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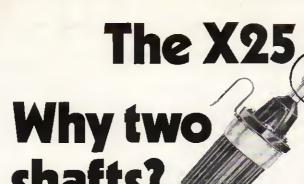
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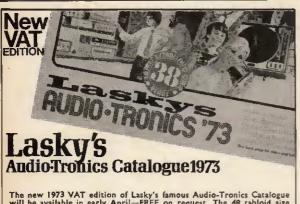
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Our August issue will be published on Friday, July 13, 1973

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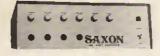
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AC117K 0-22	0-61	BC151 0-22	BD139 0-61 BD140 0-66	BF195 0-18 BF196 0-16	OC22 0-42 OC23 0-46		0-19 2N2 0-19 2N2		2N3391 0-18 2N3391A 0-18	2N4061 0-18
AC122 0-18	ADT140 0-55	BC152 0-19	BD155 0-88	3F197 0-18	OC24 0-62		0.83 2N2		2N3391 0-16	2N4062 0-13 2N4284 0-19
AC125 0-18	AF114 0-27	BC153 0-31	BD175 0-66	HF200 0-50	OC25 0-42		0-18 2N2		2N3393 0-18	2N4285 0-19
AC126 0-19 AC127 0-19	AP115 0-27	BC154 0-83	BD176 0-66	BF222 1.05	OC26 0-28		0-18 2N2	369A 0-16	2N3394 0-16	2N4286 0-19
AC128 0-19	AF116 0-27 AF117 0-27	BC157 0-20 BC158 0-13	BD177 0-72 BD178 0-72	BF257 0-50 BF258 0-66	OC28 0-55 OC29 0-56	20382	0-18 2N2		2N3395 0-19	2N4207 0-19
AC132 0-18	AF118 0-39	BC159 0-13	BD179 0-77	BF259 0-94	OC35 0-46		0-83 2N2 0-33 2N2		2N3402 0-28 2N3403 0-28	2N4288 0-19 2N4289 0-19
AC134 0-18	AF124 0-33	BC160 0-50	BD180 0-77	BF262 0-81	OC36 0-55		0-28 2N2		2N3404 0-81	2N4290 0-18
AC137 0-18	AF125 0-28	BC181 0-55	BD185 0-72	BF263 0-81	OC41 0-22	2N 388 (0.39 2N2	712 0-28	2N3405 0-48	2N4291 0-19
AC141 0-18 AC141K 0-19	AF108 0-81 AF107 0-81	BC167 0-13 BC168 0-13	BD186 0-72	BF270 0-39	OC42 0-27		0.61 2N2		2N3414 0-17	2N4202 0-19
AC142 0-16	AF139 0-33	BC168 0-13 BC169 0-13	BD187 0-77 BD188 0-77	BF271 0-88 BF272 0-88	0C44 0-17 0C45 0-14		0-22 2N2		2N3415 0-17	2N4293 0-19
AC142K 0-19	AF178 0-33	BC170 0-18	BD189 0-88	BF273 0-89	OC70 (-11		0-31 2N2 0-46 2N2	904A 0-28 905 0-23	2N3416 0-81 2N3417 0-31	2N5172 0-18 2N5407 0-85
ACI51 0-17	AF179 0-55	BC171 0-18	BD190 0-88	BF274 0-39	0071 0-11		0.54 2N2	905A 0-28	2N3525 0.88	2N5458 0-35
AC154 0-22	AF180 0-55	BC172 0.16	BD195 0-94	BFW10 0.66	OC72 0-16	2N598	0-46 2N2	906 0-17	2N3646 0-10	2N0409 0-44
AC155 0-22 AC156 0-22	AF181 0-50 AF186 0-50	BC173 0-16 BC174 0-16	BD196 0-94 BD197 0-99	BFX29 0-80 BFX84 0-24	OC74 0-16 OC75 0-17			906A 0-20	2N3702 0-11	28301 0-55
AC157 0-27	AF239 0-41	BC175 0-24	BD198 0-99	BFX85 0-83	OC76 0-17			907 0-22 907A 0-24	2N3703 0-11 2N3704 0-12	28392A 0-46 28393 0-46
AC165 0-22	AL103 0-72	BC177 0-21	BD199 1.05	BFX86 0-24	OC77 0-28		0-27 2N2		2N3705 0-11	28303 0-81
AC166 0-22	AL103 0-72	BC178 0-21	BD200 1.05	BFX87 0-27	OC81 0-17	2N699	0.89 2Ng	924 0-18	2N3706 0-10	28034 0-77
AC167 0-22 AC168 0-27	ASY26 0-28 ASY27 0-23	BC179 0-21 BC180 0-27	BD205 0-88	BFX88 0-24	OC81D 0-17		0-09 2N2		2N3707 0-12	28365 0.93
AC169 0-16	A8Y28 0-28	BC181 0-27	BD206 0-88 BD207 1-05	BPY50 0-22 BPY51 0-22	OC82 * 0-17 OC82D 0-17		0·10 2N2 0·13	926 (G) 0-14	2N3708 0-08 2N3709 0-10	25306 0-98
AC176 0-22	ASY29 0-28	BC182 0-11	BD208 1-05	BFY52 0-22	OC83 0-22			926 (Y)	2N3710 0-10	29307 0-93 29321 0-62
AC177 0-27	VRA 90 0-58	BC182L 0-11	BDY20 1-10	BFY53 0-19	OC84 0-22	2N717	0-89	0.12	2N3711 0-10	29322 0-46
AC178 0-81	A8Y51 0.28	BC183 0-11	BF115 0-27	BPX25 0-94	OC139 0-22	2N718	0-27 2N2	926 (0)	2N3819 0-31	28322A 0-46
AC179 0-81 AC180 0-19	ASY52 0.28 ASY54 0.28	BC183L 0-11 BC184 0-18	BF117 0-50 BF118 0-77	B8X19 0-17 B8X20 0-17	OC140 0-22 OC169 0-28		0.55	0-11	2N3820 0-55	28323 0-62
AC180K 0-22	ASY55 0-28	BC184L 0-18	BF119 0-77	BSY25 0-17	OC170 0-28		0-31 2N2 0-31	926 (R) 0-11	2N3821 0-89 2N3823 0-31	28324 0-77 28326 0-77
AC181 0-19	ASY56 0-28	BC186 0-81	BF121 0-50	B8Y26 0-17	OC171 0-28	25743		926 (R)	2N3903 0-81	28326 0-77
AC181K 0-22	A8Y57 0.28	BC187 0-81	BF123 0.65	B8Y27 0-17	OC200 0-28	2N744	0-22	0-11	2N3904 0-33	28327 0-77
AC187 0-24 AC187K 0-22	ASY58 0-28 ASZ21 0-44	BC207 0-12 BC208 0-12	BF125 0-50	BSY28 0-17	OC201 0-81		0-18 2N3		2N3905 0-31	28701 0-46
AC188 0-24	BC107 0-10	BC209 0-18	BF127 0-65 BF152 0-61	BSY29 0-17 BSY38 0-20	OC202 0-81 OC203 0-28		0-33 2N3 0-23 2N3		2N3900 0.30 2N4058 0.18	40361 0-44 40362 0-50
AC188K 0-22	BC108 0-10	BC212L 0-12	BP153 0-50	B8Y39 0-20	OC204 0-28	2.1930	G-23	053 0.19	2N4058 0-18	40342 0-50
ACY17 0-28	BC109 0-11	BC213L 0-12	BF154 0-50	B8Y40 0-81	OC205 0-39	2N1131				
ACY15 0-22 ACY19 0-22	BC113 0-11 BC114 0-17	BC214L 0-16	BF155 0-77	B8Y41 0-81	OC309 0-44		0-22 0-24	DIOR	ES AND RECTIF	TERS
ACY20 0.22	BC115 0-17	BC225 0-28 BC226 0-39	BF156 0-53 BF157 0-61	B8Y95 0-14 B8Y95A 0-14	OCP71 0-48 ORP12 0-48		0-16	19 0-09	BY133 0-23	OA10 0-89
ACY21 0.22	BC116 0-17	BCY30 0.27	BF158 0-61	Bu105 2-20	ORP60 0-44		0-16 AAT 0-19 AAT		BY164 0-55	OA47 0-08
ACY22 0-18	BC117 0-17	BCY31 0-29	BF159 0-68	C111E 0-55	ORP61 0-44		0-19 AAI	29 0-09	BYX38/30	OA70 0-08
ACY27 0-20 ACY28 0-21	BC118 0-11 BC119 0-38	BCY32 0-88	BF160 0-44	C400 0-88	P346A 0-22		0-28 AA1	30 0-10	0.40	OA79 0-08
ACY29 0-89	BC120 0-88	BCY33 0-24 BCY34 0-28	BF162 0-44 BF163 0-44	C407 B-28 C424 0-22	P397 0-48 ST140 0-14		0-28 AA2		BYZ10 0.39 BYZ11 0.38	0.481 0-08
ACY30 0-81	BC125 0-18	BCY70 0-18	BF164 0-44	C425 0-55	8T140 0-14 8T141 0-19		0-26 BAI 0-26 BAI		BYZ11 0-88 BYZ12 0-88	OA85 0-10 OA90 0-07
ACY31 0-31	BC126 0-20	BCY71 0-20	BF165 0-44	C426 0-89	TIS43 0-88		0-22 BA1		BYZ13 0-28	OA91 0-07
ACY34 0-28	BC132 0-13	BCY72 0-16	BF167 0-24	C428 0-22	UT46 0-80	2N1711 (0-22 BAI	48 0-18	BYZ16 0-44	OA95 0-08
ACY35 0-28 ACY35 0-81	BC134 0-20 BC135 0-13	BCZ10 0-22 BCZ11 0-28	BF173 0-24 BF176 0-39	C441 0-88	2G301 0-21		0-35 BA1		BYZ17 0-89	OA200 0-07
ACY40 0-19	BC136 0-17	BCZ12 0-28	BF176 0-39 BF177 0-39	C442 0-88 C444 0-89	2G302 0-21 2G303 0-21		0-50 BA1 0-41 BA1		BYZ18 0-89 BYZ19 0-81	OA202 0-08 SD10 0-06
ACY41 0-20	BC137 0-17	BD121 0-66	BF178 0-88	C450 0-24	2G304 0-27	2N2147 (0-79 BY		CG62	8D19 0-06
ACY44 0-38	BC139 0-44	BD123 0-78	BF179 0-33	MAT100 0-21	20306 0-44	2N2148	0-63 BY1	01 0.18	(OA91 Eq.)	1N34 0-08
AD130 0-42 AD140 0-88	BC140 0-38 BC141 0-38	BD124 0-66	BF180 0-82	MAT101 0-22	2G308 0.89	2N2180	0-66 BY		0.06	1N34A 0-08
AD142 0.58	BC142 0-88	BD131 0-55 BD132 0-68	BF181 0-88 BF182 0-44	MAT120 0-21 MAT121 0-22	2G309 0-89 2G339 0-22		0-89 BY1 0-89 BY1		CG651— (OA70-OA79)	1N914 0-07 1N916 0-07
AD143 0-42	BC143 0-38	BD133 0-72	BF183 G-44	MPF102 0-48	2G339A 0-18		0-29 BY		0-07	1N414B 0-07
AD149 0-55	BC145 0-50	BD135 0-44	BF184 0-28	MPF104 0-41	2G344 0-20	2N2217 (0-24 BY1	28 0-17	OA5 0-89	18021 0-11
AD161 0-87	BC147 0-11	BD136 0-44	BF185 0.83	MPF105 0-41	2G345 0-18	2N2218	0-22 BY1	30 0.18	OA58L 0.28	18951 0.07

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C 3	50	Precision Resistors 1%, and 01% mixed values		0-55
C 4	75			
C 5	5		* *	0-55
C 6	2		- 4	0-56
		Tuning Gange, MW/LW VHF	476	0-55
Q 7	1	Pack Wire 50 meters assorted colours		0-65
C 8	10	Reed Switches		8-55
C 3	3	Micro Bwitches		0-55
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CH	5	Jack Sockets 3 x 3.5m 2 x Standard Switch Types		0-55
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C15	4		* *	
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C17	10			0-55
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50	0.04	0-06	0.06	0-08	0-16	0.23	0.56
	0-04	0.07	0-06	0-15	0-18	0.26	0-98
200	0-06	0-10	0.07	0.18	0-22	0.27	1.10
400	0-07	0-15	0-08	0-22	0-30	0.41	1.38
	0-08	0-18	0.11	0-28	0.88	0.50	2-05
	0-11	0.19	0.12	0.28	0-41	0.81	2.20
1000	0.12	0.30	0.16	0-33	0.51	0.70	2.75
1200	-	0.37	_	0.42	0.63	0.83	_

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UIC05=12×7406	0.55	UIC53=12×7453	0-55	UIC94=5×7494	0-55
UIC06-8×7406	0.65	UIC54=12 x 7454	0.55	U1C95=5×7495	0-55
UIC07=8×7407	0-55	UIC60=12 x 7460	0-55	U1C96=5×7496	0-55
DIC10=12×7410	0.55	UIC70=8×7470	0-55	UIC100=5×74100	0-55
IIIC13=8×7418	8-55	UIC72=8 x 7472	0.55	UIC121=5 x 74121	0-55
UIC20=13×7420	0.55	UIC73=8×7473	0.55	UIC141=5×74141	0-55
UIC30 = 12 x 7430	0-55	UIC74=8×7474	0-55	UIC151=5×74151	0-55
UIC40=12 x 7440	0.55	UIC75-8×7475	0-65	UIC154=5×74154	0.65
UIC41=5×7441	0.55	UIC76=8×7476	0.55	UIC193 = 5 x 74193	0-55
UIC42=5 x 7442	0.55	UIC80=5×7480	0.55	UIC199=5×74199	0-55
UIC48=5 x 7443	0.55	UIC81=5×7481	0.55		
UIC44=5×7444	0.55	UIC82=5×7482	0.55	UICXI == 25 Assorte	:d
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BN7405	0-17	0-16	0-18	SN7493	0-74	0-71	0-64
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SN7408	0.20	0.14	0-18	SN7496	0-86	0-98	0.86
SN7409	0.20	0.19	0.18	BN74100	1.82	1.76	1.71
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MODEL	CD66	GR116	3015F Minitron	
Anode Voltage (Vde)	170min	175min	5	All
Cathods Current (mA)	2.3	14	8	0.9 + Decimal
Numerical Height (mm)	16	18	9	point, All
Tube Height (mm)	47	32	22	viewing.
Tube Diameter (mm)	19	.18.	12 wide	for all type
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SPECIFICATION

Treble Control Filters: Rumble (High Pass) Scratch (Low Pass) Signal/Noise Ratio Input overload Supply Dimensions

100Hz
8kHz
better than -65dB
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+35 volts at 20mA
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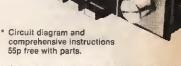


The Tourist PB is suitable for 12 volt working on both negative and positive earth vehicles. It covers the full medium and long wave bands. It is permeability tuned and sturdily constructed. Output is a full 2-5 watts into an 8 ohms speaker. But the Tourist PB will operate into any loud-speaker from 8 to 15 ohms.

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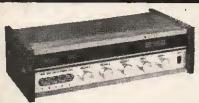
washers, screws and wire.

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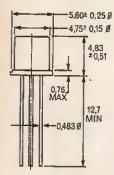
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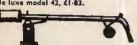
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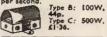


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l≩in × 5ín	32p	35p
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			0.1	0.12
ize			Matrix	Matrix
Hin	×	3#in	251p	1810
Hin	×	5in	28 p	28 p
lin.	×	34 in	28 p	28 p
l≩in.	×	5in 3≩in 5in	320	35p
7in	×	2+in	87p	66p
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BuF	450V	18 0	2500µF	25V	50p
16µF	450V	200	2500µF	50V	66p
25µF	257	710	3000µF	25V	53p
25µF	50V	710	5000µF	25V	660
32µF	450V	30p	5000µF	50Y	41-21
50µF	50V	Hp	8-8µF	450V	20p
00µF	50Y	12p	8-16µF	450V	22p
50uF	25V	151p	16-16UF	450V	30p
50µF	50V	19p	16-32µF	450V	69p
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00µF	50V	27 jp	50-50µF	350V	42p
00µF	25V	30p			

PLUGS	
Car serial	15
Co-axial	- 11
D.I.N. 2 pin (speaker)	
D.I.N. 3 pin	i.
D.I.N. 4 pin	13
D.I.N. 5 pin, 180°	14
D.I.N. 5 pin, 240°	
D.E.14. 5 pin, 240	16
D.I.N. 6 pin	16
Jack, 2 mm unscreened	- 10
Jack, 2-mm screened	- 1
Jack, 3 mm unscreened	9
Jack, 3-mm screened	- 13
Jack, gin unscreened	- li
Jack, in screened	- 2
Jack, stereo, unscreened	2
Jack, stereo, screened	2:
Speed section, selection	-

LINE SOCKETS	
Car aerial Co-axial	15ip
D.I.N. 2 pin (speaker) D.I.N. 3 pin	161P
D.I.N. 5 pin, 180° D.I.N. 5 pin, 240°	17 8
Jack, 34mm Jack, 4in screened	161p
Phono, plated metal	37p

Practical Electronics



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SOCKETS	
Car zerial	9p
Co-axial, surface	9p
Co-axial, flush	10p
D.I.N. 2 pin (speaker)	Hp
D.I.N. 3 pin	IOp
D.I.N. 5 pin, 180° D.I.N. 5 pin, 240°	10p
Jack, 24mm	10p
Jack, 34mm	iie
Jack, ±in unswitched	1610
Jack, Jin switched	iata
Jack, stereo, switched	260
Phono, single	5ip
Phono, 2 on a strip	7 p
Phono, 3 on a strip	100
Phono, 4 on a strip	IIp
Wander, single, red or bla	ck 5 p

MINIATURE **ELECTROLYTICS**

SOCKETS Car aerial Co-axial, surface Co-axial, flush D.i.N. 3 pin (speaker) D.i.N. 5 pin, 180° Do.i.N. 5 pin, 180° Do.i.N. 5 pin, 180° Jack, 3½mm Jack, 3½mm Jack, 3½mm Jack, in switched Jack, treeo, switched Phono, 2 on a strip Typ Typ Typ Typ Typ Typ Typ T	1-0µF 63V 1-5µF 63V 2-2µF 63V 3-3µF 63V 4-7µF 63V 4-7µF 63V 10µF 25V 10µF 63V 15µF 40V 15µF 63V 22µF 63V 22µF 63V 33µF 16V 47µF 10V 47µF 10V 47µF 40V 47µF 63V 68µF 63V 68µF 63V 68µF 63V 100µF 4V	790777977797779777977797779777977797779	ISOUF 25V 150UF 40V 150UF 64V 150UF 64V 220UF 64V 220UF 16V 220UF 25V 470UF 16V 330UF 16V 470UF 16*3V 470U	15p 22p 26p
Jack, Jin switched 1879 Jack, stereo, switched 269 Phono, single 519	68µF 16V 68µF 63V 100µF 4V	7p 13p 7p 7p 7p 8p	1000µF 25V 1500µF 6-3V 1500µF 10V 1500µF 16V 2200µF 6-3V 2200µF 10V	26p 15p 22p 26p 23p 25p
Wander, twin strip 7 p	150µF 6-3V 150µF 16V	7p 7p	3300μF 6-3V 4700μF 4V	26p 26p

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50-0-50μA 100μA 100-0-100μA

ō0μΑ..... δ0-0-ō0μΑ ... 100μΑ.... 100-0-100μΑ. 200μΑ....

500µA.....

5m A 10mA 50mA 100mA

200μΑ... 500μA..... 1mA 5mA

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20 amp: a.c.*	£3-10	2
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£8-95		150V d.c	#2-25
\$8-10		300V d.c	£2-25
#8-10		500V d.c	£2-25
28-10		750V d.c	42-25
£3-10		15V a.c	\$2-85
#3-10		50V d.c	22-80
	10mA #2-25	150V a.c	£2-80
	20mA \$2.25	300V a.c	22-80
	ö∂mA £2-25	500V a.o	#2.80
	100mA \$2.25	S Meter lmA.	\$8-80
42-50	150mA £2-25	VU Meter	12-65
\$2.50			
\$2.50			

TAbe with ore.	Zin. square Fronts.				
50µA	68-50	10V d.c	42		
50-0-50μA	£8-05	20V d.c	22		
100µA	#8.00	50∀ d.c	22		
100-0-100µA .	#2-B5	300V d.c	48		
500µA	28-65	15V a.c	\$2		
lmA	\$2.60	300V a.c	22		
5mA	£8-60	S Meter ImA.	#2		
10mA	\$2.50	VU Meter	28		
50mA	£2-50		42		
100mA	£2-50	5 amp. a.c	42		
500mA		10 amp. a.c.*	42		
l amp	£2-50	20 amp, a.c.*	42		
ő amp		30 amp. a.e.*	28		
	_				

Type MB.65P.			
50µA	23.70	10V d.c	\$2.60
50-0-50µA	£8-15	20V d.c	22-60
100µA	28-15	50V d.c	\$2-60
100-0-100MA .	£8-10	150V d.c	42-60
200µA	£8.05	300V d.c	\$2.60
500µA	\$2.75	15V a.c	£8-80
500-0500μA .	22-60	50V s.c	\$2.80
ImA	£2-60	150V a.c	\$2.80
δmA	22-80	300V a.c	£2-80
10mA	£2-60	500V a.c	£2-80
50mA	£2.60	S Meter ImA.	\$2.85
100mA	42-60	VU Meter	£3.70
500mA	£2-60	50mA a.c	\$2.80
1 amp	£2-60	100mA a.c.* .	42-80
δ amp	£2-60	200mA a.c.* .	42.60
10 amp.,	€2-60	500mA a.c.*	£2-60
13 amp	#2-60	1 amp. a.c.* .	42.60
20 впр	22-60	5 amp. a.c.	£8-80
30 amp. ,	£2.80	10 amp. s.c.*	£2-60
50 amp	£2-90	20 amp. s.c.*	£2-80
5V d.c	£2.60	30 amp. a.c.*	\$2.60
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0 50-0-50µA					
0 50-50µA	0	Type MR.45P. 2in. sq	u	are Fronts.	
0 50-50µA	0	50HA 42-70) ,	5 amp	â
0 1000-1001.4 22-60 20V d.c. 0 2001.4 22-60 50V d.c. 0 2001.4 22-40 20 amp. a.c. 0 201.4 201.4 22-40 20 amp. a.c. 0 201.4 22-40 20 amp. a.	0	50-0-50//A 49-65	i	10V d.c	ō
0 500-0-500µA	0	100rtA 22-80		20V d.c	8
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0 500-0-500µA	O			200V d c	ä
0 500-0-500µA	0			15V d.c	7
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1 amp \$28.40 30 amp. s.c. \$40 30 amp. s.c.		500m A 49.40			ä
"SEW" BAKELITE PANEL METERS Type MR. 65. 3 in. square Fronts. 1 amp	. 1			20 0000 000	•
"SEW" BAKELITE PANEL METERS Type ME.65. 8iin. square Fronts. 1 amp. 15					4
"SEW" BAKELITE PANEL METERS Type MR.95. 8in. square Fronts. 1 amp		1 amh	, 1	ov amp. s.c.	£
PANEL METERS Type ME.65. 8in. square Fronts. lamp.	Û	1 amp 25-40	-	ov amp. a.c.	£
Type MR.65. 8‡in. square Fronts. 1 amp	9		_		
Type MR.65. 3iln. square Fronts.	000	"SEW"	E	BAKELITE	
0 Type RR.05. \$410. square Front. 1 amp	0000	"SEW"	E	BAKELITE	
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50μΑ 50-0-50μΑ 100μΑ 100-0-100μΑ

500-0-500µA

1mA 1-0-1mA 5mA 10mA

5mA

50m A

100mA

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	20V d.c	\$2.60
_	50V d.c	28-60
	150V d.c	\$2.60
24-80	300 V d.c	#2-60
28-55	30∀ a.c."	42-85
48-05	50V s.c.*	£2.65
£3.00	150V a.c.*	#2.65
£8.00	300V a.c.*	42.65
42-70	500mA a.c.* .	\$2.60
\$2-70	l amp. a.c	22-60
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\$2.60	VU Meter	28-05
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Type 8D.460 46mm × 59-5mm Fronts

" SEW " EDGEWICE METEDS

EDGET	1131	. MITTER	.3
Type PE.70.	23in. d		
50μΑ 50-0-50μΑ 100μΑ 100-0-100μΑ 200μΑ	£8-60	300μA 1mA 300V a.e VU Meter	48-20

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العاما	Type ED.107. Size overall 100mm × 90mm				
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METERS

		ewon's stocesounts on	
	demonstrate internal	working. Available	П
	in the following range	4:	ı
		10V d.c #4-85	
		20V d.c £4.84	
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	50-0-50µA 25-10	300V d.c £4.85	1
	19-0-1mA \$4-85	Dual range	ı
	1A d.c \$4.85	500mA/5A d.c. 25-10	4
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MW1-6 60mm square MW1-8 80mm square Post extra

240° Wide Angle ImA Meters

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1004A #8-40	
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500µA \$8-05	a
1mA 28-00	-
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Brand new. 3 sets of changeover contacts at 5 amp. rating. 50p each. Poet 10p (100 lote \$40) Quantities available.

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TESO POCKET

High-precision at low-cost, Hanges: D.c. 15V, 150V, 1,000V (10,000 O.P.V.). A.c. 15V, 150V, 100V (1,000 1,000 (10,000 O.P.V.). A.c. 15V. 150V. 100V (1,000 O.P.V.). D.c. Current 150mA. Resistance 100k ohms.



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Top value 1,000 o.p.v., pocket multimeter. Ranges: 0/10/50/250/1,000V. A.C. and D.C. D.C. Current 0-1mA/ Resistance: 0/150 K obros.

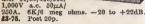
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O.P.V. pocket multimeter. 5/25/ 50/250/500/2,500V d.c. 10/50/100/500/ 1,000V a.c. 50



MODEL TH-12
20,000 O.P.V. Overload protection. Slide switch selector.
0/0-25/2-5/10/50/250/1,000V
d.c. 0/10/50/250/1,000V
d.c. 0/10/50/250/1,000V
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A first class versatile ininstrument manufactured in
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20kg/V d.o. 8kg/V
a.c. Mirror scale
06/3/12/30/120/000V
d.o. 3/30/120/000V
d.o. 3/30/120/000V
d.o. 10/100K/
10 Meg D.
-20 to +46d B. \$5.97. Post 12p.



