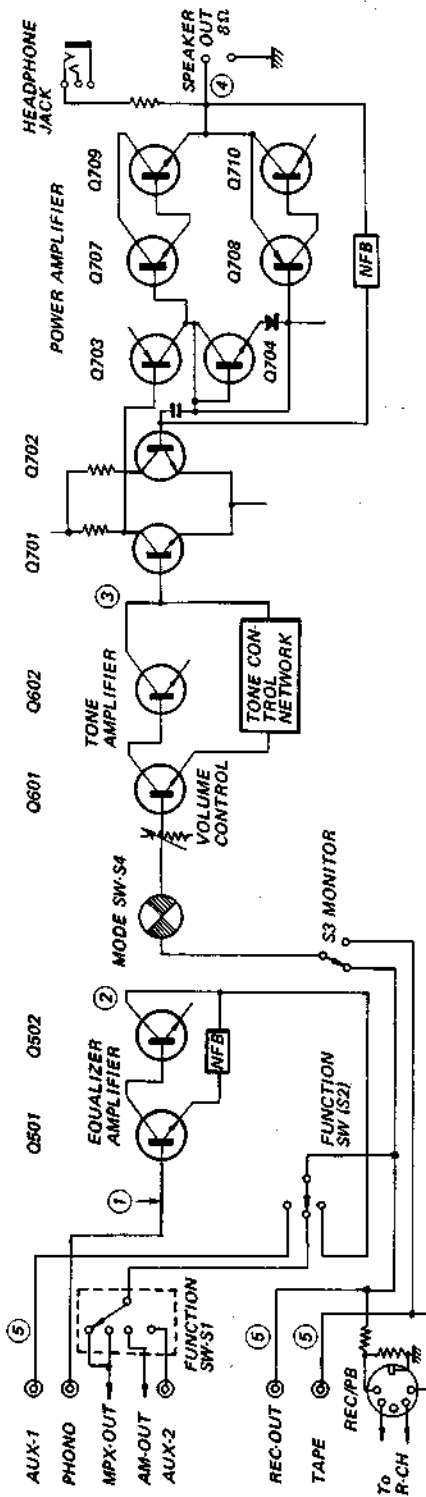
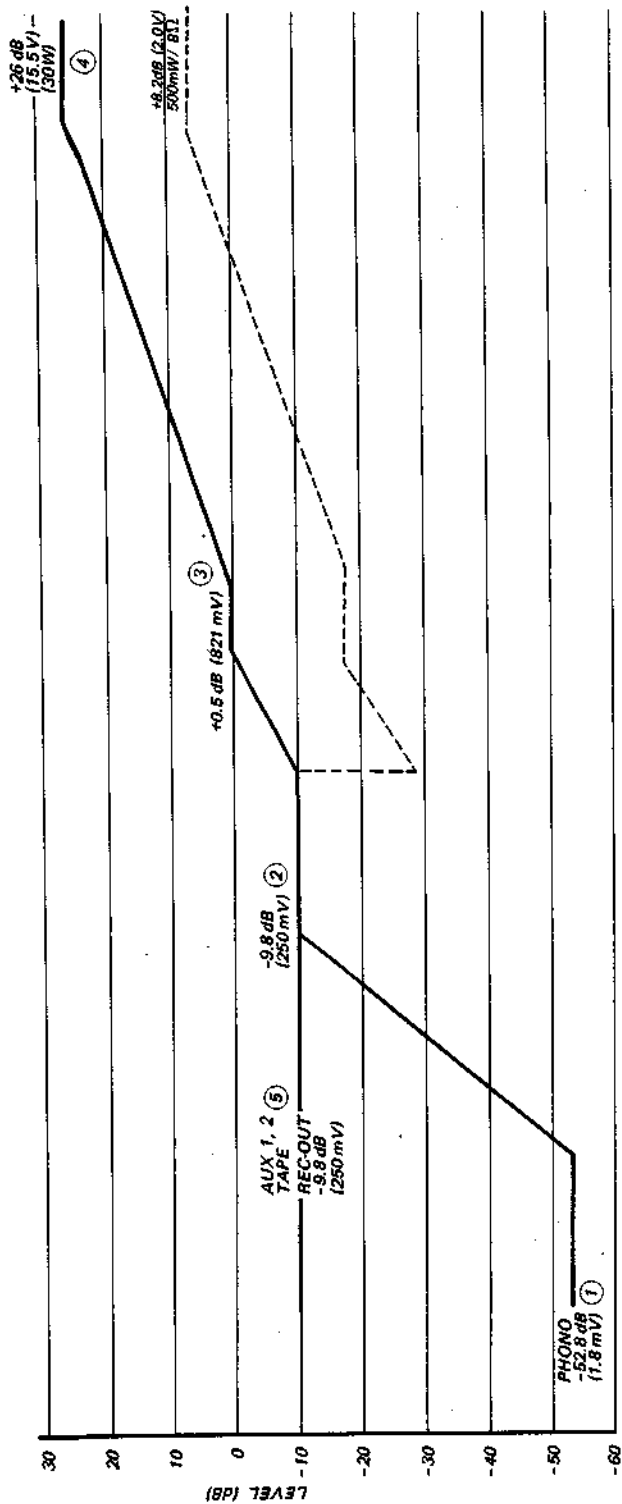
AEP Model.....900,001 and later



- Audio Amp Section -



1-4. LEVEL DIAGRAM



Note: Signal voltages are measured with an ac VTVM and expressed in dB referred to 0.775V, 1 kHz.

**SECTION 2
DISASSEMBLY AND REPLACEMENT**

WARNING

Unplug the ac power cord before starting any disassembly or replacement procedures.

2-1. TOOLS REQUIRED

The following tools are required to perform disassembly and replacement procedures on the STR-6055.

1. Screwdriver, Phillips-head
2. Screwdriver, 1/8" blade (3 mm)
3. Pliers, long-nose
4. Diagonal cutters
5. Wrench, adjustable
6. Tweezers
7. Electric drill
8. Drill bits
9. Prick punch
10. Hammer, ball-peen
11. Soldering iron, 40 ~ 150 watts
12. Solder, rosin core
13. Cement solvent
14. Contact cement
15. Thermal compound or silicone grease

2-2. HARDWARE IDENTIFICATION GUIDE

The following chart will help you to decipher the hardware codes given in this service manual.

Note: All screws in the STR-6055 are manufactured to the specifications of the International Organization for Standardization (ISO). This means that the new and old screws are not interchangeable because ISO screws have a different number of threads per mm compared to the old ones. The ISO screws have an identification mark on their heads as shown in Fig. 2-1.

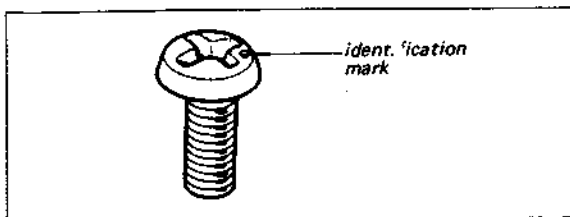


Fig. 2-1. ISO screw

— Hardware Nomenclature —

P	- Pan Head Screw		
PS	- Pan Head Screw with Spring Washer		
K	- Flat Countersunk Head Screw		
B	- Binding Head Screw		
RK	- Oval Countersunk Head Screw		
T	- Truss Head Screw		
R	- Round Head Screw		
F	- Flat Fillister Head Screw		
SC	- Set Screw		
E	- Retaining Ring (E Washer)		

W - Washer
 SW - Spring Washer
 LW - Lock Washer
 N - Nut

- Example -

2-3. TOP COVER AND BOTTOM PLATE REMOVAL

1. Remove the two machine screws at each side of the receiver, and lift off the top cover.
2. Remove the six self-tapping screws (\oplus B 3x6) at the bottom of the receiver and pull the bottom plate in the direction indicated by the arrow in Fig. 2-2.

2-4. FRONT PANEL REMOVAL

1. Remove the top cover as described in Procedure 2-3.
2. Pull all the knobs off.

- Remove the two screws ($\oplus B 3 \times 6$) and two hex-nuts securing the front panel to the front subchassis as shown in Fig. 2-3. Place a piece of cardboard or cloth between the wrench and front panel to avoid marring the panel as shown in Fig. 2-4. Now the front panel is free for servicing.

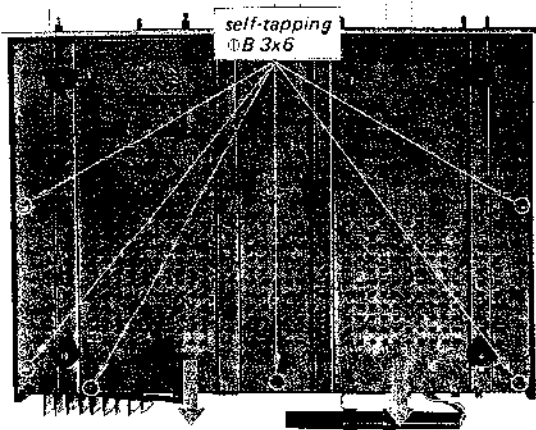


Fig. 2-2. Bottom plate removal

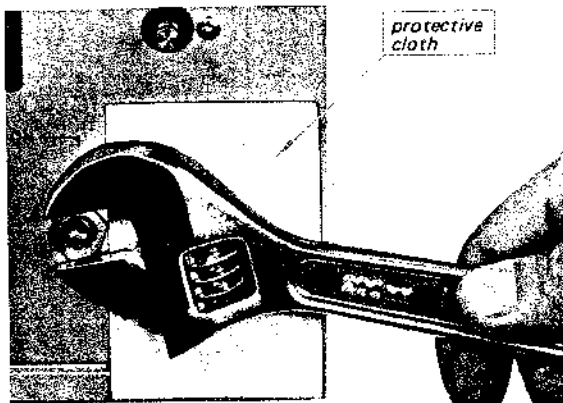


Fig. 2-4. Hex nut removal

2-5. DIAL CORD RESTRINGING

Preparation

- Remove the top cover as described in Procedure 2-3.
- Cut a 1,500 mm (59") length of dial cord.
- Tie the end of the cord to a spring as shown in Fig. 2-5.
- Rotate the tuning-capacitor drive drum fully clockwise (minimum capacitance position).

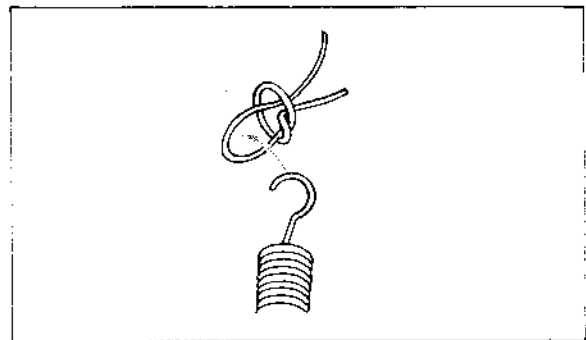


Fig. 2-5. Tying square knot to the coil spring

Procedure

While referring to Fig. 2-6, proceed as follows:

- Hook the spring to one hole of the drive drum as shown in Fig. 2-7.
- Run the cord through the slot in the rim of the drum and wrap a half clockwise turn in the inner groove.
- Run the cord over pulley "A", and then wrap two counterclockwise turns around the tuning shaft.

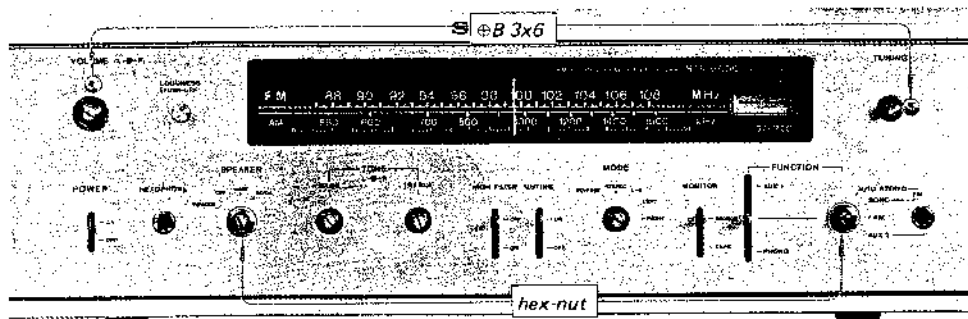


Fig. 2-3. Front panel removal

X

4. Run the cord over pulleys "B", "C" and "D", then wrap two clockwise turns around the drum from outer groove to inner groove as shown in Fig. 2-8.
5. Pass the doubled end of the cord through the eyelet, then hook it to the spring as shown in Fig. 2-9.

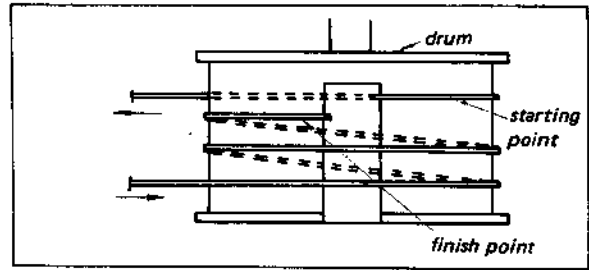


Fig. 2-8. Wrapping the dial cord

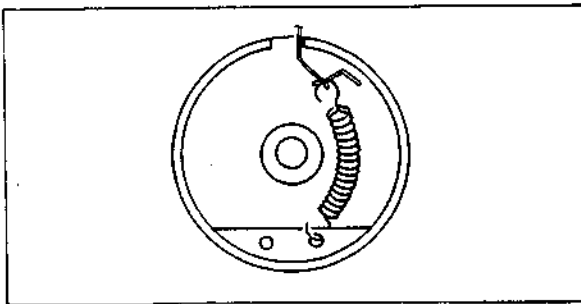


Fig. 2-7. Coil spring installation

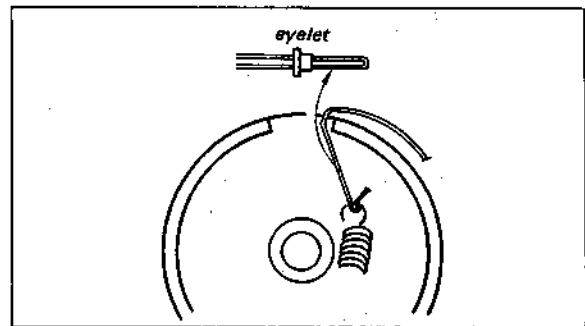


Fig. 2-9. Finishing dial cord stringing

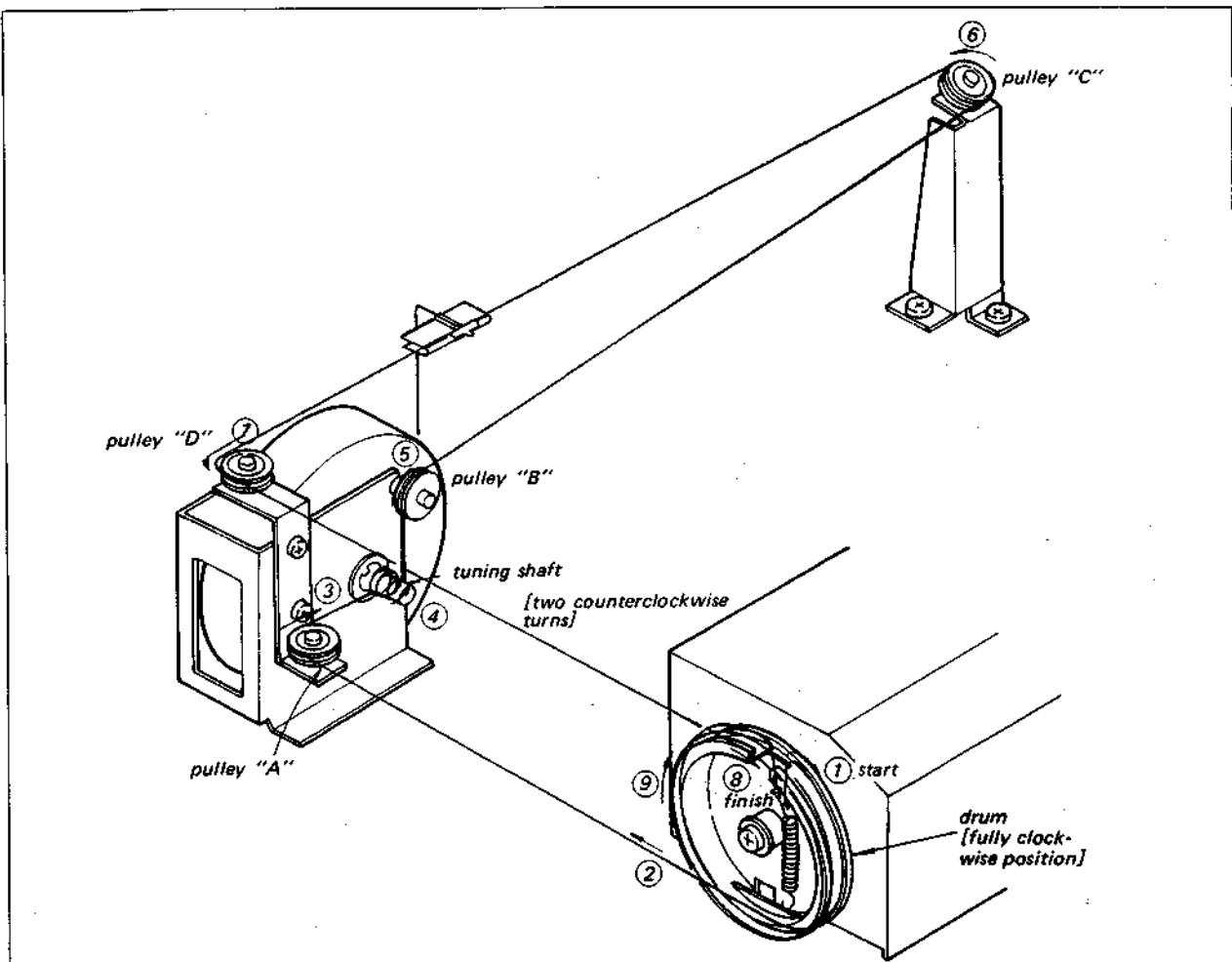


Fig. 2-6. Dial cord stringing

6. Tighten the cord, then squeeze the eyelet so that the spring is under tension. Make a knot at the cord end to keep it from slipping out of the eyelet.
7. After completing the dial-cord stringing, make sure that the tuning system works properly. Apply a drop of contact cement to the finish point.
8. Perform the mechanical dial calibration by following the procedure 2-6.

2-6. MECHANICAL DIAL CALIBRATION

Note: This is required after replacing the dial cord, dial scale or front-end assembly.

1. Put the dial pointer on the cord as shown in Fig. 2-10 and move it to a position where the pointer coincides with the left gap on the dial scale as shown in Fig. 2-11, when the tuning capacitor is set to the maximum capacitance.
2. Apply a drop of contact cement to the tab of the dial pointer.

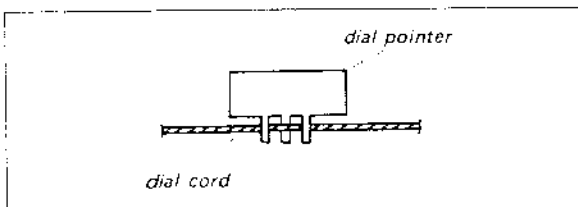


Fig. 2-10. Dial pointer installation

2. Remove the front panel as described in Procedure 2-4.
3. Remove the two screws (\ominus P 2.6x4) securing the dial-scale holder to the front subchassis as shown in Fig. 2-12.
4. Remove the defective dial scale, and then install the replacement scale.

2-8. PILOT LAMP REPLACEMENT

Prepare for replacement any of the pilot lamps by removing the top cover as described in Procedure 2-3.

Meter Lamp

1. Straighten the tab of the meter-lamp holder to permit the removal of the meter-lamp socket.
2. Pull out the meter-lamp socket, and then unscrew the lamp from the socket and install the new lamp.

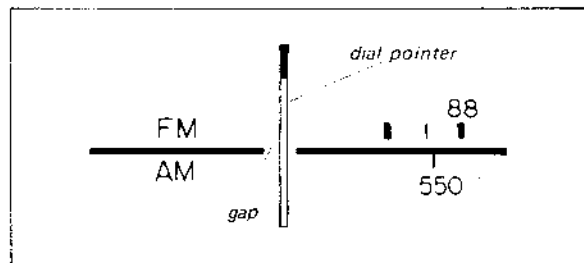


Fig. 2-11. Mechanical dial calibration

2-7. DIAL SCALE REPLACEMENT

1. Remove the top cover as described in Procedure 2-3.

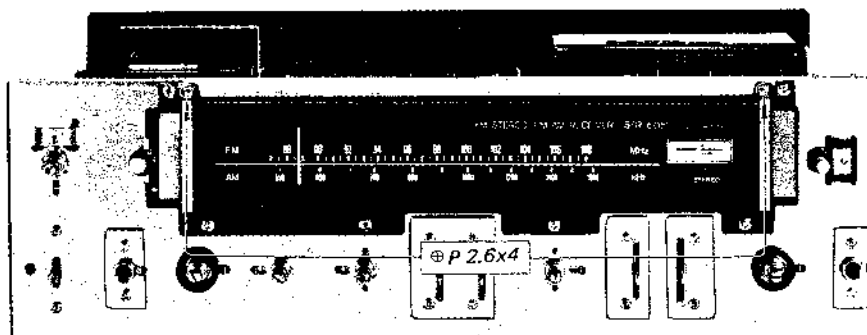


Fig. 2-12. Dial scale removal

Stereo Lamp

1. Remove the two self-tapping screws ($\oplus B 3 \times 6$) securing the meter holder to the chassis as shown in Fig. 2-13.

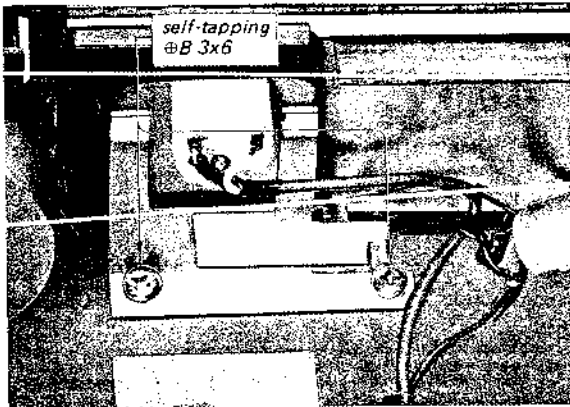


Fig. 2-13. Meter holder and stereo lamp removal

2. Pull the lamp from its holder with tweezers.
3. Cut the lamp leads and solder the lead wires to the new lamp as shown in Fig. 2-14.
4. Wrap the soldered connections with electrical tape as shown in Fig. 2-14.
5. Install the new lamp in its holder.

Dial Lamp

1. Remove the front panel as described in Procedure 2-4.
2. Pry out the fiber lamp shade, and then remove the lamp.

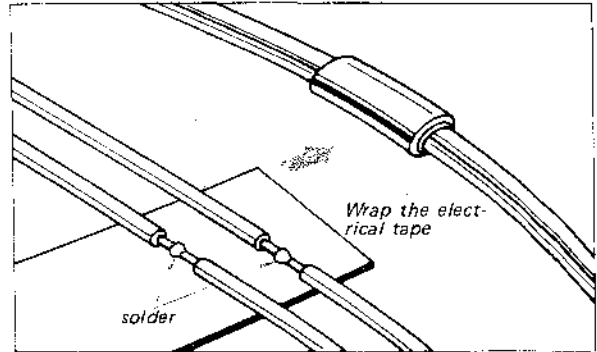


Fig. 2-14. Stereo lamp replacement

2-9. TUNING METER REPLACEMENT

1. Remove the top cover as described in Procedure 2-3.
2. Unsolder the leads from the defective meter.
3. Remove the two self-tapping screws ($\oplus B 3 \times 6$) securing the meter holder to the chassis as shown in Fig. 2-13.
4. Remove the meter, and install a new one.

2-10. CONTROL AND SWITCH REPLACEMENT

Prepare for replacing any of the controls or switches by removing the top cover and the front panel as described in Procedures 2-3 and 2-4. Refer to Fig. 2-15.

TONE Controls

1. Apply a drop of cement solvent to the ring spacer on the TREBLE control. Wait a few seconds for the cement to dissolve, and pry out the spacer with a screwdriver.

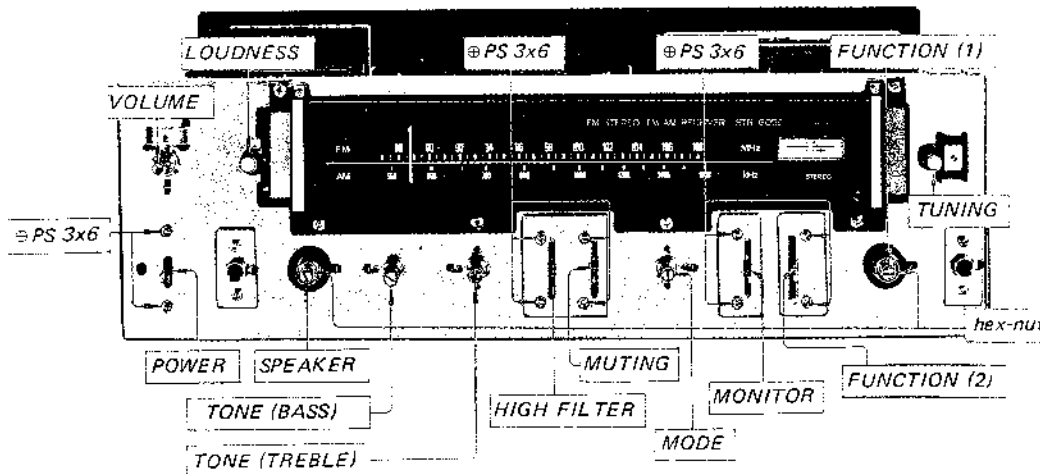


Fig. 2-15. Control and switch replacement

2. Remove the hex nuts that secure the BASS and TREBLE controls to the front-subchassis.
3. Carefully remove them along with the tone-control circuit board.
4. Cut each lug of the defective control on the board to remove the part.
5. Unsolder and remove the clipped lugs, and clean out the holes of the circuit board.
6. Install the replacement control.

POWER, HIGH FILTER, MUTING, MONITOR, FUNCTION (2) Switches

1. Remove the two screws (ØPS 3x6) securing switches to the front subchassis as shown in Fig. 2-15.
2. Unsolder the lead wires from the defective switch, and then install the replacement switch.

SPEAKER, MODE, FUNCTION (1) Switches

1. Apply a drop of cement solvent to the ring spacer on the switches. Wait a few seconds for the cement to dissolve, and pry out the spacer with a screwdriver.
2. Remove the hex nuts that secure the switches to the front-subchassis as shown in Fig. 2-15.
3. Unsolder the lead wires from the defective switch, and then install the replacement switch.

