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Our July issue will be published on Friday, June 8, 1973

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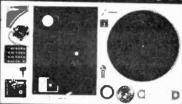
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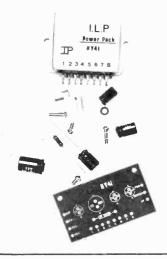
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THE HY41

The HY41 supersedes the popular HY40 introduced by ILP last year. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion (typically 0.05% at 1KHz into 8 ohms!) and is electronically and mechanically compatible with the HY40.

With this important improvement the HY41 retains all of the quality characteristics found in the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete power amplifier of extremely high performance sufficiently versatile to provide power not merely for Hi-Fi but also for public address systems and industry

The free manual gives a full circuit diagram of the HY41 and its various applications including a complete stereo amplifier.

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent years.

OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts

R.M.S. continuous.

LOAD IMPEDANCE: 4–16 ohms.

INPUT IMPEDANCE: 30K ohms at 1KHz.

VOLTAGE GAIN: 30db at 1KHz

TOTAL HARMONIC DISTORTION: less than 0.15% (typical 0.05%)

FREQUENCY RESPONSE: 5Hz-50KHz + 1db

SUPPLY VOLTAGE: + 22.5volts D.C. SUPPLY CURRENT: 0.8 amps maximum.

PRICE: inc. comprehensive manual, P.C. board, five extra components and P. & P.:-

MONO: £5:39

STEREO: £10-78

IP HY5

UNIQUE HYBRID PRE-AMPLIFIER

The EY5 has rapidly established a position in the WORLD as the sole hybrid pre-amplifier to contain all feedback and equalization networks within an integrated pre-amplifier circuit.

Supplied with the HY5 are two stabilizing capacitors and by the addition of volume, treble and bass potentiometers it is ready for use.

Internally the HY5 provides equalization for almost every conceivable input, the desired function is achieved by use of a multi-way switch or by direct interconnection

Two distinctive features of the HY5 are its inbuilt stabilization circuit, allowing it to be run off any unregulated power supply from 16–25 Volts and a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo pre-amplifier

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

INPUTS

Magnetic Pick-up (within ±1db RIAA curve)

2mV, 47K Ω Tape Replay (external components to suit head). 4mV, 47K Ω

Microphone (flat) 10mV, 47KΩ Ceramic Pick-up (equalized and compensatable) 20-2000mV. variable.

Tuner (flat) 250mV. 100K Ω Auxiliary 1 250mV. 47K Ω Auxiliary 2 2–20mV. 100K Ω

OUTPUTS

Main Pre-amp output 500mV Direct tape output 120mV

ACTIVE TONE CONTROLS (Bexendall) Treble + 12db. Treble + 12db.

INTERNAL STABILIZATION
Enables the HY5 to share an unregulated

supply with the Power Amplifier.
SUPPLY VOLTAGE

16-25 volts

PRICE: MONO £3.96

SUPPLY CURRENT

6mA approx.

OVERLOAD CAPABILITY

better than 26db on most sensitive input

infinite on tuner and auxl.

OUTPUT NOISE VOLTAGE: 0.5mV

STEREO £7:92



POWER SUPPLY PSU45

The versatile P.S.U.45 is designed to supply your HY41's +HY5's in stereo or mono format.

Specification

Input: 200-240 Volts.

Output: ± 22.5 Volts at 2 amps.
Overall Dimensions: L. 7"; D. 3.8"; H. 3.1"

PRICE: £4-95 inc. P. & P.

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TRANSFORMERS

MAINS ISOLATING SERIES
Primary 200-250 Volts Secondary 240 Volts Centre
Tapped (120V) and Earth Shielded
ALSO AVAILABLE WITH 1151/120V SEC. WINDING ALSO AVAILA

VA Weight
(Wotts) Ib oz
20 | 1 | 1
60 3 8
100 5 12
200 9 8
250 12 4
350 15 0
500 27 0
1000 40 0
1000 63 0 P & P 1 77 30 2 62 36 2 88 52 4 83 52 6 38 67 Size cm. 7·0 × 6·0 × 6·5 8·9 × 8·0 × 7·7 10·2 × 8·9 × 8·3 12·0 × 10·3 × 10·0 9·5 × 12·7 × 11·4 14·0 × 10·8 × 12·4 17·1 × 11·4 × 15·9 17·8 × 17·1 × 21·6 24·1 × 21·6 × 15·2 6.38 8.55 12.32 22.70 37.50 82 AUTO SERIES (NOT ISOLATED)
Weight Size cm. Auto Taps P & P 0 93 22 1 82 30 2 20 36 4 28 52 6 35 67 11 54 82 16 72 *

TOTALLY ENCLOSED 115V AUTO TRANSFORMERS 115V 500 Watt totally enclosed auto transformer, complete with mains lead and two 115V outlet sockets, 48-63. P & P 67p. Also available a 20 Watt version. 41-84. P & P 22p.

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LOW VOLTAGE SERIES (ISOLATED)

PRIMARY 200-250 VOLTS 12 AND/OR 24 VOLT RANGE

Ref. Amps. Weight
No. [2V 24/16 oz 7]

111 05 0:25 12 76. 57> 44 0-12V at 0:25A ×2 93 22

121 1-0 0:5 1 0 83 > 51 5-1 0-12V at 0:5A ×2 93 22

121 2 1 0 7 0 × 6-4 > 57 0-12V at 0:5A ×2 1-11 22

71 2 1 0 7 0 × 6-4 > 57 0-12V at 0:5A ×2 1-14 622

71 2 1 0 7 0 × 6-4 > 57 0-12V at 0:5A ×2 2-204 36

70 6 3 3 12 10-2 > 76 × 86 0-12V at 3A ×2 2-46 42

70 6 3 3 12 10-2 > 76 × 86 0-12V at 3A ×2 2-46 42

70 10 5 6 3 79 10-8 10-2 0-12V at 3A ×2 2-46 42

71 10 5 6 3 79 10-8 10-2 0-12V at 3A ×2 2-46 42

72 10 5 6 3 79 10-8 10-2 0-12V at 3A ×2 2-46 42

72 10 5 6 3 79 10-8 10-2 0-12V at 3A ×2 2-46 42

72 10 5 6 17 9 13 34 12-1 12-1 0-12V at 3A ×2 3-23 52

115 20 10 11 13 12-1 × 11-4 × 10-2 0-12V at 10A ×2 6-35 67

187 30 15 16 12 13 33 ×12-1 × 12-1 0-12V at 10A ×2 6-35 67

187 30 15 16 12 13 33 ×12-1 × 12-1 0-12V at 10A ×2 6-35 67

187 30 15 16 12 13 33 ×12-1 ×12-1 0-12V at 10A ×2 6-35 67

187 30 15 16 12 13 33 ×12-1 ×12-1 0-12V at 10A ×2 6-35 67 Size cm. 30 VOLT RANGE Secondary Taps

Amps. Weight 1b az 0.5 | 4 4 1.0 2 0 2.0 3 2 3.0 4 6 4.0 6 0 5.0 6 8 8.0 10 0 10.0 12 2 P & P Ref. No. 112 79 3 20 21 51 117 88 89 8:3 × 3:7 × 4:9 7:0 × 6:4 × 6:0 8.9 × 7:0 × 7:6 10:2 × 8:9 × 8:6 10:2 × 10:0 × 8:6 12:1 × 10:0 × 10:2 14:0 × 10:2 × 11:4 14:0 × 10:2 × 11:4 0-12-15-20-24-30V 1 · 1 · 1 1 · 48 2 · 2 · 72 3 · 23 4 · 02 4 · 80 6 · 20 7 · 85

50 VOLT RANGE Secondary Taps P & P £ p 1.46 30 2.13 36 2.96 42 Ambs. Weight Size cm. Weigh 10 0z 1 11 2 10 5 0 6 0 9 4 12 4 18 9 19 12 No. 102 103 7-0 × 7-0 × 5-7 0-19-25-33-40-50V 8-3 × 7-3 × 7-0 10-2 × 8-9 × 8-6 10-2 × 10-2 × 8-3 12-1 × 11-4 × 10-2 12-1 × 11-1 × 13-3 13-3 × 13-3 × 12-1 16-5 × 11-4 × 15-9 2·0 3·0 4·0 6·0 8·0 10·0 104 105 106 107 118 4.01 5.31 7.85 10.25 12.85

60 VOLT RANGE Size cm. P & P Amps. Weight ss. Weight

1b oz

2 4

3 0

5 6

8 8

10 6

16 12

23 2 124 126 127 125 123 0·5 1·0 2·0 3·0 4·0

82 6-0

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1100	AC100	0.22	100000	0.72	DC 110	0.21	RD200	1.05	1
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	ACY 18	0.22	BC113	0.11		0.12	BF154	0.50	В
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50	0.04	0-08	0.08	0.08	0.16	0.23	0.66
100	0.04	0-07	0.08	0.15	0.18	0.26	0.83
200	0.06	0.10	0.07	0.16	0.22	0.27	1.10
400	0.07	0-15	0.08	0.22	0.30	0.41	1.38
600	0.08	0.18	0.11	0.26	0.38	0.50	2.05
800	0-11	0.19	0.12	0.28	0.41	0.61	2.20
1000	0.12	0.30	0.16	0.33	0.51	0.70	2.75
1200		0.37		0.42	0.63	0.83	

ZENER DIODES 2ENER DIODES 400mW (DO-7 Case), range 2-33V, 15p each. 1-5W (Top-Hat), range 2-33V, 20p each. 10W (80-10 Stud), 2-33V, 33p each.