MODEL 500 30,000 O.P.V. with overload protection mirror scale 0/05/25/20/25/ 100/256/800/1,000V cd.c. 0/2-5/10/25/100/25/800/1,000V a.c. 0/50μλ/5/50/ 600mA. 12 amp. d.c. 0/60/k/6 Meg./60 Meg.Ω. 28-67. Poot paid.



HIORI HODEL 750X 56,000 O.P.V. 43 ranges 0-0-3 to 1,200V d.c. 0-3 to 1,200V a.c. 0-30µA/300mA. 0-3K/30 meg ohms. —10 to +17dB. 28-97. Post 20p.



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U4912 MULTIMETER

Extremely sturdy instrument
for general electrical use,
687 O.P.V.
0/0-911-5/7-5/93/69/150/390/
600/9007 d. e. and 75mV
0/0-911-5/7-5/93/69/150/390/
0/500µA/1-5/6-5/15/65/150
600/9007 h. p. d.c.
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10 amp. d.c. of 227 20 meg. -20 to +50dB. 213-95. Post 35p. 370 WTR MULTIMETER Features c.c. current ranges, 20,000 O.P.V. 0/0-5/2-5/10 / 50 / 250 / 500 / 1,000 V d.c.

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a.c. 0/10/250µA/2-5/25/250MA

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High sensitivity tester.
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Ranges: 0/0-06/0-3/3/30/
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Tests ICO and B.

Pnp/npn. Operates from

9V battery. Complete

with all instructions, etc.

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Tests PNP or NPN transistors, Tests PNP or NPN transistors, Audio indication. Operates on two 1.5V batteries. Complete with all instructions, etc.



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Checks true a.c. beta injout. Checks Ichoc. Checks Ichoc. Checks BCR, etc. Beta HI 10-500 LO 2-50 Ibco 0:5000µA, 220/240V a.c. operation. £17-50, Post 25p.



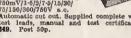
MODEL U4811 SUB-STANDARD MULTI-RANGE VOLT AMMETER

RULTI-RANGE VOLT AMMETER

Sensitivity 330 ohms/

Volt a.e. and d.e. Accuracy 0-5% d.c. 1% a.c.
Scale length 165mm,
0/300/750µA/1-6/3/7-5/
15/30/7-6/10/3800/
750mA/1-6/3/7-5 and,
300/750mA/1-6/3/7-5/
300/750mA/1-6/3/7-5/
300/750mA/1-6/3/7-5/
300/750mA/1-6/3/7-5/
300/750mV/1-6/3/7-5/
75/150/300/750V d.c. 0/
750mV/1-6/3/7-5/15/30/
75/150/300/750V a.c.
Automatic out out. Supplied complete with
test leads, manual and test certificates,

\$49. Post 50p.



TE-65 VALVE VOLTMETER

TE-65 VALVE YG 28 ranges. D.c. voits 1:5-1,500V a.c. voits 1:5-1,500V. Resistance up to 1,000 megohms. 200/240V r a.c. opera-tiom. Complete with probe and instructions. 21,7-50. Post 90p. Additional probes avail-able: R.F. \$2-12; H.V. \$2-50.



RAMODEN HM720B F.E.T. V.O.M. Laput impedance 10MΩ. Ranges: 00-26/1/2-6/10/50/ 250/1,000V d.c. 0/25/16/50/ 250/1,000V d.c. 0/25/16/50/ 25/250MA d.c. -20 to +624B. 0/56/56/50/50M 5MΩ/500MΩ. \$14-86.



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10 meg. input 10 ranges: 01/-003/1/.3/1/.3/1/.3/10/.30/.109/.309V. R. M. 8. 4cps.-1/2 Mc/s. Decibels -40 to +50dB Supplied brand new complete with leads and instructions. Operation 230V. a.c. \$17-50. Carr. 25p.



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VOLTMETER
Battery-operated, 11
meg input, 26 ranges,
Large 4fin mirror scale,
Size 5fin x 4fin x2fin,
D.e. volts 0.5-1,200V,
A.e. volts 3-900V R.M.S.
8-0-800V P-P. D.c.
Current 0.12-12MA. Resistance up to 2,000M ohn. Decibels —20 to
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417-50, P. & F. 20p.



Input impedance 10 meg ohma, 0/0-3/1-2/6/30/120/800V d.c. 0/3/12/6/00/120/800V a.c. 0/120µA/120mA d.c. 0/1K/100K/10 meg/100 meg ohma. \$15-97. Post 25p.



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SISTARGE TESTER
R ang e 0-1,000
Megohms, 500 Volt.
Battery operated.
Wide range clear
meter 4½° × 4°.
Complete with deluxe carrying case,
batteries, instructio
\$19.95. Post 30p.



BELOO AF-SA SOLID STATE SINE SQUARE WAVE C.R. OSCHLATOR Single 18-200,000 Hz: Square 18-50,000 Hz. Output max. +10dB (10 K obma) Operation Internal batteries. Attractive 2-tone case 7½ in x 5 in x 2 in. Frice \$17-50. Carr.



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For display of pulsed and periodic waveforms in electronic elecutatives. AMF, Bandwidth 100LLs. V. RMS/mm. 0-1-25; V. RMS/mm. 0-1-25; V. RMS/mm. 0-1-25; Pre-set triggered sweep 1-3,000µsec.; free running 20-200,000Hz in nine ranges. Calibrator pips. 220mm × 350mm × 430mm, 115-230V a.c. operation. 438. Carr. paid.



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8in tube, Y amp, Sensitivity
0-1V p-p/CM. Bandwidth
1-5cps-1-26MHz. Input
Imp, 2 meg 2 25pf X amp,
sensitivity 0-9V. p-p/CM.
Bandwidth 1-5cps-806kHz.
Input imp, 2 meg 1 20pf
Time base, 5 ranges 10cps300kHz. Synchronisation.
Internal/seternal. Illuminated scale 140mm
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OSCILLOSCOPE

SMHz Pass Band Separato Y1 and Y2 amplifiers. Rectangular fin x 4in C.R.T. Calibrated triggered sweep from 02µ/sec to 100 milli-see per cm. Pree running time base 50 Hz-1MHz. Built-nitime base calibrator and amplitude calibrator. Supplied complete with all saccessories and instruction manu 287. Carr. paid.



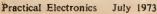
Instruction manual.

MODEL AT201

MODEL ATEODOGAN DECADE ATTENUATOR Prequency range 0-200kHz. Attenuator 0-111dB, 0-1dB step. 100 obms. Max. Input power SodBm. Size 180mm x 90mm x 55mm. 212-50, Post 37p.

ARF-800 AF/RF SIGNAL GENERATOR All transistorised, com-pact, fully portable. AF aine wave 18 Hz to 220 KHz. AF equare wave 18 Hz to 100 KHz. Output sine/square 10v. P-P. RF 100 KHz to 200 MHz. Output tv. maximum. Operation 220/240v. AC. Complete with instructions and leads. \$29-95, Post 50p.





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4 Bands covering \$50kHz-30MHz. BFO Bulit-in Speaker 220/240V s.c. Brand new with Instructions.

PRICE £15.75 CAPT.

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4 Bands covering 550kHz.30MHz. PET. 8 Meter. Variable BFO for 88B. Built-in Speaker. Band-apread. Sensitivity Control. 220/ 240V ac. or 12V dc. 122 × 44/ x 70.Brand new with instructions.

OUR £25:00 Carr.

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Solid state, 5 bands covering 200-420kHz and 0-55 to 30MHz. Islice rule dial. Bandspread. Aerial tuning, BFO, AVC, ANL, 'S' meter. AM/OW/88 B. Integrated speaker and phone socket, 220/2409 a.c. or 12V d.c. Size 235 x 266 x 150mm. Complete with instructions and circuit.

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General coverage 150-400kHz, 550kHz-30MHz. FET front end, 500kHr-30MH2. FET front end, 2 mech. filters, product detector, variable BFO, noise limiter, 8 Meter Bandspread. RF Gain. 15in x 94in x 84in. 18 ib. 220/ 240V s.c. or 12V d.c. Brand new with instructions.

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26501	0-87 0-62	BF861	0-28 0-28	NKT251	0-28 0-24 0-25	OC170 OC171	0.25	7491AN	1.00
28703 AA129	0-62	BFS98 BFX12	0-28 0-20	NKT224 NKT224 NKT251 NKT271 NKT272	0-25 0-25	OC200	0-40	7492 7493	0-75 0-75
AAZ12	0-80	BFX13	0.25	NKT273 NKT274	0-15	OC201 OC202	0.80	7494	0.80
AAZ13 ACI07	0-12 0-87	BFX29 BFX30	0.25	NKT274 NKT275	0-20	OC202 OC203 OC204	0.40	7495 7496	0.80 1.00
AC1196	0.20	BFX 35	0-50	NKT275 NKT277 NKT278	0-20 0-25	OC206	0-40 0-75 0-90	7497	6-25
AC127 AC128	0-25 0-20	BFX63	0-50	NKT278	0-25	OC206 OC206 OC207	0-90	74100 74107	2-50 0-56
	0.25	BFX84 BFX85 BFX86	0-25 0-30	NKT301 NKT304 NKT403	0-40	OC460	0.80	74110	0-80 0-80
ACV17	0.25 0.30	BFX86	0.25	NEWTAGA	0.75	OC460 OC470 OCF71	0.97	74111 74118	1.45
AC166 ACY17 ACY18 ACY19 ACY20 ACY21	0-25	BFX87 BFX88	0.25 0.20 1.00	NKT678 NKT713	0.80	ORPIZ	0-50	74119	1.90
ACY19	0.25	BFY10 BFY11	1-25	NKT713	0-25	ORP60 ORP61	0-40	74121 74122	O-MA
ACY21	0.20	BFY11 BFY17	0.25	NKT773 NKT777	0.88	S19T	0.80	74122 74123	1.85 2.70 1.00
	0.10	BYF18 BFY19	0-25	078B	0.88	SAC40 SFT308	0.25	74141	1.50
ACY27 ACY28	0.17	DETTO	O. A.E.	OAS	0-18 0-10 0-10 0-10	SFT308 ST722	0.88	74150	2-35
ACY39 ACY40 ACY41	0.50	BFY44 BFY50 BFY51	1.00 0.22 0.20	OA47 OA70	0.10	8T7231 8X68	0.88 0.20 0.30	74151 74154	1.10 2.00 1.55
ACY41	0.10	BFY51	0.20	OA71 OA73	0-10	SX631	0.30	74155	1.55
AD140	0-25 0-50	BFY52 BFY53	0.22	GA74	6-10 0-10	8X635 8X640	0-40	74156 74157	1.55 1.80 4.10
AD149	0-50	BFY64 BFY90	0.42	OA79 OA81	0.10	8X641 8X642	0-50	74170 74174	4-10 2-00
AD161 AD162	0.37	BSX27	0.50	OA85	0-12	BX644	0-60 0-75 0-75	74175	1.85
AF106 AF114	0-80 0-25	BSX60 BSX76	0-98 0-15	OA86 OA90	0-15	8X646 V15/30P	0-75	74176 74190	1-80 1-85
AF115	0.25	BSY26	0.18	OA91	0-07	V30/201P	0.75	74191	1-95
AF116 AF117	0-25	BSY27 BSY61	0-17	OA95 OA200	0-07	V60/201 V60/201P	0.50 0.75	74192 74193	2-00 2-00
AF117 AF118	0.60	BSY95A BSY95	0.12	OA202	0.10	XA101	0.10	74194	2-50
AF119 AF124	0-20 0-25	BSY95 BT102/50	0R	OA210 OA211	0.25	XA102 XA151	0-18 0-15	74195 74196	1.85
AF125	0.20	BTY42	0-75 0-92	OAZ200	0.55	XA152	0.15	74197	1.50
AF126 AF127	0-17 0-17	BTY79/1	00 R	OAZ201 OAZ202	0.50 0.48	XA161	0.25	74198 74199	4-80 4-80
AF139 AF178	0-80	BTY79/40	0.75	OAZ203 OAZ204	0-42 0-80	XA162 XB101	0.25		
A Tr170	0.85		1.25	OAZ205	0.48	XB102	0.10	Plug in se	
AF180 AF181 AF186	0.52 0.42 0.40	BY100 BY126	0.15 0.15	OAZ206 OAZ207	0.42 0.47 0.32	XB103	0.25	-low pro	[L
AF186	0.40	BY127	0-17	OAZ208	0.32	XB113 XB121	0-12		Q-15
AFY19 AFZ11	1.18 0.60	BY182 BY213	0-88 0-25	OAZ209 OAZ210	0-82	XB121 ZR24	0.68	16 pin D	0-17
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ANOTHER SENSE

This month we break new ground in applied electronics. Most of the characteristic human senses can be simulated in some fashion by electronic means, but smell and taste are much more difficult for non-biological systems to imitate. Only just recently has it become possible to simulate in some degree the sense of smell without complex and costly chemical devices. This is due to a newly developed semiconductor transducer which responds to the presence of molecules carried in the air.

This particular kind of transducer is a valuable and timely addition to the field of instrumentation and control. Its appearance is especially relevant in these days of public consciousness about the environment and concern for a curb on all objectionable and harmful pollution of the atmosphere. Electronics is now equipped to wage an increasingly effective part in the battle of the environment.

Considering the actual trandsucer itself: now that it has been realised, the gas and vapour detector seems quite simple and rudimentary in its practical form. This apparent simplicity (a common enough outward characteristic of many solid state devices) should not give cause to underestimate the technical achievement it represents. Although the powers of discrimination between different vapours and odours are limited, this transducer is obviously but the first in a series of devices of ever increasing sensitivity and selectivity that will appear in due course. In a wider context the emergence of such a device must be recognised as but one reward from a continuing process of exploration into semiconductor materials which has followed in the wake of the original transistor breakthrough of 25 years ago. This research, conducted on a world wide basis, frequently gives birth to new devices and these largely constitute the power and strength behind electronics, particularly when they enable this technology to penetrate and influence other fields.

With the creation of a kind of electronic nose, interest is likely to be directed towards a simulation of the human taste buds. This, we are given to understand, is a rather more formidable problem. But problems involving electronics usually do find their solution, as suggested in the previous paragraph.

Finally, a slight digression in order to discuss a (not too serious) question of terminology. There are some who might question the propriety of applying biological descriptions to inanimate devices or equipments. True, we have the outstanding example of the poor old computer which has long suffered from the over-generous honour accorded it by the popular title "electronic brain". We could be strictly objective and refer to this month's feature project as simply a gas and vapour detector. However, "electronic nose" seems a rather apt and certainly more colourful description for this unusual innovation in the home constructor field. Its applications are serious enough, however, as will soon be appreciated.—F.E.B.

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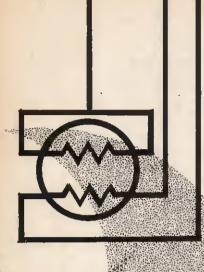
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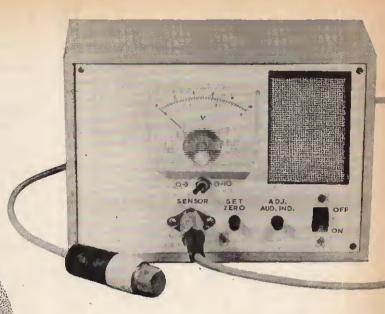
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BY DEPOSIT

This battery operated instrument is designed around a new type of semiconductor sensor which responds to certain gases, vapours, and smoke particles. The sensor cannot distinguish between different smells but it does allow an assessment to be made of the relative concentration of a contaminant in air. Applications include breathalyser, fuel vapour and firedamp detector, air pollution meter, paint dryness indicator, and gas leak locator.

The electronic nose has a meter readout scaled in arbitrary units of 0-3 and 0-10, as well as an audible indicator giving a note of rising pitch proportional to gas or smoke intake. As a rough indication of sensitivity, a measurable response is obtained from 1 cubic millimetre of alcohol evaporated in a volume of air the size of a tea-chest, this is approximately 10,000 times less than the concentration needed to form an explosive mixture.

SEMICONDUCTOR GAS SENSOR (SGS)

The gas sensor consists of two heating filaments embedded in an *n*-type semiconductor chip, see Fig. 1. Only one filament is used for heating the chip, the other merely serving as an ohmic contact. Gas or smoke impurities reduce the intrinsic resistance of the chip, as measured between the two heating filaments, and chip heat determines the rate at which an impurity is absorbed and deabsorbed.

The Japanese manufacturers of the gas sensing device used in this article are not very informative on the working principle involved, but an educated guess is that the semiconductor chip undergoes temporary doping in the presence of air impurities, and this lowers intrinsic resistance in the same way as it does with other types of semiconductor. Heat is required to convect the impurity through the porous structure of the chip, and for deabsorption when the source of contamination is removed.

Heater requirements are around 1.2V maximum at 750 milliamps, and the device runs warm to the touch.

The chip is non-polarised and will conduct in either direction. Chip resistance is in the range 10 to 50 kilohms in clean air, falling to less than 1 kilohm in the presence of moderate concentrations of gases such as hydrogen, carbon monoxide, methane, propane, coal gas, and the vapours from alcohol, petrol, acetone, etc. A whiff of tobacco smoke is also sufficient to cause a similar resistance change.

BASIC CIRCUIT

Fig. 2 shows the basic circuit of the electronic nose, where the voltage across the SGS chip is compared with a fixed voltage at the junction of R1 and R2. In the absence of an air contaminant, VR1 is adjusted so that both voltages are equal and the meter reads zero.

When gas or smoke is detected, the chip resistance decreases and "pulls down" the voltage at the junction of H2 and VR1 slider, thus causing the voltmeter to display a reading. Tests indicate that the resulting meter reading is linearly related to contaminant concentration.

With its balanced configuration, the Fig. 2 circuit is insensitive to small changes in the voltage of battery B2; when the reference voltage drops, so does the voltage across the chip, by a corresponding amount.

RESPONSE TIME

The response of the Fig. 2 circuit, to the sudden application and removal of a "smell", is shown by the

graph of Fig. 3. Absorption time is about two seconds and complete deabsorption takes about 80 seconds. However, when the instrument is employed merely for tracing the source of a contaminant, and not for measurements, there is no need to wait for complete deabsorption.

COMPLETE CIRCUIT

In the complete circuit of the nose, Fig. 4, a reference voltage is provided, as before, by R1 and R2, and VR1 adjusts meter zero. The sensor is contained in a small probe and leads connecting to the measuring circuit via plug and socket PL1 and SK1.

A 100µA meter movement, with two switch selected series resistors R3 and R4, is used as a voltmeter in Fig. 4. Because the voltmeter places a significant load on the circuit, particularly when switched to the 0-3 range, the values of R3 and R4 are compensated to give a true relationship between ranges. In the 0-10 S1 position, meter full scale is represented by 4V.

Voltage controlled oscillator TR1 and TR2 gives an audible indication of changes in SGS resistance. VR2 sets the oscillator frequency to zero when the meter reading is zero. When SGS resistance decreases, the voltage across VR2 slider and the positive supply rail increases, thus applying a bias current to TR1 base. C1 charges up via TR1 and R5 at a rate dependent on the TR1 bias current.

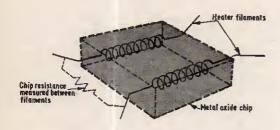


Fig. 1. Construction of the semiconductor gas sensor

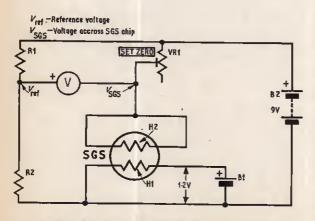


Fig. 2. Basic circuit of the electronic nose

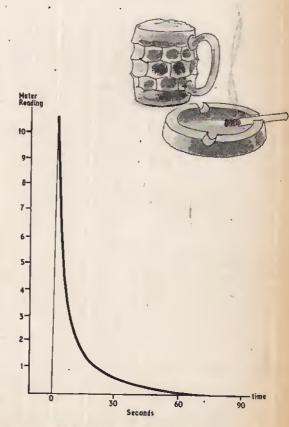
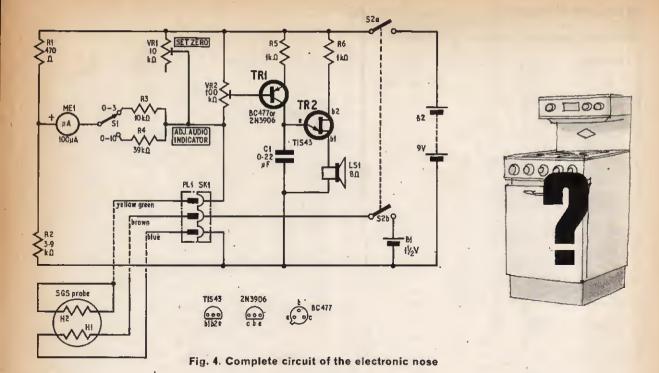


Fig. 3. Gas sensor response time graph



At the threshold voltage of the unijunction transistor TR2, CI is rapidly discharged through the loudspeaker LS1 to give an audible "click". For a small decrease in chip resistance, the audible indicator clicks like a Geiger counter, but if the chip resistance continues to decrease the clicks will run together to form a note of rising pitch which is related to the meter reading.

Battery B1 in Fig. 4 is a U14 gas lighter cell of nominal 1.5V. When loaded by the SGS heater, the internal resistance of the cell, plus the small resistance presented by interconnecting wiring, will give a voltage drop of around 0.3V, thus only 1.2V is finally applied to heater H1.

A reduction of heater voltage as B1 ages will increase absorption and deabsorption times, but will not materially affect calibration accuracy. Life of a U14 will be approximately I week when the instrument is used for two to three hours per day.

CIRCUIT BOARD CONSTRUCTION

Transistors, fixed resistors, and capacitor C1 are mounted on a 1.9in by 1.5in piece of 0.1in matrix plain circuit board, see Fig. 5. Insert terminal pins in the positions shown, so that they project from the underside of the board, then component leads can be pushed through and soldered to the pins, or other component leads. The unused portion of the circuit board is used for mounting purposes.

FRONT PANEL CONSTRUCTION

Front panel layout is shown in Fig. 6, and the photographs, but this will depend to some extent on the size of meter used.

Position the loudspeaker in the top left-hand corner at the back of the panel and draw round it with a pencil. Find the best position for MEI, to clear LSI and BI, and mark this out. Next draw in hole centres

for the controls, leaving a clear space above VR1 and VR2 for a short section cut from a plastic clip of the type used to hang posters, this makes an ideal circuit board mount.

Drill and cut all holes, including a loudspeaker aperture shaped to suit the space available, and apply lettering to the front of the panel. If VR1 and VR2 are miniature printed circuit mounting knob pre-sets, these can be glued to the front panel, along with the circuit board mounting clip, loudspeaker gauze, and loudspeaker. It is best to employ a semi-flexible glue as this will allow components to be prised off for replacement.

BOX AND BATTERY MOUNT

The U14 cell for B1 is held by a spring steel clip which is bolted to the back of the box and serves as the negative connection. A strip of brass makes contact with the positive battery pole. If the box is made of metal, the brass strip should, of course, be insulated from the box. A simple metal clip serves to hold B2 in position.

PROBE DETAILS

Details of the sensor probe are given in Fig. 7. The probe lead should not be more than 1 metre long, and all solder connections should be really secure, to minimise heater voltage drop. Try to avoid contaminating the SGS gauze with any material which might influence chip resistance, such as glue or paint. H1 and H2 heater pins are interchangeable.

ADJUSTMENT AND TESTING

Set VR1 and VR2 to the mid-position, S1 to the 0-10 range, and switch on. The meter pointer should

COMPONENTS . . .

Resistors

R1 470.0

82 3.9k O

R3 10k Ω 2% metal oxide R4 39k Ω 2% metal oxide

R6 1kΩ

All 10% \W except where stated

Potentiometers

 $\frac{10k\Omega}{VR2}$ miniature knob presets

Capacitor

C1 0.22 uF

Transistors

TR1 BC477 or 2N3906 TR2 TIS43

Sensor

SGS1 Semiconductor gas sensor (available from Trampus Electronix, P.O. Box 29, Bracknell, Berks.)

ME1 100µA f.s.d.

Batteries

B1 * U14 1.5V gaslighter B2 PP6 9V

Loudspeaker

LS1 24in dia, 8Ω

Switches

S1 Miniature single pole toggle S2 D.P.D.T. slide

Plug and Socket

PL1 3-way DIN SK1 3-way DIN

Plain 0-1in matrix board 1-9in x 1-5in. Terminal pins. 2A 3-core mains cable, 1 metre. Connecting wire.

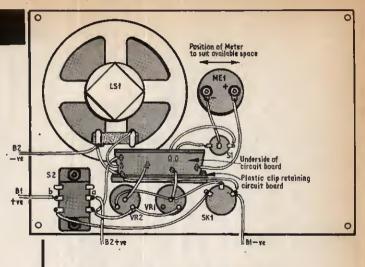


Fig. 6. Front panel interwiring details

touch the zero stop and then rise quickly to beyond full scale, accompanied by a note of rising pitch from the loudspeaker. After about a minute, the meter reading and audio note will begin to decline as the SGS decontaminates. Because absorption takes place when the SGS is not in use, and reaches a high level five minutes should be allowed for initial warm up and complete deabsorption when the instrument is first switched on.

Now trim VR1 for a reading of 0.2 on the 0-3 range and adjust VR2 for a series of slowly repeated clicks from the loudspeaker, these should cease when the

meter is zeroed.

After the initial warm up period the meter zero reading should remain reasonably stable, but remember that the SGS can detect slight traces of smoke and other contaminants in an ordinary room so some fluctuation just above zero will occur. In the event of erratic meter readings, the wiring between BI and HI should be suspected.

Out of doors, and in strong draughts, the meter reading may drop below zero due to cooling of the semiconductor chip, this effect can be minimised by

enclosing the probe in a simple windshield.

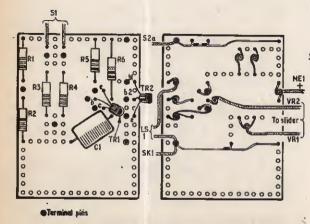


Fig. 5. Layout of the components on the perforated board

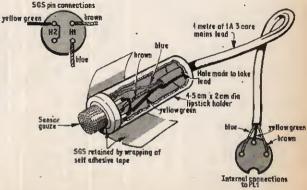
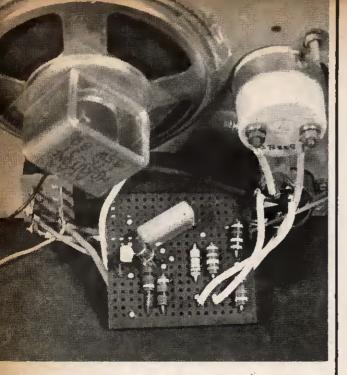


Fig. 7. Details of the sensor probe



Photograph of the rear of the front panel showing Veroboard and meter connections

To test the nose, move the probe near to an open bottle of methylated spirits or blow smoke at the sensor gauze.