LOUDNESS Switch

1. Fasten the dial cord to the drum with cellophane tape.

2. Remove the two self-tapping screws (Ø B.3x6) (■) or two screws (Ø B 3x6) (●) securing the dial pulley bracket to the chassis as shown in Fig. 2-16.

Note: ■ (AEP and UK Model)
● (USA, Canada and EP Model)

3. Put the bracket aside, and then remove the screw (Ø B 2.6x4) securing the loudness switch to the front subchassis.
4. Remove it along with the loudness control board, and then install the replacement switch or replacement mounted circuit board including loudness switch (Part No. X-48030-23-0).

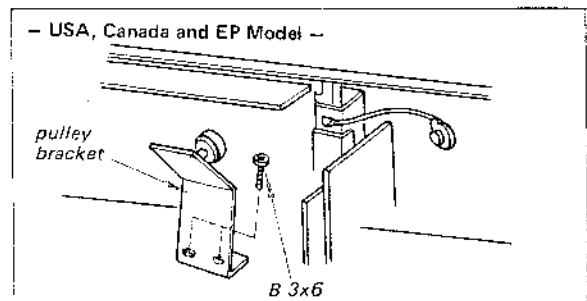
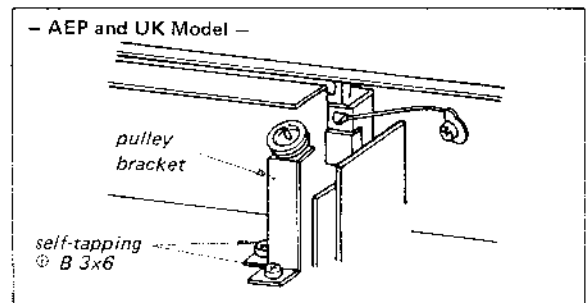


Fig. 2-16. Dial pulley bracket removal

2-11. REAR PANEL REMOVAL

1. Remove the top cover and bottom plate as described in Procedure 2-3.

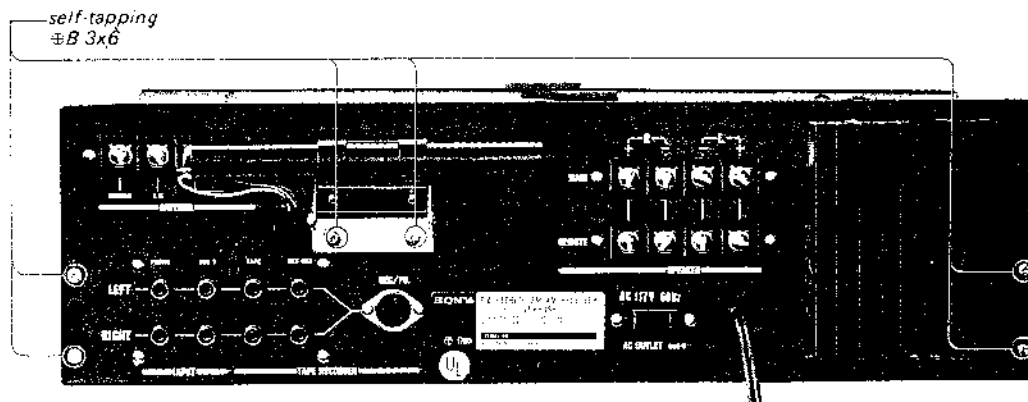


Fig. 2-17. Rear panel removal

2. Unsolder the lead wire connecting between ground terminal and chassis.
3. Unsolder the coaxial cable from fm antenna terminal.
4. Remove the six self-tapping screws ($\varnothing B 3 \times 6$), two of them secure the bar antenna holder to the chassis along with rear panel and others secure the rear panel to the chassis as shown in Fig. 2-17. This frees the rear panel.

2-12. REPLACEMENT OF COMPONENTS SECURED TO THE REAR PANEL BY RIVETS

1. Remove the rear panel as described in Procedure 2-11.
2. Bore out the rivets using a drill bit slightly larger in diameter than the rivet. See Fig. 2-18.

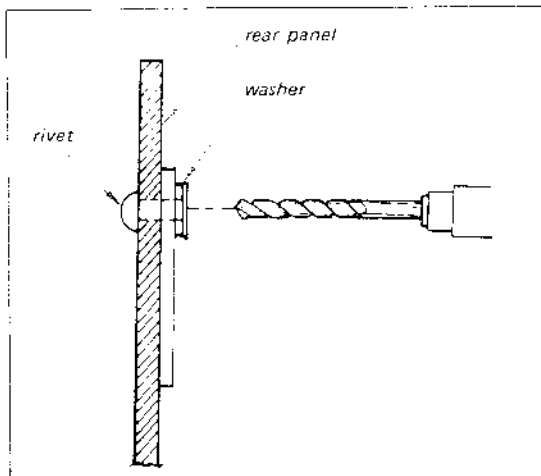


Fig. 2-18. Rivet replacement

3. Punch out the remainder of the rivet with a nail set or prick punch.
4. Remove the defective component, and install the new one.
5. Secure the new component with a suitable screw and nut, or a repair rivet screw (Part No. 3-701-402-00).

2-13. POWER TRANSISTOR REPLACEMENT

1. Remove the top cover and bottom plate as described in Procedure 2-3.
2. Remove the four self-tapping screws ($\varnothing B 3 \times 8$) securing the heat sink to the chassis as shown in Fig. 2-19.

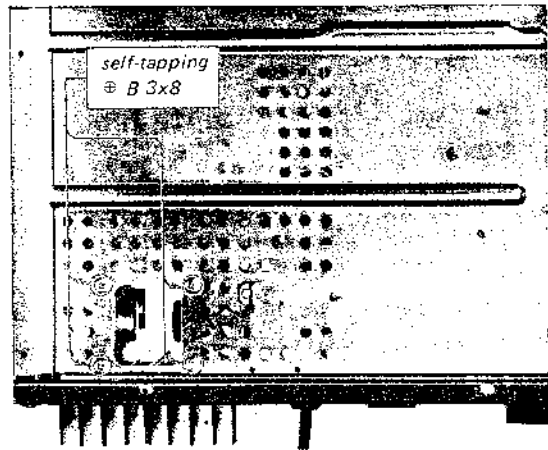
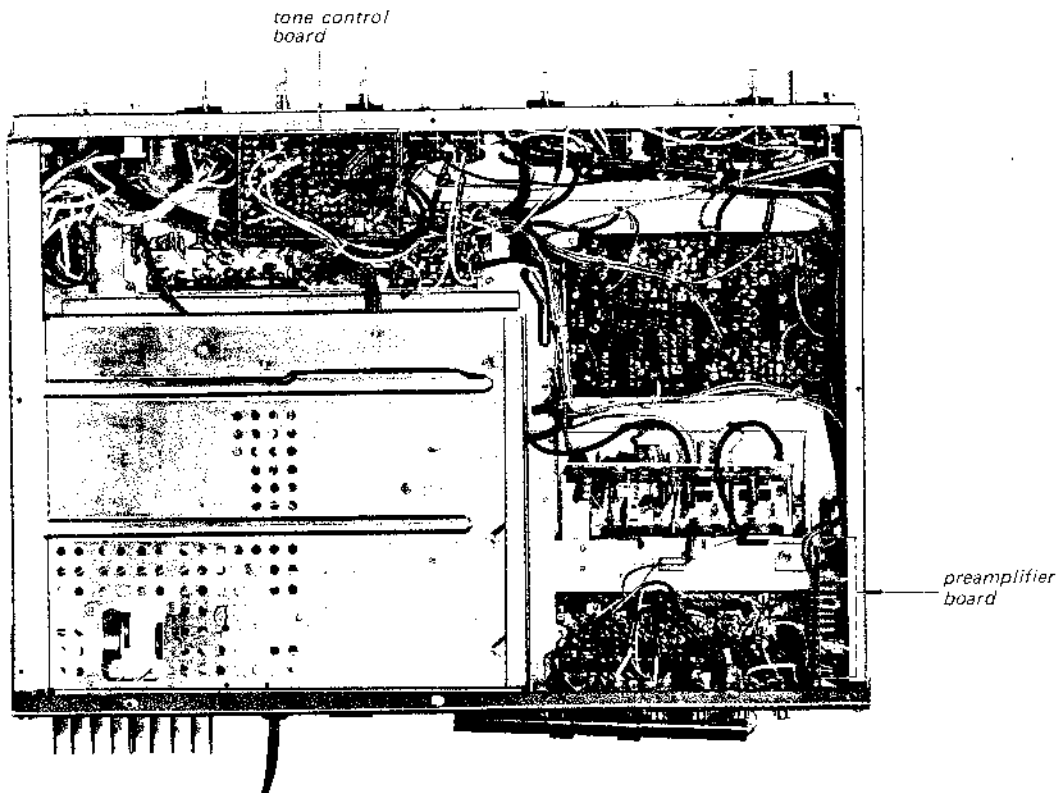
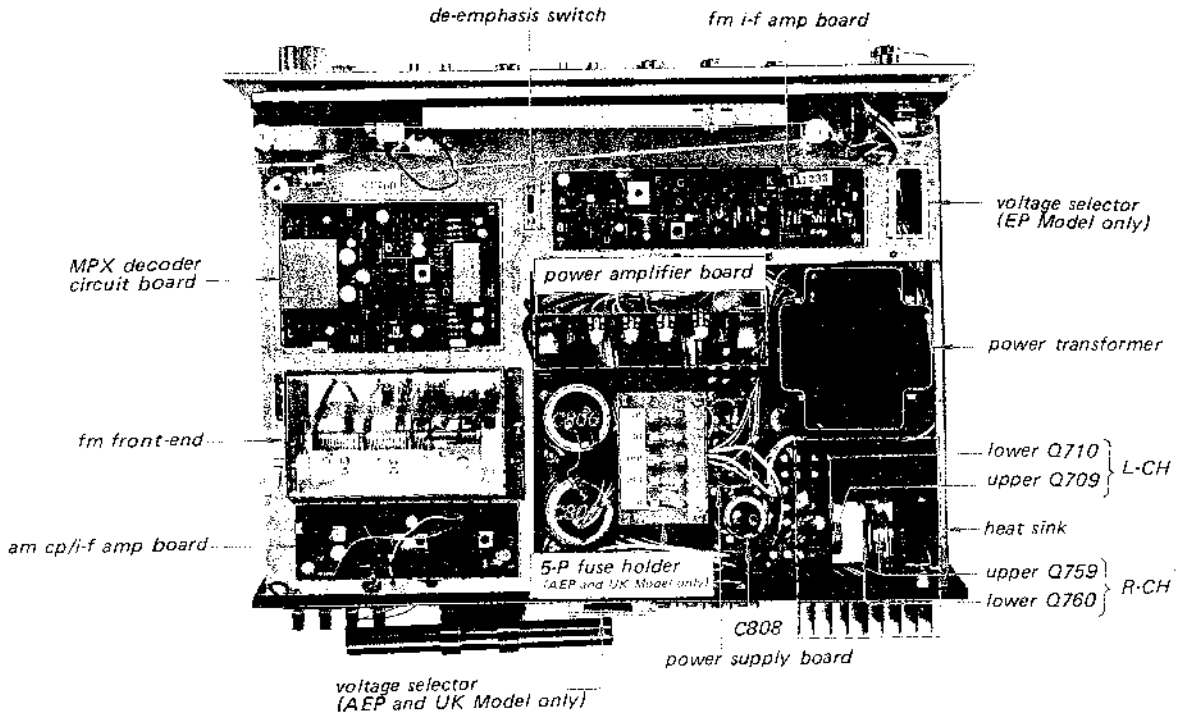


Fig. 2-19. Heat sink removal

3. Cut the emitter and base leads of the defective power transistor with a diagonal cutter. This prevents damage to the mica insulator when removing the defective power transistor.
4. Carefully draw back the heat sink, and then remove the two screws ($\varnothing B 3 \times 12$) and nuts securing the power transistor to the heat sink.
5. When replacing the power transistor, apply a coating of a thermal compound or a heat-transferring grease to both sides of the mica insulator. Any excess compound or grease squeezed out when the mounting screws are tightened should be wiped off with a clean cloth. This prevents accumulation of conductive dust particles that might eventually cause a short.

2-14. CHASSIS LAYOUT



SECTION 3 ALIGNMENT AND ADJUSTMENT

3-1. FM I-F STRIP ALIGNMENT

CAUTION

The ceramic filters in the fm i-f circuit are selected according to their specified center frequencies and color coded as shown in Fig. 3-1 and listed in Table 3-1. Check the color code of the filters to identify the same center frequency when replacing any of these filters.

TABLE 3-1. FM I-F CERAMIC FILTERS

Part No.	Color	Specified Center Freq.
1-403-562-11	red	10.70 MHz
1-403-562-21	black	10.66 MHz
1-403-562-31	white	10.74 MHz
1-403-562-41	green	10.62 MHz
1-403-562-51	yellow	10.78 MHz

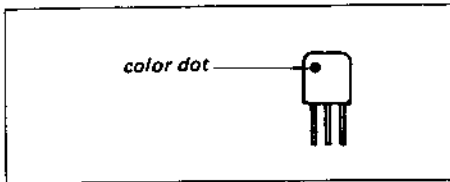


Fig. 3-1. Fm i-f ceramic filter

Signal Generator Alignment

Test Equipment Required

1. Fm signal generator
2. Ac VTVM
3. Oscilloscope
4. Alignment tools
5. Headphone (8Ω)

Note: This alignment is needed only after IFT101 in the front-end, if trimmer capacitors (CT102, CT103), or T201 (discriminator transformer) has been replaced or repaired.

All signal generator output levels specified in the fm section are for terminated output.

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.
2. Insert a headphone plug to the HEADPHONE jack.
3. Connect the input cable of the ac VTVM to the REC OUT jack (J504).
4. Connect the signal-generator output to the fm antenna terminal as shown in Fig. 3-2.
5. Short the connection point of R241 and C215 (AFC circuit) to ground as shown in Fig. 3-3.

Procedure

1. With the equipment connected as shown in Fig. 3-2, set the signal-generator controls as follows:
 - Carrier frequency..... 98 MHz
 - Modulation Fm, 400 Hz, 75 kHz deviation (100%)
 - Output level 30 μV (30 dB)
2. Set the receiver controls as follows:
 - FUNCTION switch FM MONO
 - MODE switch..... STEREO
 - VOLUME control Minimum
3. Turn the core of transformer IFT101 or T201 (bottom core) (see Fig. 3-3 or Fig. 3-5) with the alignment tool to obtain maximum output.

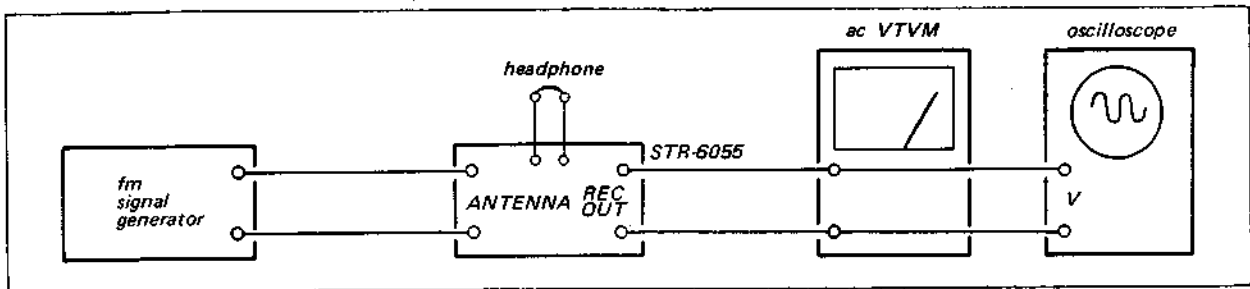


Fig. 3-2. Fm i-f, muting and front-end alignment test setup

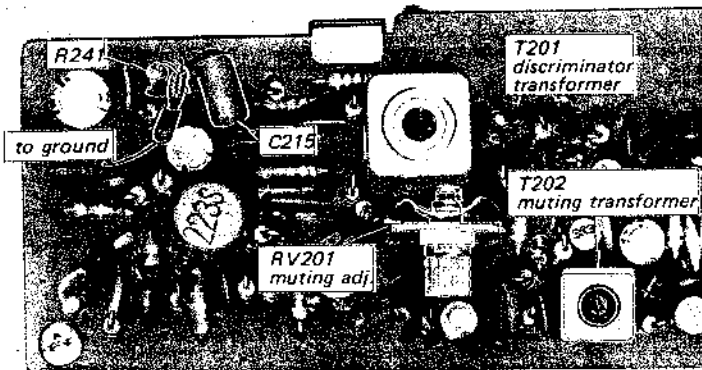


Fig. 3-3. Interruption of afc circuit and parts location

3-2. FM DISCRIMINATOR ALIGNMENT

Note: There are two or three methods of discriminator alignment, but only the simplified method using the receiver's TUNING meter is described here.

Test Equipment Required

1. Oscilloscope
2. Alignment tools
3. Headphone (8 Ω)

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.
2. Connect the input cable of the oscilloscope to REC OUT jack (J504) as shown in Fig. 3-4.
3. Short the connection point of R241 and C215 (AFC circuit) to ground as shown in Fig. 3-3.
4. Insert a headphone plug to the HEADPHONE jack.

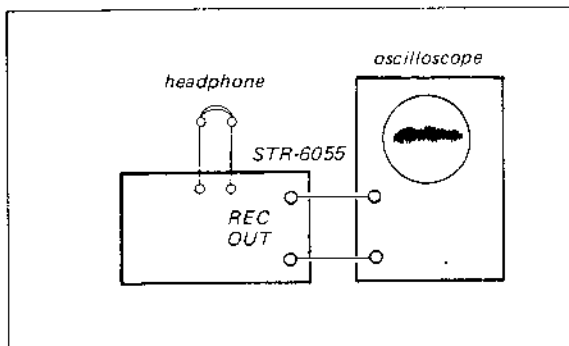


Fig. 3-4. Discriminator alignment test setup

Procedure

1. With the equipment connected as shown in Fig. 3-2, set the receiver controls as follows:
 FUNCTION switch FM MONO
 MODE switch STEREO
 No signal should be received.
2. Adjust the controls of the oscilloscope to provide a visible indication of noise. Always watch the oscilloscope to confirm that the tuner is not receiving any off-the-air signals.
3. Turn the top core (secondary side) of discriminator transformer T201 (see Fig. 3-3) with a hex-head alignment tool to obtain a null-point reading on the TUNING meter. If the discriminator transformer (T201) is not aligned correctly, some deviation on the TUNING meter will be observed.
Note: Turn the core carefully and slowly. At both extreme positions of the top core, a null point will be observed. The real null point should be obtained in the middle of the core thread length.
4. Repeat the above mentioned steps and fm i-f strip alignment two or three times.

3-3. MUTING ADJUSTMENT

Note: Two methods of muting adjustment are available, signal generator adjustment and adjustment by using an off-the-air signal. You can use either of them.

Signal Generator Adjustment

Test Equipment Required

1. Fm signal generator
2. Ac VTVM or oscilloscope
3. Alignment tool

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.
2. Turn RV201 (see Fig. 3-3) fully clockwise on the fm i-f amp board.
3. Short the connection point of R241 and C215 (AFC circuit) to ground as shown in Fig. 3-3.

TABLE 3-2 MUTING ADJUSTMENT

Coupling Between Receiver and SG	SG Frequency and Output Level	Dial Indication	Adjust	Remarks
Direct	98 MHz 400 Hz, 30% Mod.	98 MHz	T202 See Fig. 3-3	Turn the core of T202 to obtain proper muting operation.

Procedure

- With the equipment connected as shown in Fig. 3-2, set the receiver controls as follows:
 FUNCTION switch FM MONO
 MODE switch STEREO
 MUTING switch ON
- Follow the procedure given in Table 3-2. Note that the muting circuit should begin to operate at the symmetrical deflection point of TUNING meter when detuning the tuner to higher or lower than the reference carrier frequency.

Off-the-Air Signal Adjustment

Accurate muting circuit adjustment can also be performed by utilizing off-the-air local fm signals instead of the fm signal generator.

Note that a weak signal is best for this purpose.

3-4. FM FREQUENCY COVERAGE ALIGNMENT

CAUTION

Never attempt alignment of the front-end section except for the frequency-coverage and dial-calibration adjustments. The

front-end section of the tuner has been carefully adjusted at the factory, so very little adjustment is necessary in the field. Alignment need not be performed when the front-end FET is replaced since changes in FET parameters have little effect upon tuning. If an rf-stage adjustment is required, ask your nearest SONY Service Station to send your unit to the Factory Service Center for a complete front-end alignment.

Exercise caution when returning the faulty unit so that it will not be damaged during transit. The warranty will not cover damage incurrent in transit to the Factory Service Center.

Note: Before starting this alignment, the discriminator-transformer alignment should be performed.

Signal Generator Alignment

Test Equipment Required

- Fm signal generator
- Ac VTVM
- Alignment tools

TABLE 3-3 FM FREQUENCY COVERAGE ALIGNMENT

Step	Coupling Between Receiver and SG	SG Frequency and Output Level	Dial Indication	Ac VTVM Connection	Adjust	Indication
1	Direct	87.5 MHz 400 Hz 100% Mod. as low as possible	87.5 MHz	REC OUT (J504)	OSC coil L104 See Fig. 3-5	Maximum VTVM reading
2		108 MHz 400 Hz 100% Mod. as low as possible	108 MHz		OSC trimmer CT104 See Fig. 3-5	

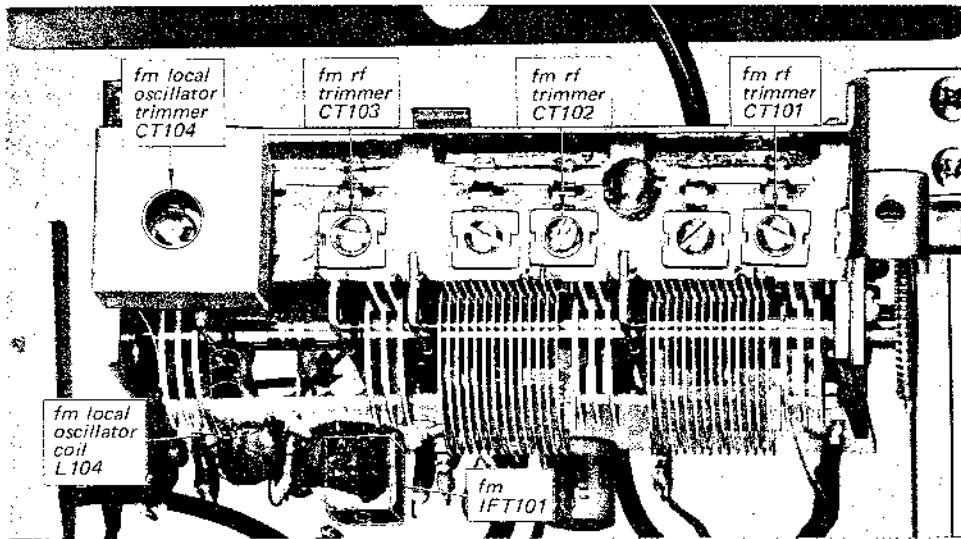


Fig. 3-5. Adjustment parts location

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.
2. Connect the equipment as shown in Fig. 3-2.
3. Set the receiver controls as follows:
 FUNCTION switch FM MONO
 MODE switch STEREO
4. Short the connection point of R241 and C215 (AFC circuit) to ground as shown in Fig. 3-3.

Procedure

Follow the procedures given in Table 3-3 when performing this alignment with an fm signal generator. Be sure that the dial is mechanically calibrated as described in Procedure 2-6 on page 19.

Off-the-Air Signal Alignment

An accurate frequency-coverage alignment can also be performed by utilizing off-the-air local fm signals. However, before performing the following procedure, be sure that the dial pointer is correctly positioned as described in Procedure 2-6 on page 19.

Test Equipment Required

1. Alignment tools

Procedure

1. Short the connection point of R241 and C215 (AFC circuit) to ground as shown in Fig. 3-3.
2. Tune the receiver to the lowest frequency station.
3. Check the dial scale for a calibration accuracy of ± 200 kHz from the carrier frequency of the

station. If the dial-accuracy deviation exceeds this limit, turn the local-oscillator coil L104 (see Fig. 3-5) slightly until optimum dial calibrations is obtained.

4. Tune the receiver to the highest-frequency station in your locality. If the dial-calibration error is excessive, adjust local-oscillator trimmer CT104 (see Fig. 3-5) to obtain maximum calibration accuracy.
5. Repeat steps 3 and 4.

3-5. FM STEREO SEPARATION ADJUSTMENT

Test Equipment Required

1. Fm stereo signal generator
2. Ac VTVM
3. Oscilloscope

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.
2. Connect the test equipment as shown in Fig. 3-6, then set the fm stereo signal generator controls as follows:
 Carrier frequency 98 MHz
 Output level $1,000 \mu\text{V}$ (60 dB)
 Mode STEREO
 Audio (400 Hz) Mod 67.5 kHz (90%)*
 Pilot (19 kHz) Mod 7.5 kHz (10%)

*Note: 75 kHz (100%) if the metering indicates total modulation (audio-pilot).

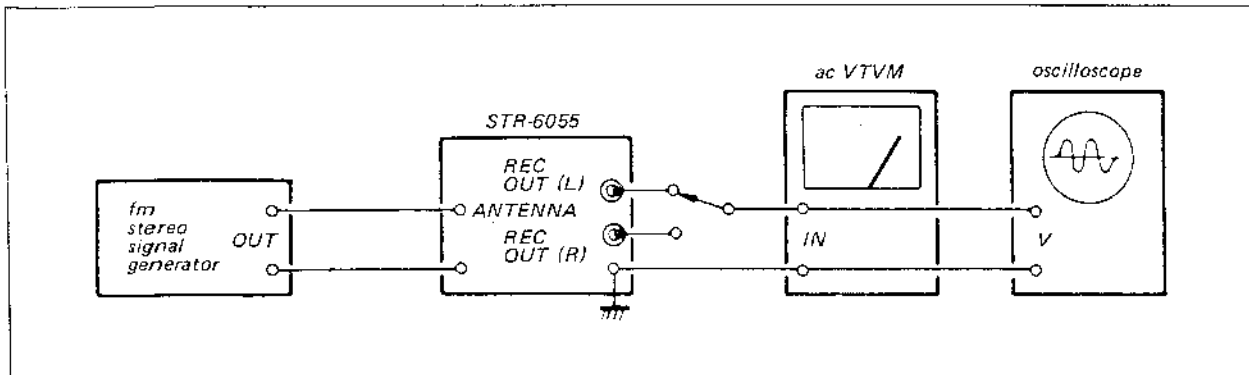


Fig. 3-6. Fm stereo separation adjustment test setup

Procedure

1. Precisely tune the set to the carrier frequency of stereo signal generator, then turn the top core of switching transformer T401 (see Fig. 3-7) to obtain maximum output at the left channel. Note that this adjustment has a close relationship with stereo distortion.
2. Record the output level of the left channel when the stereo signal generator input selector is set to the left channel.
3. Switch the stereo signal generator input selector to the right channel and read the residual signal level in the left channel.
4. The output-level to residual-level ratio represents the separation. Adjust separation adj. control RV401 (see Fig. 3-7) for minimum residual level. Check the right channel for separation. Usually, about an 8 to 9 dB difference in channel separation exists. Readjust RV401 for minimum difference between left and right-channel separation. While doing this, remember that the output level also changes according to the setting of RV401.