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14		Mixed Silicon and Germanium Diodes	
115		NPN Sil, Planar Trans. TO-5 like BFY51, 2N697	-
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117		Germanium PNP AF Transistors TO-5 like ACY 17-22	
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121		AF, Germanium Alloy Transistors 2G300 Series & OC71.	
123		The state of the s	. 0.55
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J_{25}		300MHz NPN Silicon Transistors 2N708, BSY27	
126			0.55
27		The state of the s	0.50
1-29		1 Timp Con o 1 o o out of	0.5
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C32		Mether Colonies and the control of t	0.5
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U38			0.5
U39 U40			0.5
U 40 U 41			0.5
C44 C42		VHF Germanium PNP Transistors TO-1 NKT667, AF11	
1 92		THE THE THE TAIL TO THE TOTAL TO THE TOTAL TOTAL TOTAL TOTAL THE T	

Code Nos, mentioned above are given as a guide to the type of device in the pak. The devices themselves are normally unmarked.

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		£p
Q 1 20 Q 2 16	Red spot transistors pnp	0.55
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	OA81 diodes.	0.55
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	IN69	0.55
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	18425R	0.55
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Q2.	2N697, 1 × 2N698	0.55
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Q30 7	npn	0.55
	npn Silicon switch transistors 2N708,	
Q31 6	Silicon switch translators 2.4400,	0.55
	npn	0.00
Q32 3	pnp Silicon transistors 2 × 2N1131,	0.55
		0.55
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034	Silicon npn translators 2N2369,	
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900	1 x 2N2905	0.55
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SN7409	0.20	0.19	0.18	SN74100	1.82	1.76	1.71
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Cathode Current (mA)	2:3	14	8	0.9 +
Numerical Height (mm)	16	13	9	point.
Tube Height (mm)	47	32	22	viewii Full d
Tube Diameter (mm)	19	13	12 wide	for all
I.C. Driver Rec.	BP41 or 141	BP41 or 141	BP47	on re
PRICE EACH	£1.87	£1.70	£1·09	

RTL MICROLOGIC CIRCUITS

Epoxy TO-5 case 1-24 25-99 100 up µL900 Buffer 38p 36p 29p µL914 Dual 2i/p gate gate 38p 36p 29p µL923 J-K flip-flop 55p 51p 49p Date and Circuits Booklet for IC's Price 8p.

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(14 & 16 Lead) TSO14 1-24 25-99 100+ 0.33 0.30 0.28 TBO16 16 pin type 0.34 0.35 BPS14 14 pin type 0.17 0.15 0.33 (low cost)

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50W pk 25w (RMS)

0.1% DISTORTION! HI-FI AUDIO AMPLIFIER

THE AL50

* Frequency Response 15Hz to ONLY 100,000-1dB. £3.58 each

★ Load—3, 4, 8 or 16 ohms.

★ Distortion—better than 1% at 1KHz.

* Signal to noise ratio 80dB.

* Supply voltage 10-35 Volts.

* Overall size 63mm × 105mm × 13mm.

Tailor made to the most stringent specifications using top quality components and incorporating the latest solid state circuitry and ALSO was conceived to fill the need for all your A.F. amplification needs.

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AP80 is especially designed to power 2 of the AL50 Amplifiers, up to 15 watt (r.m.s.) per channel simultaneously. This module embodles the latest components and circuit techniques incorporating complete short circuit protection. With the addition of the Mains Transformer MT80, the unit will provide outputs of up to 15 amps at 35 voits. Size: 68mm × 105mm × 30mm. These units enable you to built Audio Systems of the highest quality at a hitherto unobtainable price. Also ideal for many other applications including: —Disco Systems, Public Address, Intercom Units, etc. Handbook available, 10p PRICE £3-25

STEREO PRE-AMPLIFIER TYPE PA100

Built to a specification and NOT a price, and yet still the greatest value on the market, Built to a specification and NOT a price, and yet still the greatest value on the market, the PA100 stereo pre-amplifier has been conceived from the latest circuit techniques. Designed for use with the AL50 power amplifier system, this quality made unit incorporates no less than eight silicon planar transistors, two of these are specially selected low noise NPM devices for use in the input stages. Three switched stereo inputs, and rumble and scratch filters are features of the PA100, which also has a STEREO/MONO switch, volume, balance and continuously variable because when the part of t

bass and treble controls.



SPECIFICATION

Filters: Rumble (High Pass)
Scratch (Low Pass)
Signal/Noise Ratio
Input overload

Supply Dimensions

100Hz 8kHz better than -65dB +26dB +35 volte at 20mA 292mm × 82mm × 35m

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No need to buy a full kit-you need only purchase the parts you require at any one time.

All Electro Spares supplied components are new, branded products of reputable manufacturers, and carry full makers' quarantee.

We regret we cannot supply lists for projects published before this issue.
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"p.e." f.m. varicap stereo tuner

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Practical Electronics June 1973

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10K	10 ± 10K	or
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50K	50 + 50K	
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Ali E24 values 1p each plus p. & p. 7p for up to 50 Resistors and a further 2p for each additional 50 Deduct 33½% for 100 of one type or 25% for mixed orders over £1 in value.

1 W 10% Carbon Composition 2p each 22 W 10% Carbon Composition 3p each 21 W 5% Wire wound 9p each 5 W Wire wound 9p each 10 W Wire wound 10p each plus p & p. 7p for up to 25 resistors plus 1p for each additional 25



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log or lin less switch (& 1KΩ lin) log or lin with switch dual less switch dual with switch 10K, 100K & 1M	12p 24p 40p
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provide 6, 7½ or 9 volts	£4.99
(p & p 15p on all types)	

Presets

Vertic	al or Hori	zontal		C
0.1 wat	t 5p	0 25 wa	att 7p	4
100	1 k Ω	10K 12	100K 12	1M Ω
250	2.5K 11	25K 12	250K 12	2.5M
500	5K 🖸	50K Ω	500K Ω	5M Ω

Capacitors

dise cerami 01µF 18v		low voltage 0:1µF 30v	
022µF 18v	5p 5p	0·22μF 6v	5р 5р
047µF 12v	5p	0 47µF 3v	5p
ceramic pla	ite	30 V	
1000pf	10p	4700pf	10p
2200pf	10p	10,000pf	10p

Ceramic	- plate	63V (C33	3)	
1-8pf	8 2pf	33pf	120pf	
2-2pf	1 0pf	39pf	150pf	
3 3pf	12pf	47pf	180pf	
3-9pf	15pf	56pf	220pf	
4-7pf	18pf	68pf	270pf	
5-6pf	22pf	82pf	330pf	
6-8pf	27pf	100pf		
all 5p. ea	ch			

mylar film 1000pf 2p 01μF 2000pf 2p 02μF 5000pf 2p 04μF 05μF	3p 3p	-068μF -1μF -2μF	4р 4р 5р
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polystyrene 160V 10pf to 10,000pf in multiples of 10, 15, 22, 33, 47 & 68. 3p each

metallised	polyester	25	0V (C28	0)
01 µF 3p	068µF	3 ½ p	-47 µF	8p
015μF 3p	1µF	4p	68µF	11p
022uF 3p	15uF	4p	1µF	13p
-033µF 3p	-22µF	5 p	1 5 u F	20p
047µF 3p	·33µF	6½p	2-2µF	24p
metallised	polvester	4	100V (C2	(81)

metallised	polyester		400V (C2	81)
01µF 43p	047µF	6p	22µF	10p
015µF 41p	068µF	6р	33µF	14p
022µF 41p	-1μF	7p	-47µF	15p
033µF 5½p	15µF	8p		

 silvered mica 1% (>50pf) 500V

 2-2pf-820pf
 7p

 1nF-2-2nf
 9p

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 1pp

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01μF 7p 047μF 7p

022μF 7p 068μF 8p

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3000pf 5000pf 5000pf 10,000pf feed-through 1000pf