APPLICATIONS

If the probe is gently breathed upon one or two hours after drinking a pint of beer, this should be sufficient to cause a reading of 10. The nose can therefore act as an ultra-sensitive breathalyser. Similarly, it may be possible to assess the alcoholic content of drinks by placing the sensor at a fixed distance from a sample of given volume and temperature.

Minor gas leaks from cigarette lighters, camping stoves, and domestic cookers are readily detected, with the aid of the audible indicator, as are fumes from storage batteries and many volatile chemicals. The electronic nose is particularly useful for detecting

odourless gases.

When placed near the exhaust of an internal combustion engine, the SGS will respond to carbon monoxide, as well as oil fumes and unburnt fuel, so the reading obtained can be taken as a general measure of engine tune and condition. However, for really accurate results, the exhaust gases should be captured in a bottle or can, the contents of which are "sniffed" in a draught free environment.

For air pollution measurements, the instrument should first be zeroed in a clean air zone, preferably out in the country, and then be taken to the area to be tested. Alternatively, comparative readings can be taken in a city environment by adjusting the meter to read 1 unit instead of zero. A high reading does not necessarily suggest serious pollution as this could be caused by a wide range of harmless volatiles, such as perfumes and deodorants!



... BATTLE CHESS

Expand your concept of Chess by introducing realistic "Battle" conditions into the game. Using simple electronic concepts Battle Chess avoids the boredom of Chess played as a court intrigue and takes it into the realms of the military struggle, where factors such as the strength of a piece, its distance from its fellows, morale of a side, are all considered in each individual conflict

AUDIO COMPRESSOR ...

The Audio Compressor is a unit which gives a constant output level for a wide range of input levels. It eliminates the need for constant level adjustment when using a microphone with a tape recorder and can also be used with amateur transmitting equipment to avoid the possibility of overmodulation

MEGOHMMETER...

Both the home electrician and the electronics experimenter will find uses for this simple to build high voltage megohmmeter.

Around the home it can be used to check wiring systems, or to detect dampness in walls. In electronics it can be used to locate poor insulation which could cause trouble in high impedance circuits

SYNTHESISER-7

Ring Modulator and Spring Line Amplifier

The ring modulator enables you to create unique exciting space-age sounds within the range of 25Hz-20kHz.



AUGUST ISSUE ON SALE JULY 13, 1973



OUTER HEBRIDES ACTIVITY

In aerospace activities the more glamorous and spectacular events command most attention. The fact is that other very important projects are proceeding which seldom appear in the news. One is the work that goes on at South Uist in the outer Hebrides.

In one day during April this year seven British Petrel rockets were launched. This is the first time that more than three launches a day have been made. Each rocket carried a payload constructed at the Science Research Council's Radio and Space Research Station at Slough.

The object of the launchings was to study the ionisation of the atmosphere above 60km (40 miles) altitude. The experimental data gained will demonstrate the development of the D-region of the ionosphere through a day which will assist the checking of the theoretical work in progress at the research station.

Each payload carried a number of experiments dealing with:—

1. Measurements of the electron concentration and the collision frequency of the electrons due to Faraday rotation and differential absorption at three radio frequencies.

These frequencies are transmitted from the ground

to the rocket.

- 2. Measurements of electron concentration fine structure using a Langmuir probedesigned for this special purpose.
- 3. Measurement of the intensity of the solar ultraviolet radia-

tion at 1216Å (the Lyman alpha region) and the concentration of molecular oxygen as a result of the atmospheric absorption of this frequency of radiation.

- 4. The measurement of X-rays from the Sun in the range 2-8 Angstrom units
- 5. A new experiment which was developed in collaboration with the Max Planck institute at Lindau, West Germany and the University College of Wales at Aberystwyth, to measure the concentration of atomic oxygen in the ionosphere by the manner in which corrosion takes place of thin silver film.
- 6. The housekeeping functions of the vehicle, that is, the attitude of the rocket to the Sun, the slant range of the rocket from the launching site, the instrumentation of temperature etc.

EXERCISE ACHIEVED

The rockets all reached altitudes of some 140km (90 miles) and measurements were transmitted by each rocket to the ground station on South Uist and recorded on tape

The data shows that the objects of the exercise have been achieved although some problems were met, such as postponements due to gale force winds and spectacular, though from the project point of view unwanted, aurora displays. The performance of this rocket, which is 7½ inches in diameter and nine feet long, has shown that it can be safely launched in near gale force winds.

The project scientist Dr P. H. G. Dickinson of RSRS has praised the support he has received from the U.K. Atomic Energy Authority and from the Army which was responsible for the safety of the range and the surveillance radar on St Kilda. The Skylark rocket was designed and made by Bristol Aerojet.

ARGENTINA AND THE SKYLARK

Two test firings have been carried out by a British team for the Argentine government. This is an attempt to use Skylark as an Earth resources rocket with the aim at finding a cheaper method of carrying out such surveys

A combined team from the British Aircraft Corporation and the Royal Aircraft Establishment, Farnborough, launched the rockets from a transportable launcher to a height in excess of 240km. The payloads consisted of four cameras with different lenses and different films. The descent of the payload was by parachute.

Recovery was very rapid as heli-

copters were used.

Some 640 photographs were obtained taking in a large part of Argentina. The pictures will be analysed by teams from Reading University and the Instituto Nacional de Technologia Agropechuaria of Argentina.

The valuation made will be in terms of land use and crop inventory.

SOVIET SATELLITES

The Soviet Union launched more than three times the number of Cosmos photo-reconnaissance satellites than the United States during 1972. In spite of this only 25 per cent more time was logged by the U.S.S.R. amounting to 366 days, compared with 295 days for the United States, which used only eight satellites.

It would appear that there was some acceleration of effort after the international agreement was signed. Eighty per cent of the satellites were of an advanced design with the provision of a rocket that can be vernier controlled which makes it possible to direct the satellite over a specific target.

The cameras used are probably of high resolution and narrow angle to

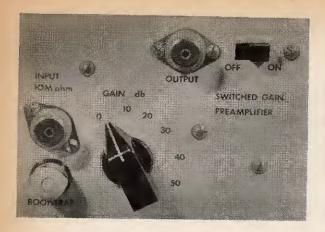
get greater detail.

The satellites are not just used for mapping strategical situated industrial plants. There is a more practical use for them. Two of the satellites were launched at 81 degree orbits and used to check ice packs and sea lanes. Another satellite enabled information about the conditions to be applied to the sea routes with the result that a flotilla of ice breakers and cargo vessels were able to make a safe trip from the Yenisei river to Murmansk. The U.S.S.R. attribute the success to the availability of accurate information from the satellites.

PIONEER

The second Jupiter probe, Pioneer G, is now on its way. It will take over from Pioneer 10 if that vehicle should have difficulty when in the vicinity of Jupiter.

The conditions there could be such that high-level radiation could damage the spacecraft. If however, the fly-by is successful the second vehicle will have a slightly different programme. Pioneer G or Pioneer II as it will be known will approach within 27,000 miles of the giant planet.



Switched Gain Preamplifier

By D. Crocker

HE amplifier to be described was designed to serve as a general purpose unit for a wide range of applications. It may be used, for example, between a signal source and a power amplifier to increase sensitivity, as an a.c. coupled oscilloscope preamplifier, a monitor amplifier to feed a crystal earpiece directly, or as an impedance matching device for coupling a crystal or ceramic pickup or a crystal microphone to an amplifier having a low input impedance.

In order to make the amplifier as versatile as possible, it was decided that a variable gain of up to 50dB was required together with a wide band width and high input and low output impedances. It should also have a low power consumption so that it may be run from

batteries, making it self contained.

CIRCUIT DESCRIPTION

The complete circuit is shown in Fig. 1. In order to obtain a high input resistance, use of a field effect transistor is made in the first stage. Unlike a bipolar (conventional) transistor, that is a voltage operated device drawing very little current from the input signal. The current flowing between the source and drain terminals is varied by the voltage between the gate (the third terminal) and the source.

This varying current is fed to the base of a p-n-p

transistor, TR2, comprising the second stage.

Since it is desirable to apply negative feedback to the source of TR1 in order to accurately determine the gain and widen the frequency response, the open loop gain (i.e. the gain before feedback is applied) should be above 3,000 for a closed loop gain of 50dB. the phase shift between TR1 source and TR2 collector is the required 180 degrees (TR2 having provided this), the gain provided would be much less than 1,000 if TR2 was given a simple resistor as a collector load. Another common emitter stage cannot be used as it would provide an extra 180 degree phase shift, making the feedback positive; an emitter follower, TR3, has therefore been used.

As this can provide no voltage gain by itself, R9, the collector load of TR2, has been bootstrapped by C4. This makes R9 appear to be in parallel with TR3 emitter-base junction, and since the effective resistance of this is lower than R9, most of TR2 collector signal current flows through TR3, giving rise to an amplified collector/emitter current in TR3 which flows through R12, developing an output voltage across it.

SPECIFICATION

Switched gains of 0, 10, 20, 30, 40, and 50dB Bandwidth Low frequency: -3dB at 10Hz High frequency: -3dB at: 200kHz (50dB gain) 500kHz (40dB gain) 800kHz (30dB gain) >800kHz (0, 10 or 20dB gain) Maximum output 12V p-p (no load) 7V p-p (10kΩ load) 1.5V p-p (1kΩ load) Slewing rate Approximately 10V/µS minimum Referred to input: approx. 150µV p-p (input open) approx. 25μV p-p (input shorted) Supply current Approximately 1.2mA Input resistance 12MΩ with S3 out.

The overall effect of the emitter follower TR3 and bootstrapped resistor R9 is to make the collector load resistance of TR2 very high, so that TR2 gives a high voltage gain.

FEEDBACK

Negative feedback is applied from the output to TR1 source through R11. As well as enabling the gain to be varied, this stabilises d.c. conditions throughout the amplifier and reduces its output impedance, When SI is set for unity gain (0dB), full negative feedback is applied; for higher gains resistors R4 to R8 are switched in to reduce the amount of feedback.

For an amplifier of this sort, the gain is given by the formula

$$G = \frac{A(Rs + Rf)}{Rf + Rs.A} \quad \text{if } A \gg 1$$

or

$$G = 1 + \frac{Rf}{Rs} \qquad \text{if } A \gg G$$

In this circuit A, the open loop gain, is between 3,000 and 7,000 depending on the load. Rf is R11 of Fig. 1 and Rs is the resistor switched in by S1. This resistor is connected in series with C3 in order not to disturb the d.c. conditions.

The current consumption of the amplifier has been reduced to about $1\cdot 2mA$ by giving R12 the rather high value of $10k\Omega$. This limits the maximum output available when a low impedance is used to the figures given in the specification. Whilst this has proved to be satisfactory, any constructor requiring a higher output current capability can decrease the value of R12 (but not below about $2k\Omega$) at the expense of reduced battery life.

an open loop gain of over 6,000, so that over 67dB of feedback is applied when the amplifier is set for unity gain, this form of instability has not occurred in the prototype.

CONSTRUCTION

In order to avoid hum pick-up at the high impedance input, the amplifier is constructed in an aluminium box which provides screening. Most of the circuitry is constructed on a piece of 0-1in matrix Veroboard measuring 25-4mm × 50-8mm.

Before mounting the components, breaks are made in the copper strips as shown in Fig. 2 and terminal pins are inserted from the copper side of the board. These should then be soldered in place and tinned.

The pins in row 7 support the screens; these are made from 2 pieces cut from tin measuring about 9mm × 33mm and 17mm × 33mm. They are flattened, cleaned, and one side of each is coated with solder. After placing small pieces of p.v.c. insulating tape between the 5 terminal pins over rows B, D and F, the

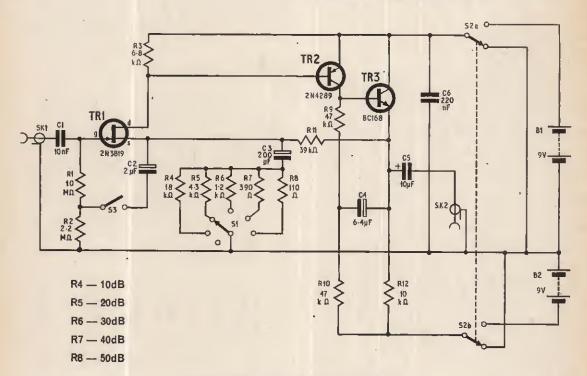


Fig. 1. Circuit diagram of preamplifier.

Since the output offset voltage may be positive or negative, the output capacitor C5 is a tantalum type; these can withstand small reverse voltages.

STABILITY

There are two ways in which an amplifier of this sort might be unstable. First, since the amplifier has a high, non-inverting gain and a high input impedance, feedback from the output to the input can easily occur causing oscillation. This is prevented by placing an earthed metal screen on the circuit board and using co-axial input and output connectors.

Secondly, due to extra phase shift in the amplifier, the feedback through R11 could become positive at high frequencies. In spite of the fact that the amplifier has

screens are soldered in place on the terminal pins in the position shown in Fig. 2. The smaller piece of tin is fitted on the copper side of the board. The components may then be wired in the positions shown in Fig. 3. Component wires should be sleeved where appropriate. An aluminium box measuring 4in × 2½in × 1½in with lid is required. The lid is drilled to accommodate the switches and sockets and these fitted.

MOUNTING THE BOARD

In order to mount the circuit board, one face of a $\frac{1}{8}$ in 6 B.A. tapped hexagonal pillar is tinned and soldered to the screen on the copper side of the board, placing it approximately below hole F8 (see Fig. 2), without allowing the screen to part from the pins. The pillar



COMPONENTS . .

Resistors 10MΩ 2-2MΩ R1 R2 R3 6.8kΩ R4 18kΩ 4.3kΩ R6 1-2kΩ R7 390Ω R8 110 $\hat{\Omega}$ R9 47kΩ R10 47kΩ R11 39kΩ 10kΩ R12. All W 5% carbon

Capacitors

C1 0.01 µF miniature polyester

C2 2μF 16V miniature electrolytic C3 200μF 6.4V miniature electrolytic

C4 6-4μF 25V miniature electrolytic C5 10μF 25V tantalum bead electrolytic

C6 0.22 F miniature polyester

Transistors

TR1 2N3819 TR2 2N4289

TR3 BC168B or BC169

Switches

S1 2 pole 6 way miniature wavechange switch (only 1 pole used)

S2 d.p.d.t. slider switch

S3 s.p.s.t. push on/push off switch (see text)

Miscellaneous

SK1, SK2 Surface mounting co-axial sockets (2 off). B1, B2 9V PP3 batteries (2 off). Aluminium box 4In × 2\frac{3}{2}in × 1\frac{1}{2}in with lid

0-lin matrix Veroboard 2in × 1in. Veropins for 0-lin matrix Veroboard (12). 6BA bolts (‡in), nuts and washers ‡in or ‡in. Terry clips (2). PP3 battery connectors (2)

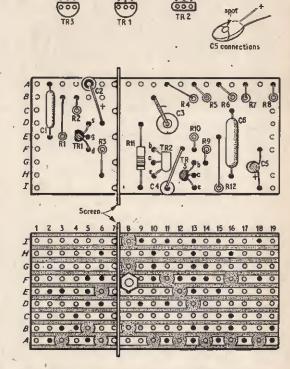


Fig. 2. Component assembly and wiring details.

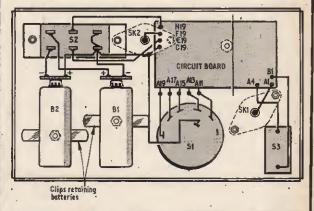
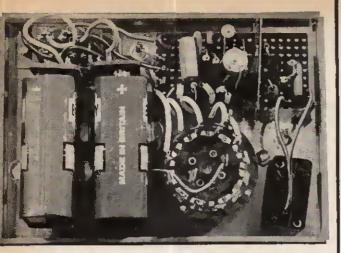


Fig. 3. Total component assembly within the aluminium box.



Rear view of the preamplifier front panel showing layout of components and position of circuit board

should not touch the board itself. It is then necessary to see if the board can be screwed into place by a 6 B.A. bolt passing through the appropriate hole in the lid into the pillar; check that the box will still fit inside the lid, and if necessary desolder and re-position the pillar. This pillar also earths the case.

Once the circuit board has been mounted the com-

ponents may be wired up as in Fig. 3.

TESTING

When the wiring has been completed and checked the batteries may be inserted. If a milliammeter is available, it may be connected in series with B1. When the circuit is switched on it should read about 1mA (ensure that S3 is out when measuring this). The amplifier can be tested using a signal generator or other signal source and an oscilloscope, power amplifier or crystal earpiece.

USING THE AMPLIFIER

This preamplifier has many uses in the workshop. As an audio preamplifier when a power amplifier has insufficient sensitivity or an input impedance too low for the signal source, as an audio signal tracer (using a crystal earpiece or a small power amplifier), as an i.f. signal tracer (using a detector diode at the output) or as a general purpose oscilloscope pre-amplifier.

In normal use a screened input lead should be used to prevent feedback or hum pick-up, this is not always necessary when the amplifier is fed from a low im-

pedance source.

It should be noted that a very low frequency damped oscillation occurs for several seconds after S3 is closed. Closing S3 gives the amplifier a very high input resistance due to the bootstrapping effect on R1; this may be several hundred megohms when a low gain is being used. However, since the 12 megohm input impedance obtained when S3 is out is usually quite sufficient, some constructors may prefer to omit this facility.

Due to the low current drain the batteries should give several months of operation. The use of silicon planar transistors should make this unit very reliable.



HANDBOOK OF I.C. CIRCUIT PROJECTS

By Jim Ashe

Published by Tab Books 224 pages, 5½in × 8½in. Price \$4.95 paperback, \$7.95 hardbound

PART from the very obvious disadvantage over American sourcing and prices, this book forms a useful addition to the shelves where one is comparatively new to work with integrated circuits.

Indeed, the fast reducing prices of i.c.s makes some sort of volume of this type almost mandatory for the

experimenter.

The current book deals with the subject from start to finish taking the reader through the basics of i.c.s and their handling and then gently introduces him to various projects. These range from audio equipment through various "novelty" items such as touch-plates

to instrumentation and security devices.

Readers of all conversions are catered for from the scientist to the Ham and indeed the author has taken the precaution of supplying a good deal of additional data on specific components, the sources of these and of data and so on. It is unfortunate that much of the source data is not valid here but most readers will be able to find products in the columns of PRACTICAL ELECTRONICS.

R.D.R.

UNDERSTANDING ELECTRONIC CIRCUITS

By Ian R. Sinclair

Published by Fountain Press M.A.P. Book Division 205 pages, 83 in × 51 in. Price £3.50

IKE Ian Sinclair's earlier book "Understanding Electronic Components", this volume is directed to the reader with some basic understanding of electronics, certainly to the extent of the properties of electronic components and the elementary laws of electrical circuits.

The author has directed his attention to the enthusiast who wishes to extend his understanding of the subject as well as to the undergraduate who wishes a wider understanding of electronics in an overall sense.

It should perhaps be emphasised that this volume does not set out to tell the reader how to construct something so much as how many of the circuits he will meet in electronics actually operate.

R.D.R.

POINTS ARISING

PUSHBUTTON VARICAP STEREO TUNER (May 1973)

Some constructors have experienced oscillation of the voltage regulator i.c. This may be cured by placing a 0.1 µF capacitor in parallel with C13. Also some constructors have reported hum picked up by the tuner. This may be cured by increasing C1 from 0.1 µF to 1.5µF or 2µF.



DEVICES ...APPLICATIONS

A Schmitt Trigger With Variable Hysteresis

As materials are evolved so new devices appear on the market at prices the experimenter can afford. In addition, existing devices come under scrutiny and are adapted to more and more applications.

In this section we present a selection of both new devices and applications, with news of applications

developed for existing devices.

Generally only basic circuit details will be given sufficient for the experimenter to create his own equipment.

THE Schmitt trigger is a two-state device which changes state each time a given input voltage is exceeded. It is one of the most widely used building blocks in electronics today, serving as a pulse shaper, voltage level monitor and voltage controlled switch.

All Schmitt triggers exhibit hysteresis, that is the input voltage at which they trigger is higher than the voltage below which they return to their initial state after being triggered. A typical hysteresis curve is shown in solid

lines in Fig. 1 (b).

The application of currently available logic circuitry to form Schmitt triggers has been investigated and one useful solution has been developed by Jim Halligan of Motorola. Using CMOS gates he has created a trigger which exhibits variable hysteresis. That is, in the curve of Fig. 1 (b) the voltage at which the trigger returns to the untriggered state can be varied.

First of all, the CMOS gate as a trigger. Fig. 1 (a) shows a Motorola MC14023 triple, three input, NAND gate

from the CMOS family.

The voltage measurement function of the trigger is performed by gate A, the point at which triggering occurs being determined by the threshold voltage of this gate.

The regenerative switching function in which the trigger returns from the triggered to the untriggered state is provided by gates B and C which are cross-connected to

form a simple bistable.

The threshold depends on the number of gate inputs which are driven, for example with a power supply $V_{\rm DD}$ at 15 volts if two inputs of a three-input gate are held high the threshold of the third gate is about 5 volts. With all three inputs driven in parallel the threshold is 8.5 volts.

Thus in Fig. 1 (a) gate A has a threshold of 8.5 volts and B a threshold of 5 volts when the output of gate C is high. In addition, with V_{in} at 0 volts the outputs of gates A and

B will be high and, due to bistable action, output of C will be low,

Increasing the value of V_{1n} will have no direct effect on gate B as all its inputs must go high if it is to change state. The truth table in Table 1 shows the various operations of the gates.

When V_{1n} reaches 8.5 volts the output of gate A will go low as all its inputs are driven in parallel. Thus the input of gate C is 11 from gate B and 0 from gate A. So the output of gate C goes high.

As a result gate B is presented with a 1 (the V_{1n} above 8.5 volts) and 11 from gate C. Thus its output goes low.

Hysteresis is created because the gate B now has two inputs high from gate C and one which is driven by V_{1n} . The threshold becomes 5 volts.

As V_{in} falls below 8.5 volts gate A output goes high but from Table 1 it can be seen that there will be no change elsewhere following this.

But, as the value of $V_{\rm in}$ falls below 5 volts the outputs of gate B goes high and that of C falls. Thus the circuit resumes its quiescent untriggered state. The transfer characteristics are shown in Fig. 1 (b) and the output waveform at 1 (c).

VARIATION OF TRIGGER

If one of the inputs of gate A is connected to $V_{\rm DD}$ the upper threshold is reduced as shown in dashed in Fig. 1 (b) to 7.5 volts.

to 7.5 volts.

The circuit of Fig. 1 (d) uses an MC14011 quad two-input CMOS gate to give true variation of the hysteresis. The value of the potentiometer is not too critical and allows the hysteresis to be set anywhere between 0 and 50 per cent VDD by adjustment of the upper level.

A small value of potentiometer will increase current consumption and a larger one will cause the sensitivity of

adjustment to suffer.

Prices of the components are:—14011, £0.52, 14023, £0.52, one off from distributors.

Table 1: NAND Gate Truth Table

Input	Output
000	1.
001	1
010	1
011	1
100	1
101	1
110	1
111	0

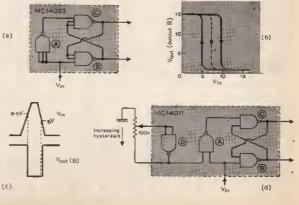


Fig. 1.

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H4 250	Mixed Resistors. Approx. quantity counted by weight	55p
H7 40	Wirewound Resistors, Mixed	
H9 2	OCP71 Light Sensitive	55p
H28 20	Photo Transistor OC200/1/2/3 PNP Silicon	55p
H30 20	1 Watt Zener Diodes, Mixed Voltages 6:8-43V	
H35 100	Miyad Diodes Germ Gold	55p
100	bonded, etc. Marked and Unmarked	55p
H38 30		55p
H39 10	Integrated circuits 6 gates BMC 962, 4 flip flops BMC 945	55p
H40 20	BFY 50/2, 2N696, 2N1613, NPN Silicon uncoded TO-5	55p
H4! 2	The state of the s	55p
UN	MARKED UNTESTED PAKS	
81 50	Germanium Transistors PNP, AF and RF	55p
866 150	Germanium Diodes Min. glass type	55p
863 200	Trans. manufacturers' re- jects all types NPN; PNP, Sil. and Germ.	55p
884 100	Silicon Diodes DO-7 glass equiv. to OA200, OA202	55p
B86 100	Sil. Diodes sub. min. IN914 and IN916 types	55p
B88 50	Sil. Trans. NPN, PNP	55p
	Sil. Trans. NPN, PNP equiv. to OC200/I 2N706A, BSY95A, etc.	
H6 40	250mW. Zener Diodes DO-7 Min. Glass Type	55p
HI5 30	Top Hat Silicon Rectifiers, 750mA. Mixed volts	55p
HI6 15	Experimenters' Pak.of Integrated Circuits, Data supplied	55p
HI7 20	3 amp. Silicon Stud Recti- fiers, mixed volts	55p
H20 20		55p
H34 15	Power Transistors, PNP, Germ, NPN Silicon TO-3	55p

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A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought. This is YOUR page and any idea published will be awarded payment according to its merits.

TAPE MOTOR CONTROLLER

The mechanical governor provided on a recently acquired portable tape recorder motor posed problems of electrical noise and mechanical wear. Thus it was decided to replace it with an electronic system to control motor speed both within a given speed setting as did the original governor and from speed to speed as is often achieved using different sized pulleys.

The circuit of the system is shown in Fig. 1 where current supply to the motor is controlled by TR3, itself controlled by the base voltage of TR1. A basic d.c. bias is applied to the base of TR1 from the Zener circuit and chain, R1, VR1/VR2. This sets the main motor speed.

Regulation of the speed about the selected value is supplied by detecting motor shaft speed using a pick-up coil and magnet assembly as shown in Fig. 2. The coil provides an a.c. voltage to D2 which thus gives a d.c. bias at the base of TR1. As the voltage fed back depends on the rotational speed of the motor speed control over a wide range is possible.

Under normal operating conditions the voltage from emitter follower TR1 appearing across R2 is about 2V. An increase in speed raises this value making D3 conduct through R3 and TR2 conducts heavily. This

cuts the bias on TR3 to slow the motor.

Similarly, a decrease in motor speed reverses the process.

Supply voltages from 8 to 15V are acceptable and in addition to exhibiting good quick-start properties, the equipment provides almost twice motor voltage on full stall condition.