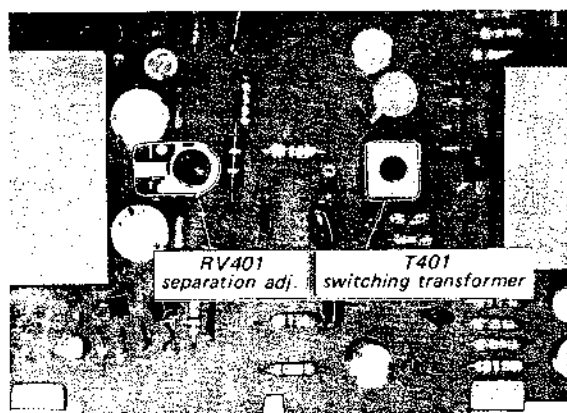


Fig. 3-7. Adjustment parts location

3-6. A-M I-F STRIP ALIGNMENT

Note: No adjustment is required in the field even if after replacing any of i-f transformers (CFT301 or IPT301) for the sets with serial numbers of 810,401 and later (USA Model) 700,301 and later (Canada Model) 553,001 and later (EP Model) 600,001 and later (UK Model) 900,001 and later (AEP Model).

Following procedures should be applied for the sets with serial numbers of 800,001 to 180,400 (USA Model) 700,001 to 700,300 (Canada Model) 550,001 to 553,000 (EP Model).

Preparation

Remove the top cover as described in Procedure 2-3 on page 16. Then set the receiver FUNCTION switch to AM.

Note: To perform this alignment, the local oscillator should be killed. To do this, shunt the local oscillator capacitor with a 0.02 μ F capacitor as shown in Fig. 3-9.

Sweep Generator Alignment

Test Equipment Required

1. Sweep generator, 455 kHz
2. Oscilloscope
3. Alignment tools

Procedure

1. Connect the sweep generator output directly to the a-m antenna terminal.
2. Connect the input cable of the oscilloscope to R321 (see Fig. 3-9) and ground on the a-m cp/i-f amp board with alligator clips as shown in Fig. 3-9.
3. Set the sweep generator controls as follows:
 Center frequency 455 kHz
 Sweep width 25 kHz
 Output as low as possible

4. With the equipment connected as shown in Fig. 3-10, adjust the oscilloscope controls and generator output to provide a visible indication.
5. Turn the top core of IFT301 (see Fig. 3-9) to obtain a maximum and symmetrical response as shown in Fig. 3-8.



Fig. 3-8. A-m i-f strip alignment by sweep generator

Rf Signal Generator Alignment

Test Equipment Required

1. Signal generator, 455 kHz, a-m modulation
2. Oscilloscope or VTVM
3. Alignment tools

Procedure

1. Set the rf signal generator controls as follows:
 Modulation INTERNAL
 Frequency 455 kHz
 OUTPUT level 1,000 μ V
2. Connect the rf signal-generator output to a-m antenna terminal.
3. With the equipment connected as shown in Fig. 3-11, turn the top core of IFT301 (see Fig. 3-9) to obtain maximum output.

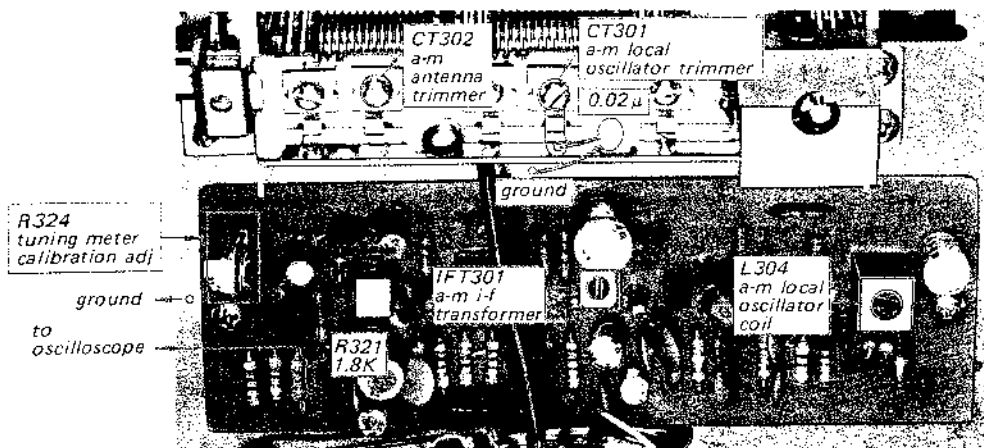


Fig. 3-9. Disabling the a-m local oscillator, a-m detector output connection and adjustment parts location

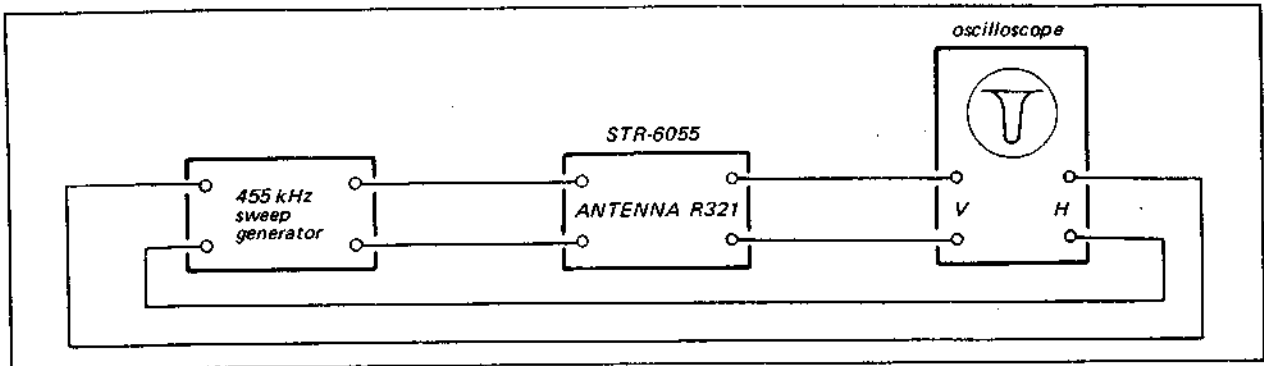


Fig. 3-10. A-m i-f strip alignment test setup by sweep generator

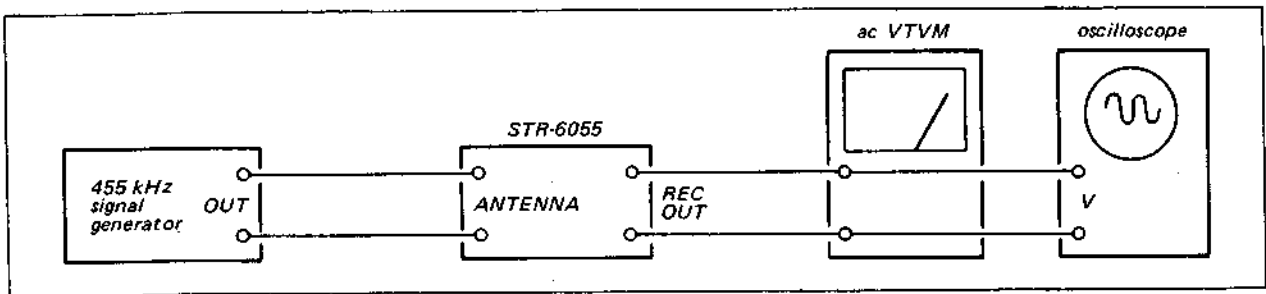


Fig. 3-11. A-m i-f strip alignment test setup by rf signal generator

3-7. A-M FREQUENCY COVERAGE AND TRACKING ALIGNMENT

Preparation

Remove the top cover as described in Procedure 2-3 on page 16. Then, set the FUNCTION switch to AM.

Signal Generator Method

Test Equipment Required

1. A-m signal generator
2. Loop antenna
3. Ac VTVM or oscilloscope

Procedure

With the equipment connected as shown in Fig. 3-12, follow the procedures given in Table 3-4 when performing this alignment with an a-m signal generator.

Off-the-Air Signal Method

An accurate frequency-coverage and tracking alignment can also be performed by utilizing off-the-air local a-m signals. However, before performing the following procedure, be sure that the dial pointer is

correctly positioned as described in the Procedure 2-6 on page 19.

Frequency Coverage Alignment

Note:

- ▲ { USA Model Serial No. up to 810,400
Canada Model ... Serial No. up to 700,300
EP Model Serial No. up to 553,000 } See Fig. 3-9.
- { USA Model..... Serial No. 810,401 and later
Canada Model ... Serial No. 700,301 and later
EP Model..... Serial No. 553,001 and later
UK Model Serial No. 600,001 and later
AEP Model Serial No. 900,001 and later } See Fig. 3-14.

Procedure

1. Tune the receiver to the lowest-frequency station in your locality.
Check the dial scale for a calibration accuracy of ± 20 kHz from the carrier frequency. If the dial calibration error exceeds this limit, turn the local oscillator-coil ^{L304(▲)} _{T301(○)} slightly until optimum dial calibration is obtained.
2. Tune the receiver to the highest-frequency station in your locality. If the dial calibration error exceeds ± 30 kHz from the carrier frequency, adjust local-oscillator trimmer-capacitor ^{CT301(▲)} _{CT302(○)} to obtain maximum calibration accuracy. Repeat the above steps two or three times.

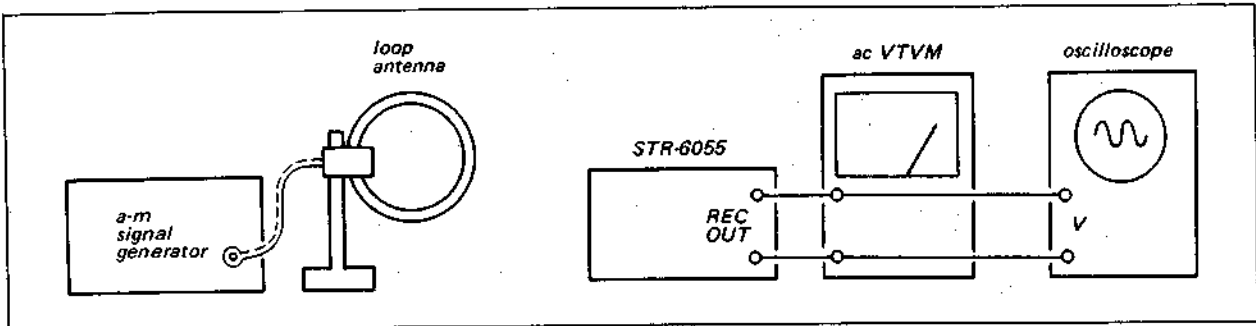


Fig. 3-12. A-m frequency coverage and tracking alignment test setup

Tracking Alignment

1. Tune the set to the station whose carrier frequency is closest to 620 kHz and adjust the position of a-m bar antenna core (see Fig. 3-13) to obtain maximum output.
2. Tune the set to the station whose carrier frequency is closest to 1,400 kHz and adjust antenna trimmer capacitor CT302(▲) to obtain maximum output.
3. Repeat the above steps two or three times.

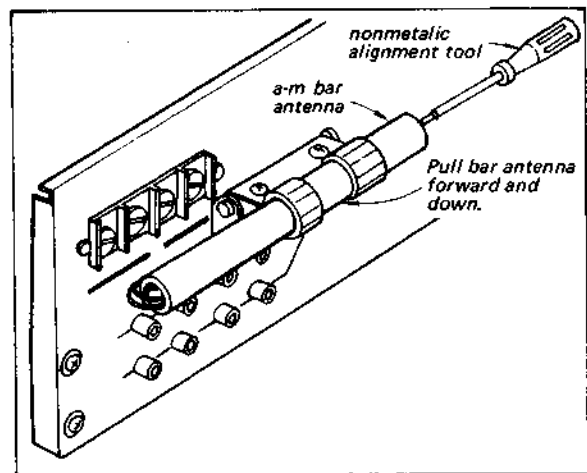


Fig. 3-13. A-m bar antenna core alignment

TABLE 3-4 A-M FREQUENCY COVERAGE AND TRACKING ALIGNMENT

A-M FREQUENCY COVERAGE ALIGNMENT					
SG Coupling Loop Antenna					
SG Output Level 400 Hz, 30% (as low as possible)					
Step	SG Frequency	Dial Indication	Ac VTVM Connection	Adjust	Indication
1	550 kHz	550 kHz	REC OUT	OSC coil L304(▲) T301(○)	Maximum VTVM reading
2	1,600 kHz	1,600 kHz		OSC trimmer CT301(▲) CT302(○)	
A-M TRACKING ALIGNMENT					
SG Coupling Loop Antenna					
SG Output Level 400 Hz, 30% (as low as possible)					
1	600 kHz	Tune to the SG signal.	REC OUT	Bar antenna coil L303(▲) L904(○) (See Fig. 3-13)	Maximum VTVM reading
2	1,400 kHz			Antenna trimmer CT302(▲) CT301(○)	

Note:

- ▲ USA Model Serial No. up to 810,400
- ▲ Canada Model Serial No. up to 700,300
- ▲ EP Model Serial No. up to 553,000
- USA Model Serial No. 810,401 and later
- Canada Model Serial No. 700,301 and later
- EP Model Serial No. 553,001 and later
- UK Model Serial No. 600,001 and later
- AEP Model Serial No. 900,001 and later

See Fig. 3-9.

See Fig. 3-14.



Fig. 3-14. Adjustment parts location

3-8. TUNING METER CALIBRATION

Test Equipment Required

1. Signal generator
2. Ac VTVM
3. Loop antenna
4. Alignment tools

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.

Procedure

1. Connect the test equipment as shown in Fig. 3-15.
2. Set the a-m signal generator and receiver controls as follows:
 - Carrier frequency 1,000 kHz
 - Output level..... 104 dB/m (160 mV/m) at a-m bar antenna
 - Modulation (400 Hz) 30%
 - VOLUME control Minimum
 - FUNCTION switch AM
 - MONITOR switch SOURCE
3. Precisely tune the set to the signal and adjust R324 (RT301) (*) to obtain the meter pointer within 2 mm (3/64") left of its maximum indication as shown in Fig. 3-16.

* **Note:** R324 (see Fig. 3-9)
 Applicable Serial Numbers
 USA Model Up to 810,400
 Canada Model..... Up to 700,300
 EP Model Up to 553,000.

RT301 (see Fig. 3-14)
 Applicable Serial Numbers
 USA Model 810,401 and later
 Canada Model..... 700,301 and later
 EP Model 553,001 and later
 UK Model..... 600,001 and later
 AFP Model 900,001 and later

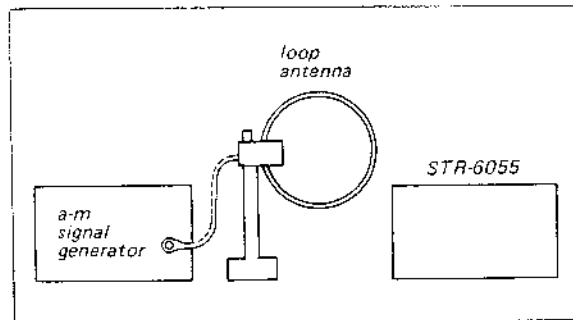


Fig. 3-15. Tuning meter calibration test setup

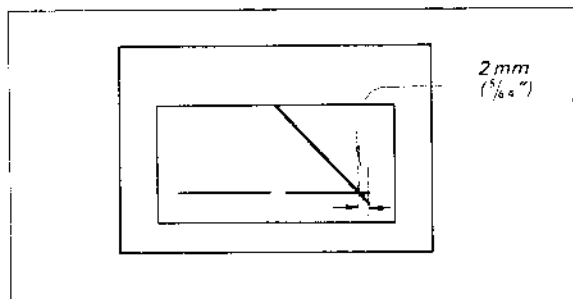


Fig. 3-16. Tuning meter calibration

3-9. POWER AMPLIFIER ADJUSTMENT

Note: There are two adjustment items in the power amplifier. One is dc-bias adjustment and the other is dc-balance adjustment. These adjustments should be repeated alternately two or three times after replacing any of the power transistors until the best operation is obtained.

1. Dc-Bias Adjustment

Serious deficiencies in performance, such as thermal runaway of power transistors, will result if this adjustment is improperly set.

CAUTION

To avoid accidental power transistor damage, increase the ac line voltage gradually, using a variable transformer, while measuring the voltage across emitter of Q709 (Q759) and collector of

Q710 (Q760) as shown in Fig. 3-17 and 3-18. Check to see that the reading does not exceed 50 mV. If it does, turn off the power immediately, then check and repair the trouble in the power amplifier board.

Test Equipment Required

1. Dc millivoltmeter
2. Variable transformer
3. Screwdriver with 3 mm (1/8") blade

Preparation

1. Remove the top cover as described in Procedure 2-3 on page 16.
2. Connect the dc millivoltmeter across emitter of Q709 (Q759) and collector of Q710 (Q760) as shown in Fig. 3-17 and 3-18.
3. Apply a drop of cement solvent to the RV701, RV702, RV751 and RV752 then wait a few seconds for the cement to dissolve.

Procedure

1. Set the adjustable resistors (see Fig. 3-19) on the power-amplifier board as follows:
 - RV702 fully
(L-CH, dc bias) counterclockwise
 - RV752
(R-CH, dc bias)..... fully clockwise
 - RV701, RV751
(dc balance) mid-position
2. Set the variable transformer for minimum output.
3. Turn the POWER switch ON, and then increase the line voltage up to the rated value.

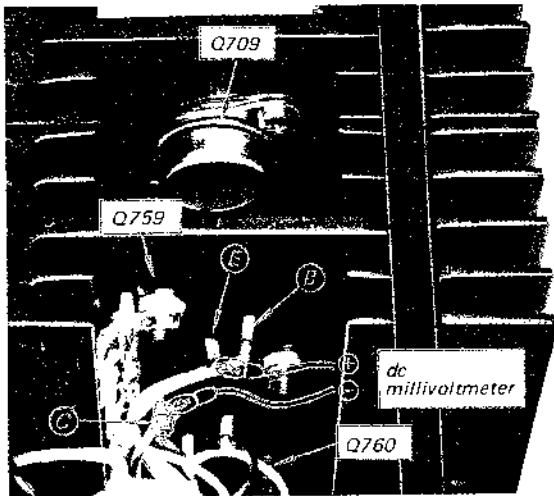


Fig. 3-17. Dc millivoltmeter connection

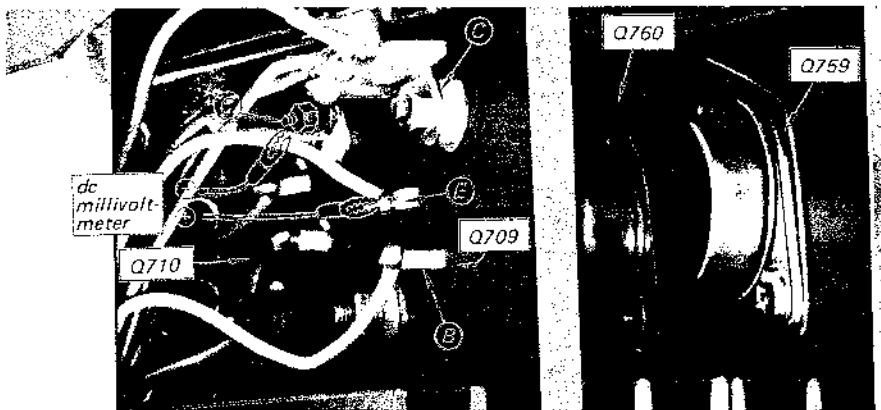


Fig. 3-18. Dc millivoltmeter connection

4. Adjust RV702 and RV752 to obtain a 50 mV reading on the meter, and then perform the dc-balance adjustment.

2. Dc-Balance Adjustment

Excessive harmonic distortion at high levels will result if this adjustment is improperly set.

Test Equipment Required

1. Dc null meter or dc millivoltmeter
2. Screwdriver with 3 mm ($1/8$ ") blade

Preparation

1. Set the SPEAKER switch to MAIN.
2. Connect the dc null meter or dc millivoltmeter to the MAIN speaker output terminal.

Procedure

1. Turn the POWER switch ON, and then adjust RV701 (RV751) for 0V reading on the meter.
2. After 10 minutes warm-up, alternately repeat this and the dc bias adjustment two or three times.
3. After completing the adjustment, apply a drop of lock paint to RV701 and RV702 (RV751 and RV752).

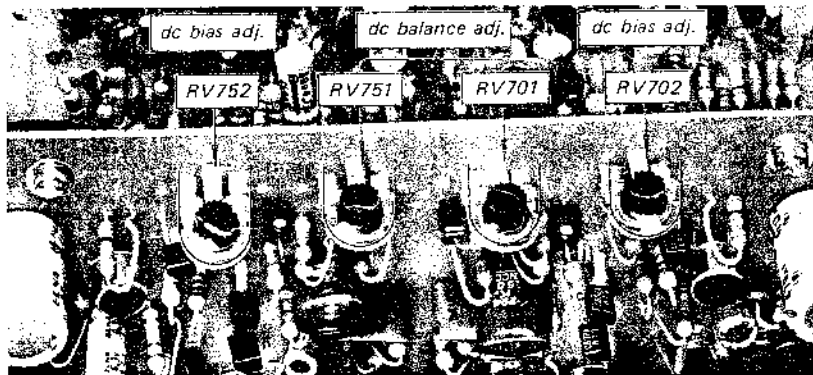


Fig. 3-19. Adjustment parts location

SECTION 4 REPACKING

The STR-6055 original shipping carton and packing material are the ideal container for shipping the unit. However to secure the maximum protection, the STR-6055 must be repacked in these materials precisely as before. The proper repacking procedures are shown in Fig. 4-1.

tection, the STR-6055 must be repacked in these materials precisely as before. The proper repacking procedures are shown in Fig. 4-1.

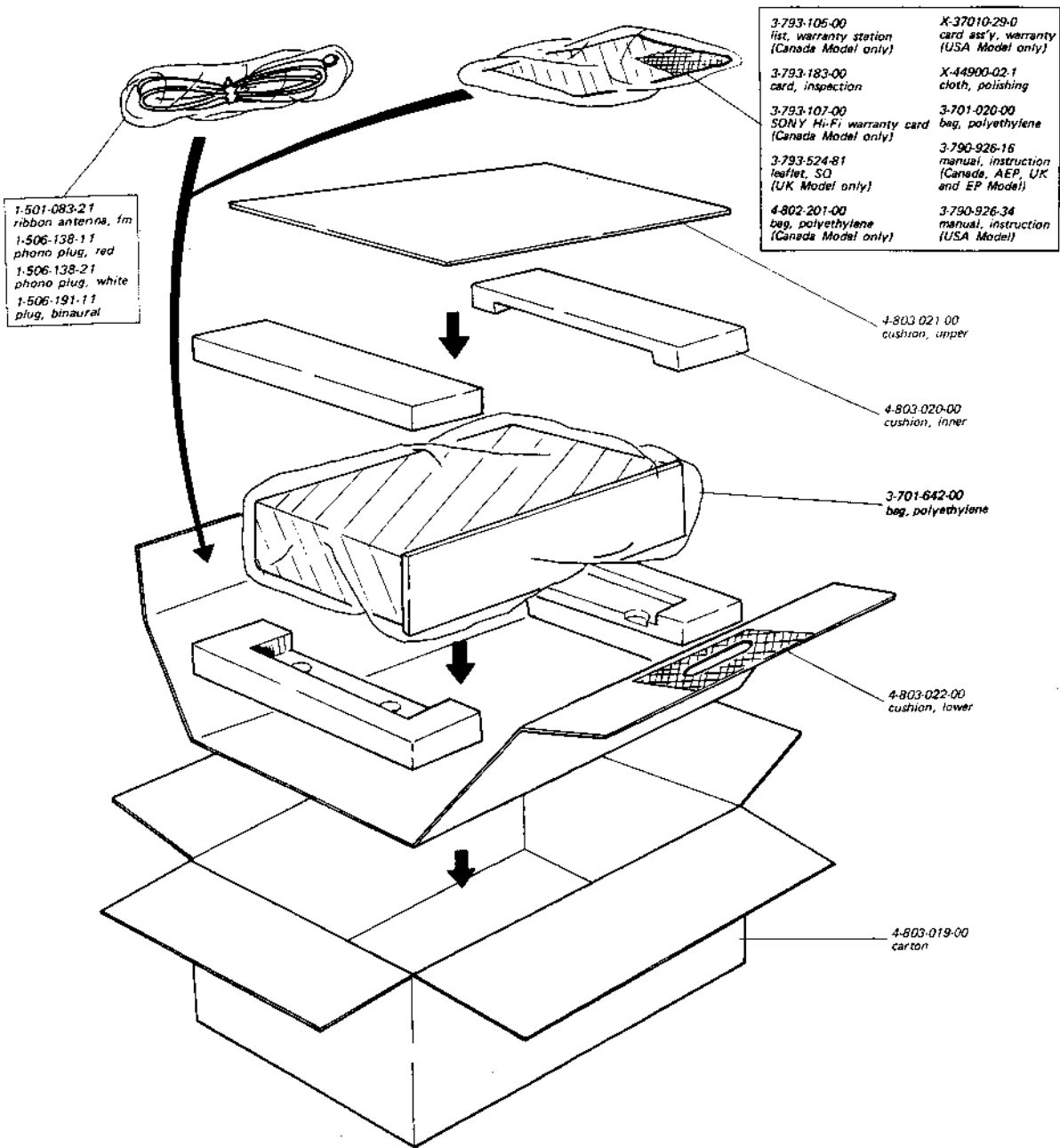
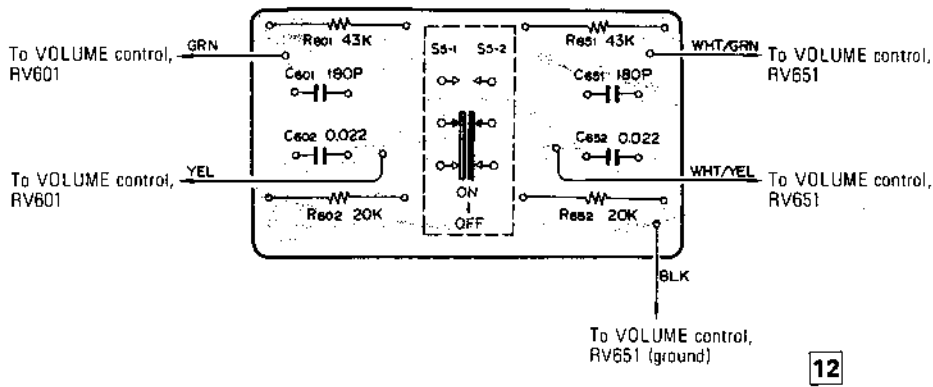