1000pf 6p 2200pf 6p 3300pf 6p 4700pf 6p

Ceramic 12KV.d.c

10pf 15pf 22pf 68pf

Heat Sinks 14p 15p 13p DIP10/2 for SL403D 15p 50p

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	Сор	perclad	Plain						
	0.1"	0.15"	0.15"						
%" %" 24"	6p 20p 24p 24p 27p 67p 90p	6p 16p (9) 21p (7) 21p (8) 27p (10) 50p 70p	10p 12p 17p 37p 52p 75p						

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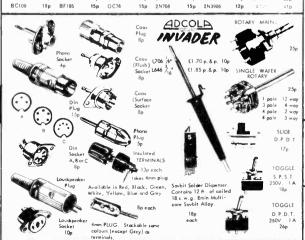
Spot-face Cutter

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No.	L.	W.	D.	Price	р. & р.
(7)	2%"	1.5%"	x 1½"	35p	8p
(8)	4"	x 4"	x 1½"	35p	8p
(9)	4"	x 2%"	x 1½"	35p	8p
(10)	4**	x 51/4"	x 1½"	40p	8p
11	4"	x 2½"	x 2"	35p	8p
12	3"	x 2"	x 1"	32p	9p
13	6"	x 4"	x 2"	50p	10p
14	7"	x 5"	x 2½"	58p	12p
15	8"	۸ 6"	x 3"	75p	I8p
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BC108	10p	BF194	15p	OC72	17p	2N706	12p	2N3905	21 p	A*	95 p
BC100	10m	BE106	150	0.076	150	25/708	150	28/2006	120		



9p 270pf 9p 270pf 9p 270pf 9p 300pf 9p 750V DISC 20pf 9p 470pf 5p 150pf 9p 1000pf 5p 150pf 9p 1000pf 5p 180pf 9p 10.000pf 9p 10.000p Mullard B Siemens Electrolytics

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-	6р		6р	6p	6р					
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Specification Projector, 150W,

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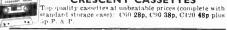
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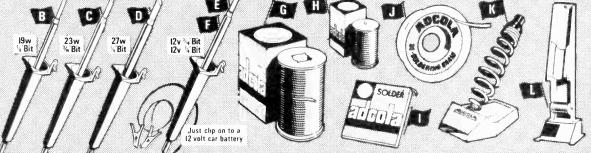


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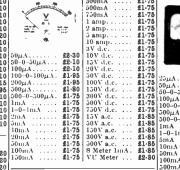
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5	Type MR.52P.	2;in. s	quare Fronts.	
)	50μA	£3.40	10V d.c £2.2	05
0	50-0-50µA	£2.85	20V d.c £2:9	95
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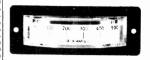
10		_		
15				
15	Type MR.65P. 3	in. ×	3 in. Fronts	
10	50μΛ	£3.70	10V d.c	£2.40
10	50-0-50μA	£3-00	20V d.c	£2.40
10	$100\mu\Lambda$	£3.00	50V d.e	£2-40
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15	5mA	£2-40	300V a.c	£2.55
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300μA	£1.95	15V a.c	£2.00	
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5mA	£1.85	VU Meter	£2.50	
10mA	£1.85	1 amp. a.c.	*£1.85	
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١	10m A	£2-15	20 amp. a.c.*	£2·15
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-	500mA	£2.15	I V C Meter	20.40

30 amp.

\$2.15 \$2.15 \$2.15 \$2.15

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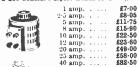
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10 amp. d.c. 10 amp. a.c. 0/20K/200K/2MEG/20 MEG. -20 +62dB. \$15. Post. 25p.

370 WTR MULTIMETER Features a.c. current ranges. 20,000 O.P.V. 0/0-5/2-5/10 / 50 /250 / 500 / 1,000 V d.c. 0/2-5/10/50/250/500/ 1,000V

0/50µA/1/10 / 100mA / 1 / 10

amp. d.c. 0/100mA/1/10 amp. a.c. 0/5K/50K/500K/5MEG/50MEG. -20 +62dB. 215. Post 25p.

RUSSIAN 22 RANGE MULTIMETER Model U437 10,000 O.P.V. A first class versatile in-instrument manufactured in U.S.S.R. to the highest standards. Ranges: 2.5/10/ 50/250/500/1,000V d.c. 2.5/ standards. Ranges: 2-5/10/ 50/250/5000/1,000V d.e. 2-5/ 10/50/250/500/1,000V a.c. D.c. current 100wA/1/10/ 100mA/1A. Resistance 300 ohms/3/30/300K/3mΩ. Complete with batteries, test leads, instructions and



V.A.T. Information

All prices quoted are subject to 10% Value Added Tax.

This must be added to the total value of goods ordered (including postage/

ROUND SCALE TYPE PENCIL TESTER MODEL T8.68



Completely portable, simple to use pocket sized tester. Ranges 0/3/30/300V a.c. and d.c. at 2,000 O.P.V. Resistance 0 20K ohns. ONLY \$1.97. Post 13p.

601 MULTI-LT 60: METER New style 20,000 O.P.V. modest New style 20,000 O.P.V. pocket multimeter. 5/25/50/250/500/2,500V d.c. 10/50/100/500/1,000V a.c. 50μA/250A. 6K/6 neg **23.75.** Post 20p.

meg ohms. -20 to +22dB.



MODEL TH-12 20,000 O.P.V. Overload protection. Slide switch selector. 0/0.25/2.5/10/50/250/1,000V d.c. 0/10/50/250/1,000V a.c. 0/50µA/25/250mA d.c. 0/3K/ 30K/300K/3 meg. -20 to +50dB. 24-97. Post 15p.

MODEL TE-300. 30,000 O.P.V. Mirror scale, overload protection 0/0-6/3/15/60/300/ 1,200V d.c. 0/6/30/120/600/ 1,200V a.c. 0/30/a/46pa/ 1,200V a.c. 0/3 60mA/300mA/600mA. 0/30µA/6mA/ 0/8K/ 80K/800K/8, meg. ohm -20 to +63dB, \$5.97, Post 15p.



MODEL PL436 20k Ω/V d.c. 8k Ω/V a.c. Mirror scale. 0·6/3/12/30 / 120 / 500 V d.c. 3/30/120/600 V a.c. 50/600μA/60/ 600μA 10/100 K/ nu0mA. 10/100 K/ 1 Meg/10 Meg \Omega. -20 to +46dB. \(\frac{26.97}{26.97}\). Post 12p. a.c. 600mA.



Decibels: to +81.5dB £8.50. Post 17p.

MODEL K228A Taut band uspension Overload pro-tection. Polarity reversing switch. 30,000 O.P.V.



0/0·5 / 2·5 / 15 / 50/250/500/1,000/2,500V 50/250/500/1,000/2,500V d.e. 0/15/50/150/ 500/1,000V a.c. 0/50μA/5/50/150/500mA/ 5A d.e. 0/3K/300K/3meg. **28-95**. Post 20p



HIOKI MODEL 700X
100,000 O.P.V. Overload
protection. Mirror scale.
0:3/0:6/1:2/1:5/8/6/1:2/30/60/
12/336/12/1:5/8/6/1:2/30/60/
0:00/11,200V a.c.
15/30/at/3/6/3/60/1:50/300mA
6/12 amp. d.c. 2K/200K/
2 Meg/20 Meg ohm —20 to
\$13-50. Post 20p.

+63dB.



MODEL C-7080 EN MODEL C-7080 EN Giant 6in mirror scale. 20,000 O.P.V. 0/0:25/1/2-5/10/50/250/ 1,000/5,000V d.c. 0/2-5/10 /50/250/1,000/5,000V a.c. 0/50µA/1/10/100/500mA/ 10 amp. d.c. 0/2K/200K/ 20 meg. -20 to +50dB. £13-95, Post 35p.