MOTOR

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Colin Anderson, Wanganui, New Zealand

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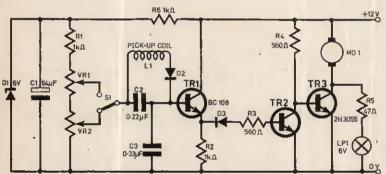


Fig. 2. Pick-up coil and magnet

Fig. 1. Circuit diagram for the tape motor controller

THE DTL integrated circuit type 9944 is a cheap dual buffer amplifier in one dual-in-line package and can be used to make a reliable, economical lamp flasher.

In the circuit, diagram Fig 1, the two amplifiers are cross-connected with capacitors C1 and C2, the values of which determine the frequency. The circuit becomes a multivibrator with the lamp LP1 in one collector.

Numbers in the circuit are the i.c. pin numbers. If C1 and C2 are $100\mu\text{F}$ (over 8V rating), a flashing rate of 30 per minute is obtained. Any 6V low current bulb may be used or the circuit could be used to operate a relay.

The Mullard equivalent i.c. is the DTL FCH 121, but the pin numbering is different. On this device, pins 1 and 8 are 6 and 8. Cross-connect pins 8 with 13 and pins 1 with 6, using the capacitors C1 and C2. Also pin 7 is +6V and pin 14 is ground (neg).

A. M. May, Weybridge.

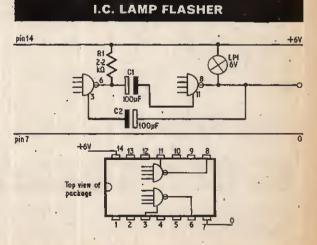


Fig. 1. Circuit diagram and i.c. connections for the simple flasher



WINDSCREEN WIPER CONTROL

HIS device, which can only be used with those windscreen wiper motors having a self-parking system since the system serves to turn-off the thyristor, can provide sweep delays of from 2 to 35 seconds.

The circuit is very simple, see Fig. 1, and uses few components. The original was constructed in a 35mm film can with switch S1 mounted on the variable resistor VR1. The control transistor is a UT46 or equivalent (2N2646) and the thyristor uses the can as a heatsink.

If the vehicle is negative earth with the windscreen wiper motor switched in the earth lead then the can may be earthed and only two leads are needed. One for positive supply and one for the positive supply to the wiper motor. As the thyristor is in parallel with the wiper motor polarity must be observed with care or triggering will not occur.

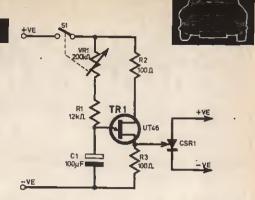


Fig. 1. Windscreen wiper delay circuit

The thyristor should be a high p.i.v. version, in the. original a 7A 50V p.i.v. type was used.

C. M. Calver. Melton Mowbray

ENGINE TEMPERATURE CONTROLLER

The circuit of Fig. 1 includes a stabiliser section, TR1, R1, R3, C1 and D1. This supplies power for the remainder which is basically a Wheatstone bridge circuit in which the thermistor TH1 forms one arm.

Balance of the bridge determines when the transistor

TR2 cuts off. The vehicle fan is switched by RLA1. For operation the normal "Temperature High" warning indicator is replaced with a thermistor. In practice the circuit is fail safe in that if TR2 goes open circuit for any reason then the fan is automatically switched on.

When the engine is cool the thermistor has a high resistance which puts TR2 in the on condition. As temperature increases and resistance decreases the bias voltage across the transistor decreases until TR1 cuts off. The level of cut-off is determined by VR1. The value of R3 is determined by measuring the bridge circuit resistance and making R3 equal it. R1 may need adjusting if R3 is high.

Manual over-ride and a form of test facility may be

provided by replacing the link by a switch.

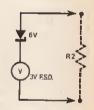
If required a temperature indicator can be added as shown in Fig. 2. The voltmeter must have a high resistance with respect to R2. The 6 volt Zener is used to hold the zero down by 6 volts. In practice the "normal" temperature will probably show as about

> R. K. Todd, Albrighton

TO FAN O-12V ₹2 500.fl 560.0 kΩ TR1 C2 I 10µF C VR1 680.0 500.0 01µF

Fig. 1. Circuit for the temperature controller

Fig. 2. Adding a temperature gauge



BRAKE LIGHT MONITOR

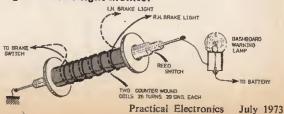
NDICATION of the state of the brake lights on a vehicle may be obtained without inserting extra lamp bulbs in the brake light circuit by using a reed switch actuated by a specially wound double coil to actuate a panel indicator lamp.

The circuit shown in Fig. 1 indicates the simplicity of the idea. Two coils serve to actuate the one reed but they are counter-wound so that as long as the pair of brake lights operate the reed remains quiescent. If

one or other bulb fails then the reed is operated to give an indication at the dash board.

S. W. F. Morum Chislehurst

Fig. 1. Brake light monitor



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	AC187	25p	BC159	120	BRY39	36 _p	ZTX108	120	2N3054		50p	LO36TI	£1,50
	AC187K	25p	BC169	15p	BSX21	260	2TX109	150	2N3055		48p	LQ37T1	£1.50
	AC168	250	B C268	15p	BSY95A	14p	2TX300	12p	2743702		12p	MC1303L	61.75
	AC188K	250	80115	67p	BU105/02	£1.75	ZTX301	150	2N3703		12p	MC1330	16p
	ACY20	200	BD131	70p	MJE340	67p	ZTX302	20m	2N3704		13p	MC1350	75p
	AD140	40p	BD132	76p	MPF192	40p	ZTX303	20p	2113705		10p	MC1351	£1.00
	AD149		80131/21		MPF103	370	ZTX304	2Sp	2N3706			MC1352	£1.00
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				21p	MPF105	48p	2TX500	15p	2N3708		12p	LMSBO	£1.45
	AF114	18p	BF167	20p	MPF106	45p	ZT X50#	15p	2N3709		12p	SL403D	€1.50
	AF115	11p	BF178	20p	OC28	40p	ZTK802	20p	2N3710		12p	TAD100	£1.50
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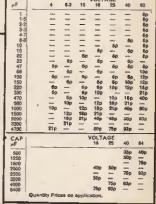
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SEMICONDUCTORS

PART 1 TRANSISTORS AND SMALL-SIGNAL DIODES

By M. J. Rose (MULLARD LTD.)

THE FIRST transistor emerged from Bell Telephone Laboratories twenty-five years ago. Since then, the whole electronics industry has been transformed. Perhaps the most important change the transistor brought about was the introduction of new technologies of manufacture.

It is from the development of these technologies that the many types of present-day semiconductor device have come. These two articles present an overall, and necessarily brief, survey of these devices.

THE FIRST TRANSISTORS

In 1948 John Bardeen and Walter Brattain of Bell Telephone Laboratories invented the point-contact transistor.

The point-contact transistor had two fine wires placed close together on a semiconductor wafer. It was shown that not only was the current flowing between one of the wires and the semiconductor wafer affected by the current between the other wire and the wafer, but under certain conditions amplification could be obtained.

Perhaps the most important result of the pointcontact transistor was that it provided experimental verification of the previous theoretical work. It paved the way for the junction transistor, a much more practicable device, invented by William Shockley also of Bell Telephone Laboratories in 1950.

JUNCTION TRANSISTOR

The junction transistor is essentially a sandwich of p-type and n-type semiconductor material containing two pn junctions. The first transistors were literally sandwiches, being made by the alloy-junction method.

A simplified cross-section of an alloy-junction transistor is shown in Fig. 1. This method of manufacture is still used for some audio-frequency transistors today.

Restrictions in minimum base width achievable with this technology gave the early alloy-junction transistors a cut-off frequency of approximately 1MHz, although refinements of the manufacturing technique raised this to approximately 5MHz.

ALLOY-DIFFUSED AND MESA TRANSISTORS

Other types of transistor were developed during the 1950's to overcome the limitations of the alloy-junction type. One of these types was the alloy-diffused transistor. The base layer of the alloy-diffused transistor was very thin, giving a cut-off frequency of approximately 800MHz.

Another type of transistor was the mesa transistor, originally developed to improve the switching characteristics. Its construction is shown in Fig. 2. The

mesa transistor could operate with switching waveforms with rise times in the region of 1µs.

The technique of mesa etching is still used today in some types of transistor.

PLANAR TRANSISTORS

The introduction of the planar transistor in 1960 marked several important changes in transistor manufacture. Silicon rather than germanium was used as the semiconductor material with considerable advantages. Silicon transistors will withstand higher junction temperatures than germanium ones, and have higher voltage ratings and lower leakage currents.

Diffusion from vapour into clearly-defined areas of the silicon wafer enabled close control to be exercised, and the diffusion areas could be constructed by photographic techniques. For the first time in transistor manufacture, true mass-production techniques could be applied, with a consequent drop in price.

In the manufacture of silicon transistors an oxide layer is grown over a wafer of *n*-type silicon. Windows can be cut in this oxide by etching in areas defined by exposure through a mask, and *p*-type silicon formed by a boron diffusion. A second oxide layer is formed, selectively etched, and *n*-type silicon formed by a phosphorus diffusion. In this way a transistor is built up, as shown in Fig. 3.

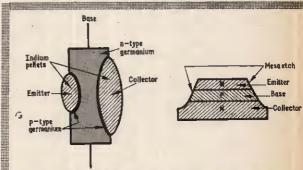


Fig. 1. Simplified cross-section of an alloyjunction transistor (left)

Fig. 2. Cross-section of mesa transistor (right)

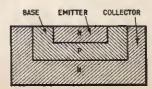


Fig. 3. Cross-section of planar transistor

Because the area of diffusion defined by a mask is the same for each transistor, the whole of a silicon wafer can be exposed at the same time. Thus up to 10,000 transistors can be manufactured on a silicon wafer 5cm in diameter in one manufacturing sequence.

Planar transistors can operate at frequencies up to the

microwave region, well above 1GHz.

PLANAR EPITAXIAL TRANSISTORS

A development of the planar transistor that improves the performance is the planar epitaxial transistor. In this type, the transistor is formed in an epitaxial layer which is simply a layer of silicon with a near-perfect crystal structure. The silicon wafer on which the layer is grown forms a substrate as shown in Fig. 4.

By forming the epitaxial layer of high resistivity material but making the collector in the layer very thin, a transistor with high breakdown voltage and high

collector current can be obtained.

FIELD-EFFECT TRANSISTORS

The transistors described so far are bipolar types, the current being conducted by two types of carriers (electrons and holes). The field-effect transistor (f.e.t.) is a unipolar transistor, the current being

conducted by one type of carrier only.

A current-carrying channel of *n*-type silicon is surrounded by the *p*-type gate, as shown in Fig. 5. When a positive voltage is applied between drain and source, a current flows through the channel. A depletion layer is formed at the *p-n* junction of the gate and channel, and by applying a negative bias between the gate and source the depletion layer moves into the channel to restrict the area available for current flow. As the bias increases, a value is reached, the pinch-off voltage, at which no current flows. Thus the current through the f.e.t. can be controlled by the gate voltage, as in a triode valve.

The f.e.t. described is a junction f.e.t. or j.f.e.t.

Another type can be constructed, the insulated-gate f.e.t., MOSFET or MOST. The initials MOS strictly speaking stand for Metal Oxide Semiconductor.

MOS refers to the construction of the transistor formed, a metal gate (aluminium or polysilicon) insulated from the silicon substrate by a layer of silicon oxide, as shown in Fig. 6.

By applying a voltage of the correct polarity to the gate, a current-carrying channel can be formed between the source and the drain, and when a voltage is applied across the source and the drain, a current flows.

The current-carrying channel in a p-type substrate is an n-type inversion layer. Similarly, a p-channel is formed in an n-type substrate. In addition, the p-channel and n-channels MOSTs can be enhancement or depletion types. Enhancement MOSTs are normally off, and require a gate voltage greater than the threshold voltage before conduction can occur. Depletion MOSTs are normally on, and require a pinch-off voltage on the gate to stop conduction.

Thus the circuit designer has four types of MOST to choose from for his application. All types of f.e.t. have very high input impedances, and so form a useful complement to bipolar transistors.

UNIJUNCTION TRANSISTOR

A unijunction transistor consists of a bar-shaped silicon slice with contacts either end and a p-n junction (the emitter) at some point along the length, see Fig. 7. It is therefore a three-terminal device but with only one junction.

With no emitter current flowing, the bar acts as a simple voltage divider with a fraction of the interbase voltage providing a reverse bias across the emitter junction. As the emitter voltage is increased, no appreciable current flows until the emitter voltage is greater than the reverse bias when there is a sudden increase in current. The unijunction transistor can therefore be used as a voltage-triggered device, in oscillators and sensors.

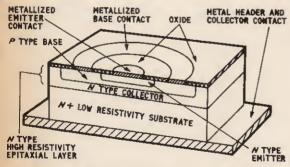


Fig. 4. Cross-section of planar epitaxial transistor

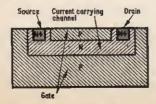


Fig. 5. Structure of junction FET

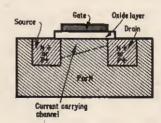


Fig. 6. Structure of MOSFET

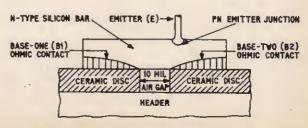


Fig. 7. Cross-section of a typical unijunction transistor

HIGHER POWER, HIGHER FREQUENCY

The development of the planar technology during the 1960's enabled the operating frequency of small-signal transistors to be raised so that by the end of the decade transistors with cut-off frequencies of 6GHz and capable of switching waveforms with rise times of Ins were available.

Germanium power transistors could handle about 10W with a suitable heatsink. This power level was increased with the introduction of silicon power transistors. Today, silicon power transistors for lowfrequency operation are manufactured by the singlediffused method, or, where high voltages are to be encountered (1kV or more), by the triple-diffused method. Powers of up to 150W typically can be handled.

The planar epitaxial technology enabled the operating frequency of power transistors to be raised considerably. More complex base-emitter configurations were developed to take operation high into the radio frequencies. A typical configuration is the interdigital one where fingers of the emitter layer interleave fingers of the base layer. Typical performance figures of present-day r.f. power transistors are 100W at 50MHz and 10W at 2GHz.

OTHER TYPES OF TRANSISTOR

The mexa transistor has an epitaxial base layer. The lightly-doped base is grown on a heavily-doped collector, the emitter layer diffused, and the complete structure mesa etched. The resulting transistor is

rugged and has a low collector resistance.

Hometaxial transistors are made by simultaneously diffusing impurities on both sides of a homogeneous base layer, followed by a mesa etch of the emitter. Because of the homogeneous base, the risk of hot spots is reduced, and the transistor has good second-breakdown properties. The wide base width also gives good voltage breakdown properties.

Triple-diffused transistors are also manufactured in this way; the base and emitter are diffused on one side of the collector wafer as usual, but a third diffusion is used to form another collector layer. The technique can be used on planar or non-planar devices, and gives an improved value of collector saturation voltage.

SMALL-SIGNAL DIODES

The modern point-contact diode uses a germanium wafer to which a fine wire contact is welded. Because the current rating is very low (only a few milliamps) these diodes are only used for applications where junction diodes cannot be used efficiently, for example, as detector and mixer diodes at microwave frequencies.

Besides the low current rating, germanium pointcontact diodes have a low inverse voltage and a high reverse current. These disadvantages were overcome by the introduction of the silicon junction diode in the mid-1950s. The junctions were made first by alloying with aluminium pellets, and later by diffusion. With the introduction of the planar technology, planar epitaxial diodes became available.

The limit of operating speed is set by the lifetime of the minority carriers formed by the reverse voltage. Gold doping reduces the lifetime of minority carriers in silicon, and by using silicon diodes with small junction areas and gold doping, recovery times of less than 5ns are obtained, making the diodes suitable for use in high

speed computers.

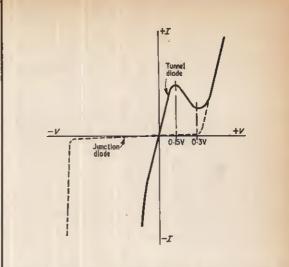


Fig. 8. Voltage/current characteristic of tunnel diode compared with junction diode

Faster switching diodes can be made by using gallium arsenide instead of silicon as the semiconductor material.

"ZENER" DIODES

As the reverse voltage across a junction diode is increased, a point is reached where the small leakage currently suddenly increases. This "breakdown" may be caused by electrons breaking from their bonds (the Zener effect), or more usually by avalanche break-

In this type of breakdown, the electron minority carriers gain sufficient velocity to dislodge other electrons which in turn are accelerated sufficiently to dislodge further electrons. There is therefore a sudden build-up of current, and this may destroy the diode.

If the breakdown can be controlled, however, a useful range of devices, generally called (incorrectly) Zener diodes but more accurately voltage regulator diodes, is obtained. The value of breakdown voltage Ois determined by the doping level of the impurities in the diode.

Voltage regulator diodes are incorporated into supply lines to provide a simple method of stabilisation. Special types with very low temperature coefficients are called voltage reference diodes, and provide a simple form of voltage standard. If the voltage regulator diode is made rugged enough to withstand the dissipation, then it can be used to suppress voltage transient surges. Such range suppression diodes can withstand surges as high as 10kV provided the duration is only a few microseconds.

TUNNEL DIODES AND VARACTOR DIODES

Two types of diode use the properties of pn junctions to give special characteristics. The tunnel diode uses a normal pn junction in which both regions are heavily doped. With a reverse bias such a large number of electrons are present at the junction that some "tunnel" their way across the junction. Similarly, with a small forward bias below that required for conduction in a

normal diode, tunnelling occurs. The forward tunnel current reaches a peak at approximately 0.15V, then decreases until normal forward conduction occurs at

approximately 0.3V.

The V/I characteristic of a tunnel diode is shown by the full line in Fig. 8 with, for comparison, the characteristic of a normal diode in broken lines. The negative resistance part of the tunnel diode characteristic can be used in oscillators.

The varactor diode uses the capacitance of a reversebiased pn junction. The reverse bias causes a depletion layer between the p and n regions, the width of the layer being proportional to the voltage. The junction area is made large to exploit this effect and to provide a reasonable change in capacitance with voltage.

Varactor diodes can be used as tuning elements to replace such devices as the parallel-plate tuning capacitor. They also have application as frequency multipliers. A string of varactor diodes and tuned circuits can be used to generate power efficiently at microwave frequencies from a lower-frequency source.

MICROWAVE DIODES

Some types of diode have properties that make them especially useful at microwave frequencies. Such types are the backward diode, PfN diode, Schottky barrier diode, and Gunn diode.

The backward diode, as the name implies, is a reverse-biased junction diode but one in which the breakdown voltage occurs very near the zero voltage point. Thus the resistance with reverse bias is lower than the forward resistance. Backward diodes can therefore be used as microwave detectors.

The PIN diode is also a pn junction but with a layer of intrinsic silicon between the p and n regions. With a forward bias, the PIN diode presents a low resistance because the bias injects carriers into the intrinsic region. With a reverse bias, however, the resistance remains very high. The PIN diode can therefore be used as a detector and also as a switch.

The Schottky barrier diode uses a metal-semiconductor junction as the rectifying element. Like the point-contact diode, the Schottky diode has the advantage of not suffering from minority carrier storage effects that limit the performance of junction diodes. Thus Schottky diodes can be used for higher-

frequency applications, and also for very fast switching.

GUNN DIODE

The Gunn diode is not, strictly, a diode at all as it does not contain a rectifying element. It is formed by a thin layer of n-type gallium arsenide grown on a gallium arsenide substrate. The substrate forms the "anode" of the Gunn diode, while a "cathode" contact is formed on the face of the n-type layer.

When a low d.c. voltage is applied across the diode, high field regions called domains are formed at the cathode and rapidly cross to the anode. Current pulses are superimposed on the steady current through the diode, the frequency of the pulses depending on the transit time and therefore the thickness of the layer. For a thickness of 10 µm, the frequency is 10 GHz.

The Gunn diode can be mounted in a tuned cavity to form a simple microwave oscillator with advantages in space and cost over such devices as klystrons. Maximum output powers up to 3W can be obtained from Gunn diode oscillators.

The second article will describe power devices, photo devices, and integrated circuits.

PHYSICS EXHIBITION

T is interesting to see that, in the issue in which PRACTICAL ELECTRONICS publishes an article on an "Electronic Nose", we are also able to describe an electronic alcohol meter which will probably replace the so-called Breathalyser in drinking-and-driving cases and prove useful in many fields of research.

BREATHALYSER?

The Alcolmeter, as the latter is called, is an electronic instrument capable of indicating the presence and concentration of certain chemicals in air. It was on show at the Physics Exhibition, Earls Court, in April together with several other interesting items which show the steady movement of electronics into the important field of measurement and control.

The Alcolmeter uses a "fuel cell" developed to detect only alcohol unlike the one used in the P.E. Electronic Nose, which is capable of much wider application. In operation the Alcolmeter cell gives an output in millivolts dependent on the alcohol content of air drawn into a cell chamber by a small single-stroke pump. This is amplified and applied to a meter calibrated in milligrams of alcohol per 100 millilitre of blood. It is understood that accuracy is up to the requirements for normal police operation here so we can perhaps expect to see the instrument in use on the roads soon.

Of course, such an instrument has many applications in research and, using somewhat different cells, can be

made to detect other types of chemical.

Light detection in the operation of power stations was well in hand on the CEGB stands with a demonstration of the Nordust smoke monitor, used to identify the presence of dust in smokestack gases. This instrument uses light scattered back to a photocell from dust particles in the airstream. Here electronics not only makes the operation possible in terms of sensitivity, as the amount of light available is very small, but also in terms of removal of anomalies due to cell variables by using balanced input operational amplifiers.

LOOKING AT FIRE

Similarly, it is only because electronic techniques such as correlation between two signals exists that the CEGB have been able to develop a flame monitor capable of identifying one flame in a furnace using perhaps 20 burners. This is effected by looking at the flame in question from two different directions, correlating the outputs of the viewing cells and indicating the amount of correlation present. Where there is a high percentage, say 80 or more, the flame is alight. Under any other conditions the amount of correlation is much lower. Such equipment can be used to initiate some form of "flame-out" warning.

On the signal generation front it looks as if there are some interesting developments afoot. The Royal Radar Establishment at Malvern have developed some devices called Surface Acoustic Wave Oscillators using an excess gain amplifier and a SAW delay line.

As both devices are very small and can be made using photoetching techniques, samples have been prepared capable of oscillation up to 490MHz as a fundamental frequency and mounted in a single TO8 can. It is envisaged that frequencies up to IGHz are well within the bounds of possibility with this device. But first of all some problems over temperature variations have to be overcome.





OHM NURSERIES

Gardening is not one of my personal predilections, although there are times when the subject manages to stir a glimmer of interest in me, as when, a few months ago, someone employed electronic means to accelerate crop-shedding (which you will, no doubt, recall) in orchards. Then there's the subject, requiring a lot more scrutiny (and perhaps something to which we shall return) called "Electrology" claiming an ability to reproduce whole outlines of leaves by previous photography of incomplete or damaged specimens!

However, the subject taking interest with a number of nurserymen in Canada right now is the result of some researches recently conducted by Professor D. S. Fenson and Dr G. Craig for their department of agriculture. Apparently, too much guesswork is associated with assessing the right time to lift plants for storage during the winter, following a hardening-off period in the autumn.

Measurement of the plants (in this case the strawberry) resistance, electrical that is, they have established a direct correlation with the fittest time to dig 'em up. This resistance changes (they don't say how) with the sudden drop in temperature immediately preceding the winter, thus permitting early removal of the less hardy plants before frost damage can occur.—Do I hear an "ohm groan" from the nongardeners?

GETTING THROUGH

Would you believe, communicationwithout-wires (apart from jungle drums, semaphore, and heliographs) was "old hat" before either Marconi, or, that great Russian, Popof, "Invented" radio? Actually, it was about 1866; even prior to the invention of the electric lamp, that a dentist, one Mahlon Loomis, demonstarted a workable electric wire-less set-up over a distance exceeding 14 miles. Just how this worked is not terribly clear, but in essence it consisted of a pair of kites, one at each station, carrying some several hundred feet of wire "line". This, at the "transmitter" was grounded to earth, while at the "receiver" it was terminated to earth via a sensitive galvanometer.

The principle of operation seems to have been one which relied upon charged air currents and, presumably, a stiff breeze to boot. But credit where it is due, he did manage in some way (by, probably, making and breaking the transmitter earth connection) to interfere with the static charge in the vicinity of the wire and so influence the overall electrical equilibrium; sufficient, certainly, to be detected well over ten miles away.

Obviously, there must have been both considerable delays, and uncertainty about the reception (particularly for changes in wind direction and speed) but his idea at the time carried enough conviction for two fair sized companies to be formed; unfortunately, poor old Loomis died without ever fully realising his ambitions and with him too did the "atmospheric communicator".

RADIO VACUUM

You might be interested to hear that even today people are experimenting with methods foreign to normal transmission techniques to overcome the many disadvantages, to name poor penetration of the walls of buildings as a typical one. Just the other day I learnt that Dr Richard Arnold of Argonne National Laboratories in Illinois managed to achieve success with the transmission of a pulse-coded signal along a muon beam of all things! Since muons are atomic particles, detection consisted of a scintilation counter arrangement.

If this were not enough, my colleague, inspired by all this chat about particulate transmitters, assures me that once he has overcome the teething troubles with the bag size of the laboratory vacuum cleaner, he is entirely convinced that we will be able to announce the birth of the worlds first dust'n'soot scatter propagation system; what do you

NASTY TASTE?

Apart from the ferry, Gosport now seems to be in the position to capture our hearts with yet another singularly cunning piece of wizardry; a firm there has brought into being an electronic (more accurately, ionic) water de-bugging device.

Designed to clip into the supply pipe, it works on the principle that bacteria have a common dislike for silver ions, and comprises simply a pair of electrodes which cause ions to go into solution and rapidly interact with the bugs, "doing them in" very quickly. One of these units can comfortably handle as much as 250 litres of brackish aqua in a little over a minute or so.

Good news for campers, but what happens once it's in the gut, I wonder? Perhaps the stomach's hydrochloric acid converts it to the chloride which ought to be very harmful, except to your pocket; have you seen the price of silver recently?



QUESTION OF FITNESS

Probably because I'm a bit of an "odd-ball" anyway, the Editor and I sometimes have difficulty agreeing about the relevance of subject matter in this column to the field of electronics; and, since this magazine is one which sets out to deal fairly exclusively with the subject, it isn't unreasonable that he should expect there to be some link. But there's the problem. Just what is some link? And by exclusively, do we really mean restrictive?