Fig. 4-1. Repacking

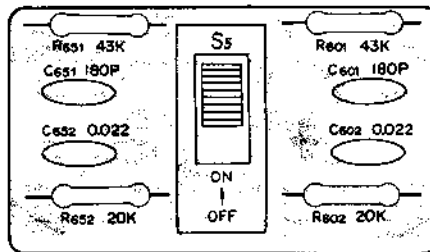
SECTION 5
DIAGRAMS

5-1. MOUNTING DIAGRAM – Loudness Control Board –
– Conductor Side –



12

– Component Side –

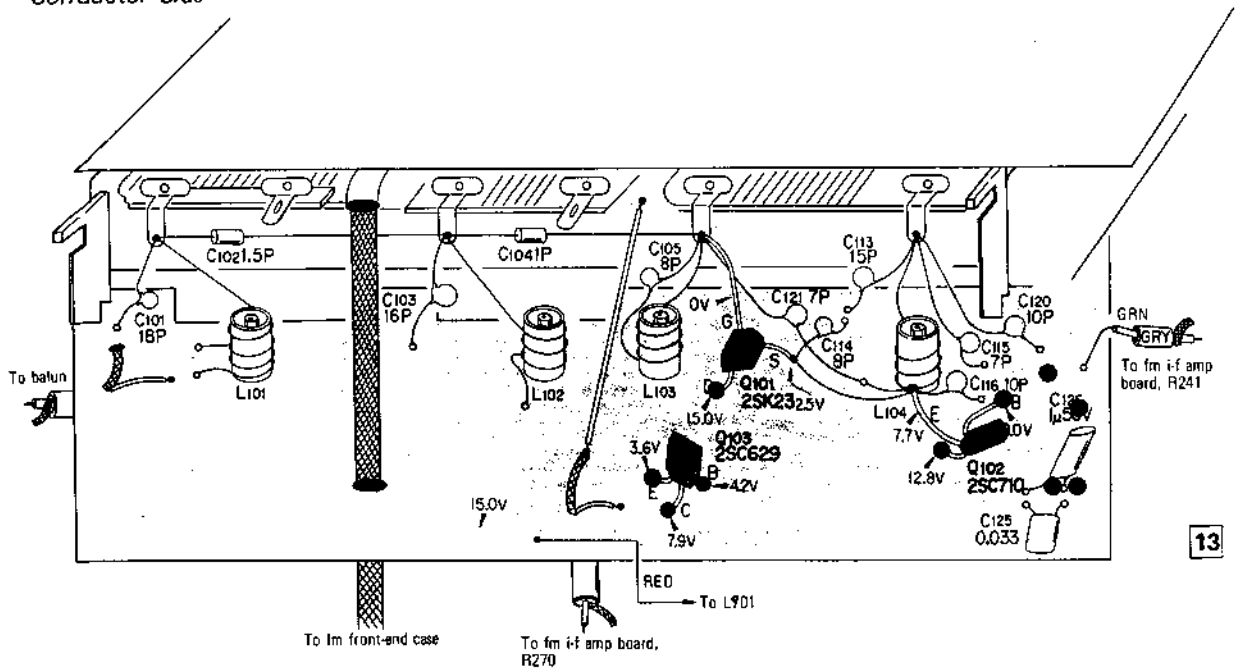


Note: The parts in the loudness control board have been changed as follows:

Ref. No.	Former value	New value
R601 (R651)	68K	43K
R602 (R652)	10K	20K
C601 (C651)	100P	180P
C602 (C652)	0.047	0.022

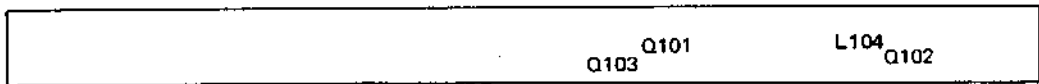
When replacing these parts, change to the new value for both channels.

5-2. MOUNTING DIAGRAM - Fm Front-End -
- Conductor Side -

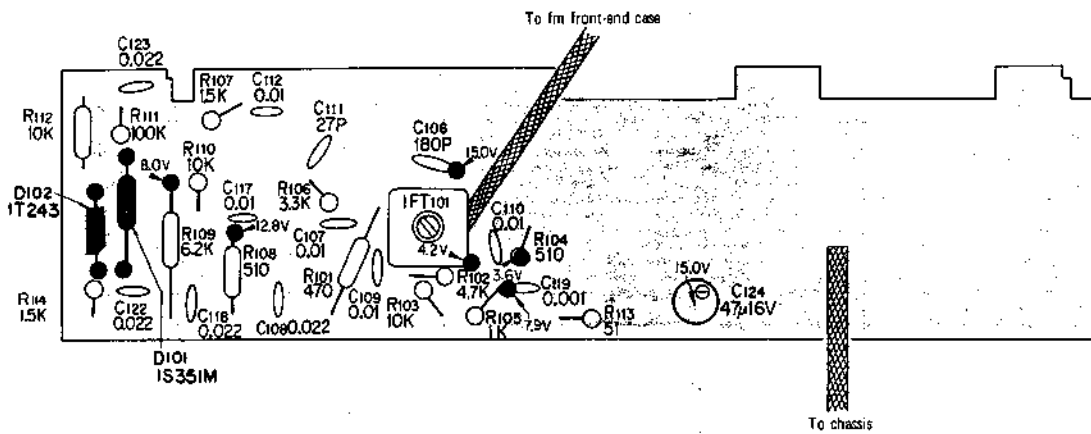


13

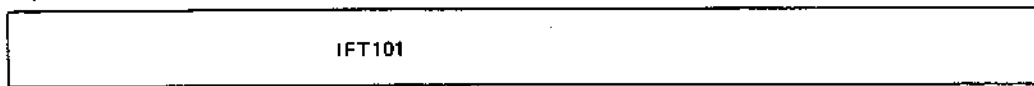
Transistor Location



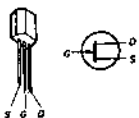
- Component Side -



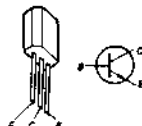
Adjustment Parts Location



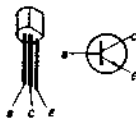
2SK23



2SC629



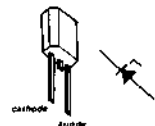
2SC710



1S351M

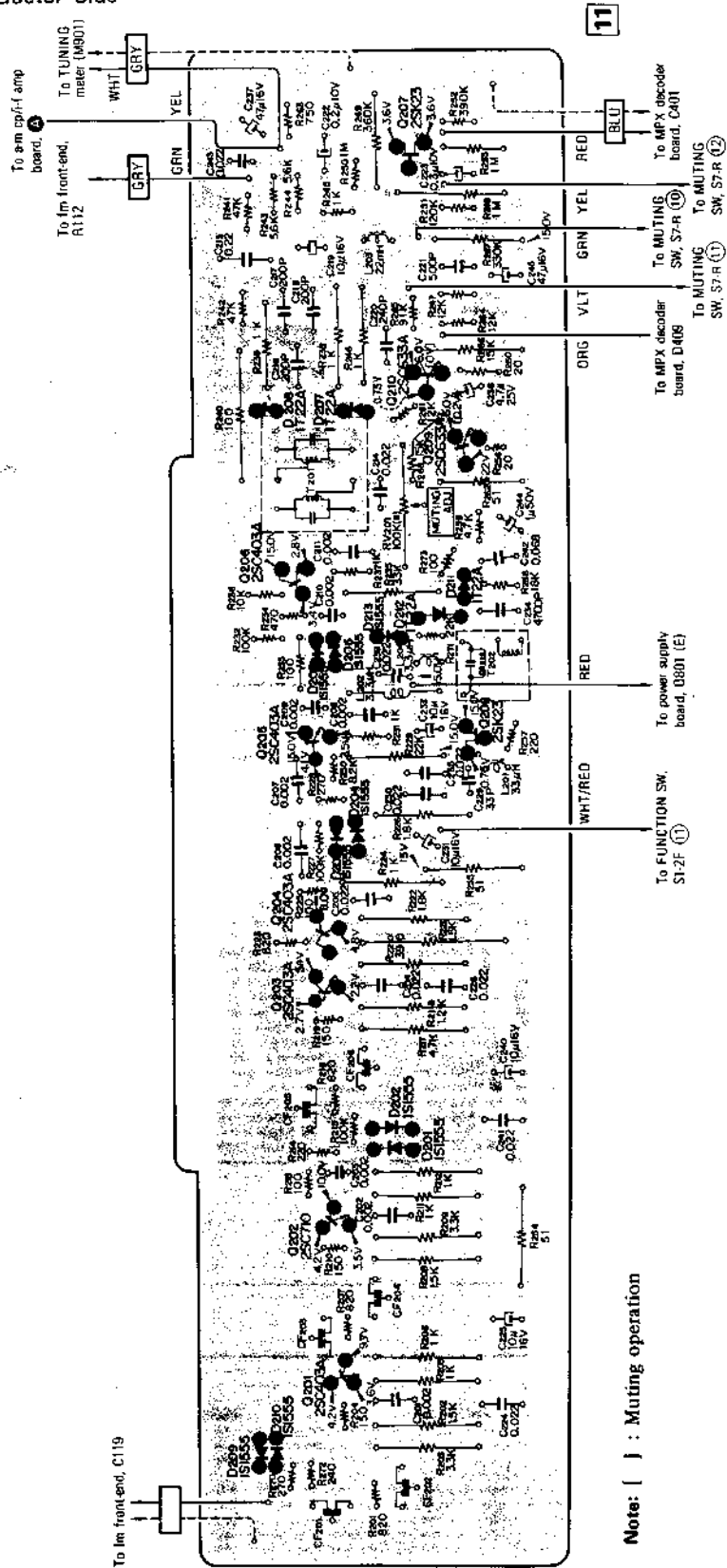


1T243





5-3. MOUNTING DIAGRAM - Fm 1-f Amp Board -
- Conductor Side -

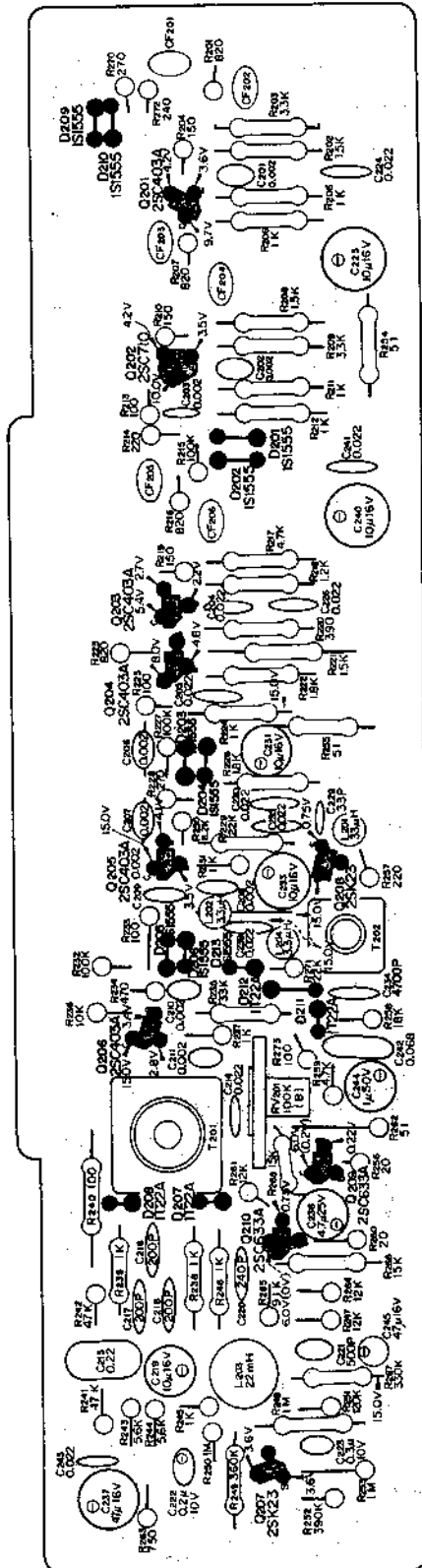


Note: [] : Muting operation

Transistors and Adjustment Parts Location

Q201	Q202	Q203	Q204	Q205	Q206	T201	Q207
						RV201	
				Q208	T202	Q209	Q210

- Component Side -

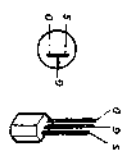


Note: [] : Muting operation

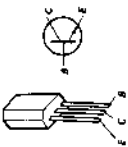
Transistors and Adjustment Parts Location

Q207	Q210	Q209	T201	Q206	Q205	Q204	Q203	Q202	Q201
			RV201						
					T202				

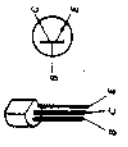
2SK23



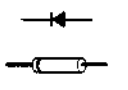
2SC403A
2SC633A



2SC710



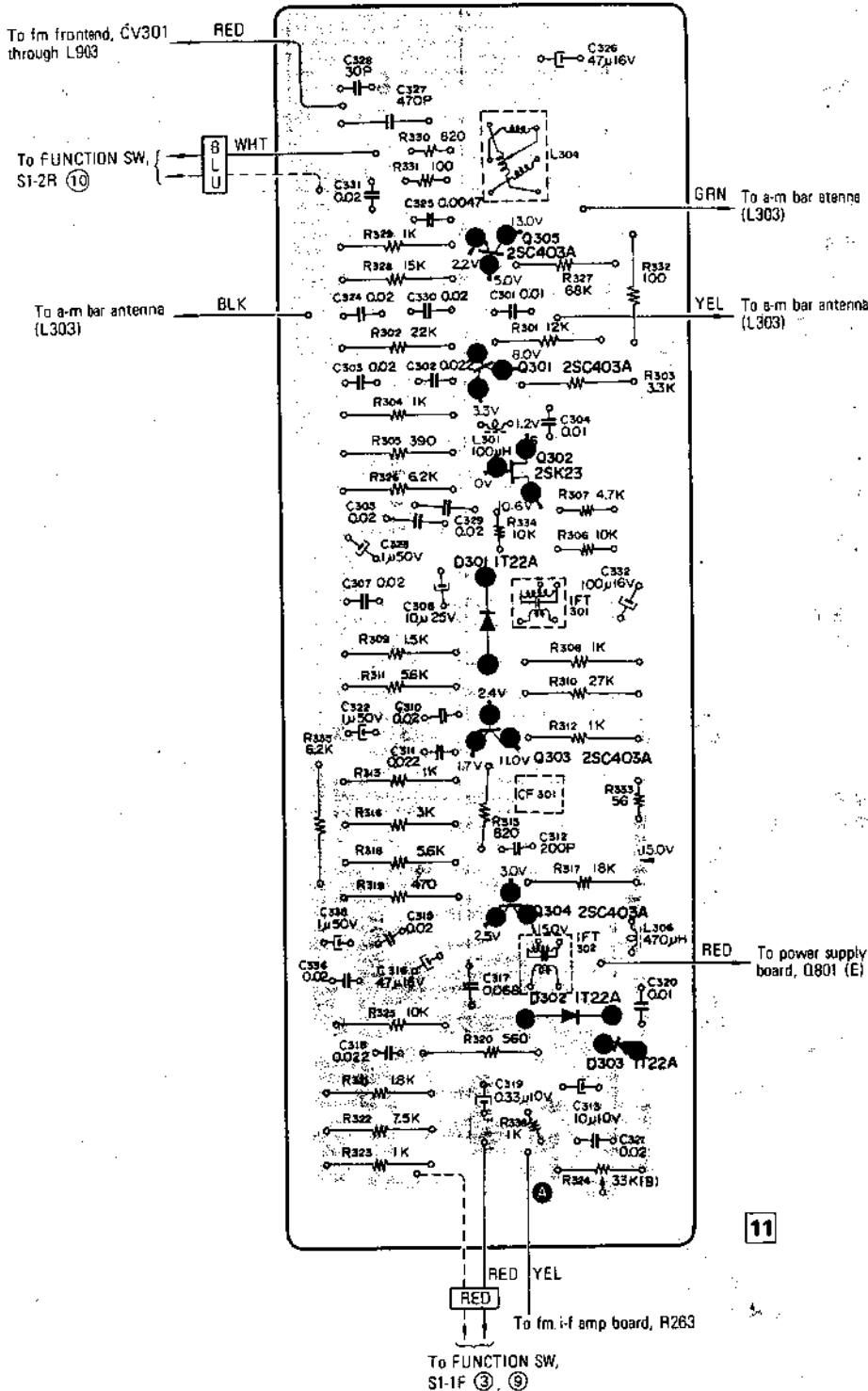
1T22A
1S1555



5-4. MOUNTING DIAGRAM - A-m Cp/l-f Amp Board -
- Conductor Side -

Applicable Serial Numbers:
 USA Model Up to 810,400
 Canada Model... Up to 700,300
 EP Model Up to 553,000

Transistors and
 Adjustment Parts Location

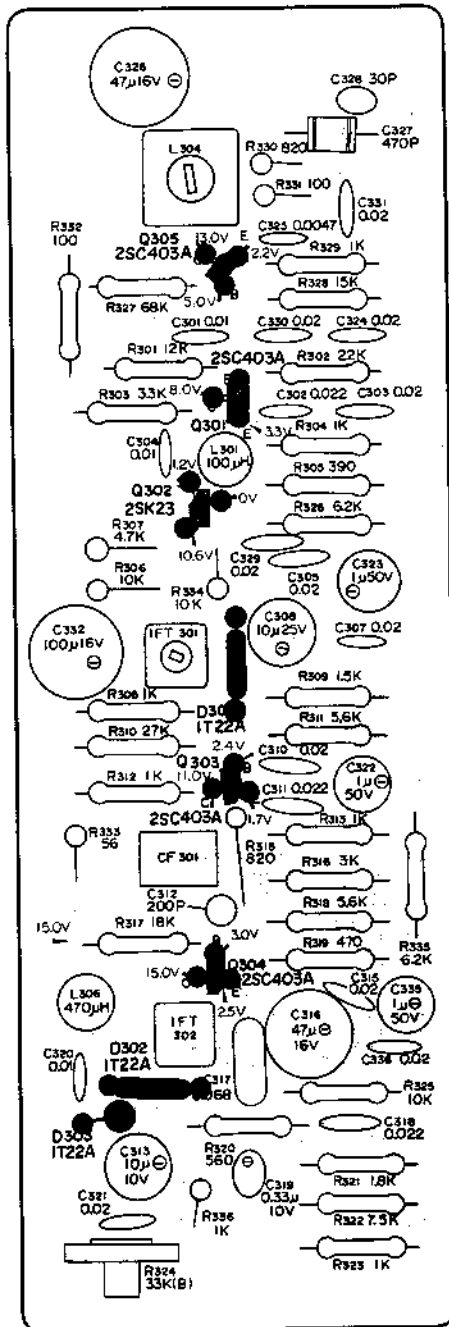


L304
Q305
Q301
Q302
IFT301
Q303
Q304
R324

— Component Side —

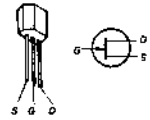
Applicable Serial Numbers:
 USA Model Up to 810,400
 Canada Model... Up to 700,300
 EP Model Up to 553,000

Transistors and
Adjustment Parts Location

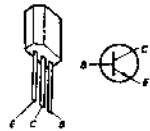


L304
Q305
Q301
Q302
IFT301
Q303
Q304
R324

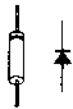
2SK23



2SC403A



1T22A

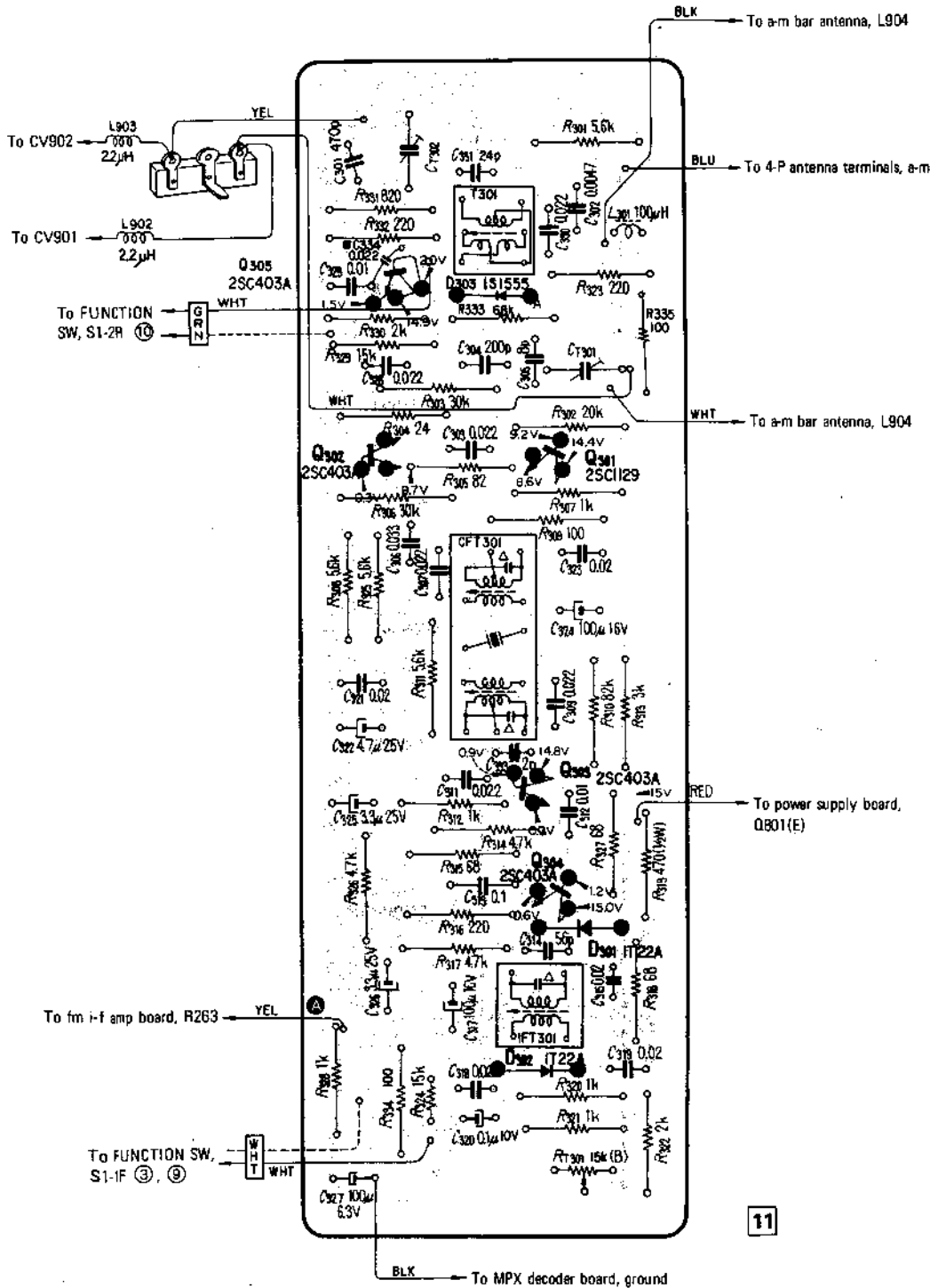




MOUNTING DIAGRAM – A-m Cp/I-f Amp Board –
 – Conductor Side –

Applicable Serial Numbers:

- USA Model 810,401 and later
- Canada Model... 700,301 and later
- EP Model 553,001 and later
- UK Model..... 600,001 and later
- AEP Model 900,001 and later



Transistors and Adjustment Parts Location

CT302	
T301	
Q305	
CT301	
Q302	Q301
Q303	
Q304	
RT301	

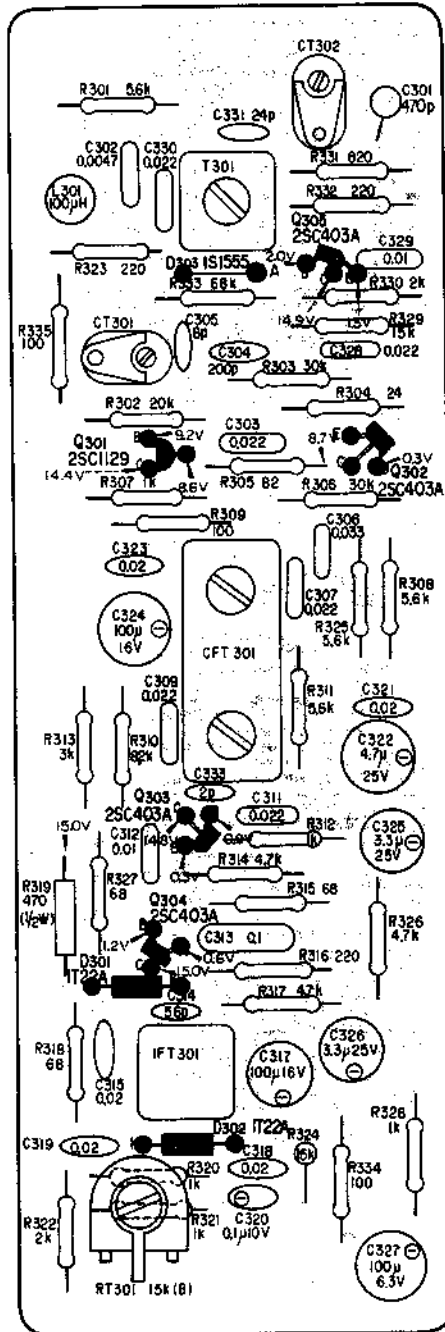
11

Note: ■ C334 is mounted on the conductor side.