U4312 MULTIMETER U4312 MULTIMETER Extremely sturdy instrument for general electrical use, 667 O.P.V. 0/0·3/1-5/7·5/30/60/150/300/ 600/900V d.c. and 75mV. 0/0·3/1-5/7·5/30/60/150/300/ 600/900V a.c. 0/300µA/1-5/6 amp. d.c. 0/1-5/15/6 amp. d.c. 0/1-5/15/60/150/600MA/ 1-5/6 amp. a.c. 0/200 ft/38/30/

 $v_1 r_{21} r_{20} r_{$

Selected TEST EQUIPMENT

PTC-401 TRANSISTOR TESTER

Full capabilities for measuring A, B and ICO. Full capabilities for measuring A, B and EO. npn or pnp. Equally adaptable for checking diodes. Supplied complete with instructions, battery and leads. \$7.50. Post 20p.



ModelS-100TB MULTIMETER/TRANSISTOR TESTER. 100,000 O.P.V. mirror scale/overload protec-tion. 0/0-12/0-6/3/12/30/120/ interior scategoverosat pieces tion. 0/0-12/0-6/31/2/50/120/
600V d.c. 0/6/30/120/600V d.c. 0/12/600µ/ 12/300m A/12
amp. d.c. 0/10 K/1 MEG/100
MEG. —20 to +50dlb, 0-01-2
MFD. Transistor tester
measures Alpha, beta and
Lo. Complete with batteries, uctions and leads £13.50. Post 25p.



MODEL 449A IN CIRCUIT TRANSIS-TOR TESTER

Checks true a.c beta in/out. Checks Icbo. Checks diodes in/out. Checks Group 1700L. Checks SCR, etc. Beta HI 10-500 LO 2-50 Icto 6-5000 μA. 220/240V a.c. operation. £17-50. Post 25p.

KAMODEN HM720B F.E.T. V.O.M. Input impedance 10M Ω. Ranges: 0/0:25/1/2:5/10/50/

Tanges: 0/25/1/25/1/25/1/4/50/ 250/1,000V d.c. 0/25/1A/25/ 25/250MA d.c. -20 to +62dB 0/5k/500k/500k/ 5M Ω/500M Ω . \$14.95. Post 30p.



MODEL L-55 FET V.O.M.

impedance Input impedance 10 meg ohnis. 0/0-3/1-2/6/ 30/120/600V d.c. 0/3/12/ 60/120/600V a.c. 0/120µA /120mA d.c. 0/1K/100K/ 10 meg/100 meg ohnis. \$15-97. Post 25p.



CI-5 PULSE OSCILLO-SCOPE

For display of pulsed and periodic waveforms in electronic circuits VERT.

AMP, Bandwidth 10MHz.
Sensitivity at 100kHz V RMs/mm. 0-1-25;
HOR. AMP. Bandwidth 500kHz. Sensitivity at 100kHz V RMs/mm. 6-3-25; Pre-set triggered aweep 1-3,000µsec; free running 20-200,000Hz in nine ranges.
Calibrator pips. 220mm × 360mm × 430mm, 115-230V ac. operation.

Calibrator pips. 220mm × 430mm, 115-230V a.c. operation. 239, Carr. paid.

TO-3 PORTABLE OSCILLOSCOPE



sin tube, Yamp, Sensitivity
0-1V p-p/CM. Bandwidth
1-Seps-1-2MHz. Handwidth
1-Seps-3-20 P X ampsensitivity 0-9V. p-p/CM.
Bandwidth 1-Seps-8-900kHz.
Input imp. 2 meg Ω 20pF X input imp. 2 meg Ω 20pF
Time base. 5 ranges 10eps300kHz. Synchronisation.
1 Illuminated scale 140mm x 215mm x 330mm. Weight 134lb. 220/
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RUSSIAN CI-16 DOUBLE BEAM OSCILLOSCOPE

OSCILLOSCO

DMHz Pass Band. Separate Y1 and Y2
amplifiers. Rectangular
Jin x 4in C.R.T. Calibrated triggered sweep
from 0-2_ifsec to 100
milli-sec per cn. Free
running time base 50
Hz-1MHz. Built-in
time base calibrator and
amplifinde calibrator.
Supplied complete with
all accessories and ins
\$87. Carr. paid.

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instruction manual.



Transistorised Signal Generator, 5 ran-ges 400kHz-30MHz. An inexpensive instrument tions and leads.

TRANSISTORISED L.C.R. A.C.



ANSISTORISED L.C.R. A.C.
MEASURING BRIDGE
A new portable
bridge offering excellent range and
accuracy at low
cost. Ranes: R.

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MODEL TE15 GRID DIP METER Transistorised. Operates as Grid Dip. Oscillator, Absorption Wave Meter and Oscillating Detector. Frequency range 440kHz_ 290MHz in 6 coils. 500kHz_ 200MHz in 6 coils. 500kHz_ total. Size 180num × 80mm × 40 mm. 415. Post 20p. £15. Post 20p.





BELCO AF-5A SOLID STATE SINE SQUARE WAVE C.R. OSCILLATOR Sine 18-200,000 Hz: Square 18-50,000 Square 18-50,000 Hz. Square 18-50,000 Hz. Output max. +10dB (10 K ohms) operation internal batteries. Attractive 2-tone case 7½ in × 5 in × 2 in. Price \$17.50. Carr. 17p.

MODEL MG-100 SINE SQUARE WAVE AUDIO GENERATOR Range 19-220,000 Hz. Sine Wave 19 -- 100.000 Hz.

19 — 100,000Hz. Square Wave. 00t. Square Wave. Out. of the Wave. Out. of the Wave. 10V. P. to P. Size 180mm × 90mm × 90mm. Operation 220/240V a.c.



DECADE ATTENUATOR Frequency range 0-200k Hz. Attenuator0-111dB, 0-1dB step.

Impedance ohnis. Max. Input power 30dBm.

1800 Post 37p.

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LVE VOLTMETER

28 ranges. D. c. volts
13-1,500 v. a.c. volts
13-1,500 v. Resistance
up to 1,000 megohms.
200/240 v r. a.c. operation. Complete with
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Additional probes available: R.F. \$2-12; H.V.
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MODEL U4311 SUB-STANDARD
MULTI-RANGE VOLT AMMETER
Sensitivity 330 ohms/Volt
a.c. and d.c. Accuracy 0-5%
d.c. 1% a.c. Scale length

d.c. 1% a.c. Scale length 163mm.
0/300/75/p1/A1/-3/3/7-5/15/30/
73/150/300/75/mA/1-3/3/7-5
amp. d.c. 0/3/7-3/13/3/7-5/
amp. a.c. 0/3/3/7-5/150/30/
75/m3/3/7-5/15-30/
75/m3/3/7-5/15/30/
75/m3/3/7-5/15/30/
75/m3/3/7-5/m3/

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KAMODEN HM-350 TRANSISTOR TESTER High quality instrument to test Reverse Leak current and D.C. current. current and D.C. current. Amplification factor of NPN, PNP, transistors, diodes, SCR's, etc. 4" × 4½" clear scale meter. Operates from internal batteries. Complete with instructions, leads and carrying handle. £12-50. Post 30p.



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Range 0-1,000 Megohins, 500 Volt. Battery operated. Wide range clear meter 42" × 4". Complete with deluxe carrying case, batteries, instructions £19.95. Post 30p.



ARF-300 AF/RF SIGNAL GENERATOR All transistorised, compact, fully portable. AF sinc wave 18 Hz to 220 KHz. AF square wave 18 Hz to 100 KHz. Output sinc/square 10v. P.P. RF 100 KHz to 200 MHz. Output 1v. maximum. Operation maximum. Operation 220/240v. AC. Complete with instructions and leads, £29-95, Post 50p.

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Industrial quality in robust metal cases Battery operation. Volume and squelch controls. Call button and press to talk button. Telescopic aerial. Complete carrying cases. £52.50

2 channel 300 mW 3 channe 2 watt.

Pair. Post 50p.

Pair. Post 50p.

£79.50 (Note: Licence required for operation in U.K.)

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Ideal for home, office, stores, fac-tories, etc. Supplied complete with bat-teries, cable and free instructions.

2 Station, £2-97, 3 Station £5-25, Post 15p. 4 Station 26-62. Post 17p.



EMI LOUDSPEAKERS

Model 350, 13in×8in with single tweeter/crossover, 20-20,000Hz, 15W RMS, Available 8 or 15 ohms, 27:25 each, Post 37p, Model 450, 13in×8in with twin tweeter/crossover, 55-13,000Hz, 8W RMS, Available 8 or 15 ohms, 23:62 each, Post 25p.