*Each time you buy your copy of P.E. have you ever wondered whether, sooner or later, the "crunch" might come when there will be literally nothing left to invent? Of course, there are the good old standbys like electronic doorbells, burglar alarms, and so forth, plus even more sophisticated ways of doing these same things in miniature. But what then?

In this inventing lark, for whatever the outlet, it seriously looks as if, to survive, we must look upon our particular subject in a less exclusive way. Indeed, to innovate in the real sense of the word must require us to never completely exclude any other discipline, just in case it turns out to be vital!

Once we (and, certainly, industry) accept this maxim, my prediction is that it will only be a matter of time before technology sees the "birth" of Interfacing Engineers capable of performing a sort of inter-disciplinary integration. Henceforth, instead of knowing an awful lot about a little, they will be in the unique position to discover myriads of, previously unheard of, applications from the suddenly limitless number of available permutations.

Let's drink to the daring, the real innovators, and to the possession of minds that never permit us the hollow excuse to be bored stiff!

PE Sound Synthesiser 6

TONE CONTROL & EXPERIMENTAL SOUND STUD

By G.D. Shaw

build up tonal correction curves to

RAMP GENS.

THIS month the Tone Control module will be described together with some notes on the setting up of an experimental sound studio.

HUTCHINSON TONE CONTROL

The tone control circuit used in the Synthesiser is based upon the design by P. B. Hutchinson which appeared in Wireless World (November 1970) and which, in turn, was inspired by R. Ambler's design for a "Tone Balance Control" which was published in the March 1970 issue of the same journal. A number of minor changes have been made to the original circuit in order to make it fully compatible with the circuitry in the Synthesiser but the frequency selective network, suggested by Hutchinson have been retained.

The type of tone control usually adopted for the majority of domestic equipments is that in which the greatest effect occurs at the extreme ends of the audio frequency spectrum while the mid-band frequencies remain relatively unchanged. The Baxandall and Siemens designs provide just such a characteristic. In recent years however there has been a tendency towards extending the scope of the tone controls in domestic equipment.

For purely domestic purposes there is no doubt that the ability to manipulate the frequencies over the whole of the audio frequency spectrum can make an enormous difference to the quality and effect of the ensuing sound. The Hutchinson tone control circuit combines many of the advantages of mid-band correction with relatively severe "end of range" effects all of which are extremely useful to the experimenter in sound.

TONE CONTROL RESPONSE CURVES

Fig. 6.1 shows a graphical representation (shown dashed) of the kind of effect obtainable by means of a conventional tone control circuit of the passive kind. Note that there is very little effect on the mid-range frequencies even when the treble and bass controls are adjusted to their maximum value positions.

On the other hand, the designed response of the Hutchinson circuit (shown solid) provides an overlap of bass and treble responses ranging from about 200Hz to 4kHz. The provision of separate boost and cut controls to both the bass and treble ranges therefore means that it is possible to tailor the overall characteristic of the curve to suit many particular applications. It is, for

example, possible to build up tonal correction curves to compensate for deficiencies in sound transmission transducers such as pick-ups or loudspeakers.

The basic principle of operation is illustrated graphically in Figs. 6.2 and 6.3. Each potentiometer in the circuit controls the 3dB frequency of a curve having a slope of about 6-7dB/octave, the function of the potentiometers being in accordance with the direction of the slope. In Fig. 6.2 the dashed lines (1) represent the effect on the audio frequencies when all the controls are set flat. Since both the boost and cut selective networks are designed to have a fairly close match a similar result is obtained if all controls are set to their maximum positions.

With all the controls set flat, the effect of advancing any one of them is to bring the related slope further into the audio spectrum. The dotted line (2) shows the result of advancing the treble boost control about halfway. So far, of course, the resultant curve is similar to that obtainable from a conventional tone control circuit but if the treble cut control is now advanced from its flat position the effect is to reduce the

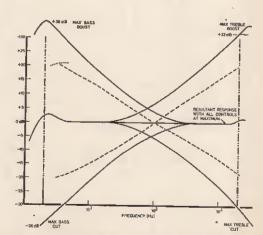


Fig. 6.1. A comparison of characteristic curves obtained from a conventional passive tone control (shown dashed) and those from the Hutchinson circuit (shown solid)

degree of boost to the upper frequencies without affecting the boost applied to the middle frequencies. This is shown as the overall resultant (4) by means of the

solid line in Fig. 6.2.

An advantage of this form of control lies in the possibility of combining the effect of boost and cut settings from opposite ends of the spectrum to provide a resultant slope of up to 12-14dB/octave. A typical curve obtainable from such combination is shown in Fig. 6.3.

TONE CONTROL BASICS

The general form of the tone control circuit is shown schematically in Fig. 6.4, from which it will be seen that two frequency selective networks are employed. One lies directly in the signal path to the operational amplifier and acts as the cut network whilst the second lies in the amplifier feedback loop and acts as the boost network.

The design of each network is identical and is shown in Fig. 6.5. C1, R1 and VR1 form a simple first-order high pass filter with a slope of about —6dB/octave. The 3dB frequency is given by I/CR where C is the value of C1 and R is the value of R1 in series with the value of VR1. With the component values shown the highest 3dB frequency is thus 1.54kHz and the lowest is 28Hz. C2, R2 and VR2 form a simple first-order low pass filter again with a slope of about —6dB/octave. In this case the highest 3dB frequency is 22.6kHz and the lowest is 415Hz.

The same network is able to provide boost characteristics since, in the feedback path of the operational amplifier, it reduces the a.c. feedback thereby increasing

the gain of the amplifier.

Depending upon the settings of the potentiometers the input impedance of the frequency selective networks can vary between 470 ohms in parallel with 470 ohms at minimum and 25.5 kilohms in parallel with 25.2 kilohms at maximum. It is thus necessary to provide a driving source having an output impedance of less than 200 ohms.

Similarly, it is necessary to compensate for changes in output impedance of the tone control circuit by

driving into a load of 50 kilohms or more.

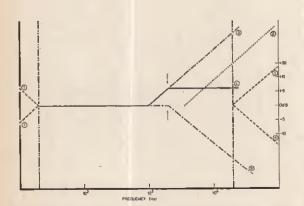


Fig. 6.2. Operation of the Hutchinson Tone Control circuit. Controls set flat (1), treble boost advanced—other controls flat (2), projected treble boost and cut with controls adjusted separately (3), resultant curve resulting from summing the boost and cut curves

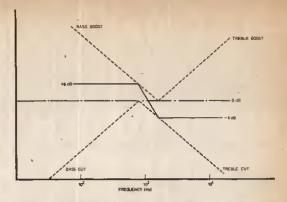


Fig. 6.3. Combining controls to give a 12dB/octave cut-off at mid-band. Dashed curves indicate responses when controls are used separately

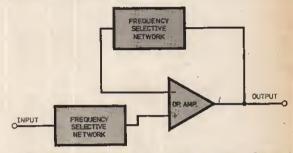


Fig. 6.4. Block diagram of Hutchinson Tone Control

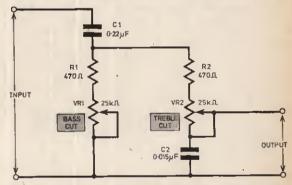


Fig. 6.5. Configuration of high pass and low pass filters used in Hutchinson Tone Control

THE COMPLETE CIRCUIT

The complete circuit of the Tone Control is shown in Fig. 6.6. TR1 is an emitter follower which provides a low impedance drive. R10 and R11 provide d.c. feedback to maintain the output of the amplifier at 0 V. Capacitor C6 decouples audio frequencies from the d.c. feedback loop and R12 provides a limit to this decoupling so that the a.c. closed-loop gain of the amplifier is limited to about +36dB. In the original circuit this was found necessary to avoid resonances at the extreme ends of the audio spectrum under conditions of maximum boost. The resonances were caused by interaction between the boost characteristic and the d.c. feedback loop at the bass end, and the high frequency compensation of the amplifier at the treble end.

TONE CONTROL MODULE

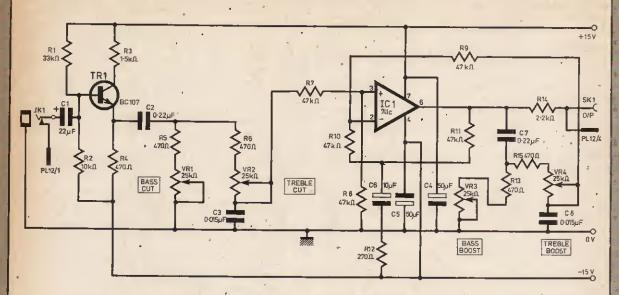


Fig. 6.6. Circuit diagram of the tone control

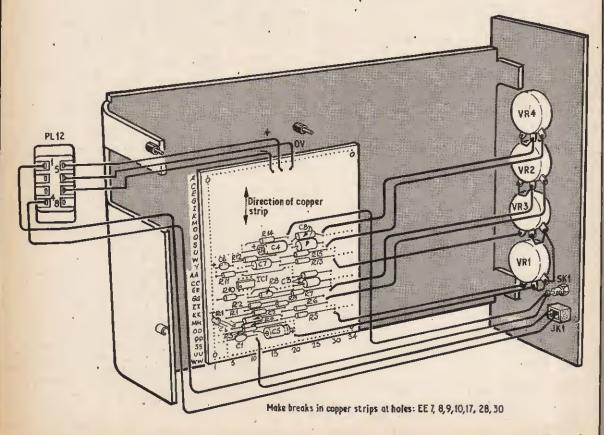


Fig. 6.7. Veroboard component assembly and wiring details for Tone Control module

The original circuit utilised a 709 operational amplifier as the active element and substitution of a 741 has resulted in the virtual elimination of high frequency resonance problems. However there is the possibility that low frequency resonance may remain under conditions of maximum bass boost although this can be tuned out by careful adjustment of the value of R12.

ASSEMBLY

Construction of the module should generally follow the pattern already outlined, that is, assembly and wiring of the front panel components followed by assembly of the module hardware and finally the wiring in of the circuit board.

Layout of the components on the circuit board is not too critical and the recommended circuit board layout

is shown in Fig. 6.7.

Some care should be taken in wiring up the potentiometers since it is quite easy to wire these in the wrong way round and this will give rise to some very odd results during testing out the finished module. Fig. 6.7 also shows the component layout and wiring on the front panel and McMurdo plug.

TESTING

Testing the module covers only the establishment of performance characteristics at various points in the audio spectrum and it is recommended that a sine wave oscillator be used for this purpose. Starting with the oscillator at 20Hz the frequency should be advanced in multiples of 20Hz up, to and including 100Hz and thereafter in multiples of 200Hz up to 1,000Hz.

From 1,000Hz the frequency advance should be in steps of 2,000Hz. This procedure will give about 20 points on the graph and should be quite sufficient

to establish the circuit characteristics.

COMPONENTS . . .

TONE CONTROL

Resistors	•		
R1	33kΩ	R7-R11	47kΩ (5 off)
R2	10kΩ	R12	270Ω
R3	1.5kΩ	R13	470Ω
R4	470Ω	R14	2·2kΩ
R5-R6	470Ω (2 off)	R15	470Ω
	watt metal oxide		

Capacitors

C1 22μF tantalum 16V
C2 0.22μF polyester
C3 0.015μF polystyrene
C4-C5 50μF elect. 25V (2 off)
C6 10μF elect. 25V
C7 0.22μF polyester
C8 0.015μF polystyrene

Integrated Circuit IC1 741C 8pin d.i.l.

Transistor TR1 BC107

Potentiometers VR1-VR4 25kΩ linear min, moulded carbon (4 off)

Miscellaneous

JK1-3-5 mm jack socket, SK1-2 mm miniature
socket, 0-1 matrix Veroboard

The measuring procedure should be repeated five times, the first being with all controls set flat and subsequent measurements with each of the individual boost and cut controls set to maximum in turn. If component values have been chosen as specified the resultant graph should bear a very close resemblance to that shown in Fig. 6.1.

There is little to be said concerning the actual usage of the module since, as with all tone control circuits, the position of the various controls will be dictated by a variety of circumstances not least of which is the individual constructor's preference. The little extra trouble involved in the manipulation of four controls is well rewarded by the greater versatility of the system in the provision of exactly the sound required. As a general guide to module usage there are two basic methods of operation.

Tonal correction curves are provided by setting both the boost and cut slopes to the middle of the audio spectrum and then moving one or the other out until

the required effect has been achieved.

Simple boost or cut characteristics are obtained by setting all the slopes to the extreme ends of the audio spectrum and then moving the appropriate one in towards the middle until the desired effect is achieved. If only one control is manipulated in this latter case the resultant slope will, of course, be about 6dB/octave.

FORMING A STUDIO

In part four of this series the reader was introduced to the idea that the tape recorder was a very necessary adjunct to the Synthesiser if the full potential of the instrument were to be realised. This idea may be extended by considering the Synthesiser as the central focus around which may be constructed a simple yet versatile sound studio capable of providing a wide range of facilities to the keen experimenter. Thus, in addition to the tape recorder, there are a number of other items of so-called peripheral equipment which may be added to and usefully extend the versatility of the system. Fig. 6.8 illustrates one possible arrangement which utilises a fairly wide range of external apparatus.

In common with the basic concept of modular construction for the Synthesiser, the establishment of a studio may also be considered on the basis of a form of modular growth in which individual pieces of apparatus, including the Synthesiser, are treated as separate modules. In the Synthesiser itself all the modules plug into a bus-bar system which provides power supplies and signal paths for the interconnected, prewired modules. In a similar fashion the studio must have a central signal routing system and this is provided by what is known as the main patch. The power supplies for the studio apparatus are not, however, routed via the main patch, and the provision of a power supply distribution system will be dealt with later in these notes.

THE MAIN PATCH

The main patch in any studio is thus a very important piece of apparatus. Ideally it should provide a rapid means of interconnecting the signal pathways of all the apparatus incorporated in the studio in virtually every possible configuration and do so with a minimum of noise and crosstalk addition.

There are a number of possibilities for the construction of the main patch, the actual method adopted depending largely upon the means and inclination of individual constructors. It is worthwhile stressing,

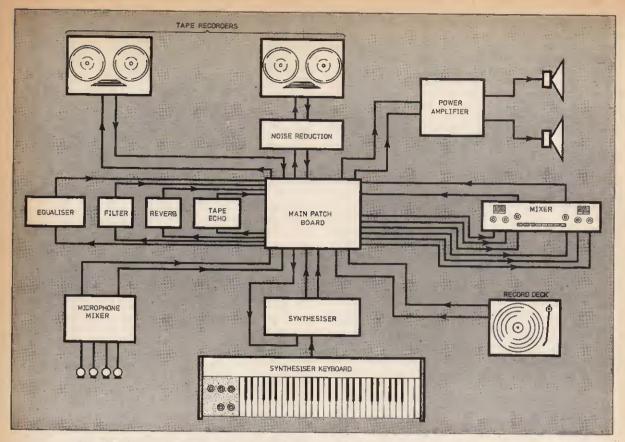


Fig. 6.8. One possible arrangement of apparatus to form an experimental sound studio

however, that a little extra quality in components chosen and effort expended in construction will be amply repaid in terms of reduction in the problems associated with poor contact assemblies.

One method of construction involving very little effort on the part of the constructor lies in the employment of ready-built patching matrices such as those manufactured by Sealectro Ltd. Matrices of a similar type are supplied "built-in" in a number of commercially available synthesisers and have proved to be quite

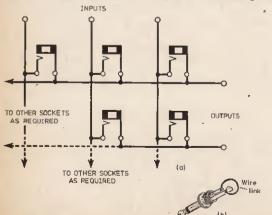


Fig. 6.9. Arrangement of jack sockets to form a patch board

successful in this application. The adoption of a readybuilt matrix has the advantage that the patch can be assembled very quickly. In addition, the space taken by the patch is very small indeed although this could perhaps be something of a disadvantage if the screened signal cables for connection to the matrix are of any size at all. Another idea for the construction of a patching system is the use of 3.5mm jack sockets arranged generally as shown in Fig. 6.9. It is important to note that the type of jack sockets employed should be of the open-circuit variety.

The type of main patch employed in professional circles frequently makes use of a system based upon standard G.P.O. type jacks. Although rather larger than either of the other possibilities mentioned the greater contact area and general robustness of the components involved makes this latter system worthy of consideration if any serious work of high quality is envisaged. In terms of cost the system is not significantly more expensive than those previously mentioned.

OVERCOMING HUM

Perhaps one of the greatest problems associated with the interconnection of a range of apparatus lies in the formation of ground or hum loops within the system. In some circumstances the source of the hum can be quite difficult to trace and it is therefore worthwhile expending a little time in establishing a system which eliminates, from the outset, as many sources of hum as possible.

By far the greatest source of hum lies in the formation of loops through the earthing system of mains powered apparatus. The frequency of the hum can vary between 50Hz and 120Hz depending upon the mains frequency and type of rectification employed. Until quite recently the majority of mains powered apparatus employed the live chassis method of construction in which the signal ground was common with the mains earth. This type of apparatus provides the greatest hazard as far as the formation of hum loops is concerned.

COMMON POINT EARTH

Connection of a number of pieces of mains powered apparatus into the main patch may be accomplished as shown schematically in Fig. 6.10. In this case power supplies are distributed from a panel which has one common earth to the mains. Screened signal cables are used where the screen is disconnected at one end of the cable only. Thus the connections at the main patch are made on high signal levels only while the circuit is completed through the common earth in the power distribution panel.

Although these latter remarks apply particularly to apparatus in which live chassis contruction has been employed, a similar form of interconnection may be employed with apparatus which has a fully floating signal ground, i.e. there is no electrical connection between the signal ground and the mains earth.

In circumstances where all the apparatus involved is of this latter form of construction it is necessary to provide a common ground return on all signal lines and this is best done by connecting the screens of all signal cables to a common point at the main patch. This latter point is left floating except in the case where mixed types of apparatus are being used and, in these circumstances, the signal cable screens of apparatus with floating outputs only are commoned at the main patch and this point then returned direct to mains earth.

CHOOSING THE RIGHT CABLE

A second prime source of hum lies in the type of cable used to route signals and its relationship to the

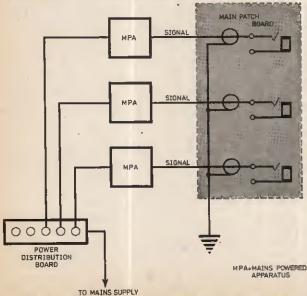


Fig. 6.10. Connections of peripherals into the main patch

impedance of the inputs to which it is connected. It is best to choose a signal cable which is compatible with both low and high signal level inputs and thus one should steer clear of the majority of commercially available so called miniature signal cables which are prewired to a range of plugs and sockets. Unfortunately many of these, whilst perfectly suitable for the transmission of relatively high level signals into high impedance loads, are quite unsuited to low impedance networks due to the relatively high impedance and core/screen capacitance of the cable itself.

A signal cable which should prove suitable for most purposes in the studio will have a nominal impedance of about 50 ohms and a core screen capacitance of not more than 100 picofarads/metre.

CHOICE OF PERIPHERALS

Referring again to Fig. 6.8 the choice of peripherals for the Synthesiser must rest primarily with the individual, the only limiting factor being the compatibility of one unit with another. Although all the equipment shown is available commercially in several different forms a large proportion of it could be constructed at home with the aid of designs published. For example the Synthesiser Reverberation Amplifier (to be published next month) can be employed, without modification, to drive the spring line unit type HR162 available from Henry's Radio Ltd. Commercially available reverberation units include those manufactured by Eagle, Grampian and Sansui ranging in price from about £20 to £70 and offering a variety of facilities.

The choice of mixers is legion and can range from the four channel T.T.C.-B2005 distributed by Adastra at about £4 to the four channel Unimixer by Soundex at prices from about £90. Again, however, the constructor may opt to build his own mixer from a range of

published designs.

The choice of filter units is rather more restricted for domestic users but for those with about £200 to spend the Levers-Rich A501 offers full audio spectrum equalisation in seven bands.

THE TAPE RECORDER

As with the case of the mixers, the choice of tape recorder is extremely wide but can be narrowed down by application of a list of specific requirements which are desirable although not absolutely mandatory. In professional circles the production of a Synthesiser recording may entail a mix-down from perhaps as many as twenty-four discrete tracks. In these circumstances the final master tape will have been involved in only one tape to tape transfer and consequently the recorded sound is likely to have suffered only the very slightest loss in quality.

If the amateur attempts to simulate these professional methods (in terms of tracks involved) with the equipment normally available to him there may be as many as fifteen to twenty tape transfers involved in a very complex recording. Thus the quality of sound on the tapes involved in a number of transfers is likely to be much reduced in comparison with the original. This factor alone provides the first and foremost requirement of any tape recorder to be used for creative work and

that is quality.

In general terms the purchaser should invest in a machine which offers the best possible specification coupled with a minimum number of features designed to ease the problems entailed in the production of good quality recordings. These may be summarised as follows.

SPECIFICATION AND MACHINES

Three speed machines are preferable with 38cm/sec: 19cm/sec and 9.5cm/sec providing the ideal. Speeds below 9.5cm/sec are unlikely to be very useful. Three or four heads with provision for "on" and "off" tape monitoring. Stereo quarter track machines having sound-with-sound facilities, ideally through the medium of an input mixer, provide the minimum facilities for just about the only form of multitracking available to the average amateur.

There are a number of tape recorders on the market which combine an excellent technical specification with the features listed above. Amongst these are: Akai 1720L and GX220D; Beocord 1600; Ferrograph 702 and 704; Philips N4418 and N4450; Pye 9137; Rovex Series 77 1104; Sony TC630, and TC270; Telefunken M204TS; Uher Royal de Luxe.

As far as sound-with-sound or track transfer is concerned most of the recorders listed employ slightly different methods of providing this facility involving the use of one or more switches and, in some cases, an external signal routing cable. In this respect the Philips N4450 is notable in that, after laying down the first recording, this may then be transferred back and forth between tracks, with or without the addition of further material, by the manipulation of a single switch. This latter recorder also incorporates facilities for the addition of echo/reverberation again at the throw of a switch.

DECK CONVERSION

Recorders offering the facility of more than two tracks, or four tracks if one includes the quadraphonic types that are beginning to make their appearance, would normally be expected to be priced out of the reach of most amateurs. The determined constructor, however, could do worse than consider the conversion of some of the computer decks which become available from time to time. Several of the capstan driven decks now available from about £40 are superbly engineered devices and suitable for conversion. The speed of computer decks is usually quite high but this need not necessarily be a disadvantage since it would be used for record and replay of its own material only.

Heads are available from a number of sources. Gresham, for example, manufacture an eight-track, half-inch head for record/replay purposes at a price of around £70-£80. For the purposes of erasing, economies can be made by stacking a number of stereo erase heads. The electronics for the recorder could be based on the Hi-Fi Tape Link (recently published in this magazine) or, alternatively, on the range of Motorola integrated circuits designed specifically for tape record/replay purposes.

OTHER ADDITIONS

The addition of a Dolby B noise reduction unit should be a prime consideration and for a studio system it is best to choose a discrete unit rather than have the system built-in to a recorder.

Other useful additions to the studio include a separate power amplifier with speakers for monitoring purposes and a suitable turntable unit. These latter items could generally be "borrowed" from the domestic Hi-Fi system as and when required.

Next Month: The Meter Unit; Voltage Controlled Reverberation Amplifier and Ring Modulator.

NEWS BRIEFS

Telephone Scrambler for Commercial Use

For businessmen who require absolute security in their communication, EMI have introduced a new electronic scrambler for use with both private and public telephone networks. Known as the Privateer it is suitable for a wide number of applications in business, commercial and industrial organisations.

Once a telephone call has been set up in the conventional manner the Privateer telephone is brought into operation simply by pressing a separate switch. Subsequent conversation is then electronically scrambled as it is transmitted to the telephone line, and only becomes intelligible when it is unscrambled by matching equipment at the receiving end.

Commercial Radio

STUDIO automation is expected to be a major feature in many of the radio stations forming Britain's new commercial network which will start broadcasting within the next 12 months. The trend is being set by London's general entertainment channel Capital Radio, which is aiming for equipment which can automatically select and play a high proportion of the station's daily programme output from material stored on magnetic tapes.

Negotiations are going ahead with EMI to supply equipment from their extensive range of Schafer radio automation systems.

Equipped with a solid-state memory, this system is able if required, to provide a full 24 hours of continuous broadcasts—the memory being programmed to select on a time or sequential basis pre-recorded "events", including entertainment, commercials, and station announcements, from its vast repertoire. Using a keyboard, an operator can insert a complete day's programme in the memory, building up the programme on a minute by minute basis if required.

Data Transmission

Many different kinds of information, from heartbeats to motorway hazards, can now be transmitted over Britain's public telephone network in a new communications service introduced by the Post Office on June I.

During trials, the new Datel 400 service has already been used to send heartbeat readings from a small hospital to a large one a mile away, and information on river pollution has been sent to a Government laboratory. The latest development being studied at the moment is the sending of information on motorway hazards and weather conditions to a central monitoring point.

Golden Jubilee

N April the well-known components manufacturers, A. F. Bulgin & Co. Ltd., celebrated their 50th anniversary with an announcement of unbroken profit figures stretching back over the full 50 years to 1923 when Arthur Bulgin started the company on £100 capital.

It is perhaps indicative of the value of "service" that this organisation who hold as their aim to provide customers with a service through a family company, that they have managed to succeed so well in the face of growing and

aggressive international competition.

At the celebrations Arthur Bulgin, managing director, presented three new members of the company's 25 Year Club with gold watches including Albert Ridpath, foreman of the injection moulding section, who becomes the 50th member of this club. Quite a record for a company started on £100 and still running as a family concern.



NAND GATES

O avoid having to make, and use, three different types of circuit (AND, OR and NOT gates) to provide logic functions; the modern logic designer uses exclusively NAND gates. By suitable arrangement of these it is possible to generate any logic function that is desired and this type of gate is to be used throughout this series.

NAND VARIANTS

A NAND gate is made by inverting after an AND; the simplest possible circuit is shown in detail in Fig. 3.1 and depicts a basic form of DTL (Diode Transistor Logic). The DTL used in the Logic Tutor is only a minor variant of this circuit; do not confuse this with the basic TTL (Transistor Transistor Logic) NAND gate-Fig. 3.2-which, although it carries out the same function, has entirely different input gating and output stages.

Though NAND is a compound of AND with NOT it is so commonly used that it is considered to be yet another fundamental gate and as such has a standard symbol of its own (Fig. 3.3.). It is possible to obtain 1, 2, 3, 4 and 6 input NAND gates as standards and certain DTL gates have facility for the number of inputs to be expanded by adding extra

diodes at the expander node.