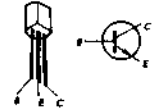
- Component Side -

Applicable Serial Numbers:
 USA Model 810,401 and later
 Canada Model... 700,301 and later
 EP Model 553,001 and later
 UK Model 600,001 and later
 AEP Model 900,001 and later

Transistors and
Adjustment Parts Location



2SC1129

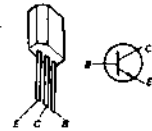


CT302

T301

Q305

2SC403A



CT301

Q301

Q302

1S1555
1T22A
1T40

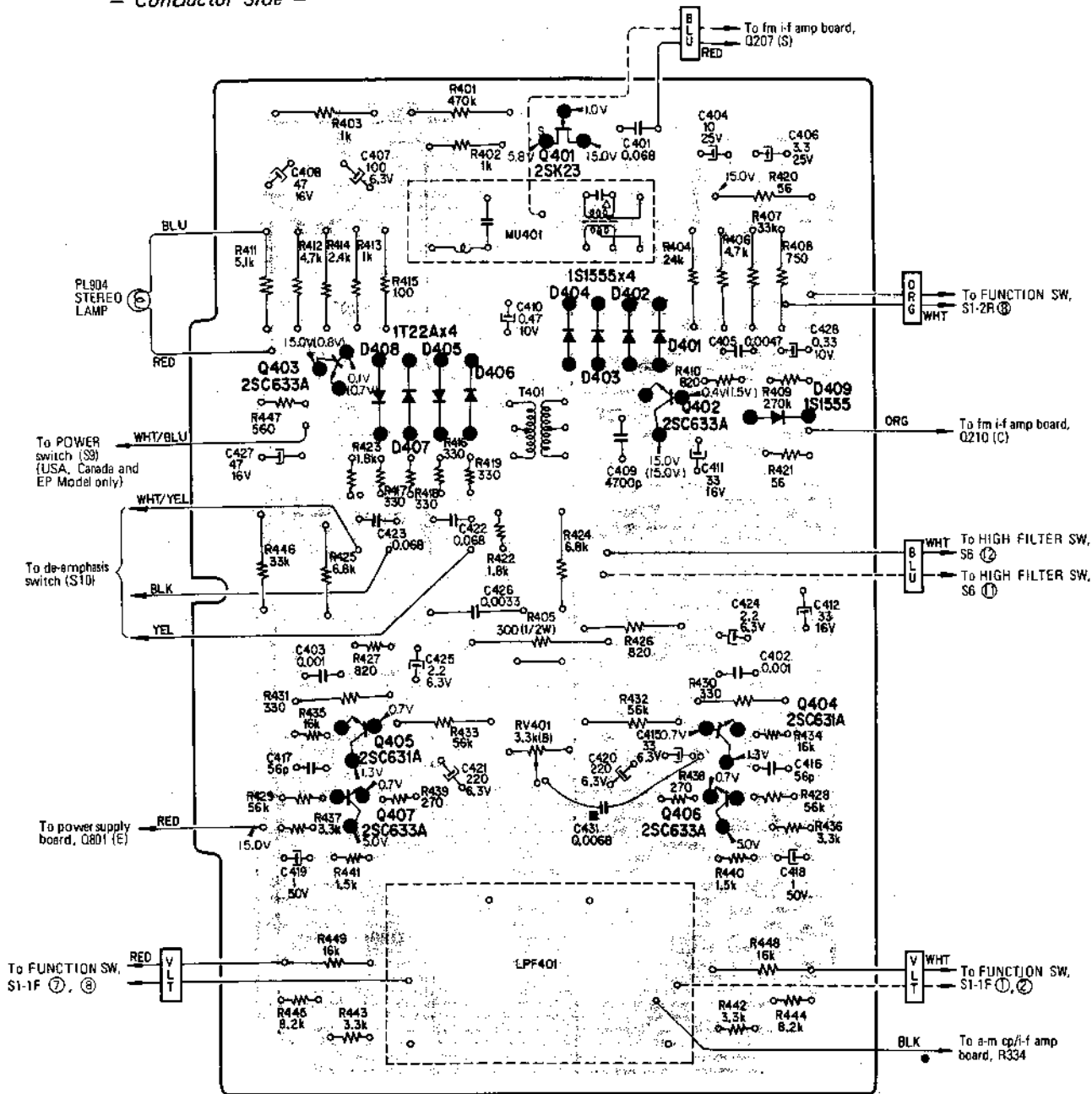
Q303

Q304

RT301

5-5. MOUNTING DIAGRAM - MPX Decoder Board -

- Conductor Side -



12

Transistors and Adjustment Parts Location

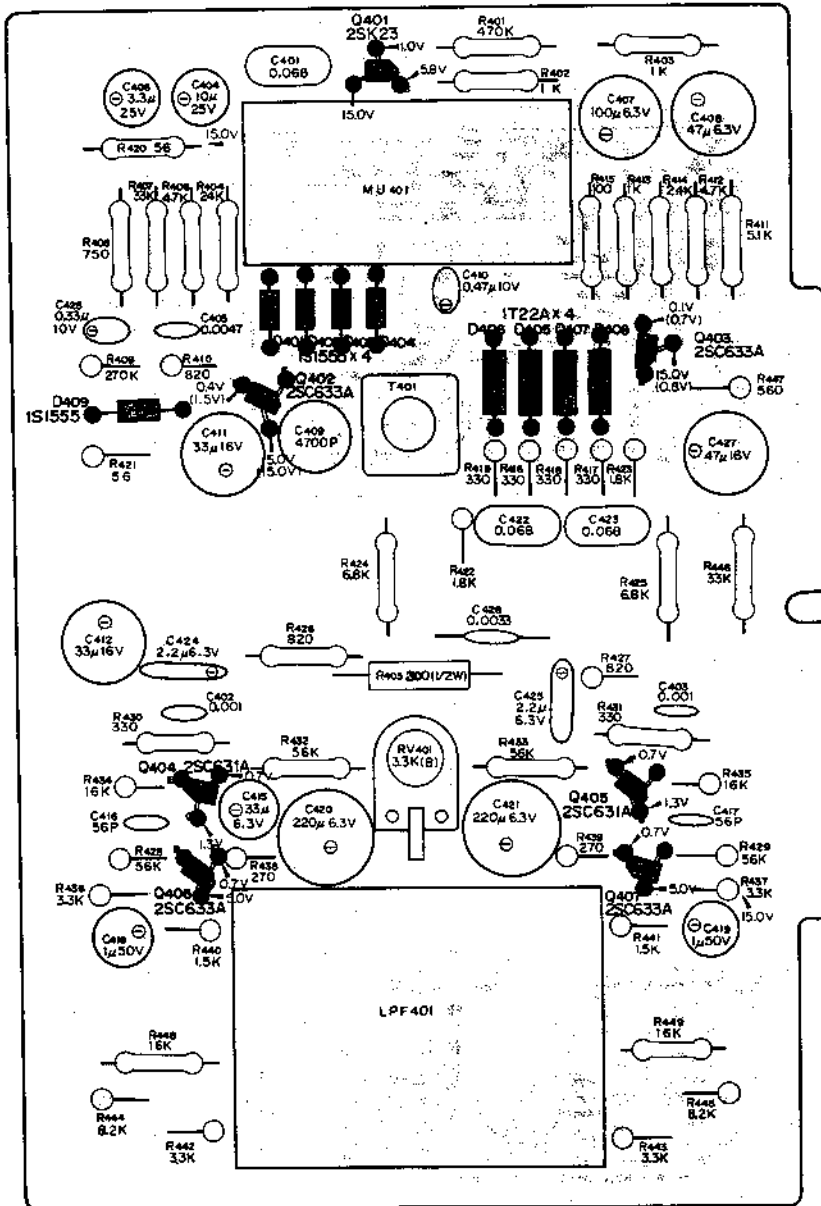
Q403	Q401	
	T401	Q402
Q405		
Q407	RV401	Q404
		Q406

Note: Lead wire marked * is applied for the sets with the serial number of 810,401 and later (USA Model)
 700,301 and later (Canada Model)
 553,001 and later (EP Model)
 600,001 and later (UK Model)
 900,001 and later (AEP Model).

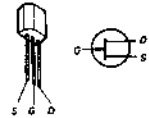
■ C431 is mounted on the conductor side.

() Stereo operation

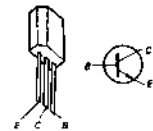
- Component Side -



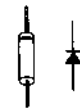
2SK23



2SC631A
2SC633A



1T22A
1S1555



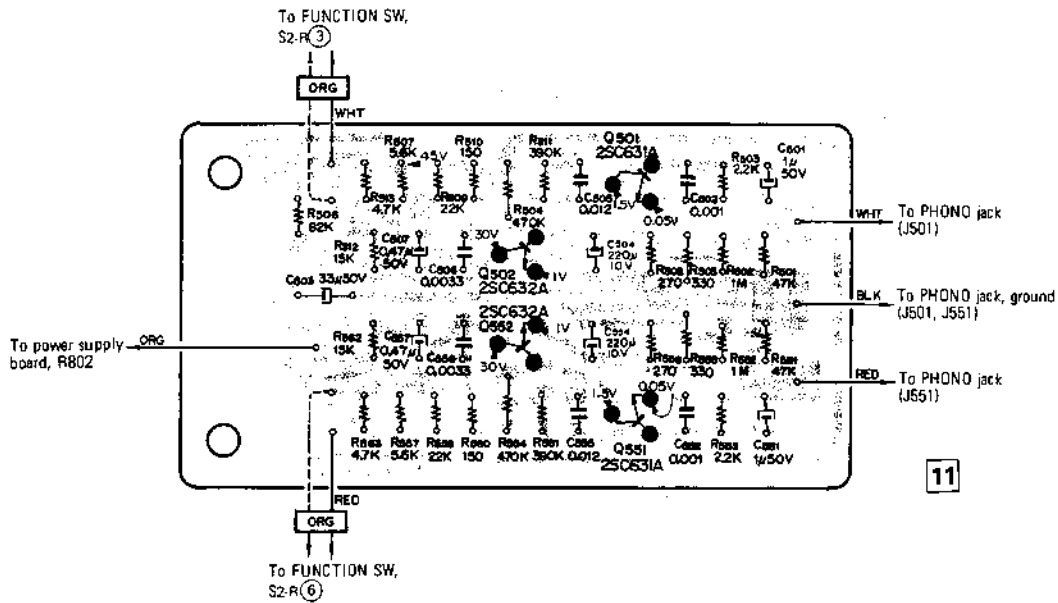
Transistors and Adjustment Parts Location

	Q401	Q403
Q402	T401	Q405
Q404	RV401	Q407
Q406		

Note: () Stereo operation



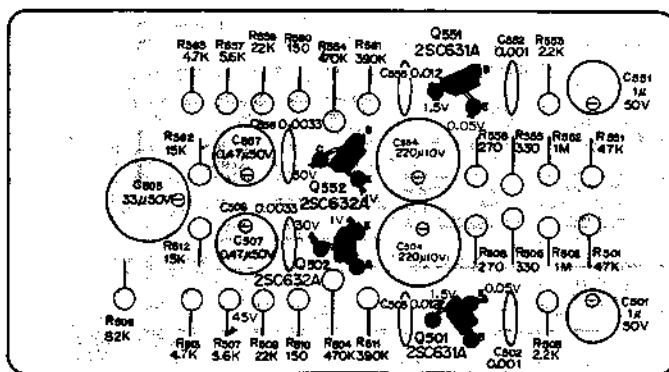
5-6. MOUNTING DIAGRAM — Preamplifier Board —
 — Conductor Side —



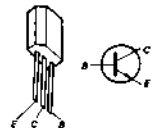
Transistor Location

	Q501
Q502	
Q552	
	Q551

— Component Side —



2SC631A
 2SC632A

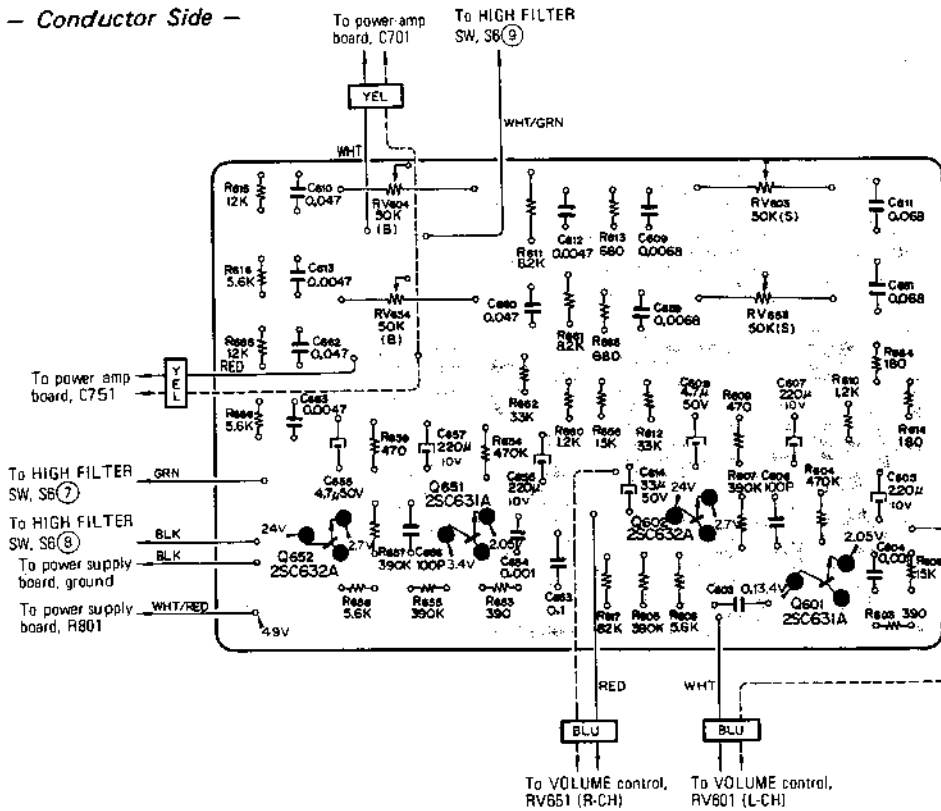


Transistor Location

	Q551
Q552	
Q502	
	Q501

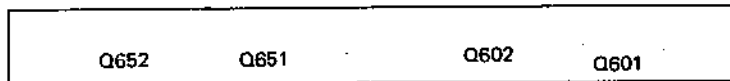
5-7. MOUNTING DIAGRAM - Tone Control Board -

- Conductor Side -

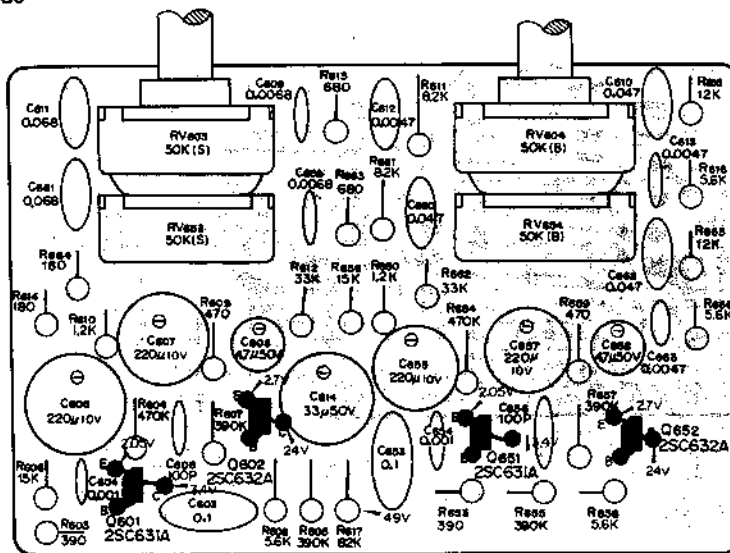


11

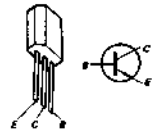
Transistor Location



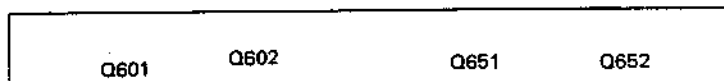
- Component Side -



2SC631A
2SC632A

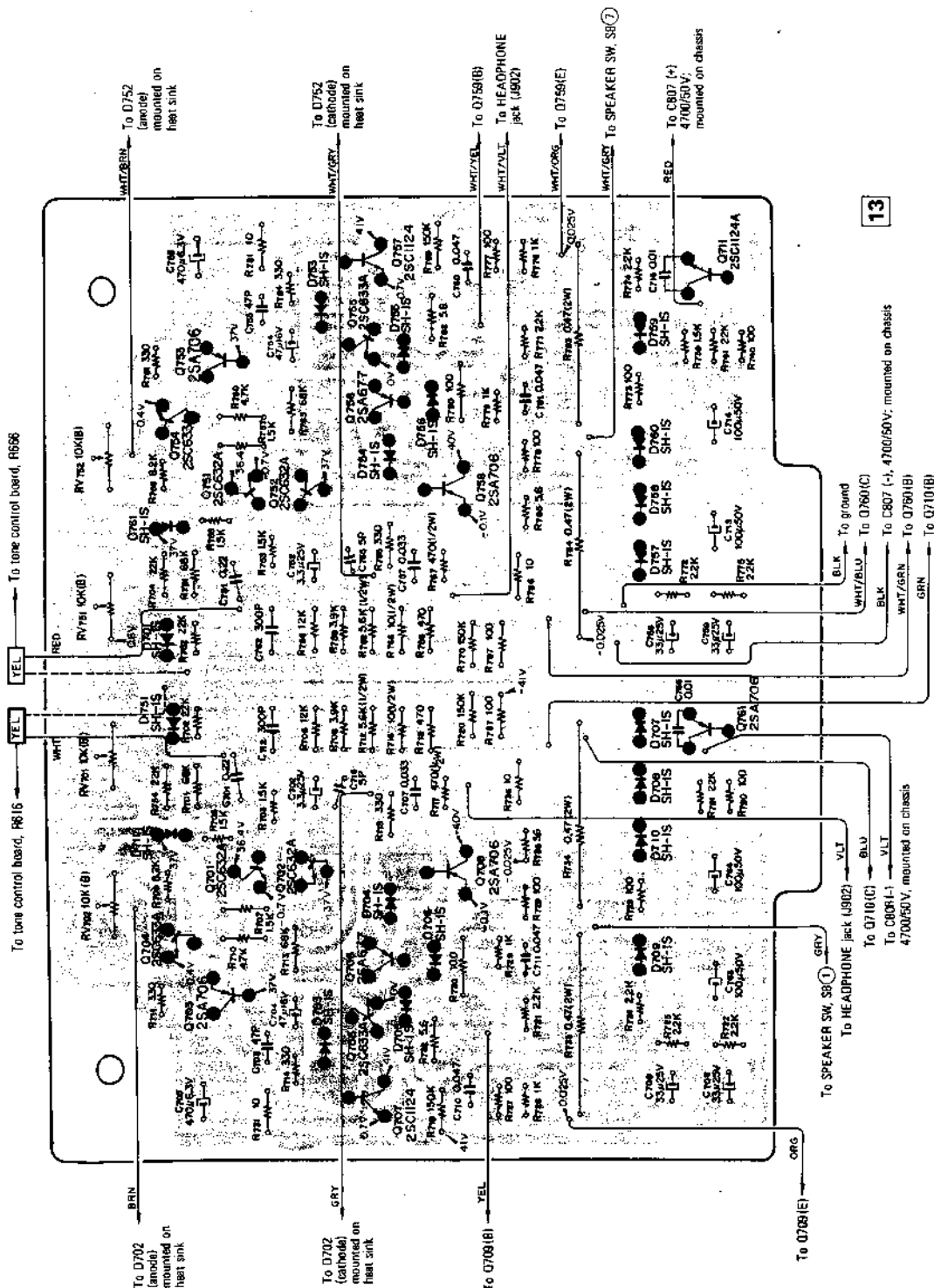


Transistor Location





5-8. MOUNTING DIAGRAM — Power Amplifier Board —
— Conductor Side —

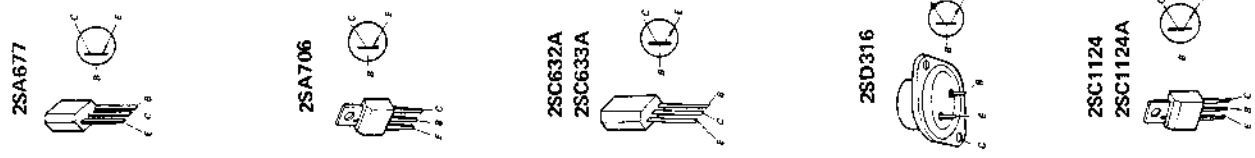
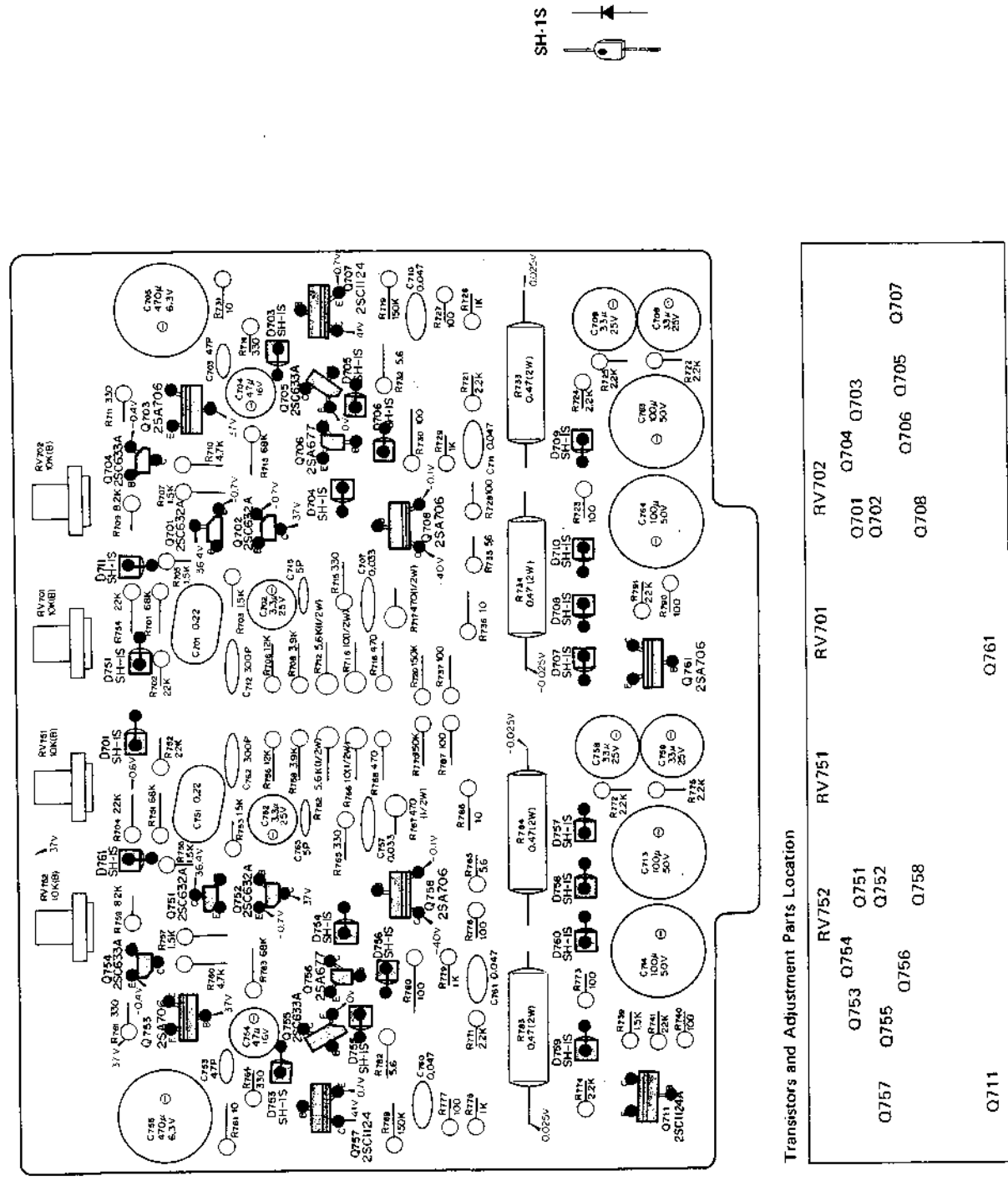


Transistors and Adjustment Parts Location

RV702	RV701	RV751	RV752
Q703	Q704	Q705	Q706
Q707	Q708	Q709	Q710
Q711	Q712	Q713	Q714
Q715	Q716	Q717	Q718
Q719	Q720	Q721	Q722
Q723	Q724	Q725	Q726
Q727	Q728	Q729	Q730
Q731	Q732	Q733	Q734
Q735	Q736	Q737	Q738
Q739	Q740	Q741	Q742
Q743	Q744	Q745	Q746
Q747	Q748	Q749	Q750

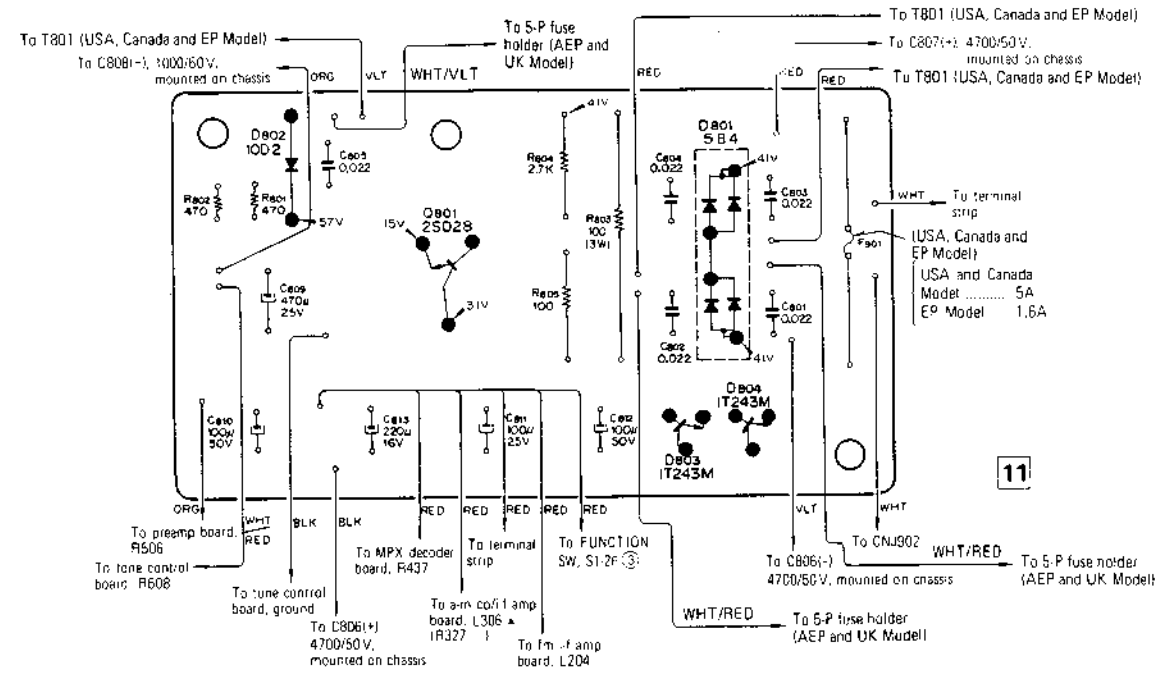
STR-6055 STR-6055

— Component Side —



5-9. MOUNTING DIAGRAM — Power Supply Board —

— Conductor Side —

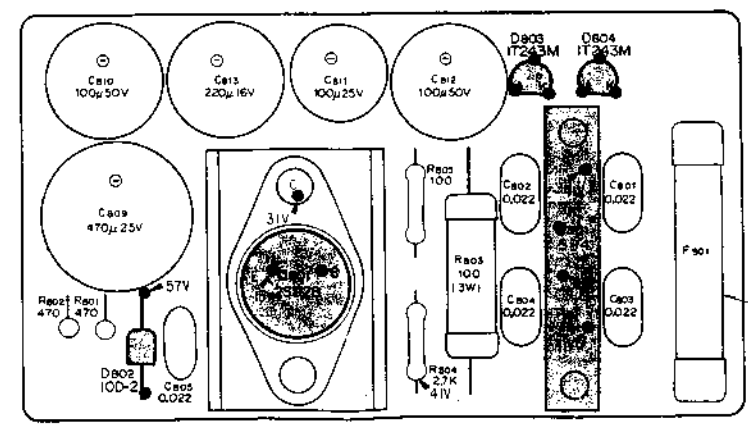


Transistor Location

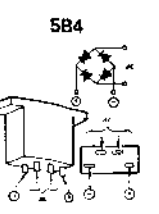
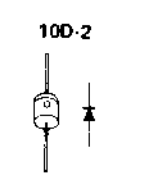
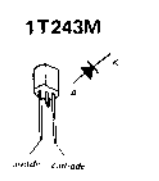
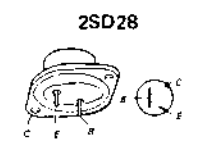
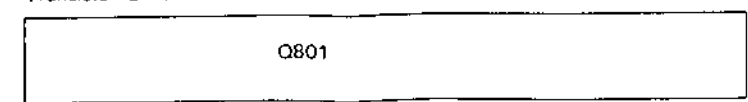