SPECIAL OFFER! **SPEAKERS**

Matched pair of stereo bookshelf speakers. De-luxe teak veneered finish. Size: 14½in×9in×7½in. 8 ohms. 8W RMS. 16W peak. Complete with DIN lead. £12-95. Carr. 50p.

EA.41 REVERBERATION

Self contained, transistor self contained, transistorised, battery operated.
Simply plug in microphone, guitar, etc., and output into your ampli-



fier. Volume control, depth of reverberation control. B walnut cabinet. 7% in × 3in × 4% in. Beautiful

TINE-30 RECEIVER



4 Bands covering 550kHz-30MHz. BFO. Built-in Speaker 220/240V a.c. Brand new with instructions. £15.75. Carr. 37p.



UR-1A SOLID STATE COMMUNICATION

4 Bands covering 550kHz-30MHz. FET. 8 Meter. Variable BFO for SSB, Built-in Speaker, Bandspread, Sensitivity Control. 220/240V a.c. or 12V d.c. 121n×41n×7in. Brand new with instructions. £25. Carr. 37p.

SKYWOOD CX203 COMMUNICATION



Solid state. Coverage on 5 bands 200–420 kHz and 0.55 to 30MHz. Illuminated slide rule dial. Bandspread. Aerial tuning. BFO, AVC, ANL, "S" meter. AM/CW/SSB. Integrated speaker and phone socket. Operation 220/240V a.c. or 12V d.e. Size 325 × 266 × 150 mm. Complete with instructions and circuit. £32-50. Carr. 50p.

LAFAYETTE HA-600 SOLID STATE RECEIVER



General coverage z. 550 General coverage 150-400kHz, 550 kHz-30MHz. FET front end. 2 mech. filters, product detector. variable

ter. S Meter, Bandspread. RF Gain, Jöin. 89in. 88in. 18lbs. 220/240V a.c. or 12V d.c. Brand new with instructions. 250, Carr. 50p.



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4 band cover-ing 550kHz-30MHz con-bandspread on doubtinuous and electrical 30MHz components on 10, 15, 20, 40 and 80 metres. 8 valve plus 7 diode circuit. 4/8 ohm output and phone jack. 88B-CW. ANL. Variable BFO. 8 meter. 8ep. bandspread dial. IF frequency 455kHz. Audio output 1-3W. Variable RF and AF gain controls 115/250V a.c. 8ize 7in × 13in × 10in with instruction manual. 249 50. Carr. paid.



HA-10 STEREO HEADPHONE

AMPLIFIER
All silicon transistor amplifier operates from magnetic, ceramic or tuner inputs with twin stereo headphone outputs and separate volume controls for each channel. Operates from 9V battery. Inputs 5MV/100MV. Output 50MW. 25-97, Post 15p.



1021 STEREO

LISTENING STATION For balancing and gain select... loudspeakers with

for stere headphone switching. 2 gain controls, speaker on-off slide switch, stereo headphone sockets. 6in×4in×2in. 22.25. Post 15p.



5 microphone inputs each with control facilities. Battery operated. 91/m 50/m. Phono ceramic 100mV 1 meg. Output 250mV 100K.

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17+17W r.m.s. stereo amplifier with inputs for Magnetic and Crystal phono, Tuner, Tape, Aux and Tape Monitor. Outputs for two pairs of stereo speakers and tape. Stereo headphone socket, Full range of controls including loudness control, scratch filter. etc. Size 13in × 91in × 31in.
Unrepeatable offer—limited stocks!

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NIKKO TRM50 17+ 17W rms stereo amplifier, Garrard AP76 plinth and cover, Goldring G800E cartridge, pair of Linton 2 speakers and all leads.

£93.95

LEAK DELTA 30 SYSTEM



Leak Delta 30 stereo amplifier, Goldring GL75, plinth, cover and G800 cartridge. Pair of Leak 150 speakers and all speakers leads.

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TELETON SAQ206B SYSTEM

MP60

Teleton SAQ206B 8+8W amplitier BSB plinth and cover, Goldring (3800 cartridge, pair of Apollo speakers and all leads. OFR PRICE

£50.95

Ins. £1.50

AUDIOTRONIC ACP-8 8-TRACK CAR PLAYER



Attractive black and silver finish, 12V neg-carth. Slider controls for volume, tone and balance. Channel selector button with red pilot lamp. Complete with speakers, mounting brackets and instructions.

£12.50

AUDIOTRONIC ACR3500 CAR RADIO

Manual tuning of Medium and Long waves 12V pos. or neg. earth. Complete with speaker, mounting brackets and instructions.

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Push button tuning of one LW and five MW stations of your choice. 12V Pos. or Neg. earth. Complete with speaker, mounting brackets and instructions.

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MONOTONE 6750 SYSTEM



PRICE

£30.50

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Wharfedale Linton Amplifier, Turntable, Linton Linton 2 speakers and all leads.

Carr. and Ins. £1-25 £95.95 LINTON RECEIVER SYSTEM \$125. Carr. and Ins. £1.50.

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amstrad 8000 II 7+7W amplifier. BSR MP60, plinth and cover, Goldring (800 cartridge, pair of Apollo speakers and all leads. OUR OUR PRICE

£43.50





B.S.R. TD8S 8-TRACK STEREO TAPE PLAYER DECK

Integrated preamps (output 125mV) to feed into any stereo amplifier. Automatic and manual programme selector. 4 pole synchronous motor, 210/240V a.c.

OUR PRICE £12.75

Carr. 50p.

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× Z30/8tereo 60/PZ5 15.95. P. & P. 37p £15-95. 2× Z30/Stereo 69/PZ6 218.00, P. & P. 37p 2×Z50/8tereo 60/PZ8 220-25. P. & P. 37p Transformer for PZ8 \$3.65 extra Active Filter Unit Pair of Q16 Speakers £10.70 extra

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TRANSISTORISED FM TUNER

6 TRANSISTOR

HIGH QUALITY

TUNER, SIZE

ONLY 6in×4in× SIZE

TUNER, SIZE
ONLY 6inx 4inx
2jin. 3 I.F. stages,
Double tuned discriminator. Ample
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TE 1018 DE-LUXE MONO HIGH IMPE-DANCE HEADSET Bensitive, soft earpads, adjustable headband. Magnetic, Impedance 600 ohms. \$1.07 Post 15p.





KOSS BARGAIN! SP3XC STEREO HEADPHONES HEADPHONES
Extremely sensitive high
quality headphones.
Impedance 4-16 ohms.
10-15,000 Mz. Soft
sponge ear cushions.
10ft coiled lead. Brand
new and boxed.
(List £9-50). OUR PRICE £6.50 P. & P.



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Wonderful value
and excellent performance comblined. value

excellent performance combined.
Adjustable headband. 8 ohn impedance. 20-12,000
cps. Complete with
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HEADPHONES
Low cost high performance stereo headphones.
Foam rubber ear cups.
Adjustable head-band. 8 ohm impedance, 25-18,000Hz, With lead and stereo jack plug. ONLY £1-97. Post 12p.

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AA6300 AM/FM STEREO TUNER AMPLIFIER

20 + 20 watts rms. Magnetic, ceramic and tape inputs. FM 88-108 MHz. AM 535-1605 kHz. Dual stereo speaker outputs. Headphone socket. (Rec. List Price £117-46)

P. & P. 75p £61.95

AKAI AA6300 SYSTEM



£132.50

Carr. & ' Pack. £2-00

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STEREO AMPLIFIERS

RA310	15 + 15 watt	£34·95
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RA1210	60 + 60 watt	£89-95
	P & P 50n extra	

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ROTEL **RXI50 SYSTEM**



AM/FM 71 + 71 watt stereo tuner amplifier, BSR MP60, plinth and cover, G800, pair of Denton 2 speakers and all leads.

OUR PRICE £75.95 Carr. & Ins. £1.25

STEREO **PHONES** Soft leather ear cushions.

Impedance 8-16 ohm. 20-20.000Hz. 15ft coiled lead. Brand new and boxed. OUR £6.75 PRICE

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HEADTHURBE Features unique mech-anical 2 way units and fitted adjustable level, controls. 8 ohm im-pedance 20-20,000eps Complete with spring lead & stereo jack plug 47.07 Post 12n

P. & P 30p

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Improves the performance of cassette and semi-professional recorders, Reduces tape hiss by 3dB at 600Hz, 6dB at 1200Hz and 10dB for all frequencies above 3000Hz. Controls for for an irequencies above above 72. Controls for input levels and noise reduction on record and replay. 2 meters for Dolby level. Off tag monitoring. Frequency response: 20fz to 13kHz+1dB 19kHz-35dB. Size 15iin x 9in x 3iin. A.c. 200/250V.