Select a two input NAND on the Logic Tutor and connect the inputs to two logic level switches monitoring the output level on one of the lamps. The output will be the inverse of AND as shown in the truth table (Q = A·B) which means that the output will always be level I until all the inputs are at level I and at that condition the output will fall to 0. Verify this for yourself and then do the same with a 4 input gate.

INVERT FROM NAND

We demonstrated INVERT last month, but now look at the truth table and Fig. 3.4 to see how this came about. Consider a gate with inputs A and B shorted together and the shorted node called X. If X is I both A and B will be 1-refer to the truth table and see that two ones on the input of a NAND give an output 0. If X is made 0 both A and B will be 0 and this is shown to give an output of I. Thus whatever X is the output is the opposite—an INVERT or NOT function.

Another way of obtaining INVERT is to make input B permanently a level I. Whatever A is, the output will be

the opposite.

A point worth noting is that in all forms of current sinking logic (DTL and TTL) a floating input to a NAND gate will be presumed (by the gate) to be at level I however it is bad practice to leave spare inputs of a gate disconnected as they are liable to pick up stray signals and cause severe logic ambiguities.

AND FROM NAND

If the NAND function is obtained by inverting AND we can get back to AND by inverting again after a NAND gate as shown in Fig. 3.5. The output is A·B inverted—the double negate cancels and we are left with simply A.B.

As an exercise try and simulate a six input AND using any of the NAND gates available on the Logic Tutor (expanders are not permitted!). We shall publish an answer next month and at the same time describe De Morgan's Theorem and show how to use this in conjunction with NAND logic to produce OR.

by M. Hughes

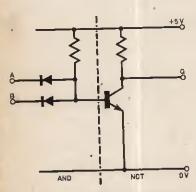


Fig. 3.1. Basic DTL NAND gate

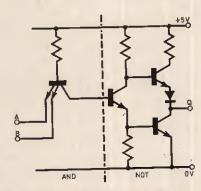


Fig. 3.2. Basic TTL NAND gate

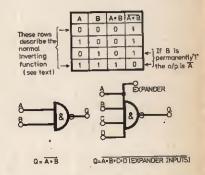


Fig. 3.3. Two forms of NAND gate together with a truth table

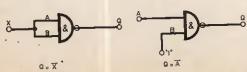


Fig. 3.4. Two ways of getting INVERT from NAND

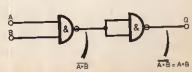
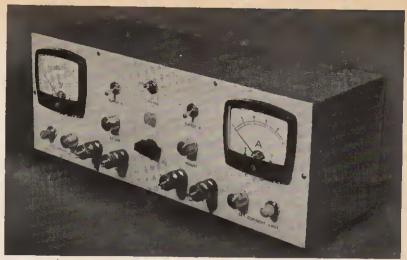


Fig. 3.5. The AND function from NAND gates



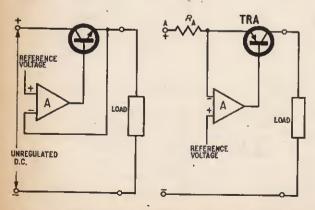
TWIN POWERS

By G. S. SCHAJER

HE USE of integrated circuits has made it possible to make a high quality regulated supply incorporating a current limit facility using a surprisingly small number of components. Since the supplies are so simple and inexpensive, a twin power supply unit is a practical possibility, and is extremely useful especially in development work.

The current limit facility is one which is not found except in the most expensive power supply units, and it will be found to be extremely useful-sometimes in rather unexpected ways. Apart from its use as an overload protection and as a constant current source, the author has found it to be an excellent way of testing Zener diodes!

The current limit facility is nearly as accurate as the voltage regulator and, if required, the unit may be used as a constant current source.



regulator circuit

Fig. 1. The basic voltage Fig. 2. The basic current limiting circuit

CIRCUIT DESCRIPTION

The integrated circuits used are the 741 operational amplifiers (op amps) which are cheap and readily The operational amplifier differs from the ordinary amplifier in as much as that it has two inputs, an inverting one (marked with a minus sign) and a non-inverting one (marked with a plus sign).

If there is a small difference between the voltages applied to these inputs, the output will change according to the magnitude and polarity of the difference. The op amp may therefore be used as a difference amplifier. This is the case in the circuit shown in Fig. 1, which shows a voltage follower circuit.

The output of the op amp is controlling a transistor in the supply rail of an unregulated supply. It is comparing a reference voltage and the output voltage and amplifying the difference. If the output voltage falls, the input to the transistor is increased, tending to restore the output to its original value. Similarly an increase in output will cause less current to the transistor base and thus decrease the output.

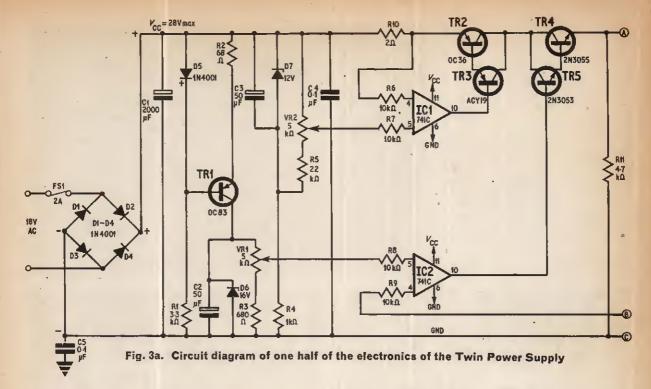
Since the op amp has a very high gain, the input need only change by a small amount for it to produce a substantial restorative action. The output thus follows the reference voltage with a high degree of accuracy.

CURRENT LIMITING

Fig. 2 shows a basic current limiting circuit. It functions in exactly the same way as the voltage regulator except that instead of the output voltage, the voltage drop across RA (and thus the current) is compared with the reference.

The current at which the circuit limits is dependent on the reference voltage and the value of the resistor according to the relationship:

Current limit = Reference Voltage



SPECIFICATION

Two identical supplies with complete isolation Fully variable output voltage 2 to 15V Fully variable current limit 35mA to 1A Ripple at 15V, 1A 0.4mV Voltage regulation 0.1%

If the circuit is operated in the region of current limiting it makes a good constant current source. Its accuracy is nearly as good as that of the voltage follower. The slight deterioration is due to the thermal instability of the resistor.

FULL CIRCUIT DIAGRAM

The voltage follower and the current source may be used together as shown in the full circuit diagram, Fig. 3a. The reference voltages are made adjustable so that the output voltage and the current limit may be adjusted.

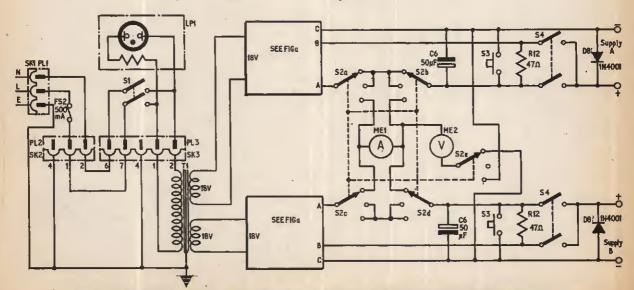


Fig. 3b. Mains input and switching arrangement for the complete Twin Power Supply

COMPONENTS ...

Resistors

3-3kΩ **R1** R₂ 68Ω R3 680Ω R4 1kΩ

R5 $22k\Omega$

 $10k\Omega$ (4 off) 2Ω 10W wirewound R6-R9 R10

R11 4.7kΩ R12 47Ω

All ±10% ‡W carbon except R10

Potentiometers

VR1 5kΩ linear VR2 5kΩ linear

Capacitors

C1 2,000 µF 50 V elect.

C2 50µF 25V elect. 50μF 15V elect. C3

C4 0-1 μF 250V polyester C5 0-1 μF 250V polyester

C6 50µF 25V elect.

Integrated Circuits IC1, IC2 741C (2 off)

Transistors

TR1 OC83 TR2 OC36 TR3 ACY19 TR4 2N3055 TR5 2N3053

Diodes

D1-4 1N4001 (4 off) or 2A bridge rectifier D5 1 N4001 D₆ 16V 400mW Zener

D7 12V 400mW Zener

DB 1N4001

Switches

S1 D.P.D.T. mains toggle

S2 5 pole 3 way, make before break, rotary wafer Push to make release to break, pushbutton

S4 D.P.D.T. toggle

All the above components are for one half of the Twin Power Supply except S1 and S2, only one each of these being required

Miscellaneous

Douglas MT127 0-24-30-40-48-60V 2A T1 -ME1

1A f.s.d. 3in meter

ME2 20V f.s.d. 3in meter (or 15V if available) PL1, SK1 Bulgin P360 mains plug and socket

PL2, SK2 PL3, SK3 B7G plug and socket B7G plug and socket

LP1 Mains neon with integral resistor FS1 2A fuse and holder (2 off)

FS₂ 500mA fuse and holder

Terminals (4 off)

0.15in matrix edge connectors and Veroboard

Case, plastic feet, paxolin panel, Veroboard, printed circuit board, power transistor insulating set (4 off)

A germanium transistor (TR1) biased by a silicon diode (D5) is used as a constant current source to feed a Zener diode (D6) thus giving a very stable reference voltage. Since the output voltage cannot be taken below 2V a resistor, R3, is connected in series with the potentiometer VR1 to make the minimum reference voltage also 2V.

The current limit reference Zener diode (D7) does not need similar stabilisation as it only operates when the load current is constant and hence the rail voltage is not changing. A maximum of 2V is required for this voltage and this is obtained from a 12V Zener (D7) and a resistor R5 in series with VR2 to produce the required voltage at the potentiometer wiper.

The outputs of the op amps do not run directly to the output transistors but via driver transistors in the super alpha pair configuration. This prevents overloading of the op amp output which can only supply 25mA. The 741 has indefinite short circuit protection which is comforting during testing.

Under extreme conditions one or other of the output transistors is dissipating nearly the full power input of the supply from the transformer, i.e. about 20W. The need for heatsinking is obvious and should be taken into account when deciding the output power required.

OUTPUT VOLTAGE MONITORING

The connection from the inverting input of IC2 is connected directly to S4 (Fig. 3b) so that the voltage drop in meters, switches and wiring does not affect the regulation of the supply. When S4 is closed the input to the op amp is connected to the output terminal and so any voltage drop across the switch due to contact resistance will similarly not affect the regulation.

When the switch is off the supply is disconnected from the output but the high impedance op amp input is still connected to the supply by R12. This enables the supply to operate so that the voltage and current limit may be set when S4 is off. The current limit set switch S3 is independent of S4.

A reverse biased diode D8 is connected across the output terminals to protect the supply from back e.m.f. when disconnecting a reactive load.

Resistor R11 is included so that the supply has a load in the absence of any external load. C6, protects against high frequency oscillation and C5 reduces mains borne interference.

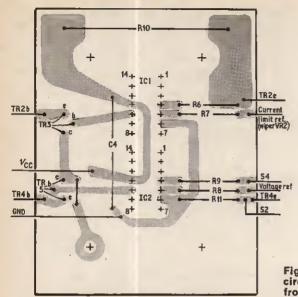
THERMAL EFFECTS

The 2Ω resistor in the current limiting circuit should be a high wattage type to prevent excessive heating which causes resistance variations. A 10W type drifts by about five per cent which is reasonable; a higher wattage should be used for greater stability.

On the prototype the rail voltage off load was 28V. Under no circumstances should this voltage be exceeded by using a higher voltage transformer as this will damage the integrated circuit. On a 1A load the rail voltage drops to 21V,

Meters are provided to measure the voltage and current output of the two supplies. It is possible to switch them from one supply to the other, but without breaking the continuity of either supply, and yet maintaining the isolation between the supplies.

The circuit shows the switching arrangement required. . A five pole, three way, make before break switch is The centre position is "off". A make-beforebreak switch is essential, otherwise there will be a break in continuity during switching.



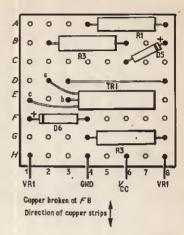


Fig. 4. Printed circuit board carrying the integrated circuits and associated components shown full size from copper side. Also shown is the Veroboard carrying the Zener diode circuit

CONSTRUCTION

The method of construction is a matter of personal choice, dictated to some extent by the materials available. Some thought should be put into planning the layout of the components and a little extra care when making up the unit is well worth the effort.

Some details are given of the construction of the prototype, together with some of the points to remember when building the supply. They are given merely as a guide and need not be adhered to rigidly.

The integrated circuits and associated components are mounted on a printed circuit board which is shown in Fig. 4. The two rows of holes for the i.c.s are easily drilled by clamping a small piece of Veroboard (0·lin matrix) to the printed circuit board and using its holes as a template. Holders for the i.c.s may be used if desired.

The Zener diodes and associated components are mounted on the front panel, attached to the tags of the potentiometers. The stabilisation circuitry for the

Zener diode D6 is made on a small piece of Veroboard (0·15in matrix) which is mounted on the back of VR1 by means of its connecting wires. If the potentiometer has a metal case steps should be taken to prevent the Veroboard shorting.

The layout of the front panel is shown in the photograph (Fig. 8). This was made from a piece of white Perspex Thin thick. Detailed drilling details are not given as some of the components will vary according to availability. The meters shown are not generally available but SEW series 65 meters are a suitable type to use.

It is a good idea to connect the spindles of potentiometers VR1 and VR2 to the negative rails of the supplies as this reduces hum pickup. Use thick wire for the current carrying sections of the supply to reduce heating effects.

TRANSFORMER

The power is supplied by a single transformer with two isolated 18V, 2A secondary windings. This is not

Photograph of the completed Twin Power Supply



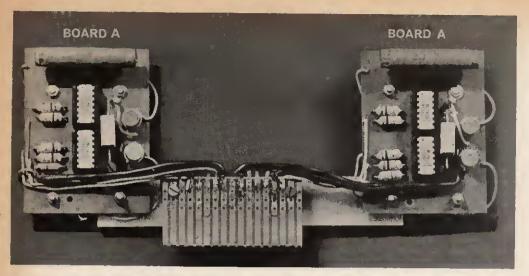


Fig. 5. Layout of the components on the back panel. Details of board A are given in Fig. 4. TR2 and TR4 are mounted under the boards on the panel

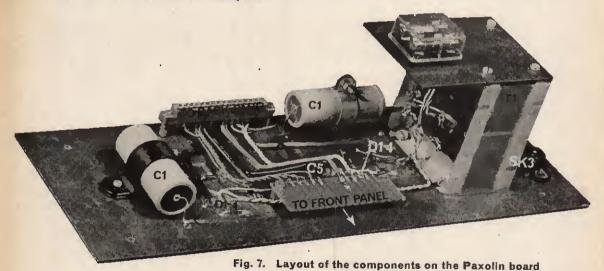


Fig. 8. Layout of the components on the front panel. Board B is shown in Fig. 4

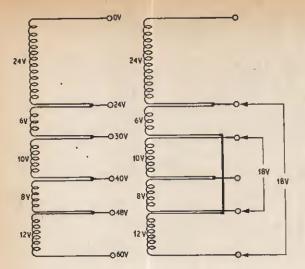


Fig. 6. Diagram showing modification to the Douglas MT127 transformer

available in this form but a simple modification to a Douglas MT127 will give the required result. The MT127 has taps at 0-24-30-40-48-60 volts and these are connected to tags.

The connections to the tags marked 30V and 48V should be carefully unsoldered and the end of the 30V winding should be connected to the beginning of the 48V winding. This is shown diagrammatically in Fig. 6.

One 18V supply is available at the tags of the 30V and 48V and the other at the tags of 24V and 60V. Check that the correct windings are being used with a continuity meter and make sure that there is complete isolation between the two windings; Alternatively two 18V, 2A transformers may be used.

PAXOLIN PANEL

The transformer, rectifiers and capacitors are mounted on a Paxolin panel which fits into the bottom of the case, Fig. 7. Discrete diodes soldered to squares of stick-down copper sheet form the bridge rectifiers but it is probably more convenient to use a ready-made 2A bridge.

The back panel is a piece of 14 s.w.g. aluminium painted matt black and this acts as a heatsink for the four power transistors, Fig. 5. The p.c.b.s carrying the i.c.s are also mounted on this panel. Mica washers to insulate the transistors as well as plastic bushes on the bolts should be used to ensure that the transistors do not make contact with the panel.

On the prototype, both the front and back panels were made removable by using offcuts of Veroboard and edge connectors to form plugs and sockets. The mains input is kept completely separate by using B7G plugs and sockets to make the connections between the front and back panels of the transformer.

TESTING THE SUPPLY

Should the supply not function check the reference voltages as any fault here will be reflected in the output. If they are alright then the fault must lie in the op amps or associated components.

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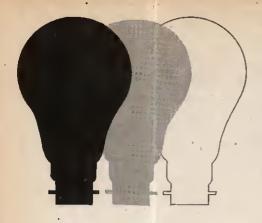
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CONTROL UNIT By G. A. Davies

ANYONE who has helped to organise a discotheque group performance or even put lights on a Christmas tree, must have felt the need for a simple flashing light controller. Many circuits, perhaps using a multivibrator, exist. But these are severely limited in their flashing rates if on and off periods are kept approximately equal, and if strobe-like effects and very slow changes are required from the same unit range switching will be needed, leading to expensive and complex wiring and tedious operation.

The circuit described here uses a similar method but a quite different principle to produce a variety of effects all available from settings of a single control. In place of continual OFF/ON switching, gradual changes enhance the effects, which range from a fast strobe to

flashes lasting up to and over one minute.

COSTS

The cost of the unit if all parts including case are purchased new will be a little over £2.70. By making the case and shopping around this may be reduced. An experimental set-up may be made from common items in the spares box with the possible exception of the thyristor.

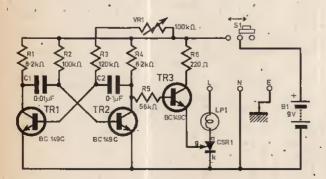


Fig. 1. The circuit diagram of the light controller. TR1/TR2 form a multivibrator driving CSR1 through driver TR3

GENERAL

The circuit of Fig. 1 shows a multivibrator, TR1/TR2 actuating a driver circuit TR3 which in turn actuates a thyristor CSR1.

The components of the multivibrator are selected to give a repetition rate of about 50Hz with TR1 in the on state for about 1ms and TR2 on for between 10 and 25ms dependent on the setting of VR1.

THE DRIVER CIRCUIT

Output pulses from the multivibrator are fed to the base of TR3 via R5 which is large enough to prevent the following circuit from interfering with the multivibrator. Each pulse turns TR3 on, allowing an amplified pulse of current through the collector into the gate of the thyristor. R6 limits this pulse to about 30mA which should be sufficient to operate most thyristors. If a low current thyristor is to be used some saving in battery drain can be made by increasing the value of R6 to 390 Ω .

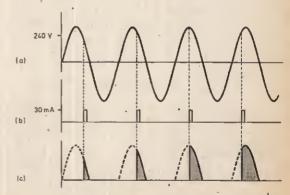
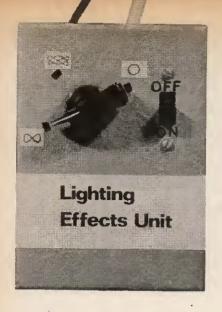


Fig. 2. Graphic representation of the relationship between mains cycles, control pulses and load current, at a given multivibrator frequency



reached when the pulse has finished before the cycle The power then falls to zero until the next pulse arrives at the end of the cycle and power increases slowly again.

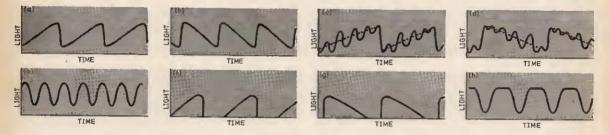
If a 100W bulb is used as the load the light will be seen to change intensity as in Fig. 3 (f), the rate of change depending on the frequency difference between the mains and multivibrator pulses. The reversed version, Fig. 3 (g), is obtained by setting the pulse frequency lower than 50Hz.

With large frequency differences a strobe effect is obtained and with careful adjustments an almost constant light output can be set. It must be borne in mind that the maximum light output is only half that given by the bulb connected straight to the mains.

The other effects shown in Fig. 3 can be obtained with the controller remembering that thermal inertia, i.e. the time taken for the lamp filament to heat up and cool down, will slow down some of the faster changes.

CONSTRUCTION

To the inexperienced constructor it is impossible to overstress the need for great care over safety. The whole circuit is connected to the mains, and although



rig. 3. Variations in light output obtainable with the circuit of Fig. 1

THE THYRISTOR

This semiconductor device can be considered as a combination of a rectifier and a latching relay. It will not under any circumstances (other than breakdown) pass a current from cathode to anode, and also it will not pass a current from anode to cathode unless a sufficiently large current is supplied to the gate. Once triggered in this way, however, the anode current will not stop, even after the gate triggering current has stopped, until the anode voltage is reduced to zero, or changes sign to negative if a.c. is applied.

In this particular application a.c. mains voltage is connected to the anode through the controlled load. When the mains is negative no current can pass due to rectifying action. Thus any current pulses arriving

at the gate from TR3 are ignored.

When the mains is positive the thyristor can conduct, but will not do so until a trigger pulse has turned it on:

Fig. 2 (a) shows the mains input voltage. For a given frequency of the multivibrator (b) is the train of current pulses in the gate of the thyristor. All negative half cycles of the a.c. are rectified out and positive half cycles are let through only after an input pulse, so that the voltage supplied to the load looks like Fig. 3 (c).

If the multivibrator is set as shown at a slightly higher frequency than the mains 50Hz, that part of the half cycle allowed through will change. As the drive pulse comes earlier in each cycle, the power in the load slowly rises to a maximum of about half the rated power (all negative half cycles are lost) until a point is the live is isolated through the lamp and is thus proof against short circuits, great care must be taken over insulation. The isolation of the case from the thyristor poses the greatest problem here and the solution depends upon the component chosen.

The prototype was housed in an aluminium box measuring about 4in × 23in × 13in deep and fitted with an overlapping lid held by a self-tapping screw at either end. The switch and the potentiometer are fixed to the lid and the remaining parts are assembled in the box. Firstly the top and bottom should be drilled to accept the components.

· A small aluminium clip shown in Fig. 4 is needed to secure the battery with small blocks of polystyrene stuck to the lid and sides to hold it firmly in place. The clip can be fixed with Araldite.

The circuit board and terminal block are secured by in 6 B.A. tapped pillars screwed or Araldited (preferably both, since the screws will have to be very short) to the case. Countersunk screws are used in the prototype, the heads being filled in with Araldite and painted over but this is of course a matter of individual

COMPONENT BOARD

Details of this are given in Fig. 4. The six 6 B.A. clearance holes should be drilled carefully first and then the components pushed through and soldered, the transistors being soldered last to avoid overheating.



LIGHTING CONTROL UNIT

COMPONENTS . .

Resistors
* R1 * 8-2kΩ
R2 100kΩ R3 120kΩ R4 : 8-2kΩ

220Ω

All IW 10% carbon

Potentiometer

VR1 100kΩ carbon linear

Capacitors

250V poly.

C1 0.01µF C2 0.1µF 250V poly.

Thyristor CSR1 BO226 (see text)

Transistors

TR1, 2, 3 BC149C (or BC169C, BC108C, BC109C)

Miscellaneous

S1 d.p.d.t. slider switch or on/off toggle
B1 9V PP3 or equivalent

2in x 3½in s.r.b.p. 0-1in matrix, 2A terminal block (5 sections required), 6 ½in 6BA tapped pillars, Case see text, 6BA screws, 18swg aluminium for battery clip and heatsink, grommets, wire, solder, sleeving.

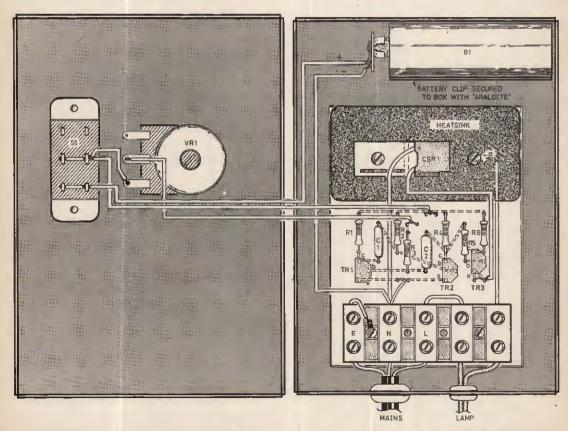


Fig. 4. General layout of the controller showing main wiring board and component positioning. NOTE: Ensure that the heatsink is well isolated from case

Four loops of wire are made on the component side so that connections can be made. Next screw two in pillars to the inside holes; the board can then be

bolted down onto the pillars.

The heatsink in Fig. 4 is mounted on the board using two further pillars, the thyristor being fixed down under one screw (a smear of silicone grease will be helpful here under the tab to improve heat transfer). The lead from the terminal block can be soldered to, or clamped by the other screw/pillar to act as the anode connection. Make certain the heat-sink mounting is insulated from the case.

The wiring can be completed with p.v.c. insulated stranded wire, colours being chosen to suit. Sleeving the thyristor leads is advisable or tape may be used.

COMPONENTS

The components specified in the list are all quite easily obtainable. The transistors can be any of the listed types, BC 149C being most suited to the mode of construction used and also the cheapest.

If surplus transistors are tried TR3 should be selected for low leakage and high gain. The C suffix on the

listed types ensures a gain of over 400.

The thyristor will depend on the application; for mains it must be rated at 400V peak forward and reverse voltage, the current will depend upon the power of the lamps used. At 240V a 3A thyristor with a good heatsink will take a total load of 1000W and give 500W of light. A 0-8A version in a TO5 or similar case will overcome the insulation problem and make the unit possibly smaller but will not take more than 250 to 275W load and give 100 to 150 watts of light.

The 2in × 1in panel used cannot be considered a good heatsink and also it is inside a sealed box. In addition large powered bulbs also complicate things because their high thermal inertia causes large surges on switching on. The unit should be able to take a 500W lamp at all control settings to give 250W of light. With lower powered lamps in parallel it should be possible to go up to 700 to 800W. However, due to the cost of such a setup this has not been tried. A few hundred watts will be found adequate for most purposes but care should be taken to ensure there is no overheating with the components specified.

TESTING

When the wiring is complete and checked, connect the battery in circuit and, using a meter set to a current range, put this across the switch. A current of about 3mA should register, switch the range down if necessary to read this. Rotate VR1 and the current should

change slightly.