Note: USA Model Serial No. 810,401 and later
 Canada Model Serial No. 700,301 and later
 EP Model Serial No. 553,001 and later
 UK Model Serial No. 600,001 and later
 AEP Model Serial No. 900,001 and later

— Component Side —



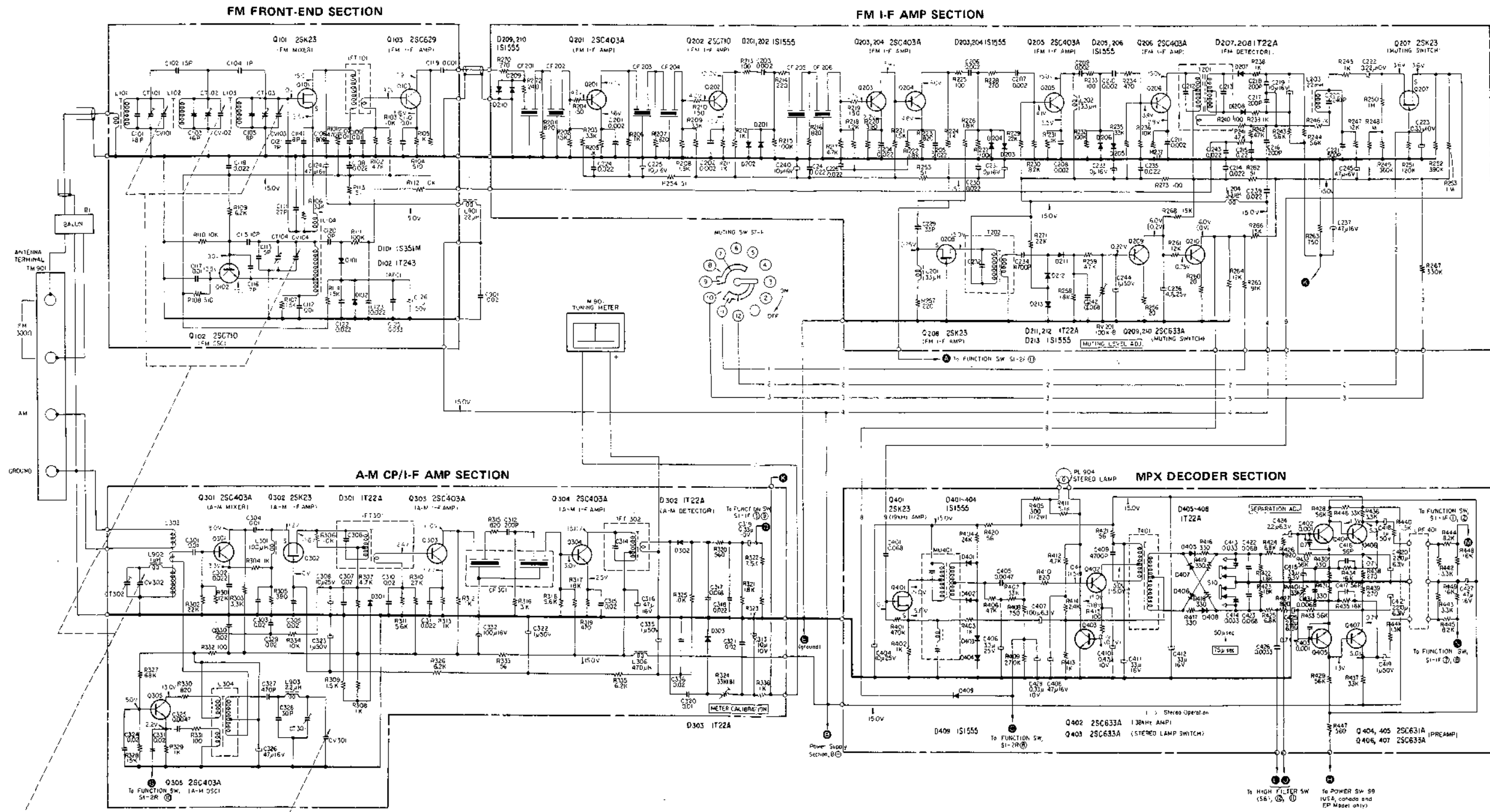
Transistor Location



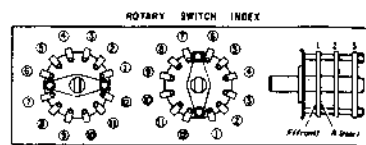
STR-6055 STR-6055

5-10. SCHEMATIC DIAGRAM – Tuner Section –

Applicable Serial Numbers:
 USA Model Up to 810,400
 Canada Model ... Up to 700,300
 EP Model Up to 553,000



Ref. No.	Function	Position
S7	MUTING SW	ON
S10	DE-EMPHASIS SW (50 μsec - 75 μsec)	75 μsec



Note:

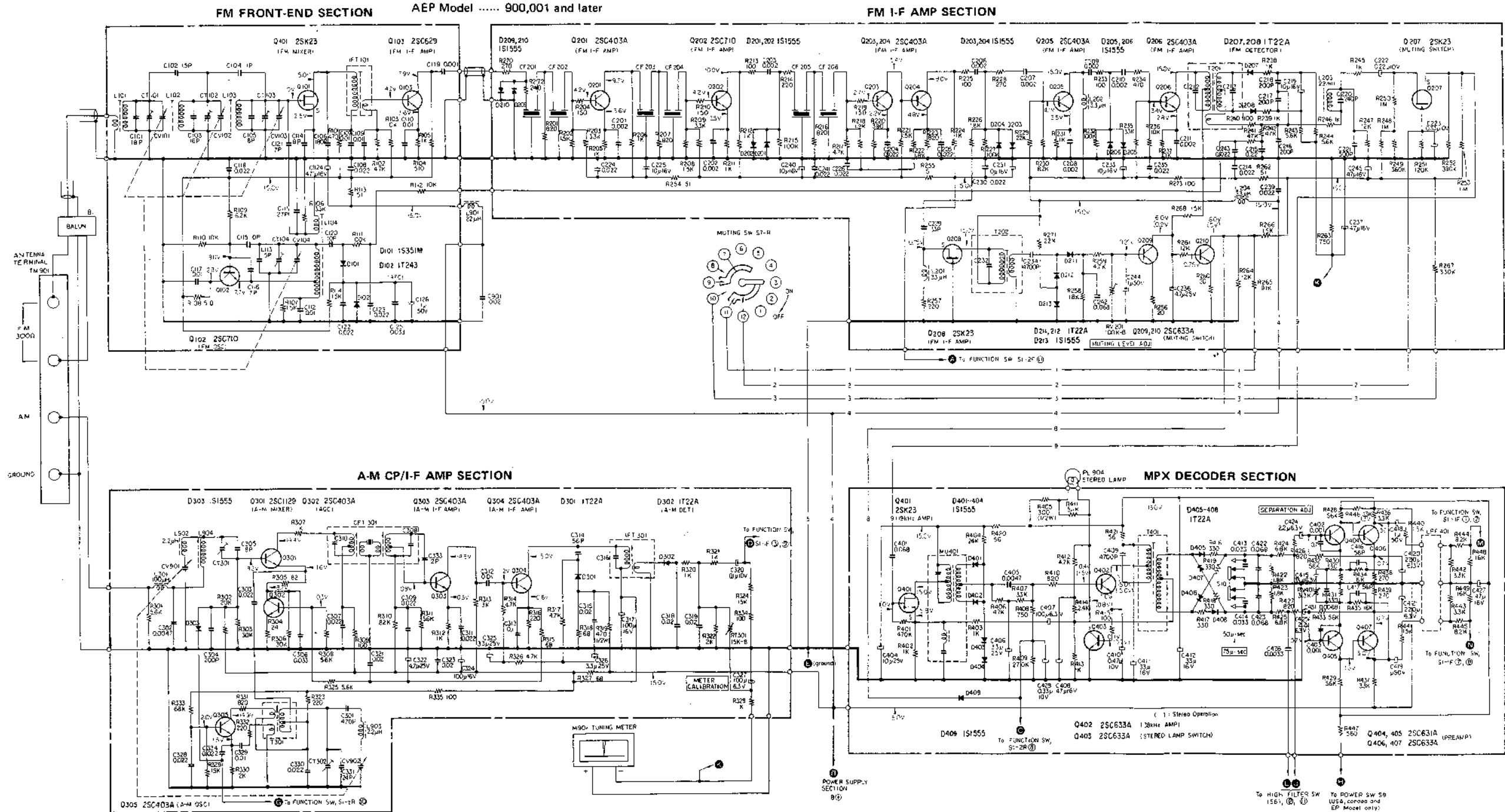
All resistance values are in ohms, k = 1,000, M = 1,000k
 All capacitance values are in μF except as indicated with p,
 which means μF.
 All voltages are dc measured with a VOM which has an
 input impedance of 20k ohms/volt. No signal in.
 All voltages represent an average value.
 () : Stereo operation
 [] : Muting operation

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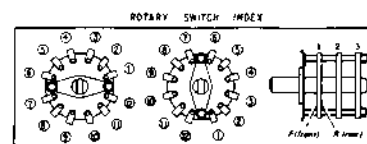
SCHMATIC DIAGRAM – Tuner Section –

Applicable Serial Numbers:

- USA Model 810,401 and later
- Canada Model... 700,301 and later
- EP Model 553,001 and later
- UK Model 600,001 and later
- AEP Model 900,001 and later



Ref. No.	Function	Position
S7	MUTING SW	ON
S10	DE-EMPHASIS SW (50 μsec – 75 μsec)	75 μsec

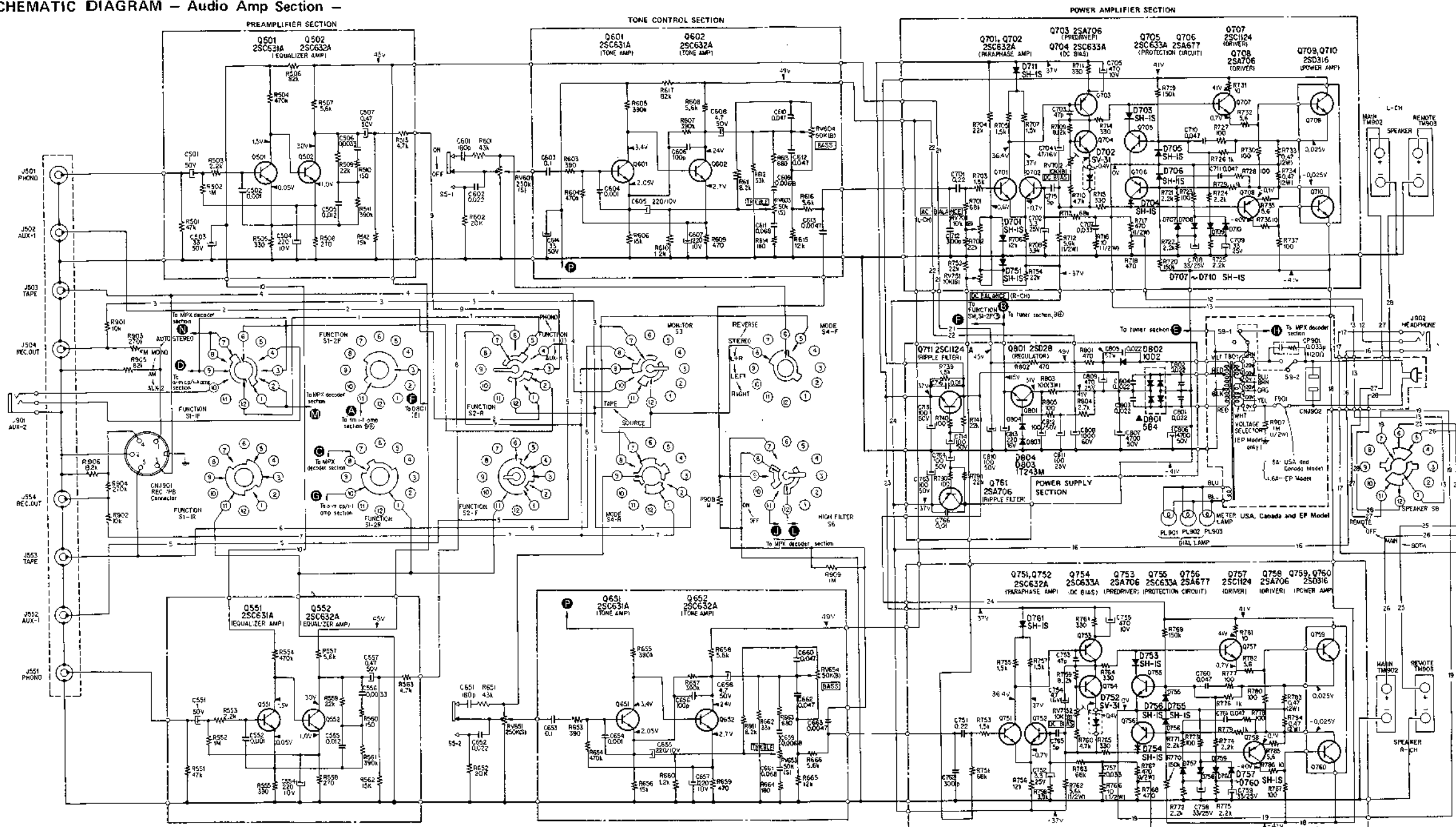


Note:

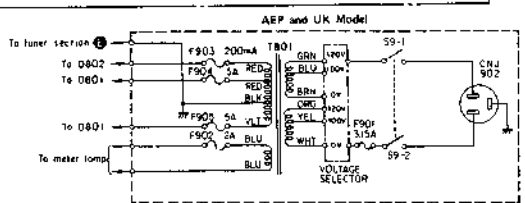
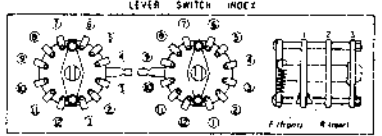
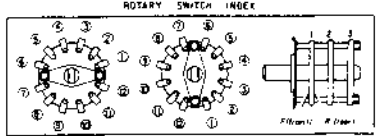
All resistance values are in ohms. k = 1,000, M = 1,000k
 All capacitance values are in μF except as indicated with p, which means μF.
 All voltages are dc measured with a VOM which has an input impedance of 20k ohms/volt. No signal in.
 All voltages represent an average value.
 () : Stereo operation
 [] : Muting operation

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5-11. SCHEMATIC DIAGRAM - Audio Amp Section -



Ref. No.	Function	Position	Ref. No.	Function	Position
S1-1~2	FUNCTION (1) SW (AUTO STEREO - MONO - AM - AUX-2)	AUTO STEREO	S5-1~2	LOUDNESS SW	ON
S2	FUNCTION (2) SW (AUX-1 - FUNCTION (1) - PHONO)	FUNCTION (1)	S6	HIGH FILTER SW	OFF
S3	MONITOR SW (SOURCE - TAPE)	SOURCE	S8	SPEAKER SW (REMOTE - OFF - MAIN - BOTH)	BOTH
S4	MODE SW (REVERSE - STEREO - L+R - LEFT - RIGHT)	STEREO	S9-1~2	POWER SW	OFF

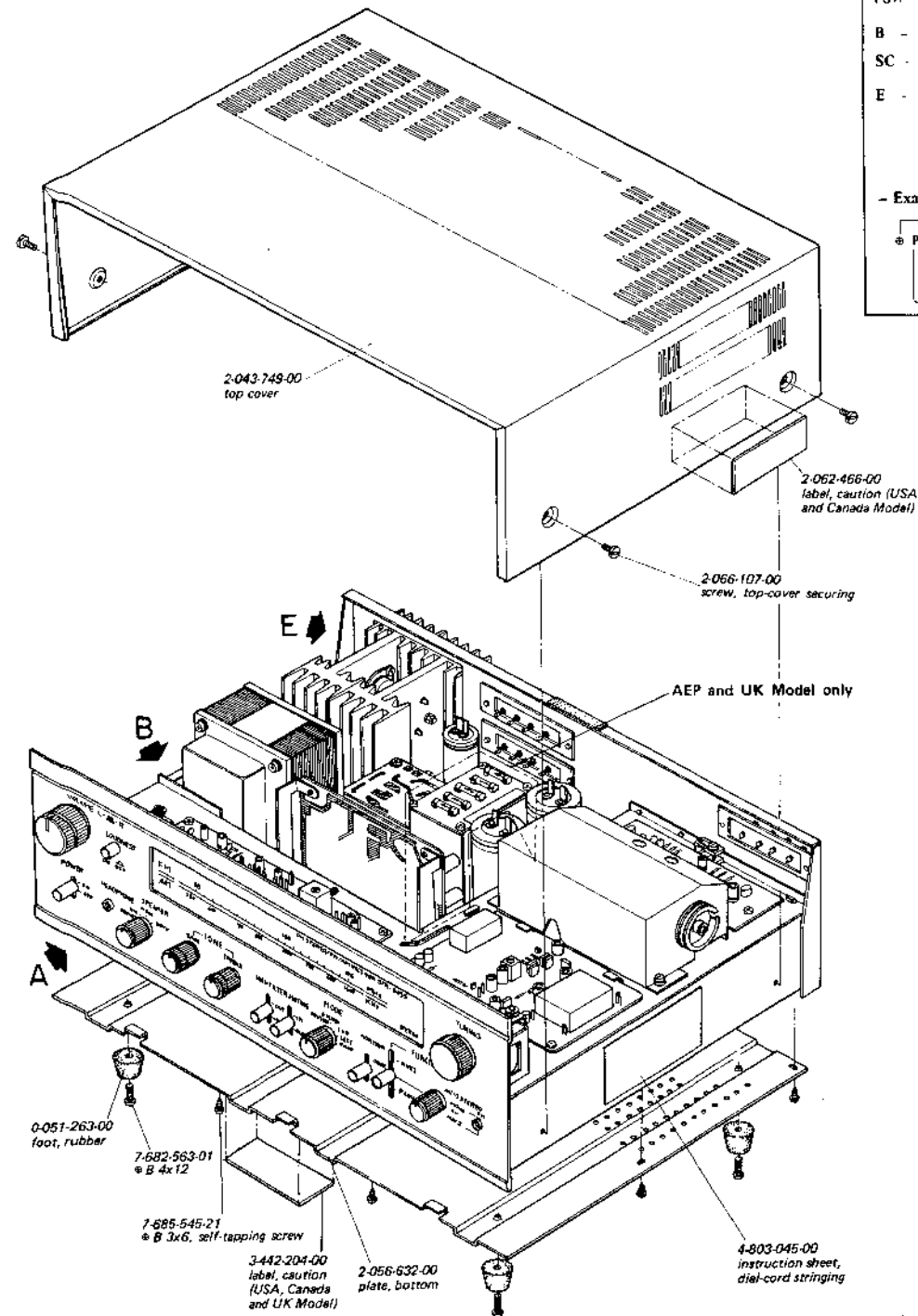


Note:
 All resistance values are in ohms. k = 1,000, M = 1,000k
 All capacitance values are in μF except as indicated with p, which means μM .
 All voltages are dc measured with a VOM which has an input impedance of 20k ohms/volt. No signal in.
 All voltages represent an average value.

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SECTION 6 EXPLODED VIEWS

(1)



— Hardware Nomenclature —

P	Pan Head Screw	
PS	Pan Head Screw with Spring Washer	
PSW	Pan Head Screw with Spring Washer and Washer	
B	Binding Head Screw	
SC	Set Screw	
E	Retaining Ring (E Washer)	
W	Washer	
SW	Spring Washer	
LW	Lock Washer	
N	Nut	

— Example —

Type of Slot:

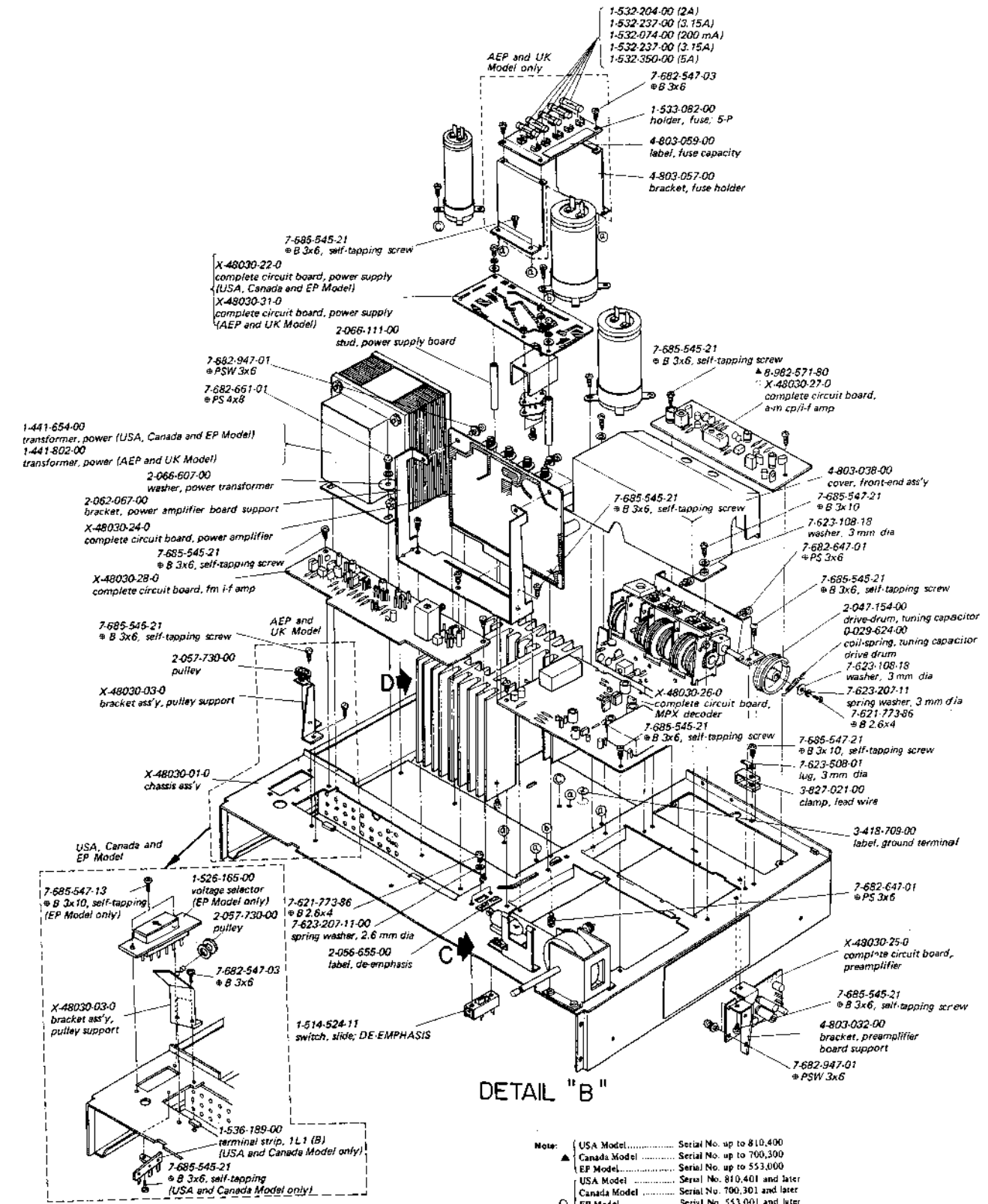
Length in mm (L):

Diameter in mm (D):

Type of Head:

Note:
USA Model: Serial No. 800,001 and later
Canada Model: Serial No. 700,001 and later
EP Model: Serial No. 550,001 and later
UK Model: Serial No. 600,001 and later
AEP Model: Serial No. 900,001 and later