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Top Hi-Fi quality
in library cases.
Type 5
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	TAPE	CART	RIDGES	
Type		1	5	10
40M		75p	£3.50	\$6-50
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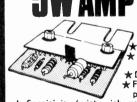
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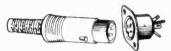
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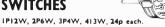
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Code	Power	Tolerance	Range	Values	I to 9	10 to 99	100 up
_		==.		available	()	ee note belo	w)
_	1/20W	5%	82 Ω-220K Ω	E12	9	8	7.5
Ċ	1/8W	5%	4-7 Ω ~470 Κ Ω	E24	i i	0.9	0.75
C	1/4W	10%	4·7 Ω=10M Ω	E12	1	0.9	0.75
C	1/2W	5%	4-7 Ω - IOM Ω	E24	1.2	i i	0.6
C	IW	10%	47Ω-10MΩ	E12	2.5	ż	1.6
MO	1/2W	2%	10 Ω – I M Ω	E24	4.7	ī	2
ww	IW	10% ± 1/20 Ω	0 22 Ω - 3 9 Ω	E12	7	ž	- A
ww	3 W	5%	Ι Ω-Ι0Κ Ω	E12	ź	ż	7
ww	7W	5%	ΙΩ-ΙΟΚΩ	Ēiž	é	ó	ă
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Ef2 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 el 2 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51 62, 75, 91 and their decades.

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BURIED FROM VIEW

The irresistible march of microelectronics means that more and more circuitry is disappearing from view. Circuitry that often is highly interesting in itself is being buried, not just metaphorically but literally, within the confines of black boxes. An oft-posed question is, should the user attempt to prise the secrets of the black boxes, or should he be content to accept them quite simply at their face value like any other circuit components? The latter is probably the most sensible thing to do. Yet sometimes it is essential to have a certain amount of inside information, though there are degrees of delving, of course.

Two articles in this issue illustrate different approaches to a fairly complex i.c., in order to suit (a) the builder of a

detailed project and (b) the designer or experimenter.

The article decribing the General Purpose Timer gives all the information needed to build this complete instrument. The i.c. upon which the design is based is treated purely and simply as a black box. So far as the constructor is concerned it is just one of the 28 circuit components employed.

This approach is perfectly satisfactory in a constructional article where a proven design is presented for the reader to copy, right down to the final detail, circuit-wise. But should the constructor wish to modify or depart from the specified design in any way, he then obviously needs to know quite a bit more about the i.c. As indeed do all those interested

in designing and experimenting for themselves.

Such requirements we have met on this occasion by a separate article which describes the technical characteristics of this i.e. and its possible applications. The device is discussed in practical terms and all relevant parameters are given that need to be taken into account when designing a system around it. Even for this purpose, this "closer look" at the i.e. need not extend to a detailed examination of the actual circuitry of the chip. A functional block diagram is perfectly adequate.

Now this brings us back to that interesting and arguable point. Is there any need to peer deeper into the anatomy of

an integrated circuit?

Microcircuit manufacturing processes allow quite unusual innovations in the creation of circuit elements. For example, it is quite normal for the active pn junction to be employed as a maid-of-all-work, including serving in the humble role of a passive element. Not surprisingly, any normally well-recognised classic circuit configuration becomes less obvious, maybe entirely unidentifiable, to the average eye scanning the equivalent circuit diagram of some monolithic device.

The internal circuit of an integrated device is not likely to be of any practical value to the general user. Admittedly it can be of academic interest and even offer some reward as a technical brain teaser. But we imagine most users will be content for the i.c. to remain an inscrutable black box. The external discrete circuitry which these devices invariably stimulate and within the web of which they become enmeshed, singly or severally, provides enough for designer, experimenter, or constructor to concentrate upon.—F.E.B.

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"HE GENERAL PURPOSE timer discussed here is based on 555 i.e. timer from Signetics which is the subject of the feature commencing on page 514. The unit is constructed in an Eddystone die-cast box of external dimensions $7.39 \times 4.703 \times 2.062$ inches. There is enough space for a mains power pack or a battery to be included inside the box. Whilst construction of the prototype is described in detail, many variations are possible to suit individual needs.

CONTROLS

The control arrangement is shown in the photograph. The eleven-position switch S1 is used to select the first digit of the desired number of seconds, whilst the second digit is selected by the potentiometer VRI. The range switch, S2, is arranged so that the total time selected by S1 and VR1 can be multiplied by 0·1, 1 or 10.

SELECTION OF DELAY

If a time delay of 860 seconds is required, S1 is set to "80", VR1 to "6" and S2 to "×10". A delay of 1:1 seconds can be obtained by setting S1 to "10", VR1 to "1" and S2 to "x0-1". Delays of less than I second can be obtained by setting S1 to "0" and S2 to "x01". The maximum delay with this unit is about 1100 seconds (18 minutes 20 seconds).

In general the delays are accurate to a few per cent.

OPERATION

The timing period commences at the instant the START button is released after it has been pressed. The internal relay closes at this time, but automatically opens again at the end of the desired delay. It may be found that the timing starts when the START button is pressed down owing to contact bounce. This will not matter provided that the button is pushed quickly and then released.

If the START button has been pressed and one does not wish the timing operation to continue, the RESEI button may be pressed. A fresh timing operation can then commence from the beginning when the START button is pushed again. The use of the reset facility prevents having to wait or alter the timing setting after commencing a fairly long timing operation and wishing to terminate it.

THE CIRCUIT

The circuit of the timer is shown in Fig. 1. It is essentially the same as the basic circuit of Fig. 1 of the article commencing on page 514, but some refinements have been added.

The resistance of R_A has been replaced by R1 to R11 in series with VR1. The resistor R11 is of low value and is included to prevent a fairly high current (about 60mA) from flowing to pin 2 of the 555 if SI and VRI should both be set for zero resistance. The resistors around S1 are each 1M\O and their total value (as set by S1) is added to the setting of VRI.

THE TIMING CAPACITORS

The timing capacitor is selected by S2a. It cannot be emphasised too strongly that the electrolytic capacitors C1 and C2 must be good quality components which have a low leakage current. The writer would have expected electrolytic capacitors with a working voltage of about 30V to pass a lower leakage current when 12V is applied to them than similar capacitors with a working voltage rating of 15V. However, measurements of the leakage current of a number of capacitors seems to indicate that this may not be the case.

An additional range of "x100" could have been added using a 1,000 µF electrolytic capacitor to give time delays of up to 11000 seconds (over 3 hours), but a capacitor selected for low leakage current would probably be required. A 4-way 2-pole switch

would then be required for S2.

COMPONENTS . . .

Resistors

1M Ω , 5% (preferably 2% R1 to R10

4·7kΩ, 10% R11 22kΩ, 10% R12, R13

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Capacitors

C1 100μF, 15 to 30V, electrolytic

C2 10µF, 15 to 30V, electrolytic

C3 1µF, 63V, polyester (WIMA or RS Components Ltd.)

Switches S1

11 way, 1 pole rotary switch

3 way, 2 pole rotary switch

Single pole push-to-make switches S3, S4

(RS Components Ltd.) Single pole, single throw toggle switch **S5**

Semiconductors

ICI NE555V integrated circuit (SDS Components Ltd., Gunstore Rd., Hilsea Trading Estate, Portsmouth, Hants.)

DI OA47 Germanium diode

Miscellaneous

RLA MS1B 12V micro-switch relay (Keyswitch

Relays Ltd.)

8 pin dual-in-line socket (RS Components Ltd.) Eddystone die-cast box $7.39 \times 4.703 \times 2.062$ in

external dimensions

1 Lektrokit board

4 BA bolts 11 to 11 in long

12 4BA nuts

6 8BA nuts and bolts

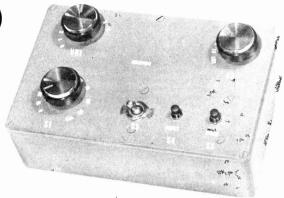
TIMER

By J.B. DANCE M.Sc.