Following this or, alternatively if no meter is available, connect the mains lead to a plug fitted with a 3A or 5A fuse and the output lead to a lamp of 40 to 100W power. (In the prototype a cable socket was connected so that lamps could be plugged in and removed freely.) Switch the mains on. If nothing happens all is well; now switch on the unit with S1. The bulb should now light and flash in a pattern depending upon the setting of VR1. Explore the range of the potentiometer and see that it is similar to that indicated in Fig. 4. If this is not so, adjustment of R3 may be necessary.

LAMP HOUSING

For lighting effects the lamps will have to be fitted in some way to achieve the desired effect. Coloured pearl bulbs or bulbs fitted in boxes with coloured filters can be used but these give very directed illumination and an improvement is made with more diffuse lighting. Frosted screens can be tried but an excellent result is given by a box with lamps, reflectors and filters placed at the bottom of a wall so that the light is projected up onto the wall. Car headlamp reflectors can also be adapted to produce a focused beam for special effects.

FAULT FINDING

Three conditions will indicate faults in the unit which might occur.

Firstly the lamp may not light. Check the operation of the multivibrator with a high impedance earpiece or an oscilloscope after first unplugging the unit from the mains and putting S1 to on. With one side of the earpiece connected to battery negative look for a signal at 40 to 100Hz at TR2 collector. If this is present check for the same at the gate of CSR1. If the signal is also present here the thyristor or the mains wiring is at fault. Look for breaks in the relevant part of the circuit or for faulty components.

Secondly if the lamp lights continually at half brilliance then check for a signal at TR2 collector with a voltmeter. If this reads about 1V then check TR3 and its wiring. If it is 2V or more the multivibrator is the cause of the trouble. Check for correct positioning

of C1 and C2.

Finally if the lamp lights at its normal full brilliance this indicates a short circuit in the unit. If none can be found then the thyristor may be shorted. Check that the fuse and bulb are not also shorted before replacing.

THE FINISH

Finishing the unit is left entirely to the constructor, the prototype was painted with a silver "hammer" paint and the words "LIGHTING EFFECTS UNIT" painted in black across a white stripe. The "critical settings" of the control were marked approximately with Lissajous figures corresponding to the ratio of the mains and multivibrator frequencies. Any method of calibration will suffice and as long as you can understand it that is all that really matters.

SUPPRESSION

If the unit interferes with radios or TVs on high power operation a suppressor may have to be fitted. This will consist of inductors in series and/or capacitors in parallel with the power lines. This was not included in the prototype because of space consideration and since no interference has arisen anyway.

USE

Every constructor will rapidly become accustomed to the effects produced by the unit and will soon know how to use its effects to the greatest advantage. Positioning and colouring will follow logically to give the best results.

A word of warning, however. The strobe effects produced should not be used as dominant lighting in an otherwise darkened room as it can have a bad effect on the subconscious mind after prolonged exposure. Headaches can also be caused, probably due to the sensation of rapid blinking. If strobes are used make sure that the other lighting is at least equal in strength to the strobed light.

Finally, when fault finding as described earlier make certain the equipment is not connected to the mains.

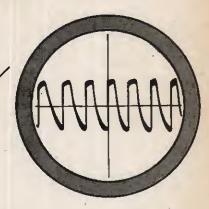
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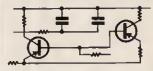
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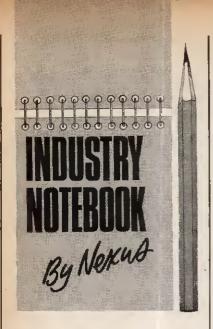
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PACE SETTERS

This month I tip two private companies worth watching for future growth. First is Sinclair Radionics currently moving into a strong position with nearly £2 million turnover and a confident forecast of at least £3 million next year. Clive Sinclair at the age of 32 still has the great bulk of his working life ahead of him and now he has the sort of cash flow needed to finance the development of new projects.

The Sinclair "Executive" pocket calculator has proved itself to be the big business breakthrough he had hoped for. It came in a higher price bracket than his previous products and quickly established the volume of production that pushed his company forward, almost overnight, into a new and exciting situation. The "Executive" is now rolling out of final test at the rate of 10,000 per month and 40 per cent of production goes

overseas.
Sinclair is investing a lot of his additional turnover in R and D. Expect to see in the next few months or so some new calculators, a range of test instruments to follow up the recently introduced digital multimeter, a pocket-sized TV set, a digital watch and something exciting in digital computers. But he is not going to neglect his army of faithful followers in the audio and hi-fi sector. There will be something new for them too.

Sinclair is clearly widening his market in the consumer sector and starting to move into the professional area, initially through instrumentation. There's nothing that succeeds like success.

My number two tip is Jermyn Manufacturing led by managing director Graham Rattcliff. I last saw him towards the end of April when he was running one of a series of seminars promoting the use of heat pipes, an old idea which has now won a new lease of life as an efficient heat dissipator in modern equipment.

The company's main line is in moulded components and heat pipes are comparatively new. Nonetheless, Rattcliff sees them as an important new line although he confesses that more people want "specials" than from the standard range—always a manufacturer's headache.

I tip both Sinclair Radionics and Jermyn Manufacturing for big growth not only because the products are right but also because both companies are strong on marketing. Sinclair has just established a New York office. Jermyn has sales offices in Munich, Paris and San Francisco.

DISTRIBUTOR NEWS

Component distributors are looking distinctly more cheerful these days now that orders are flowing in faster than supplies rather than the other way round.

Farnell Electronics, among Britain's biggest, has broken through the £1 million pre-tax profit barrier for the first time. Last month SASCO celebrated, its 21st Birthday with

press advertising wishing founder George ("Stewie") Stewart best wishes.

John Zuber, new chairman and managing director of Guest International has opened a Distribution Division which, if it works as well as his computerised stock control and order system—about the best I've seen—should prosper exceedingly well. And Swift Hardman's new Industrial Electronics and Automation Division, now under the control of ex-TI man Peter Linforth, is aiming at an initial turnover of £1.5 million and was on target when I visited the company a few weeks ago. New franchises include Greenpar connectors), Beswick (fuses), and OKI (opto-electronics) bringing the franchise total to 20 in electronics plus 16 in the electromechanical sector.

it was rumoured that Motorola Semiconductors might hire or fire one or two distributors this year but the five old faithfuls (Jermyn Distribution, Celdis, Semicomps, A.M. Lock and GDS) all got a nice pat on the back from vice-president Tom Connors at the annual re-appointment ceremony and there were no

new appointments.

Leicester-based Atlantic have won a valuable new franchise from National Semiconductors and this should give a big boost to one of our newest distributor companies.

SELLING SERVICES

You don't necessarily have to manufacture or sell goods to make money. Quite a few companies make a good living selling services. Companies like International Aeradio and Cable and Wireless have been doing it for years. A division within Plessey has substantial business on advising and supervising warship re-fits. This outfit works in conjunction with shipyards and although you don't get orders on a daily or even weekly basis, when they come they're big.

Then there is EASAMS which, if my memory serves me correctly, started off in life in 1962 with the name being an acronym for Elliott Automation Space and Military Systems. Since those days there has been some diversification to include consultancy and planning projects in all sorts of fields such as hospitals, transportation and even, recently, an investigation on how best to integrate the coconut industry in Malaya.

FIRST AWARD

EASAMS is the first consultancy firm to win the Queen's Award to Industry for exports with overseas contracts up 300 per cent since 1970. But one of EASAMS toughest jobs must still be the management of the integration of the electronic systems on the European Multi-Role Combat Aircraft (MRCA).

Another unusual Queen's Award winner, again for exports, is fouryear-old Racal-Milgo. I write 'unusual' because this company is 50 per cent owned by the Milgo Corporation U.S.A. and I'm wondering (jokinglyl) whether Milgo Corp. will be flying the flag and wearing the ties equally with their British partners. Although the equipment comes from the U.S.A. it's the British end of the partnership that provides the international consultancy service. feasibility studies, planning and implementation of some of the most complex data transmission networks in the world today.

Our congratulations go to EASAMS and Racal-Milgo together with English Electric Valve Company who got the only award in the electronics industry for technological innovation, the product being the ceramic hydrogen

thyratron.

MOBILE RADIO RATIONALISATION

The long war of attrition in the twoway mobile radio market with lots of little fellows trying to compete with one big one appears to be over following the announcement that Cossor's mobile business has been acquired by Burndept Electronics. This new move follows Burndept's acquisition of the Ultra business in mobile radiotelephones last year, and we must now assume that Burndept will be in a position to seriously start challenging Pye Telecommunications who hold well over 50 per cent of the market.

PATENTS BEVIEW...

WATER LEVEL ALARM

BP1 271 654

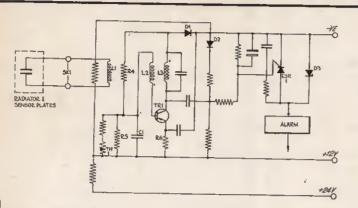


Fig. 1

In BP 1 271 654 Smiths Industries Limited describes circuitry for using an oscillator to sense the loss of cooling water from a car radiator.

A tuned collector oscillator centres round transistor TR1 which has its emitter connected via resistor R6 to a 12 or 24V positive line. The transistor collector connects with the negative line through a transformer winding L3 and diode D1.

The transformer has two further windings L1 and L2, of which winding L2 is connected between the base of TR1 and the junction of R4, R5 and C1. Two electrodes mounted inside the car radiator are connected to L1 through the input socket SK1.

The transformer windings have a turns ratio such that the effective impedance in parallel with L3 is too high to adversely affect oscillation, hence the oscillator will oscillate at audio frequency. But when the radiator water is high enough to short the two electrodes together, the effective impedance in parallel with L3 will drop and so damp down and stop oscillation.

By feeding the oscillator output via a rectification circuit to an alarm system it is possible to sound an alarm or trigger a light when the water level falls and the undamped oscillator starts to produce an audio

frequency oscillation.

Various temperature sensitive components, such as a thermistor TH1, provide temperature compensation for the oscillator. This is obviously necessary bearing in mind the high temperature levels encountered in running vehicle engines.

AMPLIFIER PROTECTION

The Marconi Company Limited in BP 1285 685 has some simple circuitry to offer which, according to some fairly complicated theory, provides protection for transistors against high voltages resulting from generated harmonics.

Fig. 1 shows a well known A-B push-pull wide band amplifier for high frequency working. Transistors TR1 and TR2 are fed by input transformer T1 with a centre-tapped secondary and feed the centre-tapped primary of an output transformer T2. (The circuit is simplified for clarity.)

The amplifier will produce a comparatively high level third harmonic and to eliminate output power at the frequency of this harmonic a filter network (VC1, 2 and L1) is fed via a coaxial cable, the final output being fed to the output terminal. Assuming an applied voltage of 28V, with perfect coupling the voltage on the transistor should never exceed 56 volts. But nothing in this world being perfect, third harmonics produce excessive voltage levels.

The coaxial cable disturbs the ideal phase relationship between the fundamental frequency wave and the third harmonic wave and causes the harmonic to increase the voltage swing in the collector voltage rather than reduce it.

The Marconi invention centres round the provision of a voltage limiting means for modifying the

phase of the third harmonic frequency component of the voltage present at the transistor collectors, in relation to the fundamental frequency of the voltage there. But there is no change in amplitude of the fundamental frequency component. Usually a diode clipper is used to limit the voltage wave at the transistor collector so as to produce a flattopped voltage wave limited to a maximum level such that the fundamental is virtually unaffected by the clipping.

Circuit Fig. 2 shows their novel circuit. This generally resembles the Fig. 1 circuit except that diodes D1 and D2 are used with a d.c. blocking capacitor C1, the diodes receive a d.c. voltage of 60 volts. As a result, the voltage wave form at the collectors of the transistors have a flat top limited to 60 volts.

Although the diodes appear to function as an ordinary amplitude clipper, in fact they operate to clip to a clipping level and merely modify the phase of the third harmonic. Fig. 3a shows the fundamental Y and third harmonic Z which together make up curve X of Fig. 3b. Thus there is no distortion and the clipping is linear.

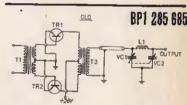


Fig. 1

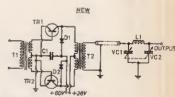
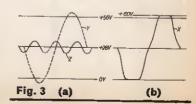
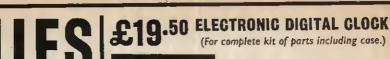


Fig. 2



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301 301A 301A 301A

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AC126	14p	BC142	88p	BF329	18p	OC45	14p	Rectific	
AC127	18p	BC143	83p	BF890	18p	OC70	28p		
AC128	18p	BC144	80p	BF390	87p	OC71	14p	1N914	8p
AC142K		BC145	28p	BFX84	28p	OC72	140	1N016	8p
AC141K	200	BCI47	9 p	BFX85	85p	OC81	14p	1N4148	8p
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AC187	13p	BC149	9p	BFX87	282	OC84	28p	1N4007	220
AC187K		BC153	16p	BFX88	26p	TIP29A		16113	179
AC188	13p	BC154	17p	BFY50	21p	TIP30A		18120	17p
AC188K		BC157	18p	BFY51	17p	TIP31A		18121	15p
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ACY18	21p	BC159	140	BFY64	89p	TIP33 &	1.82	18191	18p
ACY19	25p	BC167	180	BFY90	720	TIP34A		18132	
ACY20	220	BC168	110	BSX20	100		1.54	18920 18922	8p
ACY21	28p	BC169	11p	C407	22p	TIP35A	4 40	18923	18p
ACY22	18p	BC177	150	C426	880		2.53	18940	9b
ACY39	68p	BC179	15p 9p	C428	81p	TIP36A	8-19	AAll9	115
AD140	44p	BC182L		C450		TIP4LA		AA129	119
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AD161	290	BC213L	11p	MP8122		2N1131	220	BA100	18p
AD162	29p	BC214L	îÎp	MP8123		2N1132	280	BA102	25p
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AF114	14p	BC259	95	NKT21		2N1711	26p	BA144	145
AF115	14p	BC267	140	NKT21		2N2904	40n	BA145	22p
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A8726	319	BD123	66p	NKT61		2N3663	57p	OA5	18p
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BC108	97	BD1414		NKT77		2N8707	8p	QA73	11p
BC109	θр	BD142	50p	0019	66p	2N3708	9p	OA79	80
	1519	BF159	88p	0020	öğp	2N3709	9p		
BC116	169	BF178	29p	0023	38p	2N3710	9p	OA81	9p
BC125	16p	BF177	28p	OC25 OC28	289	2N3711	8p	OA85	llp
BC126	25p	BF178 BF179	29p 35p		38p	2N3794 2N3819	17p 28p	OA90	8p
BC132 BC134	16p	BF194	150	OC29 OC35	88p	40361	50p	OA91	8p
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MARKET PLACE

items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

SLIDE RULE CALCULATOR

Electronic calculators of the pocket variety fall generally into two types. The first is the fairly simple device the operation of which is limited to the more normal mathematical operations, and the price of which is steadily falling to around the £40 mark.

The alternative and much more sophisticated device is capable of carrying out complex maths functions such as rooting or calculating to specific laws. Currently there are only one or two of these latter and their price tends to rest in the £200 region.

Predictably, just as the cost of the smaller version has dropped swiftly over the last two years, the more complex device is now showing reduction tendencies:

Thus Texas Instruments have just announced a pocket "slide-rule" type of instrument capable of calculating reciprocals, squares and roots, true credit balance and all the more usual functions.

It will also convert an arithmetic entry to scientific notation when the entry exceeds eight digits.

Capable of handling positive and negative numbers over a 200 decade range, the SR-10 is selling at £87.50.

DIGITAL MULTIMETER

Another contender in the low cost digital multimeter field has just been announced by Advance. Called the Alpha, it will retail at £55, a price which makes it a real challenge to the analogue meter as well as other digital multimeters.

The Alpha features a total of 24 ranges enabling it to measure d.c. voltages from 1mV to 1kV, a.c. voltages from 1mV to 500V, d.c. and a.c. current from 0·1μV to 1A and resistance from 0·1Ω to 10MΩ. The reading is given on a three digit l.e.d. display which can be adjusted in brightness to conserve power. Accuracy is claimed to be 0·2% ±1 digit on the 1V d.c. range and 0·5% ±1 digit on other d.c. voltage ranges. Input impedance is a high 10MΩ and the instrument can stand overloads of 350V on the 1V range and 1kV on other voltage ranges.

The unit consumes 90mW to 400mW dependent on the setting of the brightness control. It is normally supplied by a 9V dry battery which will last several months in normal use. There is also an optional mains unit at £7 extra and a rechargeable battery unit at £17 extra.

The circuit uses a simple single ramp analogue-to-digital conversion system and all the digital circuitry is contained in a single m.o.s. l.s.i. integrated circuit, custom-designed

by Advance.

The unit is available direct from the maker, Advance Electronics Ltd., Raynham Road, Bishop's Stortford. Hertfordshire.

WORKSHOP

Several items for the workshop seem to be worth further consideration this month.

For the work bench there's a new adjustable vice from W. Greenwood Electronic. Called the Oryx, the vice is ideal for holding circuit boards during soldering.

The vice has a rotating base which is also part of a steel ball joint in the bench clamp. This arrangement allows the user to position his workpiece in practically any position and the fibre faced jaws of the vice will hold delicate components without damage

Addresses of nearest stockists and price list can be obtained from W. Greenwood Electronic Ltd., 21, Germain Street, Chesham, Bucks.

With the decrease in size of components, tool manufacturers have had to scale down a great number of their tools to meet the demands of handling these components. With this problem in mind Tele-Productions are now marketing a range of Swedish Stirex hand tools.

Available as side cutters, pliers and scissors the tools have been designed for close work and the complete tools measure a maximum length of only 175mm, including handle. A feature of the tools are plastics spring-loaded handles which are available in varying colours. This enables particular tools to be colour coded for easy identification.

Full details of tools available and price lists can be obtained from Tele-Production Tools, 28-B, Hamlet Court Road, Westeliff-on-Sea, Essex.

One of the more unusual items is a set of multi-blade screwdriver packs from F. Parramore and Sons (1924) Ltd. For electrical requirements the Multi-3 pack comprising wallet, chuck and three blades is recommended. For general purpose use the Multi-5 pack should be able to tackle most jobs.

Complete with a robust 4-jaw chuck handle the cruciform interchangeable blades are suitable for slotted, Pozidriv and Phillips screws.

One of the reasons for the shape of the blade is that when dealing



Oryx adjustable vice from W. Greenwood Electronic

with over tight screws extra torque can be applied to the blade with a spanner or a special ratchet, available as an extra.

The price of the Multi packs vary from approximately £2.25 to £3.55 and addresses of nearest stockists can be obtained from F. Parramore & Sons (1924) Ltd., Caledonian Works, Chapeltown, Sheffield.

Finally, to help keep the workshop tools tidy readers may be interested in a new range of "Clipco" nylon general purpose clips from Herbert Terry & Son Ltd.

Ideal for holding all kinds of tools, the Clipco is made from unbreakable nylon and having a special toothed stem the clips can be easily fixed (no screws or screwing being required) to walls, tiles, wood and pegboard.

to walls, tiles, wood and pegboard.

Available from most hardware and do-it-yourself stores the clips are sold in packets containing 8, 9 or 10 clips, dependent on size required, and retail at 21p to 23p per pack. The clips can also be purchased in bulk quantity at £2·18 per 100 direct from Herbert Terry & Son Ltd., Redditch, Worcs.



Parramore Paramo screwdriver packs
Alpha digital multimeter from Advance



Correspondents wishing to have a reply must enclose a stamped addressed envelope. We regret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

Linear or Log Law

Sir-I would like to congratulate your author on the P.E. Sound Synthesiser, on the obviously very impressive results obtained. However, if I might add a few criticisms, I feel that they could prove helpful to your readers, and even Mr Shaw himself. These points are based on a few years' experience in the field of voltage-controlled Synthesiser circuits, and on the successful marketing of such circuits and complete Synthesisers which I have designed.

Firstly, the use of integrated circuits for some applications can be very satisfactory. However, one must remember that one often has to use the most suitable i.c., which is not always (in fact is rarely) exactly made for the job. As a result, one ends up putting the makers' recommended current through the device, say 5 to 10mA, when if one designed a circuit from discrete components one would often find that a few micro-amps is all that is needed for the same job! This means that you can under-run the components at a tiny fraction of the makers' recommended current . . . the life expectancy is thus made infinitely

Secondly, use of most i.c.s involves dual-rail power supply, and also adds dramatically to the overall costs, as i.c.s are going up in price (the Dewtron modules cost, using discrete components, would still be less than £200 for more versatile design than that published, incidentally)

Thirdly, and alas, I feel more importantly, the decision to abandon the use of the proper log-law for the Voltage Controlled Oscillator design. This is really naughty, as it means:

1. No chording facility.

2. No variable keyboard pitch, which is essential if you want the range of use one expects from a synthesiser.

3. Difficulties in tuning the tuning ladder—each note must have its own preset, the adjustment of which will vary vastly from one end of the ladder to the other.

 Poor frequency stability, because, should voltage or resistance wander, the drift will be concentrated more to one end of the keyboard than the other, due to the non-linearity (or non-log'arity) of the tuning network, whereas, if logarithmic tuning

voltage law were used, the variation would be distributed evenly over the whole keyboard, so making it less noticeable, in terms of pitch.

5. Uneven swing in pitch control when driven by a slow-oscillator, unless special (and difficult) compensation is deliberately introduced into the output of the slow oscillator. Imagine a siren that starts climbing in pitch rapidly and then slows down to a slow whine when it reaches the higher frequencies.

Admittedly, the building of a stable, accurate law circuit is difficult, if the effects of temperature are to be minimised, and the circuit shown in the first article is very prone to temperature drift, as matched transistors from a thermal standpoint are virtually impossible to select due to the cycling time involved. None-the-less, circuits are possible (a special unique one is used in the Dewtron-2 module), which will not change even if a hot soldering iron is applied to them! Admittedly, the one in question did take years to perfect, and perhaps Mr Shaw did not have this sort of time at his disposal.

It would be interesting to learn if other readers have further ideas on the above or other points arising from the P.E. Synthesiser design.

B. H. Baily, Dewtron Design Engineering

Computer Club

Sir-You and your readers may be interested to hear that following my letter in P.E. last year an Amateur Computer Club has been formed.

The club is for those interested in the construction, design or programming of computers as a hobby. At present the main activity of the club is the production of a newsletter which acts as a databus to distribute information on hardware and software techniques of interest to the members.

Any one in joining the club may receive further details from me (SAE appreciated).

> M. Lord. 7 Dordells, Basildon, Essex

BACK NUMBERS WANTED

Anyone who can supply the undermentioned are asked to communicate directly with the reader.

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Mr. A. V. Barker, 35, Broadwalk, Westhoughton, Nr. Bolton, BL5 2AA.
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V. G. Dennison, 24, Ledbury Street, Aspley, bane, Australia, 4034, vember 1968, July 1972
C. H. M. Limb, 68, Westminster Way, North ksey, Oxford.

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D. Wood, 52, Egerton Road, Streetly, Sutton idfield, Warks bruary 1967

C. Calvert, 34, Renwick Road, Blyth, strhumberland, squet 1970

T. Barclay, Allan Glen's School, 190, Cathedral est, Glasgow, C.4.

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We regret that back numbers of Practical Electronics can no longer be supplied. We will try to publish announcements of readers' requirements (without a guaranteed date) free of charge. January to March 1972 Mr. P. Daswani, 276, Satara Road, Poons—I, India.

February 1969
Mr. P. Couzens, 30, Napier Road, Wembley, Middlesex, HAD 4VA
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Mr. S. Clements, 17, Semple Street, Hamilton, Lanarks, Mil 91L.
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VA10665	15;
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THYRISTORS 2N5060 50V 0-8A 30p, 2N5064 200V 0-8A 47p, CRS1/40 400V 1A 25p, 106F 50V 4A 40p, 106D 400V 4A 55p. MULLARD POLYESTER CAPACITORS C296 SERIES 400V: 0.001μF, 0.0015μF, 0.0015μF, 0.0015μF, 0.0015μF, 0.0022μF, 0.0022μF, 0.0033μF, 10.0047μF, 0.0047μF, 0.0047μF, 0.0047μF, 0.0047μF, 0.0047μF, 0.0047μF, 0.0047μF, 0.005μF, 0.005μF, 0.005μF, 10.0047μF, 0.005μF, 10.0047μF, 0.005μF, 10.0047μF, 0.005μF, 10.0047μF, 0.005μF, 10.0047μF, 0.005μF, 10.0047μF, 10.004

MYLAR FILM CAPACITORS 100V 0·001μF, 0·002μF, 0·005μF, 0·01μF, 0·02μF, 2½p. 0·04μF, 0·05μF, 0·068μF, 0·1μF, 3½p.

CERAMIC DISC CAPACITORS 100pF to 10,000pF, 2p each.

ELECTROLYTIC CAPACITORS—MULLARD 015/6/7 RANGE REPLACES C426, C457 RANGES, (µF/v) 1-0/63, 1-5/63, 2-2/63, 3-3/63, 4-7/63, 6-8/40, 10/25, 10/63, 15/16, 15/40, 15/63, 22/10, 22/25, 22/63, 33/6-3, 33/40, 47/4, 47/10, 47/25, 47/40, 47/63, 68/6-3, 68/16, 100/4, 100/10, 100/25, 100/40, 150/6-3, 150/16, 150/25, 220/4, 220/16, 230/4, 330/10, 330/4, 330/10, 470/6-3, 59 each, 68/63, 150/40, 220/25, 330/16, 470/10, 680/5-3, 1000/4, 9p. 100/63, 150/63, 220/40, 470/25, 680/16, 1,000/10, 1,500/6-3, 12p. 220/63, 470/40, 680/25, 1,000/16, 1,500/10, 2,200/6-3, 15p. 330/6-3, 680/40, 1,000/25, 1,500/16, 2,200/10, 3,300/6-3, 4,700/4, 18p.

30LID TANTALUM BEAD 0·1μF 35V 0·22μF 35V 0·47μF 35V 1·0μF 35V	2-2μF 35V 4-7μF 35V 6-8μF 25V 10μF 25V	22µF 16V 33µF 10V 47µF 6-3V 100µF 3V	12p

VEROSOARD		JACK PLUGS AND SOCKETS
17 × 5 (plain) 17 × 3½ (plain) 17 × 2½ (plain) 2½ × 5 (plain) 2½ × 3½ (plain) Pin insertion tool	22p 16 24p 24 24p 24 27p 27, 75p 57, 100p 78; — 82; — 60; — 42; — 12; — 11	Stereo screened 35p 3-3mm screened 13p 3-3mm screen
Spot face cutter Pkt. 50 pins	42p 42g 20p 20g	BATTERY ELIMINATOR

t. 50 pins	20p	20p	9V main	powe	F Supply.	Same siz	te as PP9	battery.
RGE (CAN)	ELEC	TROLY	TICS					

2800µF 100V £2:60 3200µF 16V 50p

POLYSTYRENE CAPACITORS 160V 2½% 10pF to 1,000pF E12 Series Values 4p each.