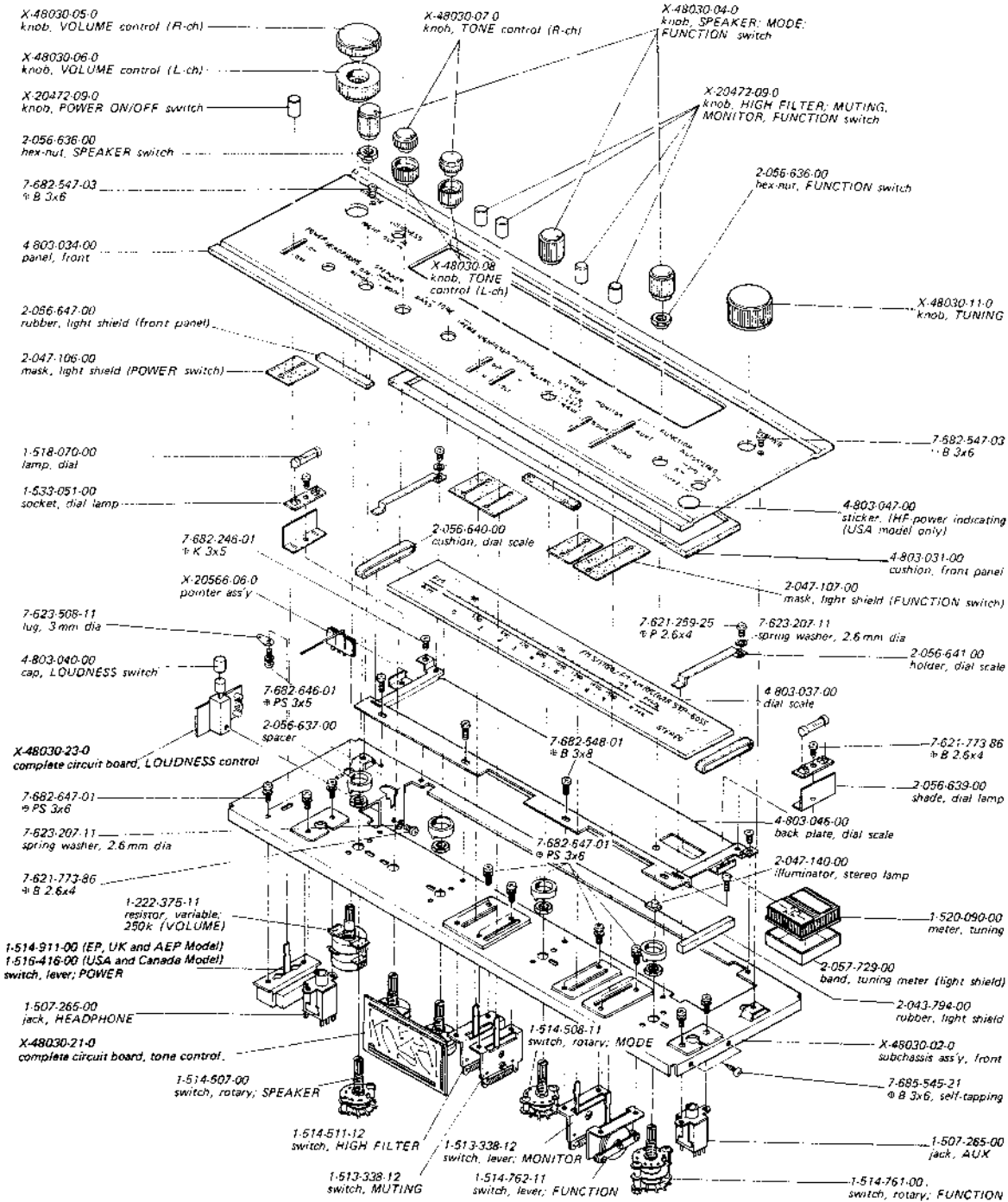
(2)



Note:
USA Model: Serial No. up to 810,400
Canada Model: Serial No. up to 700,300
EP Model: Serial No. up to 553,000
USA Model: Serial No. 810,401 and later
Canada Model: Serial No. 700,301 and later
EP Model: Serial No. 553,001 and later
UK Model: Serial No. 600,001 and later
AEP Model: Serial No. 900,001 and later

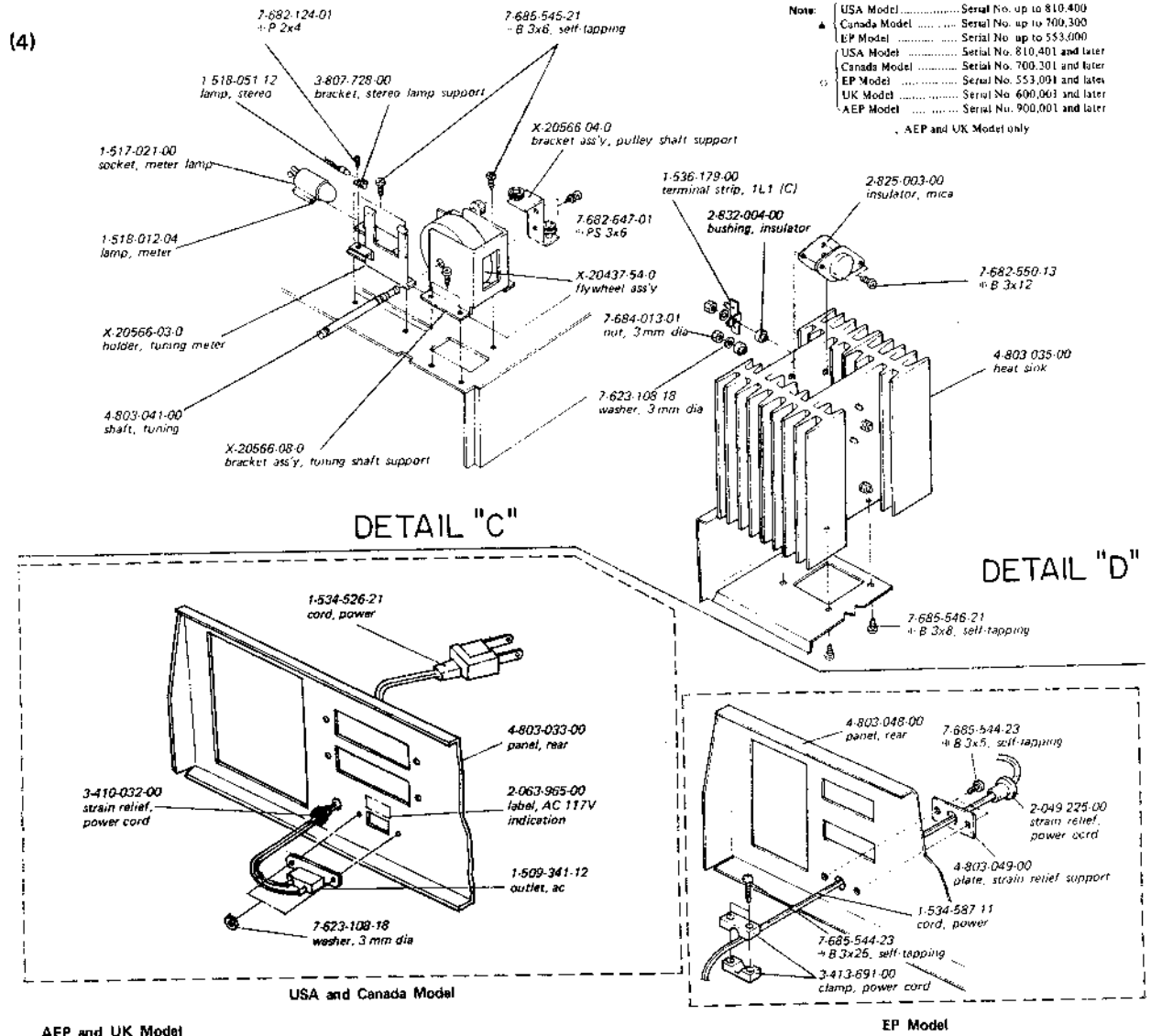
STR-6055 STR-6055

(3)



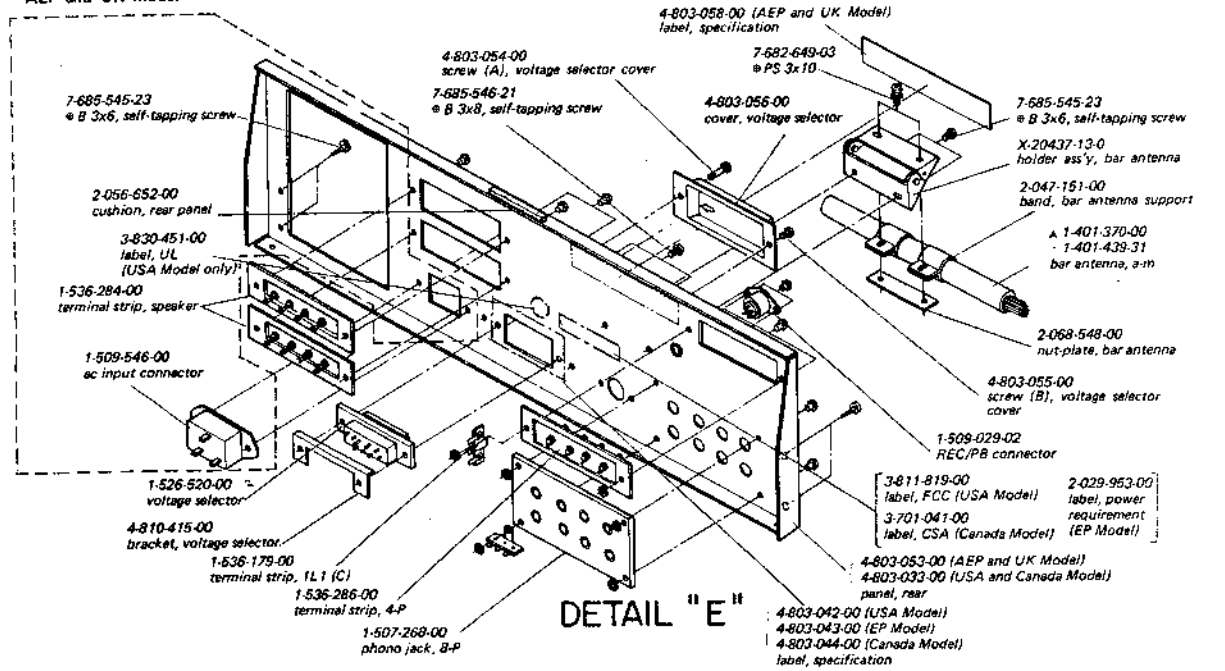
DETAIL "A" (TURN CLOCKWISE 90°)

(4)



Note: (USA Model) Serial No. up to 810,400
 (Canada Model) Serial No. up to 700,300
 (EP Model) Serial No. up to 553,000
 (USA Model) Serial No. 810,401 and later
 (Canada Model) Serial No. 700,301 and later
 (EP Model) Serial No. 553,001 and later
 (UK Model) Serial No. 600,001 and later
 (AEP Model) Serial No. 900,001 and later
 AEP and UK Model only

AEP and UK Model



DETAIL "E"

SECTION 7 ELECTRICAL PARTS LIST

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>	<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
COMPLETE CIRCUIT BOARDS					
	X-48030-21-0	tone control	D705 (D755)		diode SH-1S
	8-982-571-10	fm front-end	D706 (D756)		diode SH-1S
	X-48030-28-0	fm i-f amp	D707 (D757)		diode SH-1S
	X-48030-25-0	preamp	D708 (D758)		diode SH-1S
	X-48030-24-0	power amp	D709 (D759)		diode SH-1S
	X-48030-22-0	power supply (USA, Canada and EP Model)	D710 (D760)		diode SH-1S
	X-48030-31-0	power supply (AEP and UK Model)	D711 (D761)		diode SH-1S
	X-48030-26-0	MPX decoder	D801		diode 5B4
	▲ 8-982-571-80	a-m cp/i-f amp	D802		diode 10D2
	○ X-48030-27-0	a-m cp/i-f amp	D803		diode 1T243M
	X-48030-23-0	loudness control	D804		diode 1T243M
Note:	▲ USA Model Serial No. up to 810,400 Canada Model Serial No. up to 700,300 EP Model Serial No. up to 553,000 ○ USA Model Serial No. 810,401 and later Canada Model Serial No. 700,301 and later EP Model Serial No. 553,001 and later UK Model Serial No. 600,001 and later AEP Model Serial No. 900,001 and later		Q101		FET 2SK23
			Q102		transistor 2SC710
			Q103		transistor 2SC629
			Q201		transistor 2SC403A
			Q202		transistor 2SC710
			Q203		transistor 2SC403A
			Q204		transistor 2SC403A
			Q205		transistor 2SC403A
			Q206		transistor 2SC403A
			Q207		FET 2SK23
			Q208		FET 2SK23
			Q209		transistor 2SC633A
			Q210		transistor 2SC633A
			▲ Q301		transistor 2SC403A
			○ Q301		transistor 2SC1129
			▲ Q302		FET 2SK23
			○ Q302		transistor 2SC403A
			Q303		transistor 2SC403A
			Q304		transistor 2SC403A
			Q305		transistor 2SC403A
			Q401		FET 2SK23
			Q402		transistor 2SC633A
			Q403		transistor 2SC633A
			Q404		transistor 2SC631A
			Q405		transistor 2SC631A
			Q406		transistor 2SC633A
			Q407		transistor 2SC633A
			Q501 (Q551)		transistor 2SC631A
			Q502 (Q552)		transistor 2SC632A
			Q601 (Q651)		transistor 2SC631A
			Q602 (Q652)		transistor 2SC632A
			Q701 (Q751)		transistor 2SC632A
			Q702 (Q752)		transistor 2SC632A
			Q703 (Q753)		transistor 2SA706
			Q704 (Q754)		transistor 2SC633A
			Q705 (Q755)		transistor 2SC633A
			Q706 (Q756)		transistor 2SA677
			Q707 (Q757)		transistor 2SC1124
D701 (D751)		diode SH-1S			
D702 (D752)		diode SV-31			
D703 (D753)		diode SH-1S			
D704 (D754)		diode SH-1S			
D101		diode 1S351M			
D102		diode 1T243			
D201		diode 1S1555			
D202		diode 1S1555			
D203		diode 1S1555			
D204		diode 1S1555			
D205		diode 1S1555			
D206		diode 1S1555			
D207		diode 1T22A			
D208		diode 1T22A			
D209		diode 1S1555			
D210		diode 1S1555			
D211		diode 1T22A			
D212		diode 1T22A			
D213		diode 1S1555			
D301		diode 1T22A			
D302		diode 1T22A			
▲ D303		diode 1T22A			
○ D303		diode 1S1555			
D401		diode 1S1555			
D402		diode 1S1555			
D403		diode 1S1555			
D404		diode 1S1555			
D405		diode 1T22A			
D406		diode 1T22A			
D407		diode 1T22A			
D408		diode 1T22A			
D409		diode 1S1555			

Ref. No.	Part No.	Description
Q708 (Q758)		transistor 2SA706
Q709 (Q759)		transistor 2SD316
Q710 (Q760)		transistor 2SD316
Q711		transistor 2SC1124A
Q761		transistor 2SA706
Q801		transistor 2SD28

TRANSFORMERS, COILS AND INDUCTORS

B101	1-41 7-014-00	balun
IFT101	1-403-295-00	transformer, i-f; 10.7 MHz
o CFT301	1-403-150-00	CFT, 455 kHz
▲ IFT301	1-403-152-00	transformer, i-f; 455 kHz
o IFT301	1-403-149-00	transformer, i-f; 455 kHz
▲ IFT302	1-403-128-00	transformer, i-f; 455 kHz
L101	1-401-351-00	coil, fm antenna
L102	1-425-446-00	coil, fm rf
L103	1-425-446-00	coil, fm rf
L104	1-405-377-00	coil, fm osc
L201	1-407-163-00	inductor, micro; 33μH
L202	1-407-184-00	inductor, micro; 3.3μH
L203	1-407-408-00	inductor, micro; 22mH
L204	1-407-184-00	inductor, micro; 3.3μH
L301	1-407-169-00	inductor, micro; 100μH
▲ L302	1-407-178-00	inductor, micro; 1.0μH
▲ L303	1-401-370-00	bar antenna, a-m
▲ L304	1-405-359-00	coil, a-m osc
▲ L305		-----
▲ L306	1-407-177-00	inductor, micro; 470μH
L901	1-407-161-00	inductor, micro; 22μH
▲ L902	1-407-178-00	inductor, micro; 1.0μH
o L902	1-407-182-00	inductor, micro; 2.2μH
L903	1-407-182-00	inductor, micro; 2.2μH
o L904	1-401-439-31	bar antenna, a-m
MU401	1-425-548-00	MPX unit
T201	1-403-291-00	transformer, discriminator
T202	1-403-299-00	transformer, i-f; 10.7 kHz
o T301	1-405-459-00	coil, a-m osc
T401	1-425-260-00	transformer, switching; 38 kHz
T801	1-441-654-00	transformer, power (USA, Canada and EP Model)
	1-441-802-00	transformer, power (AEP and UK Model)

Note:

- ▲ USA Model Serial No. up to 810,400
- ▲ Canada Model Serial No. up to 700,300
- ▲ EP Model Serial No. up to 553,000
- o USA Model Serial No. 810,401 and later
- o Canada Model Serial No. 700,301 and later
- o EP Model Serial No. 553,001 and later
- o UK Model Serial No. 600,001 and later
- o AEP Model Serial No. 900,001 and later

CAPACITORS

All capacitance values are in μF except as indicated with p, which means μμF.

C101	1-101-862-11	18p	±5%	50V	ceramic
C102	1-101-938-11	1.5p	±10%	500V	ceramic

Ref. No.	Part No.	Description
C103	1-101-952-11	16p ±5% 50V ceramic
C104	1-101-937-11	1p ±10% 500V ceramic
C105	1-101-945-11	8p ±5% 50V ceramic
C106	1-102-985-11	180p ±5% 50V ceramic
C107	1-101-072-11	0.01 ± ⁸⁰ / ₂₀ % 50V ceramic
C108	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C109	1-101-072-11	0.01 ± ⁸⁰ / ₂₀ % 25V ceramic
C110	1-101-072-11	0.01 ± ⁸⁰ / ₂₀ % 25V ceramic
C111	1-101-869-11	27p ±5% 50V ceramic
C112	1-102-077-11	0.01 ±20% 50V ceramic
C113	1-101-873-11	15p ±5% 50V ceramic
C114	1-101-958-11	8p ±0.5p 50V ceramic
C115	1-101-978-11	10p ±5% 50V ceramic
C116	1-102-875-11	7p ±5% 50V ceramic
C117	1-101-072-11	0.01 ± ⁸⁰ / ₂₀ % 25V ceramic
C118	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C119	1-101-918-11	0.001 ± ⁸⁰ / ₂₀ % 25V ceramic
C120	1-101-978-11	10p ±5% 50V ceramic
C121	1-101-957-11	7p ±0.5p 50V ceramic
C122	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C123	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C124	1-121-353-11	47 16V electrolytic
C125	1-105-679-12	0.033 ±10% 50V mylar
C126	1-121-391-11	1 50V electrolytic
C201	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C202	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C203	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C204	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C205	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C206	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C207	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C208	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C209	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C210	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C211	1-101-919-11	0.002 ± ⁸⁰ / ₂₀ % 25V ceramic
C212		
C213		built in T201
C214	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C215	1-105-689-12	0.22 ±10% 50V mylar
C216	1-101-030-11	200p ±5% 50V ceramic
C217	1-101-030-11	200p ±5% 50V ceramic
C218	1-101-030-11	200p ±5% 50V ceramic
C219	1-121-471-11	10 16V electrolytic
C220	1-107-140-11	240p ±10% 50V silvered mica
C221	1-101-424-11	500p ±20% 250V ceramic
C222	1-127-020-11	0.22 10V solid aluminum
C223	1-127-021-11	0.33 10V solid aluminum
C224	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C225	1-121-471-11	10 16V electrolytic
C226	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C229	1-101-872-11	33p ±5% 50V ceramic
C230	1-101-924-11	0.022 ± ⁸⁰ / ₂₀ % 25V ceramic
C231	1-121-471-11	10 16V electrolytic
C232		built in T202
C233	1-121-471-11	10 16V electrolytic
C234	1-101-922-11	4,700p ± ⁸⁰ / ₂₀ % 50V ceramic

Ref. No.	Part No.	Description			
C235	1-101-924-11	0.022	$\pm 80\%$	25V	ceramic
C236	1-121-395-11	4.7		25V	electrolytic
C237	1-121-409-11	47		16V	electrolytic
C239	1-101-924-11	0.022	$\pm 80\%$	25V	ceramic
C240	1-121-471-11	10		16V	electrolytic
C241	1-101-924-11	0.022	$\pm 80\%$	25V	ceramic
C242	1-105-683-12	0.068	$\pm 10\%$	50V	mylar
C243	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
C244	1-121-391-11	1		50V	electrolytic
C245	1-121-409-11	47		16V	electrolytic

Note:
 ▲ USA Model Serial No. up to 810,400
 ▲ Canada Model Serial No. up to 700,300
 ▲ EP Model Serial No. up to 553,000

▲ C301	1-101-072-11	0.01	$\pm 80\%$	25V	ceramic
▲ C302	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
▲ C303	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C304	1-101-072-11	0.01	$\pm 80\%$	25V	ceramic
▲ C305	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C306				built in IFT301	
▲ C307	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C308	1-121-398-11	10		25V	electrolytic
▲ C309				-----	
▲ C310	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C311	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
▲ C312	1-103-608-11	200p	$\pm 5\%$	50V	styrol
▲ C313	1-121-469-11	10		10V	electrolytic
▲ C314				built in IFT302	
▲ C315	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C316	1-121-409-11	47		16V	electrolytic
▲ C317	1-105-683-12	0.068	$\pm 10\%$	50V	mylar
▲ C318	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
▲ C319	1-127-021-11	0.33		10V	solid aluminum
▲ C320	1-101-072-11	0.01	$\pm 80\%$	25V	ceramic
▲ C321	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C322	1-121-391-11	1		50V	electrolytic
▲ C323	1-121-391-11	1		50V	electrolytic
▲ C324	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C325	1-105-829-12	0.0047	$\pm 20\%$	50V	mylar
▲ C326	1-121-409-11	47		16V	electrolytic
▲ C327	1-103-617-11	470p	$\pm 5\%$	50V	styrol
▲ C328	1-101-871-11	30p	$\pm 5\%$	50V	ceramic
▲ C329	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C330	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C331	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
▲ C332	1-121-415-11	100		16V	electrolytic
▲ C335	1-121-391-11	1		50V	electrolytic
▲ C336	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic

Note:
 ○ USA Model Serial No. 810,401 and later
 ○ Canada Model Serial No. 700,301 and later
 ○ EP Model Serial No. 553,001 and later
 ○ UK Model Serial No. 600,001 and later
 ○ AEP Model Serial No. 900,001 and later

○ C301	1-103-617-11	470p	$\pm 5\%$	50V	styrol
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Ref. No.	Part No.	Description			
○ C302	1-105-829-12	0.0047	$\pm 20\%$	50V	mylar
○ C303	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ C304	1-102-977-11	200p	$\pm 5\%$	50V	ceramic
○ C305	1-102-945-11	8p	$\pm 5\%$	50V	ceramic
○ C306	1-105-679-12	0.033	$\pm 20\%$	50V	mylar
○ C307	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ C308				built in CFT301	
○ C309	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ C310				built in CFT301	
○ C311	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ C312	1-105-673-12	0.01	$\pm 20\%$	50V	mylar
○ C313	1-105-685-12	0.1	$\pm 20\%$	50V	mylar
○ C314	1-101-884-11	56p	$\pm 5\%$	50V	ceramic
○ C315	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
○ C316				built in IFT301	
○ C317	1-121-415-11	100		16V	electrolytic
○ C318	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
○ C319	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
○ C320	1-127-019-11	0.1		10V	solid aluminum
○ C321	1-102-073-11	0.02	$\pm 80\%$	50V	ceramic
○ C322	1-121-395-11	4.7		25V	electrolytic
○ C323	1-101-073-11	0.02	$\pm 80\%$	25V	ceramic
○ C324	1-121-415-11	100		16V	electrolytic
○ C325	1-121-456-11	3.3		25V	electrolytic
○ C326	1-121-456-11	3.3		25V	electrolytic
○ C327	1-121-413-11	100		6.3V	electrolytic
○ C328	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ C329	1-105-673-12	0.01	$\pm 20\%$	50V	mylar
○ C330	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ C331	1-102-960-11	24p	$\pm 5\%$	50V	ceramic
○ C332				-----	
○ C333	1-102-935-11	2p	± 0.25 pF	50V	ceramic
○ C334	1-105-837-12	0.022	$\pm 20\%$	50V	mylar
○ CT301					
○ CT302	1-141-095-11			capacitor, trimmer	

C401	1-105-683-12	0.068	$\pm 10\%$	50V	mylar
C402	1-105-661-12	0.001	$\pm 10\%$	50V	mylar
C403	1-105-661-12	0.001	$\pm 10\%$	50V	mylar
C404	1-121-398-11	10		25V	electrolytic
C405	1-105-669-12	0.0047	$\pm 10\%$	50V	mylar
C406	1-121-344-11	3.3		25V	electrolytic
C407	1-121-413-11	100		6.3V	electrolytic
C408	1-121-409-11	47		16V	electrolytic
C409	1-103-575-11	4,700p	$\pm 5\%$	50V	styrol
C410	1-127-022-11	0.47		10V	solid aluminum
C411	1-121-403-11	33		16V	electrolytic
C412	1-121-403-11	33		16V	electrolytic
C413	1-105-679-12	0.033	$\pm 10\%$	50V	mylar
C414	1-105-679-12	0.033	$\pm 10\%$	50V	mylar
C415	1-121-402-11	33		6.3V	electrolytic
C416	1-101-884-11	56p	$\pm 5\%$	50V	ceramic
C417	1-101-884-11	56p	$\pm 5\%$	50V	ceramic
C418	1-121-391-11	1		50V	electrolytic
C419	1-121-391-11	1		50V	electrolytic
C420	1-121-419-11	220		6.3V	electrolytic

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>		
C421	1-121-419-11	220		6.3V electrolytic
C422	1-105-683-12	0.068	±10%	50V mylar
C423	1-105-683-12	0.068	±10%	50V mylar
C424	1-127-013-11	2.2		6.3V solid aluminum
C425	1-127-013-11	2.2		6.3V solid aluminum
C426	1-105-667-12	0.0033	±10%	6.3V mylar
C427	1-121-409-11	47		16V electrolytic
C428	1-127-021-11	0.33		10V solid aluminum
C431	1-105-671-12	0.0068	±10%	50V mylar
C501 (C551)	1-121-912-11	1		50V electrolytic
C502 (C552)	1-105-661-12	0.001	±10%	50V mylar
C503	1-121-405-11	33		50V electrolytic
C504 (C554)	1-121-420-11	220		10V electrolytic
C505 (C555)	1-105-674-12	0.012	±10%	50V mylar
C506 (C556)	1-105-667-12	0.0033	±10%	50V mylar
C507 (C557)	1-121-911-11	0.47		50V electrolytic
C601 (C651)	1-107-137-11	180p	±10%	50V silvered mica
C602 (C652)	1-105-677-12	0.022	±10%	50V mylar
C603 (C653)	1-105-685-12	0.1	±10%	50V mylar
C604 (C654)	1-105-661-12	0.001	±10%	50V mylar
C605 (C655)	1-121-420-11	220		10V electrolytic
C606 (C656)	1-107-131-11	100p	±10%	50V silvered mica
C607 (C657)	1-121-420-11	220		10V electrolytic
C608 (C658)	1-121-396-11	4.7		50V electrolytic
C609 (C659)	1-105-671-12	0.0068	±10%	50V mylar
C610 (C660)	1-105-681-12	0.047	±10%	50V mylar
C611 (C661)	1-105-683-12	0.068	±10%	50V mylar
C612 (C662)	1-105-681-12	0.047	±10%	50V mylar
C613 (C663)	1-105-669-12	0.0047	±10%	50V mylar
C614	1-121-405-11	33		50V electrolytic
C701 (C751)	1-105-689-12	0.22	±10%	50V mylar
C702 (C752)	1-121-392-11	3.3		25V electrolytic
C703 (C753)	1-107-123-11	47p	±10%	50V silvered mica
C704 (C754)	1-121-409-11	47		16V electrolytic
C705 (C755)	1-121-425-11	470		10V electrolytic
C707 (C757)	1-105-679-12	0.033	±10%	50V mylar
C708 (C758)	1-121-404-11	33		25V electrolytic
C709 (C759)	1-121-404-11	33		25V electrolytic
C710 (C760)	1-105-681-12	0.047	±10%	50V mylar
C711 (C761)	1-105-681-12	0.047	±10%	50V mylar
C712 (C762)	1-107-142-11	300p	±10%	50V silvered mica
C713 (C763)	1-121-417-11	100		50V electrolytic
C714 (C764)	1-121-417-11	100		50V electrolytic
C715 (C765)	1-102-942-11	5p	±0.5p	50V ceramic
C716 (C766)	1-105-673-12	0.01	±10%	50V mylar
C801	1-105-917-12	0.022	±20%	200V mylar
C802	1-105-917-12	0.022	±20%	200V mylar
C803	1-105-917-12	0.022	±20%	200V mylar
C804	1-105-917-12	0.022	±20%	200V mylar
C805	1-105-917-12	0.022	±20%	200V mylar
C806	1-121-815-11	4,700		50V electrolytic
C807	1-121-815-11	4,700		50V electrolytic
C808	1-121-330-11	1,000		60V electrolytic

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>		
C809	1-121-733-11	470		25V electrolytic
C810	1-121-417-11	100		50V electrolytic
C811	1-121-416-11	100		25V electrolytic
C812	1-121-417-11	100		50V electrolytic
C813	1-121-358-11	220		16V electrolytic
C901	1-101-073-11	0.02	±80%	25V ceramic
CV101	1-151-191-13	capacitor, tuning		
CV102				
CV103				
CV104				
CV301				
CV302				

RESISTORS

All resistors are in ohms, ±5%, ¼W and carbon type unless otherwise indicated.