Capacitors of a large value have wide tolerances. In addition they pass different leakage currents. Three trimmer potentiometers (VR2 to VR4) are therefore incorporated in the circuit so that each range can be calibrated. S2a and S2b are on the same switch wafer and select the appropriate trimmer potentiometer together with the appropriate capacitor.

CALIBRATION

Adjustment of the preset potentiometers VR2 to VR4 alters the control voltage applied to pin 5 of the 555 timer. When each potentiometer has been adjusted so that the delay is correct at one point



on each range, all of the other delay values should be approximately correct. It is best to make the final adjustments to each potentiometer at a setting near to the maximum delay for the range concerned where the timing errors are greatest. A stop watch is desirable (but by no means essential) for the accurate calibration of the x0·1 range.

The tolerance of R1 to R10 in the prototype was a nominal ±5 per cent. The accuracy of the delays was found to be a few per cent, as would be

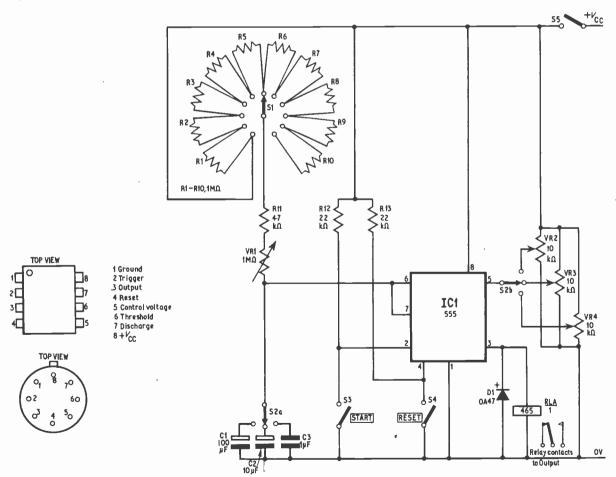
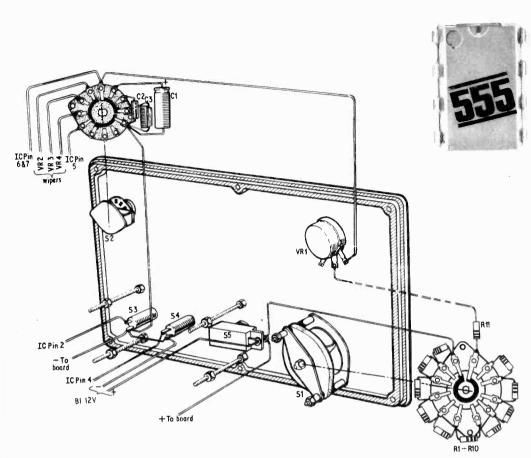
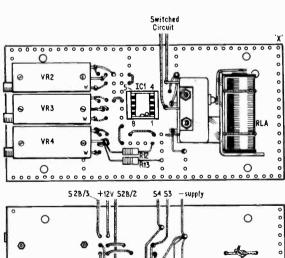


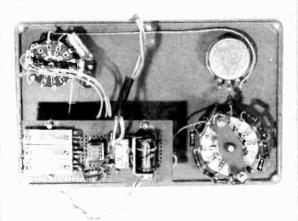
Fig. 1. Circuit diagram of the general purpose timer and the pin connections for the 555 i.c. in both DIL and TO99 case versions





Hole made to relieve

Fig. 2. The front panel layout with switch wiring detached for clarity. The top and bottom views of the circuit board showing disposition of components and interwiring is shown on the left



The completed timer with board and controls mounted in position

0

528/1

52B(w)

SZA(w)

expected. An error of 30 seconds in 1,000 seconds is therefore reasonable. Resistors of closer tolerance should be used if it is desired that the timing delays

shall be as accurate as possible.

The tolerance of the capacitors C1 to C3 can be quite wide, since the potentiometers VR2 to VR4 allow a timing adjustment of over 10:1 on each range. For example, when the time was set to 20 seconds on the x1 range, adjustment of VR3 allowed any delay between 3 and 43 seconds to be obtained with the capacitor C2 used by the writer.

The 26-turn trimming potentiometers enable a very accurate setting to be obtained. The cheaper "skeleton" preset potentiometers could be tried, but it would then be necessary to connect fixed resistors between each side of these potentiometers and the power supply lines to make their adjustment less

In selecting the component values, the factor of 1.1 in equation 1 of the previous article was neglected so that component values with round whole numbers could be employed.

START AND RESET

The trigger pin 2 and the reset pin 4 are normally biased to the $+V_{\rm cc}$ potential through R12 and R13 respectively. This prevents unwanted triggering of the START or RESET functions by spurious voltage peaks.

It is especially important that pin 2 should normally be at the $+V_{cc}$ potential, since the triggering action is extremely sensitive. It has been mentioned earlier that the circuit exhibits spurious triggering at the end of each delay period if R12 is omitted; the relay then fails to open at the end of the delay

THE RELAY

The relay is connected so that it is normally open and is energised only during the timing delay. This minimises the current consumption of the circuit.

The relay used is a miniature micro-switch 12V, 4650 relay, type MS1B which is available (through retailers) from Keyswitch Relays Ltd. The tolerance of the operating voltage is 20 per cent, so it will function with a coil operating voltage of 9.6 to 14.4V. The writer has found that a relay of this type will operate with less than 7V across the coil and the prototype circuit functions satisfactorily from a 9V battery. However, a 12V supply is ideal.

The MSiB is a printed circuit relay, but other versions are available including the totally enclosed plug-in version type MS1P. The MS1B can switch up to 5A at up to 250V in a.c. circuits; this is adequate for most purposes, since it can switch over a kilowatt at the normal mains voltage. In d.c. circuits the maximum recommended currents are 2.5A at up to 24V, 0.25A at up to 100V and 0.2A at up to 250V. Higher alternating voltages can be switched, since these voltages fall to zero many times per second and any arc which is formed is then broken.

THE DIODE

The diode in parallel with the relay coil shorts out the back e.m.f. transient voltages developed when the current ceases to flow through the coil. This prevents damage to the 555 timer.

It should be noted that either the type of diode specified or a similar gold bonded germanium diode should be used. The writer has tried a number of other types of diode in the circuit but many of these do not suppress the transient voltage across the relay coil adequately enough to prevent this pulse from re-triggering the circuit. If, therefore, another type of diode is employed and the relay does not open at the end of the delay period, re-triggering is almost certainly the cause.

CONSTRUCTION

All of the components are mounted on the lid of the die-cast box, since this provides maximum accessibility and ease of adjustment of the trimming potentiometers. The switch S1 is mounted in the position shown in Fig. 2 and the resistors R1 to R10 inclusive are mounted directly onto this switch. VR1 is mounted near to S1 with R11 mounted directly between these components.

The timing capacitors C1 to C3 mount directly onto the switch S2. The push-button switches S3 and S4 automatically open when they are released. They are mounted in line with the on/off switch S5 under the circuit board containing the other com-

ponents.

CIRCUIT BOARD

The circuit board used is a piece about 4×1.7 in sawn off a Lektrokit board with holes spaced at 0-lin intervals into which metal pegs can be inserted. The board is supported by four 4BA long bolts at a little over one inch under the lid of the box clearing S3, 4 and 5.

The three trimmers VR2 to VR4 are mounted using 8BA bolts on one edge of the board where they

can be easily adjusted as in Fig. 2.

An 8-pin dual-in-line i.c. socket mounts on the board, the solder on the pins holding it in position. R12 and R13 mount by the side of this socket. Care should be taken to ensure that the NE555V is always inserted with the correct orientation in this socket, since it is symmetrical and will fit in either way.

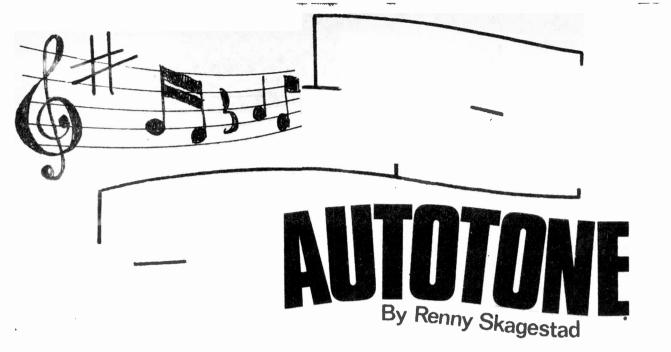
It is convenient to mount the relay on its side, cutting holes in the circuit board to accommodate any small projecting parts of the relay. In fact the wire connections to the relay can hold one side of this component to the board, the coil side being fixed by passing pieces of thin wire around the coil and tying the wire under the board.