SMOKE AND COMBUSTIBLE GAS DETECTOR—GD!

The GDI is the world's first semiconductor that can convert a concentration of gas or smoke into an electrical signal. The sensor decreases its electrical resistance when it absorbs deoxiding or combustible gases such as hydrogen, carbon monoxide, methane, propane, such as hydrogen, carbon monoxide, methane, propane, study large enough to be utilized without amplification. Full details and circuits a exapplied with each detector.

Detector GDI, £2. Kit of experience of the detector
LARGE RANGE ITT/TEXAS IC's NOW IN STOCK

		PRI	CES ARI	CALCL	JLATED	ON TOT	AL NUN	IBER OR	DERED	REGARDLE	SS OF	MIX		
7400 7401 7401 7402 7403 7404 7405 7406 7407 7408 7409 7411 7412 7413 7416 7420 7421	- P B B B 20 20 50 56 36 18 23 36 45 45 36	12-24 P 16 16 16 18 18 18 18 18 18 18 18 21 21 28 43 16 21 28 43 16 21 28 43 16 21 28 43 28 43 43 43 43 43 43 43 43 43 43	25-99 P 14 14 14 16 40 44 27 27 26 39	100 + p 13 13 13 13 14 14 35 23 23 22 22 22 23 23 23 22 22 23 23 22 22	7448 7450 7451 7453 7454 7460 7470 7472 7473 7474 7475 7476 7480 7481 7482 7483 7484 7485	185 18 18 18 18 18 30 .40 40 55 40 100 125 100 120 250	175 16 16 16 16 16 16 28 38 38 52 36 52 36 95 175 97 175 245	170 14 14 14 14 14 25 27 36 32 90 110 90 95 110	165 13 13 13 13 24 23 30 28 49 30 85 105 85 92 105	74118 74121 74141 74145 74150 74151 74153 74154 74156 74180 74180 74190 74191 74192 74192 74197	100 43 100 150 110 120 200 150 130 155 195 195 195 195 200 200 200 200	82 40 95 140 280 100 110 180 120 136 190 190 190 180 190	73 38 90 135 250 95 105 170 100 112 185 180 180 180	64 36 85 130 220 89 95 160 86 96 105 180 164 170
7426 7430 7432 7440 7441 7442 7443 7447	32 20 40 20 80 80 125 175	29 18 36 18 75 75 75 120 165	27 23 16 32 16 70 70 115	20 14 28 14 65 65 15 120	7490 7491 A 7492 7493 7494 7495 7496 74100	45 75 100 75 75 95 105 100 250	67 92 70 68 90 100 95 240	37 60 85 65 60 85 95 90	33 52 79 60 52 80 90 85 230	709 741 741 723 747 748 Dil. socket		INEAR IC 14 pin DIL 8 pin DIL 14 pin DIL 14 pin DIL 14 pin DIL 8 pin DIL 14 pin and		32p 14p 18p 95p 85p 35p



MAINS MOTOR

Precision made—as used in record decks and tape recor-dera—ideal also for extractor fan, blower, heaters, etc. New and perfect. Snip at 729. Postage 20p for first one then 10p for each one ordered.

MUSIC ON TAPE A further buy enables us t

A further buy enables us to offer these at an even lower price—namely 65p each or 6 for 42-50, Send for list of titles. We can't repeat when sold out...

ANCED ARMATURE

500 ohn, operates as speaker or microphone, so useful in intercom or similar circuits, 87p each, 10 for \$8.78.





CAR ELECTRIC PLUG Fits in place of cigarette lighter, Useful method for making a quick

system, 38p each or 10 for \$2-42.



12 VOLT II AMP

This comprises double-wound
230/240V mains transformer
with full wave rectifier and
2000 mF amoothing. Price
22:20, plus 20p post a packing.

Heavy Duty Mains Power Pack. Output voltage adjustable from 15-40V in steps-maximum Beavy July Mains Fower Fack. Output voltage adjustable from 15-40V in steps-maximum load 230W—that is from 6 amp at 40V to 15amp at 15V. This really is a high power heavy duty unit with dozens of workshop uses. Output voltage adjustment is very quick—simply interchange press on leads. Silleon rectifiers and smoothing by 5,000mF. Frice 26-38 pine 65p post.



MAINS RELAY BARGAIN
Special this month are some single, double and tretble pole changeover relays, Contacts rated at 15 amps. Operating coil wound for 240V A.C. Good British Make. Unused. Size approx.-14' × 1'. Open construction. Size Birgle pole 88p each. 10 for £8.67.



WAFER SWITCHES

2 pole, 2 way—4 pole, 2 way—2 pole, 3 way—4 pole, 3 way—2 pole, 4 way—2 pole 6 way—1 pole, 12 way. All at 22p

CONNECTING WIRE

CONNECTING WIRE
7-0076 Copper conductors. 600 metre drums
available in the following colours: Red/Brown,
Yellow, Green/Grey, Blue/Green, Red/Gronge,
Green/White, Grey/White, Blue/Grange, Brown/
Red, Brown/White, Red/Grey, Blue/Grey, Blue/
Brown. Price \$2-20 per drum plus 40p post.
Bate alternative colours. Ditto—but with 200
metres. Price \$1-28 per drum plus 20p. Ditto—but
with 100 metres. Price 83p per drum.

DOOR SWITCH

As fitted to refrigerators, etc. Switches off as door closes. White 17p each. Black 15p each.

GOODMANS P.M. SPEAKERS

8in × 5in hi flux 15 ohm coil very suitable for use with Mullard Unilex stereo amplifier or with the EP9000 on its own, \$1.65 plus 20p post each.

7in x 4in also 15 ohm and sultable for use with the EP9000—not quite such good quality of course. Price \$1.05 each.

U.Y. LIGHTING

Useful for flaw detection in metals and for looking for water marks, etc., also for fitting over tropical flash tanks—African violets and other indoor plants which must have U.V. for healthy growth. The outlift comprises of a 12in U.V. tube—choke—starter holder and tube ends. Price \$2.20 plus

SOIL HEATING KIT



Suitable for garden frame or propogating shelf, etc. Comprises 40W mains transformer, heating wire, connection strip and insulated wire with connection diagram. \$1.85 plus 20p post.

80 WATT TUBULAR ELEMENTS
Brass encased with beaded flex ends. Standard
replacements in most absorbtion type retrigerators
but also has dozens of other uses. 28p each or
10 for \$1.45.

DOUBLE ENDED MAINS MOTOR
This is a very quiet running
induction motor, with two opecial
features—ones is that it has for the casy fixing and the other is it has a
good length of spindle coming from
each end. So it could drive two
things at once—or with simple
elide attachment could reverse direction. Sip each
or 10 for \$4.96.



MULLARD AUDIO AMPLIFIERS
All in module form, each ready built complete with heat
sinks and connection tags, data supplied,
Model 1153 500mW power output 72p,
Model 1172 735mW power output 94p,
Model EP9008 4 wait power output \$1.60.
EP9001 twin channel or stereo pre amp. \$1.92.
10% discount if 10 or more ordered.

CENTRIFUGAL BLOWER
Miniature mains driven blower centrifugal type blower
unit by Woods. Powerful but specially built for quiet
running—driven by cushioned induction motor with
specially built low noise bearings. Overall size 4½ × 4½
× 4". When mounted by finage, air is blown into the
equipment but to suck air out, mount it from centre
using clamp. Ideal for cooling electrical equipment or
fitting into a cooker hood, film drying cabinet or for
removing flux smoke when soldering etc. etc. A real
bargain at 28-06.



GOOD COMPANION I.C. VERSION

We can now offer these again in I.C. version using Ferranti ZN444 and Mulard AF Module 1172. Cabinet is approx. Illn widex & Bin high x 3in deep. Complete assembly instruction—Excellent tone wood cabinet \$5-75 plus 2000.

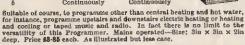
MIGHTY MIDGET
Probably the tiniest possible radio, as described in Practical Wireless, 4 January
73. All electronic paris \$2.20 post paid.

THIS MONTH'S SNIP-

HORSTMANN 24-HOUR TIME SWITCH

With six position programmer, When fitted to hot water systems this could programme as follows:
Programme Hot Water Central Heating

rogramme	Hot water	Central meating
0	Off	Off
1	Twice dally	Off
2	All day	Off
3	Twice dally	Twice daily
4	All day	All day
5	Continuously	Continuously





PORTABLE ELECTRIC DRILL

Very superior quality made by a famous Dutch Toolmaker. Model No. A8M839. 300W—2 speed 2,200/3,000. With fin chuck and chuck key, also separate side handle and hammer facility for dealing with concrete, etc. An equivalent British made drill would cost £15. £10.90—smillar model but without the hammer attachment £8.76, real for 7 days.

Have either model on approval for 7 days.

A 4 4

DISTRIBUTION PANELS

Just what you need for work bench or isb. 4 × 13 amp sockets in metal box to take standard 13 amp fused plugs and on/off switch with neon warning light. Supplied complete with 7 feet of heavy cable. Wired up ready to work, £2.50 plus 20p post and insurance.



CAPACITOR DISCHARGE CAR IGNITION

This system which has proved to be amazingly efficient and reliable was first described in the Wireless World about a year ago. We can supply kit of parts for an improved and even more efficient version (Practical Wireless, June). Price \$6.55 plus De-luze model including printed circuit board, etc. \$7.65.



24-HOUR TIME SWITCH Made by fmiths, these are A.C. mains operated, NOT CLOCKWORK. Ideal for mounting on rack or shelf or can be built into box with 13A socket. Two completely adjustable time periods per 24 hours, 5A changeover contacts will switch circuit on or off during these periods. 28-75 post and ins. 25p. Additional time contacts 55p pair.

THYRISTOR LIGHT DIMMER
For any lamp up to lkw. Mounted on switch plate to fit in place of standard switch. Virtually no radio interference. Price 28-95, plus 20p poet and insurance. Industrial model 5A 28-30.



BATTERY CONDITION TESTER
Made by Mallory but sultable for all batterles made by
Ever Ready and others, most of which are zinc carbon
types but also mercury manganese—nicad—silver oxide
and alkaline batterles may be tested. The tester puts a
dummy load on the battery and the meter scale indicates
the condition depending upon which section the pointer
rests. The section reads "replace" "weak" or "good". The
tester is complete in its case, size \$3 \times 2 \times 2 \times 2 \times 4 \times 2 \times 4 \times 4 \times 2 \times 4 \tim



SPIT MOTOR

200-256V Induction Motor, driving a Carter gear box with 11 of output drive shaft running at 5 revs per minute. Intended for rozating chickens also suitable for driving models—windmills, coloured disc lighting effect, etc. etc. 82-04 plus 20p post and insurance.

Where postage is not stated then orders over 25 are post free. Below 25 add 20p. Semiconductors add 5p post. Over £1 post free. S.A.E. with enquiries please.

CONTROL

CONTROL CONTROLLER NEW IKW MODEL
Electronically changes speed from approximately 10 revs. to maximum. Full power at all speeds by finger-tip control. Kit includes all parts, case, everything and full instructions. \$1.65. Made up model also available. \$2.75.

MAINS TRANSISTOR POWER PACK

PACK
Designed to operate transistor sets and amplifiers. Adjustable output 8v., 9v., 12 volts for up to 500mA (class B working). Takes the place of any of the following batteries: PPI, PP3, PP4, PP6, PP7, PP9, and others. Kit comprises: mains transformer rectifier, monothing and load resistor condensers and instructions. Real only only \$1.10.

SWITCH TRIGGER MATS



Bo thin is undetectable under carpet but will ewitch on with slightest pressure. For burglar alarms, ahop doors, etc. 24in x 18in 21.54.

18in x 10in \$1.54.

PP3 BATTERY CHARGER
Almost 3 times the life can be obtained from PP3
battery If you re-charge it from the mains—
this ready to use charger with instructions only



15A ELECTRICAL PROGRAMMER

PROGRAMMER
Learn in your sleep:
Have radio playing and kettle boiling as you awake—switch on lights to ward off intruders—have a warm house to come home to. All these and many other things you can do If you invest in an electrical programmer. Clock by famous maker with 15 amp. on/off switch. Switch-on time can be set anywhere to stay on up to 6 hours. Independent 66 minute memory logger. A beautiful unit. Price \$2.15 + 20p p. & p. or with glass front chrome bezel \$3p extra.



ROCKER SWITCH

13 amp self-fixing into an oblong hole. Size approximately 1" × 2" Sp each, 10 for 81p.



THERMAL CUTOUT
A miniature device in dis. on one screw fixing
mount—can be used for motor overload protection,
fire alarm, soldering iron, switch off, etc., etc.
15 amp contacts open with fiame—radiant or
conducted heat. 11p each or 10 for 60p,

HONEYWELL PUSH BUTTON
MICRO SWITCHES
As used in some vending machines, etc. Each
switch is a changeover type rated at 15A 250V a.c.
Intended for panel mounting with fixing nuts and
single black push knob. Single changeover
switch 829, Double switch 42p, Triple switch 51p.

KETTLE ELEMENTS

KETILE ELEMENTS
Made by the famous A.B.I. Co. Complete with
washers and combined fixing ring and plug
shroud. Normal 2 round pin and flat pin earth
connection and overload reset push button.
2 Models—life (approx.) suitable for Swan and
other similar models—life (approx.) suitable; for
G.E.C. Hotpoint, etc. All quick boil 2½kW elements
at 240V. Price \$1.38 each.

MAINS TRANSFORMER BARGAIN Standard mains 240V input Secondary 2.4V 9A intermittent 5A continuous. Price \$5p.

BULGIN 4 PIN PLUG AND

SOCKET
Plug has shielded pin and flex entry and socket for panel mounting. 88p per pair. Ditto—but 3 pin,

MACLAREN THERMOSTAT

Make and break 20A a.c. with the sensor probe coupled by 2 feet capillary covering range of 10-100°C—complete with large engraved control knob. Price 35p.

J. BULL (ELECTRICAL) LT (Dept. P.E.), 7 Park Street, Croydon CRO IYD Callers to 102/3 Tamworth Road, Croydon

Sinclair Project 60

Now-the Z.50 Mk.2

with built-in automatic transient overload protection

When originally introduced, the Sinclair Z.50 proved how it was possible to design and produce a popularly priced modular power amplifier having characteristics to challenge the world's costliest amplifiers. Many thousands of Z.50's are now giving excellent service day in, day out. But we have also learned that constructors do not always use their Z.50's ideally. That is why we have introduced modifications whereby risk of damage through mis-use is greatly reduced and performance further enhanced. The Z.50 Mk.2 has improved thermal stability, more accurately regulated D.C. limiting to ensure more symetrical output voltage swing and clipping and still less distortion at lower power. Z.50 Mk.2 is compatible with all other Project 60 modules, and may be incorporated to advantage in existing systems. Eleven silicon epitaxial planar transistors are now used, two more than in the original Z.50; circuitry has been re-designed, making this versatile high performance amplifier better than ever.



with free manual £5.48

Z.30 the power amplifier for quality and economy



with free manual

The Z.30 provides excellent facilities for the constructor requiring a high fidelity audio system of less power than that available from Z.50's. Using a power supply of 35 volts, Z.30 will deliver 15 wets RMS into 8 ohms, or 20 watts RMS into 3 ohms using 30 volts. Total harmonic distortion is a fântastically low 0.02% at 15 watts into 8 ohms with signal to noise ratio better than 70 dB unweighted. Input sensitivity 250mV into 100K ohms. Size 80 x 57 x 13 mm (3½ x 2½ x ½) Z.30, Z.50 and Z.50 MK.2 modules are compatible and interchangeable.

Guarantee

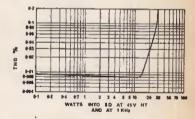
II, within 3 months of purchasing any product direct from Sinclair Radionics Ltd., you are dissatisfied with it, your money will be refunded at once. Many Sinclair appointed Stockists also offer this same guerantee in co-operation with Sinclair Radionics Ltd.

Stockists also offer this same guerantee in co-operation with Sinclair Radionics Ltd.
Each Project 60 module is tested before leaving our factory and is guaranteed to work perfectly. Should any defect arise in normal use, we will service it at once and without any charge to you, if il is returned within two years from the date of purchase. Outside this period of guarantee a small charge (typically £1,00) will be made. No charge is made for postage by surface meil. Alt Meil is cherged at cost.



Brilliant new technical specifications

Input impedance 100 K Ω Input (for 30 ν into 8 Ω) 400mV Signal to noise ratio, referred to full o/p at 30 ν HT 80dB or better Distortion 0.02% up to 20W at 8 Ω . See curve Frequency response 10Hz to more than 200 KHz \pm 1dB Max. supply voltage 45 ν (4 Ω to 8 Ω speakers) (50 ν 15 Ω speakers only) Min. supply voltage 9 ν Load impedance — minimum: 4 Ω at 45 ν HT Load impedance — maximum: safe on open



Typical Project 60 applications

System	The Units to use	together with	Units cost
Simple battery record player	Z.30	Crystal P.U., 12V battery volume control, etc.	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control, etc.	£9.45
12W. RMS continuous sine wave stereo amp, for average needs	2 x Z.30s, Stereo 60; PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers	2 x Z.30s, Stereo 60; PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms)	2 x Z.50s, Stereo 60; PZ.8, mains transformer	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43

Sinclair Radionics Ltd., R.O. London Rd., St. Ives, Huntingdonshire PEI7 4HJ (Phone: St. Ives 64311). Regd. Bus. No. England 699483.

the world's most advanced high fidelity modules

Stereo 60 Pre-amp/control unit



Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout, a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly easy to mount.

SPECIFICATIONS—Input sensitivities: Radio — up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A. curve ±1dB:20 to 25,000 Hz. Ceramic p.u. — up to 3mV: Aux — up to 3mV. Output: 250mV: Signal to noise retio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE+12 to —12dB at 10MHz: BAS+12 to —12

Built, tested and quaranteed.

£9.98

Project 60 Stereo F.M. Tuner



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range; 87.5 to 108MHz. Sensitivity: 7µV for lock-in over full deviation. Squeich level: Typically 20µV. Signal to noise ratio: >65dB. Audio fraquency response: 104tz – 15kHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2 µV. Cross talk; 40dB. Output voltage: 2 x 150mV R.M.S. maximum Operating voltage: 25–30VDC. Indicators: Stereo on: tuning. Size: 93 x 40 x 207mm.

Built and tested, Post free.

£25

Super IC.12 Integrated circuit high fidelity amplifier



Having introduced Integrated Circuits to hi-ficonstructors with the IC.10, the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the Super IC.12, a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up, F.M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC. make possible, It is the equivalent of a 22 transactions.

sistor circuit contained within a 18 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC.12 is compatible with Project 60 modules which would be used with the Z.50 and Z.30 amplifiers. Complete with free manual and printed circuit board.

SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak), 6-80. Frequency Response: 5Hz to 100KHz±1dB. Total Harmonic Distortion: Less than 1%. (Typical 0·1%) at all output powers and frequencies in the audio band (28V). Load Impedance: 3 to 15 ohms, Input Impedance: 250 Kohms nominal. Power Gain: 90dB (1,000,000,000 times) after feedback. Supply Voltage: 6 to 28V. Quiescent current: 8mA at 28V. Size: 22×45×28mm including pins and heat sink.

Manual available separately 15p post free,

With FREE printed circuit board and 40 page manual.

£2.98 Post free

Power Supply Units The new PZ.8 Mk.3

The most reliable power supply unit ever made available to constructors. Brilliant circuitry makes failure from over load and even direct shorting of the output impossible. This is due to an ingenious re-entrant current limiting principle which, as far as we know has never before been available in any comparable unit outside the most expensive laboratory equipment. Ripple and residual noise have been reduced to the point of almost total elimination. This is, of course, the perfect unit for Project 60 assemblies, particularly where the new Z.50 MK.2 amplifiers are used. Nominal working voltage—45.

PZ.8 Mk.3-£7.98

(Mains transformer, if required) £5.98

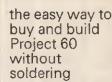
PZ.5 30v. unstabilised

(not suitable for Project 60 tuner) £4.98

PZ.6 35v. stabilised

(not suitable for IC. 12) £7.98

Project 605





Project 605 in one pack contains: one PZ.5, two Z.30's, one Stereo 60 and one Masterlink, which has input sockets and output components grouped on a single module and all necessary leads cut to length and fitted with clips to plug straight on to the modules thus eliminating all soldering.

Complete with comprehensive manual, post free All you need for a superb 30 watt high fidelity stereo amplifier £29.95

Order form

Please send

I enclose cash/cheque/money order.

Name

Address

PE/7/73

SINCLAIR RADIONICS LTD., LONDON ROAD ST. IVES, HUNTINGDONSHIRE PEIT 4HJ



Add 10% V.A.T. to all prices quoted on page opposite and above

IN ELECTRONIC SUPPLIFS



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RESISTORS
Carbon film 5% from 12 to 1M 0, 10% from 1-2M 0 to 10M 0. E12 series, 4W 1p; †W 12p.
Metal Oxide 2% 10 0 to 1M 0. E24 series †W 4p.
Also IW, 2‡W, 5W, 10W types stocked.

7 SEGMENT DISPLAY Ministron 3015F £1-98 each.

POTENTIOMETERS

Miniature carbon track with 4in. spindles. 5k G, 10k G, 25k G, 500k G, 10k G, 250k G, 500k G, 1M G, 250k G, 500k G, 1M G, 20 or lin with switch 21e. Log or lin with switch 21p. Dual gang less switch 35p.

P.E. SOUND SYNTHESISER

If this project seems expensive YOU HAVEN'T SEEN

We shall be stocking all the parts for this exciting project, from the spacial I.C.'s right down to the nuts, bolts and spacers for mounting the Veroboards. Send S.A.E. NOW for our detailed price lists.

YOU SIMPLY MUST SEE OUR PRICES!

741C

8 pin DIL OP AMP PRICE REDUCED ON THESE TOP QUALITY I.C.: 16p.

MeMURDO Socket, RS8, 52p Plug, RP8, 36p, (As used in P.E. Sound Synthesiser.) SEMI-CONDUCTORS

CONDUCTORS

We stock a large range of transistors and LCs, for details please see our free extended to the control of the cont

DIN PLUGS. 3 pin, 9p. 5 pin A (180°), 5 pin B (240°), 10p each.

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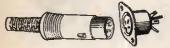
Other semiconductors: AC128, 17p; AF117, 35p; BFYSI, 19p. Full lists and technical data will be found in Catalogue No. 6. See also amendments list.

SIEMENS' THYRISTORS 0-8A 400V 65p, 600V 50p. 3A 400V 70p, 600V £1-02.

ZENER DIODES full range E24 values: 400mW: 2-7V to 36V, 14p each; IW: 6-8V to 82V, 21p each: 1-5W: 4-7V to 75V, 48p each. Clip to increase 1-5W rating to 3 watts (type 266F) 4p.

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TRANSISTOR ACCESSORIES
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SWITCHES

1011 SPST toggle, 20p; 40p OPDT toggle, 29p (these are chrome placed, 25A rating); 7201 sub-miniature DPDT 250V a.c./2A, 48p.



ROTARY SWITCHES

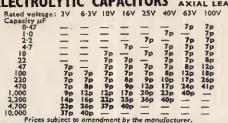
Radiospares Miniature Maka-switch (in assembly kit form). Shaft, 48p. Wafers, MBB-2P5W, IP11W: BBM1P12W, 2P6W, 3P4W, 4P3W, 6P2W, 32p each.

WAVECHANGE **SWITCHES**



IPI2W, 2P6W, 3P4W, 413W, 24p each

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POLYCARBONATE—5% TOLERANCE

250V up to 0-1 µF: 100V, 0-1 µF and

250V up to 0·lμF: 100V, 0·lμF and above, 0·01; 0·012; 0·015; 0·018; 0·022; 0·033; 0·047; 0·056; 4p each, 0·666; 0·082; 0·l; 0·12; 0·15; 4p each, 0·18; 0·22; 5p each, 0·27; 0·33; 4p, 0·39 7p, 0·47 8p, 0·56 10p, 0·68 11p, 1μF 13p, Prices subject to amendment by the manufocturer.

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10 to 99 | (see note below) 8 0.9 0.9

100 up

7·5 0.75 0·75 0·8 1·6

RESISTORS - 10%, 5%, 2%

		/0/	/0/ /0	,	
Code	Power	Tolerance	Range	Values	I to 9
_				available	
С	1/20W	5%	82 Q-220K D	E13	. 9
С	1/877	5%	4-7 E -470K D	E24	1
C	1/4W	10%	4-7 Ω ~ 10M Ω	E12	
C	1/2W	5%	4-7 D-10M D	E24	1.2
000	1W	10%	4-7 n = 10M n	E12	2.5
MO	1/2W	2%	10 D-1MD	E24	Ä
ww	IW	10% ± 1/20 0	0-22 0-3-9 0	E12	,
ww	3W	50/	1 B-10K B	Ei2	ź
ww	7W	5%	1Ω-10KΩ	ĒI2	
Codes:		film, high stability		EIZ	,
Codes:	L = carbon	film, high stability	f. DW DOISE		

MO = metal oxide, Electrosil TR5, ultra low noise.

MO = metal oxide, Electrosil TR5, ultra low noise.

WW = wire wound, Plessey.

Yalues:

values: E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51 62, 75, 91 and their decades.

Prices are in pence each for quantities of the same ohmic value and power rating. NOT mixed values, (Ignors fractions on total value of resistor order.)

We regret that where substantial price increases have occurred in some items, they are due to increased costs of imported goods. S.A.E.

KNOBS

All grub screw fixing for ±in shafts. Black. For other types rese Catalogue No. 6, p. 54, F.14 (20mm) pack of 2, 32p; F.13 (25mm) pack of 2, 32p; F.19 (20mm) pack of 2, 32p; F.19 (20mm) pack of 2, 32p; F.19 (20mm) pack of 2, 40p; F.84 (20mm) pack of 4, 40p; K.84 (20mm) pack of 4, 40p; K.80/3 (17mm) aluminium, 24p erch.













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2.2M 0, 12p; P20 log, 47K 0 to
2.2M 0, 12p; D20 log, 47K 0
to 2.2M 0, 12g; Dual gang
lin, 47K 0 to 2.2M 0
to 100, 12g; Dual gang
lin, 47K 0 to 2.2M 0
to 2.2M 0
to 2.2M 0
to 2.2M 0
to 2.2M 0
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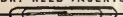
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