R101	1-244-665-11	470
R102	1-244-689-11	4.7 k
R103	1-244-697-11	10 k
R104	1-244-666-11	510
R105	1-244-673-11	1 k
R106	1-244-685-11	3.3 k
R107	1-244-677-11	1.5 k
R108	1-244-666-11	510
R109	1-244-692-11	6.2 k
R110	1-244-697-11	10k
R111	1-244-721-11	100 k
R112	1-244-697-11	10 k
R113	1-244-642-11	51
R114	1-244-677-11	1.5 k
R201	1-244-671-11	820
R202	1-244-677-11	1.5 k
R203	1-244-685-11	3.3 k
R204	1-244-653-11	150
R205	1-244-673-11	1 k
R206	1-244-673-11	1 k
R207	1-244-671-11	820
R208	1-244-677-11	1.5 k
R209	1-244-685-11	3.3 k
R210	1-244-653-11	150
R211	1-244-673-11	1 k
R212	1-244-673-11	1 k
R213	1-244-649-11	100
R214	1-244-657-11	220
R215	1-244-721-11	100 k
R216	1-244-671-11	820
R217	1-244-689-11	4.7 k
R218	1-244-675-11	1.2 k
R219	1-244-653-11	150
R220	1-244-663-11	390
R221	1-244-677-11	1.5 k
R222	1-244-679-11	1.8 k
R223	1-244-671-11	820
R224	1-244-673-11	1 k
R225	1-244-649-11	100

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
R226	1-244-679-11	1.8k
R227	1-244-721-11	100k
R228	1-244-659-11	270
R229	1-244-705-11	22k
R230	1-244-695-11	8.2k
R231	1-244-673-11	1k
R232	1-244-721-11	100k
R233	1-244-649-11	100
R234	1-244-665-11	470
R235	1-244-709-11	33k
R236	1-244-697-11	10k
R237	1-244-673-11	1k
R238	1-244-673-11	1k
R239	1-244-673-11	1k
R240	1-244-649-11	100
R241	1-244-713-11	47k
R242	1-244-713-11	47k
R243	1-244-691-11	5.6k
R244	1-244-691-11	5.6k
R245	1-244-673-11	1k
R246	1-244-673-11	1k
R247	1-244-699-11	12k
R248	1-244-745-11	1M
R249	1-244-734-11	360k
R250	1-244-745-11	1M
R251	1-244-723-11	120k
R252	1-244-735-11	390k
R253	1-244-745-11	1M
R254	1-244-642-11	51
R255	1-244-642-11	51
R256	1-244-632-11	20
R257	1-244-657-11	220
R258	1-244-703-11	18k
R259	1-244-689-11	4.7k
R260	1-244-632-11	20
R261	1-244-699-11	12k
R262	1-244-642-11	51
R263	1-244-670-11	750
R264	1-244-699-11	12k
R265	1-244-720-11	91k
R266	1-244-701-11	15k
R267	1-244-733-11	330k
R268	1-244-701-11	15k
R269	-----	-----
R270	1-244-659-11	270
R271	1-244-705-11	22k
R272	1-244-658-11	240
R273	1-244-649-11	100

Note: { USA Model Serial No. up to 810,400
 { Canada Model Serial No. up to 700,300
 { EP Model Serial No. up to 553,000

▲ R301	1-244-699-11	12k
▲ R302	1-244-705-11	22k
▲ R303	1-244-685-11	3.3k
▲ R304	1-244-673-11	1k

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
▲ R305	1-244-663-11	390
▲ R306	1-244-697-11	10k
▲ R307	1-244-689-11	4.7k
▲ R308	1-244-673-11	1k
▲ R309	1-244-677-11	1.5k
▲ R310	1-244-707-11	27k
▲ R311	1-244-691-11	5.6k
▲ R312	1-244-673-11	1k
▲ R313	1-244-673-11	1k
▲ R314	-----	-----
▲ R315	1-244-671-11	820
▲ R316	1-244-684-11	3k
▲ R317	1-244-703-11	18k
▲ R318	1-244-691-11	5.6k
▲ R319	1-244-665-11	470
▲ R320	1-244-667-11	560
▲ R321	1-244-679-11	1.8k
▲ R322	1-244-694-11	7.5k
▲ R323	1-244-673-11	1k
▲ R324	1-222-984-11	33k(B), adjustable
▲ R325	1-244-697-11	10k
▲ R326	1-244-692-11	6.2k
▲ R327	1-244-717-11	68k
▲ R328	1-244-701-11	15k
▲ R329	1-244-673-11	1k
▲ R330	1-244-671-11	820
▲ R331	1-244-649-11	100
▲ R332	1-244-649-11	100
▲ R333	1-244-643-11	56
▲ R334	1-244-697-11	10k
▲ R335	1-244-692-11	6.2k
▲ R336	1-244-673-11	1k

Note: { USA Model Serial No. 810,401 and later
 { Canada Model Serial No. 700,301 and later
 { EP Model Serial No. 553,001 and later
 { UK Model Serial No. 600,001 and later
 { AEP Model Serial No. 900,001 and later

○ R301	1-244-691-11	5.6k
○ R302	1-244-704-11	20k
○ R303	1-244-708-11	30k
○ R304	1-244-634-11	24
○ R305	1-244-647-11	82
○ R306	1-244-708-11	30k
○ R307	1-244-673-11	1k
○ R308	1-244-691-11	5.6k
○ R309	1-244-649-11	100
○ R310	1-244-719-11	82k
○ R311	1-244-691-11	5.6k
○ R312	1-244-673-11	1k
○ R313	1-244-684-11	3k
○ R314	1-244-689-11	4.7k
○ R315	1-244-645-11	68
○ R316	1-244-657-11	220
○ R317	1-244-689-11	4.7k
○ R318	1-244-645-11	68
○ R319	1-202-565-11	470
○ R320	1-244-673-11	1k

±10% ½W composition

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
o R321	1-244-673-11	1 k
o R322	1-244-680-11	2 k
o R323	1-242-657-11	220
o R324	1-242-701-11	15 k
o R325	1-244-691-11	5.6 k
o R326	1-244-689-11	4.7 k
o R327	1-244-645-11	68
o R328	1-244-673-11	1 k
o R329	1-244-701-11	15 k
o R330	1-244-680-11	2 k
o R331	1-244-671-11	820
o R332	1-244-657-11	220
o R333	1-244-717-11	68 k
o R334	1-244-649-11	100
o R335	1-244-649-11	100
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R401	1-244-739-11	470 k
R402	1-244-673-11	1 k
R403	1-244-673-11	1 k
R404	1-244-706-11	24 k
R405	1-202-560-11	300 ±10% ½W composition
R406	1-244-689-11	4.7 k
R407	1-244-709-11	33 k
R408	1-244-670-11	750
R409	1-244-731-11	270 k
R410	1-244-671-11	820
R411	1-244-690-11	5.1 k
R412	1-244-689-11	4.7 k
R413	1-244-673-11	1 k
R414	1-244-682-11	2.4 k
R415	1-244-649-11	100
R416	1-244-661-11	330
R417	1-244-661-11	330
R418	1-244-661-11	330
R419	1-244-661-11	330
R420	1-244-643-11	56
R421	1-244-643-11	56
R422	1-244-679-11	1.8 k
R423	1-244-679-11	1.8 k
R424	1-244-693-11	6.8 k
R425	1-244-693-11	6.8 k
R426	1-244-671-11	820
R427	1-244-671-11	820
R428	1-244-715-11	56 k
R429	1-244-715-11	56 k
R430	1-244-661-11	330
R431	1-244-661-11	330
R432	1-244-715-11	56 k
R433	1-244-715-11	56 k
R434	1-244-702-11	16 k
R435	1-244-702-11	16 k
R436	1-244-685-11	3.3 k
R437	1-244-685-11	3.3 k
R438	1-244-659-11	270
R439	1-244-659-11	270
R440	1-244-677-11	1.5 k

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
R441	1-244-677-11	1.5 k
R442	1-244-685-11	3.3 k
R443	1-244-685-11	3.3 k
R444	1-244-695-11	8.2 k
R445	1-244-695-11	8.2 k
R446	1-244-709-11	33 k
R447	1-244-667-11	560
R448	1-244-702-11	16 k
R449	1-244-702-11	16 k
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R501 (R551)	1-242-713-11	47 k
R502 (R552)	1-242-745-11	1M
R503 (R553)	1-242-681-11	2.2 k
R504 (R554)	1-242-737-11	470 k
R505 (R555)	1-242-661-11	330
R506	1-242-719-11	82 k
R507 (R557)	1-242-691-11	5.6 k
R508 (R558)	1-242-659-11	270
R509 (R559)	1-242-705-11	22 k
R510 (R560)	1-242-653-11	150
R511 (R561)	1-242-735-11	390 k
R512 (R562)	1-242-701-11	15 k
R513 (R563)	1-242-689-11	4.7 k
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R601 (R651)	1-244-712-11	43 k
R602 (R652)	1-244-704-11	20 k
R603 (R653)	1-242-663-11	390
R604 (R654)	1-242-737-11	470 k
R605 (R655)	1-242-735-11	390 k
R606 (R656)	1-242-701-11	15 k
R607 (R657)	1-242-735-11	390 k
R608 (R658)	1-242-691-11	5.6 k
R609 (R659)	1-242-665-11	470
R610 (R660)	1-242-675-11	1.2 k
R611 (R661)	1-242-695-11	8.2 k
R612 (R662)	1-242-709-11	33 k
R613 (R663)	1-242-669-11	680
R614 (R664)	1-242-655-11	180
R615 (R665)	1-242-699-11	12 k
R616 (R666)	1-242-691-11	5.6 k
R617	1-242-719-11	82 k
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R701 (R751)	1-242-717-11	68 k
R702 (R752)	1-242-705-11	22 k
R703 (R753)	1-242-677-11	1.5 k
R704 (R754)	1-242-705-11	22 k
R705 (R755)	1-242-677-11	1.5 k
R706 (R756)	1-242-699-11	12 k
R707 (R757)	1-242-677-11	1.5 k
R708 (R758)	1-242-687-11	3.9 k
R709 (R759)	1-242-695-11	8.2 k
R710 (R760)	1-242-689-11	4.7 k
R711 (R761)	1-242-661-11	330
R712 (R762)	1-202-591-11	5.6 k ±10% ½W composition
R713 (R763)	1-242-717-11	68 k
R714 (R764)	1-242-661-11	330
R715 (R765)	1-242-661-11	330

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>	<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
○ R321	1-244-673-11	1k	R441	1-244-677-11	1.5k
○ R322	1-244-680-11	2k	R442	1-244-685-11	3.3k
○ R323	1-242-657-11	220	R443	1-244-685-11	3.3k
○ R324	1-242-701-11	15k	R444	1-244-695-11	8.2k
○ R325	1-244-691-11	5.6k	R445	1-244-695-11	8.2k
○ R326	1-244-689-11	4.7k	R446	1-244-709-11	33k
○ R327	1-244-645-11	68	R447	1-244-667-11	560
○ R328	1-244-673-11	1k	R448	1-244-702-11	16k
○ R329	1-244-701-11	15k	R449	1-244-702-11	16k
○ R330	1-244-680-11	2k	R501 (R551)	1-242-713-11	47k
○ R331	1-244-671-11	820	R502 (R552)	1-242-745-11	1M
○ R332	1-244-657-11	220	R503 (R553)	1-242-681-11	2.2k
○ R333	1-244-717-11	68k	R504 (R554)	1-242-737-11	470k
○ R334	1-244-649-11	100	R505 (R555)	1-242-661-11	330
○ R335	1-244-649-11	100	R506	1-242-719-11	82k
R401	1-244-739-11	470k	R507 (R557)	1-242-691-11	5.6k
R402	1-244-673-11	1k	R508 (R558)	1-242-659-11	270
R403	1-244-673-11	1k	R509 (R559)	1-242-705-11	22k
R404	1-244-706-11	24k	R510 (R560)	1-242-653-11	150
R405	1-202-560-11	300 ±10% ½W composition	R511 (R561)	1-242-735-11	390k
R406	1-244-689-11	4.7k	R512 (R562)	1-242-701-11	15k
R407	1-244-709-11	33k	R513 (R563)	1-242-689-11	4.7k
R408	1-244-670-11	750	R601 (R651)	1-244-712-11	43k
R409	1-244-731-11	270k	R602 (R652)	1-244-704-11	20k
R410	1-244-671-11	820	R603 (R653)	1-242-663-11	390
R411	1-244-690-11	5.1k	R604 (R654)	1-242-737-11	470k
R412	1-244-689-11	4.7k	R605 (R655)	1-242-735-11	390k
R413	1-244-673-11	1k	R606 (R656)	1-242-701-11	15k
R414	1-244-682-11	2.4k	R607 (R657)	1-242-735-11	390k
R415	1-244-649-11	100	R608 (R658)	1-242-691-11	5.6k
R416	1-244-661-11	330	R609 (R659)	1-242-665-11	470
R417	1-244-661-11	330	R610 (R660)	1-242-675-11	1.2k
R418	1-244-661-11	330	R611 (R661)	1-242-695-11	8.2k
R419	1-244-661-11	330	R612 (R662)	1-242-709-11	33k
R420	1-244-643-11	56	R613 (R663)	1-242-669-11	680
R421	1-244-643-11	56	R614 (R664)	1-242-655-11	180
R422	1-244-679-11	1.8k	R615 (R665)	1-242-699-11	12k
R423	1-244-679-11	1.8k	R616 (R666)	1-242-691-11	5.6k
R424	1-244-693-11	6.8k	R617	1-242-719-11	82k
R425	1-244-693-11	6.8k	R701 (R751)	1-242-717-11	68k
R426	1-244-671-11	820	R702 (R752)	1-242-705-11	22k
R427	1-244-671-11	820	R703 (R753)	1-242-677-11	1.5k
R428	1-244-715-11	56k	R704 (R754)	1-242-705-11	22k
R429	1-244-715-11	56k	R705 (R755)	1-242-677-11	1.5k
R430	1-244-661-11	330	R706 (R756)	1-242-699-11	12k
R431	1-244-661-11	330	R707 (R757)	1-242-677-11	1.5k
R432	1-244-715-11	56k	R708 (R758)	1-242-687-11	3.9k
R433	1-244-715-11	56k	R709 (R759)	1-242-695-11	8.2k
R434	1-244-702-11	16k	R710 (R760)	1-242-689-11	4.7k
R435	1-244-702-11	16k	R711 (R761)	1-242-661-11	330
R436	1-244-685-11	3.3k	R712 (R762)	1-202-591-11	5.6k ±10% ½W composition
R437	1-244-685-11	3.3k	R713 (R763)	1-242-717-11	68k
R438	1-244-659-11	270	R714 (R764)	1-242-661-11	330
R439	1-244-659-11	270	R715 (R765)	1-242-661-11	330
R440	1-244-677-11	1.5k			

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
○ R321	1-244-673-11	1 k
○ R322	1-244-680-11	2 k
○ R323	1-242-657-11	220
○ R324	1-242-701-11	15 k
○ R325	1-244-691-11	5.6 k
○ R326	1-244-689-11	4.7 k
○ R327	1-244-645-11	68
○ R328	1-244-673-11	1 k
○ R329	1-244-701-11	15 k
○ R330	1-244-680-11	2 k
○ R331	1-244-671-11	820
○ R332	1-244-657-11	220
○ R333	1-244-717-11	68 k
○ R334	1-244-649-11	100
○ R335	1-244-649-11	100

R401	1-244-739-11	470 k
R402	1-244-673-11	1 k
R403	1-244-673-11	1 k
R404	1-244-706-11	24 k
R405	1-202-560-11	300 ±10% ½W composition
R406	1-244-689-11	4.7 k
R407	1-244-709-11	33 k
R408	1-244-670-11	750
R409	1-244-731-11	270 k
R410	1-244-671-11	820
R411	1-244-690-11	5.1 k
R412	1-244-689-11	4.7 k
R413	1-244-673-11	1 k
R414	1-244-682-11	2.4 k
R415	1-244-649-11	100
R416	1-244-661-11	330
R417	1-244-661-11	330
R418	1-244-661-11	330
R419	1-244-661-11	330
R420	1-244-643-11	56
R421	1-244-643-11	56
R422	1-244-679-11	1.8 k
R423	1-244-679-11	1.8 k
R424	1-244-693-11	6.8 k
R425	1-244-693-11	6.8 k
R426	1-244-671-11	820
R427	1-244-671-11	820
R428	1-244-715-11	56 k
R429	1-244-715-11	56 k
R430	1-244-661-11	330
R431	1-244-661-11	330
R432	1-244-715-11	56 k
R433	1-244-715-11	56 k
R434	1-244-702-11	16 k
R435	1-244-702-11	16 k
R436	1-244-685-11	3.3 k
R437	1-244-685-11	3.3 k
R438	1-244-659-11	270
R439	1-244-659-11	270
R440	1-244-677-11	1.5 k

<u>Ref. No.</u>	<u>Part No.</u>	<u>Description</u>
R441	1-244-677-11	1.5 k
R442	1-244-685-11	3.3 k
R443	1-244-685-11	3.3 k
R444	1-244-695-11	8.2 k
R445	1-244-695-11	8.2 k
R446	1-244-709-11	33 k
R447	1-244-667-11	560
R448	1-244-702-11	16 k
R449	1-244-702-11	16 k
R501 (R551)	1-242-713-11	47 k
R502 (R552)	1-242-745-11	1 M
R503 (R553)	1-242-681-11	2.2 k
R504 (R554)	1-242-737-11	470 k
R505 (R555)	1-242-661-11	330
R506	1-242-719-11	82 k
R507 (R557)	1-242-691-11	5.6 k
R508 (R558)	1-242-659-11	270
R509 (R559)	1-242-705-11	22 k
R510 (R560)	1-242-653-11	150
R511 (R561)	1-242-735-11	390 k
R512 (R562)	1-242-701-11	15 k
R513 (R563)	1-242-689-11	4.7 k
R601 (R651)	1-244-712-11	43 k
R602 (R652)	1-244-704-11	20 k
R603 (R653)	1-242-663-11	390
R604 (R654)	1-242-737-11	470 k
R605 (R655)	1-242-735-11	390 k
R606 (R656)	1-242-701-11	15 k
R607 (R657)	1-242-735-11	390 k
R608 (R658)	1-242-691-11	5.6 k
R609 (R659)	1-242-665-11	470
R610 (R660)	1-242-675-11	1.2 k
R611 (R661)	1-242-695-11	8.2 k
R612 (R662)	1-242-709-11	33 k
R613 (R663)	1-242-669-11	680
R614 (R664)	1-242-655-11	180
R615 (R665)	1-242-699-11	12 k
R616 (R666)	1-242-691-11	5.6 k
R617	1-242-719-11	82 k
R701 (R751)	1-242-717-11	68 k
R702 (R752)	1-242-705-11	22 k
R703 (R753)	1-242-677-11	1.5 k
R704 (R754)	1-242-705-11	22 k
R705 (R755)	1-242-677-11	1.5 k
R706 (R756)	1-242-699-11	12 k
R707 (R757)	1-242-677-11	1.5 k
R708 (R758)	1-242-687-11	3.9 k
R709 (R759)	1-242-695-11	8.2 k
R710 (R760)	1-242-689-11	4.7 k
R711 (R761)	1-242-661-11	330
R712 (R762)	1-202-591-11	5.6 k ±10% ½W composition
R713 (R763)	1-242-717-11	68 k
R714 (R764)	1-242-661-11	330
R715 (R765)	1-242-661-11	330

Ref. No.	Part No.	Description
R716 (R766)	1-202-525-11	10 ±10% ½W composition
R717 (R767)	1-202-565-11	470 ±10% ½W composition
R718 (R768)	1-242-665-11	470
R719 (R769)	1-242-725-11	150k
R720 (R770)	1-242-725-11	150k
R721 (R771)	1-242-681-11	2.2k
R722 (R772)	1-242-681-11	2.2k
R723 (R773)	1-242-649-11	100
R724 (R774)	1-242-681-11	2.2k
R725 (R775)	1-242-681-11	2.2k
R726 (R776)	1-242-673-11	1k
R727 (R777)	1-242-649-11	100
R728 (R778)	1-242-649-11	100
R729 (R779)	1-242-673-11	1k
R730 (R780)	1-242-649-11	100
R731 (R781)	1-242-625-11	10
R732 (R782)	1-242-619-11	5.6
R733 (R783)	1-205-802-11	0.47 ±10% 2W wirewound
R734 (R784)	1-205-802-11	0.47 ±10% 2W wirewound
R735 (R785)	1-242-619-11	5.6
R736 (R786)	1-242-625-11	10
R737 (R787)	1-242-649-11	100
R739	1-242-677-11	1.5k
R740 (R790)	1-242-649-11	100
R741 (R791)	1-244-705-11	22k
R801	1-244-665-11	470
R802	1-244-665-11	470
R803	1-206-147-11	100 ±10% 3W metal-oxide
R804	1-244-683-11	2.7k
R805	1-244-649-11	100
R901	1-244-697-11	10k
R902	1-244-697-11	10k
R903	1-244-739-11	560k
R904	1-244-739-11	560k
R905	1-244-719-11	82k
R906	1-244-719-11	82k
R907	1-202-645-11	1M ±10% ½W composition (USA, Canada Model only)
RV201	1-221-966-11	100 k(B), adjustable
RT301	1-222-952-11	15 k(B), adjustable
RV401	1-222-948-11	3.3 k(B), adjustable
RV601 (RV651)	1-222-375-11	250 k(S), variable (VOLUME control)
RV603 (RV653)	1-222-373-11	50 k(S), variable (TONE control)
RV604 (RV654)	1-222-374-11	50 k(B), variable (TONE control)
RV701 (RV751)	1-222-981-11	10 k(B), adjustable
RV702 (RV752)	1-222-981-11	10 k(B), adjustable
Note:	USA Model	Serial No. 810,401 and later
	Canada Model	Serial No. 700,301 and later
	EP Model	Serial No. 553,001 and later
SWITCHES		
S1	1-514-761-00	rotary (FUNCTION (1))
S2	1-514-762-11	lever (FUNCTION (2))

Ref. No.	Part No.	Description
S3	1-513-338-12	lever (MONITOR)
S4	1-514-508-11	rotary (MODE)
S5	1-513-149-00	push (LOUDNESS)
S6	1-514-511-12	lever (HIGH FILTER)
S7	1-513-338-12	lever (MUTING)
S8	1-514-507-00	rotary (SPEAKER)
S9	1-514-911-00	lever (POWER) (EP, UK and AEP Model)
	1-516-416-00	lever (POWER) (Canada and USA Model)
S10	1-514-524-11	slide (DE-EMPHASIS)

FILTERS

			(Color)	(Specified Center Freq.)
CF201	}	1-403-562-11	fm i-f, ceramic	red 10.70MHz
CF202		1-403-562-21	fm i-f, ceramic	black 10.66MHz
CF203		1-403-562-31	fm i-f, ceramic	white 10.74MHz
CF204		1-403-562-41	fm i-f, ceramic	green 10.62MHz
CF205		1-403-562-51	fm i-f, ceramic	yellow 10.78MHz
CF206				
CF301		1-403-153-00	filter, ceramic; 455 kHz	
LPF401		1-231-088-00	filter, low-pass	

Note: USA Model Serial No. up to 810,400
 Canada Model Serial No. up to 700,300
 EP Model Serial No. up to 553,000

MISCELLANEOUS

	1-231-057-12	encapsulated component, 120 Ω + 0.033 μF (USA, Canada and EP Model)
J901, J902	1-507-265-00	jack, AUX, HEADPHONE
	1-507-268-00	phono jack, 8-P
CNJ901	1-509-029-02	REC/PB connector
CNJ902	1-509-341-12	AC outlet (USA and Canada Model)
	1-509-546-00	connector, ac input; 3-p (AEP and UK Model)
	1-517-021-00	socket, meter lamp
PL904	1-518-012-04	lamp, meter 8V 0.15A
	1-518-051-12	lamp, STEREO 4.5V 40mA
	1-518-070-00	lamp, dial 8V 0.3A
M901	1-520-090-00	meter, TUNING
	1-526-165-00	voltage selector (EP Model)
	1-526-520-00	voltage selector (AEP and UK Model)
F901	1-532-237-00	fuse, 3.15A (AEP and UK Model)
	1-532-214-00	fuse, 5A (USA and Canada Model)
	1-532-250-00	fuse, 1.6A (EP Model)
F902	1-532-204-00	fuse, 2A
F903	1-532-074-00	fuse, 200mA
F904	1-532-350-00	fuse, 5A
F905	1-532-350-00	fuse, 5A
	1-533-051-00	socket, dial lamp
	1-533-082-00	holder, fuse; 5-p (AEP and UK Model)
	1-534-587-11	cord, power (EP Model)
	1-534-526-21	cord, power (USA, Canada Model)
	1-536-179-00	terminal strip, 1L1 (C)
	1-536-181-00	terminal strip, 2L1 (C)
	1-536-189-00	terminal strip, 1L1 (B) (USA, Canada Model only)
TM901	1-536-286-00	terminal strip, 4-p (ANTENNA)
TM902, 903	1-536-284-00	terminal strip, 4-p (SPEAKER)