The only component placed under the board is the diode in parallel with the relay coil. The wires connected to the board are made long enough to allow the board to be easily removed and turned over if it should require attention at any time.

USES IN PHOTOGRAPHY

The circuit is suitable for use as an enlarger timer in photography. The relay contacts are used to control the power to the enlarger lamp directly. The lamp is illuminated when the START button is released and is automatically switched off after the required time. This is much more convenient than having to estimate times or to peer at a watch in a dark room.

The circuit can also be used for timing the development of plates and films. The relay contacts are used to operate a buzzer or a small bell at the end of the required time.



The making of music automates has been an attractive challenge to inventors for many years. In both the mechanical and electronic versions a variety of ingenious ideas have been applied. However, in general most of the systems are complex and expensive in their realisation.

The circuit to be described has none of these drawbacks.

BLOCK DIAGRAM

A block diagram (Fig. 1) indicates the operating principles of the Autotone. Here an astable pulse generator of fixed frequency switches on and off the supply feeding a shift register of four so advancing conduction from one stage to the next in sympathy with the pulse input.

Each conducting stage in the counter is preadjusted to provide a particular bias voltage to a tone generator so that variable pitch tones are available. The sequenced tones produced are further amplified and fed to loudspeaker.

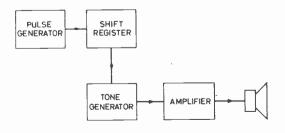


Fig. 1. Block diagram of Autotone

COMPONENTS . . .

Resistors R1 R2 R3 R4 R5 R6 R7 R8 R9 R10-R11 R12 R13-R14 AII ½ wa	220 Ω	R15 R16 R17 R18 R19 R20 R21 R22 R23 R24 R25 R26	$\begin{array}{c} \textbf{220} \ \Omega \\ \textbf{1k} \ \Omega \\ \textbf{56k} \ \Omega \\ \textbf{56k} \ \Omega \\ \textbf{470} \ \Omega \\ \textbf{120} \ \Omega \\ \textbf{820} \ \Omega \\ \textbf{330} \ \Omega \\ \textbf{33k} \ \Omega \\ \textbf{10k} \ \Omega \\ \textbf{1.5k} \ \Omega \\ \textbf{220} \ \Omega \\ \end{array}$
Potentiometers VR1-VR3 220k \(\Omega\) vertical presets			
Capacitors			
C1 C2 C3 C4	0.5µF to 20µF (sec 0.01µF elect. 9V 100µF elect. 9V 10µF (5 off) 0.01µF (5 off) 0.1 to 0.01µF 100µF elect. 9V 100µF elect. 9V 100µF elect. 9V 200µF elect. 9V 200µF elect. 9V	e text) elect.	15V
Diodes D1-D9 ISJ50 (9 off)			
Transistors TR1, TR3, TR4, TR6, TR8, TR10 TR2, TR5, TR7, TR9 TR11, TR12 TR13, TR15, TR16 TR14 Miscellaneous BY1-BY2 9V batteries (PP9) (2 off) Veroboard as required, LS1, 5in 3 Ω loudspeaker			

PULSE GENERATOR

The complete circuit of the Autotone for four note generation is shown in Fig. 2. This can be increased for an eight note diatonic scale or twelve note chromatic scale by adding extra stages to the ring counter.

The pulse generator is made up of transistors TR1 and TR2. When voltage is first applied C1 starts to charge through the emitter of TR1. This transistor switches on TR2 and a regenerative switching action takes place until the voltage on C1 exceeds the base voltage of TR1 when conduction ceases.

C1 now discharges through R1 until conditions are right for the conduction cycle to start again.

The pulsating voltage at TR2 emitter switches TR3 on and off which itself is in series with one of the supply rails to the ring counter so this is also switched.

SHIFT REGISTER

When power is first applied to the shift register none of the stages normally conducts. But when C4 is switched in it starts to charge thereby giving TR5 the necessary base current drive to make the stage

CIRCUIT DIAGRAM

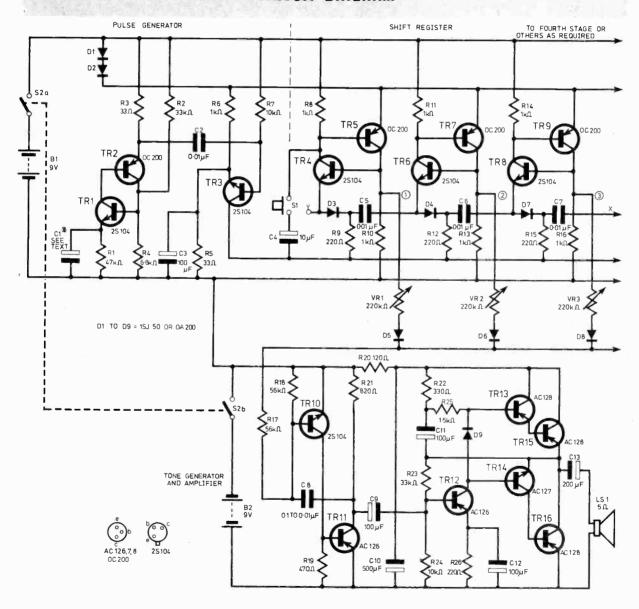


Fig. 2. Circuit diagram of Autotone. Only three stages of the shift register are shown but other identical stages can be added to as required

(TR4/TR5) conducting and an output voltage of about 8V appears at the top of R10. For this condition to exist TR3 must also be switched on.

When TR3 is pulsed off the first stage of the register ceases to conduct. However, since C5 is charged the bias conditions are correct for TR6 to conduct and hence TR7 when TR3 next switches on. This means that the 8V output now advances one stage to appear at the top of R13.

After the fourth shift, to maintain a cycling output, point X can be connected to point Y.

TONE GENERATOR AND AMPLIFIER

The outputs taken from the shift register via the presets (VR1-VR4) are used to give base bias to TR10. Both this and TR11 are wired as a high gain amplifier with C8 providing regenerative feedback for oscillation.

The frequency is partly determined by the base voltage at TR10 which depends in turn on the resistance of the presets with the given value of C8 this frequency can only be varied within a certain range.

WIRING DETAILS

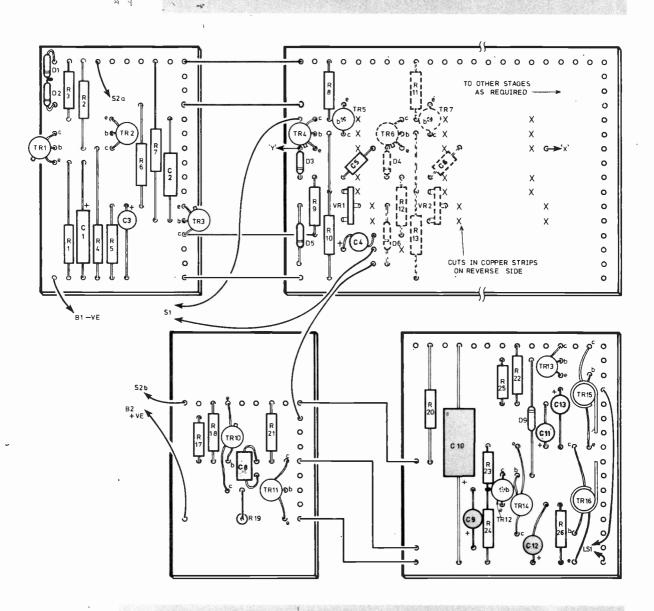
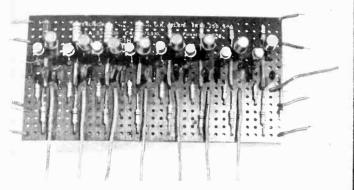


Fig. 3. Assembly and wiring details of Autotone Veroboards



Shift register board containing the eight stages for a diatonic scale. In this example the output potentiometers and diodes have been omitted and transferred to another board

To change this, values of capacitor from 0.1 to $0.1 \, \mu \mathrm{F}$ can be tried.

Tones generated are fed to a four transistor power amplifier. For experimental work the amplifier can be left out and an 80 ohm loudspeaker substituted for R21. In the circuit diagram two power supplies are used to avoid interaction.

CONSTRUCTION

The prototype unit was constructed on separate Veroboards as in Fig. 3. Of course, there is no reason why it could not be assembled on a single board, but it does make testing and servicing simpler to make the circuit elements modular.

TESTING

To test the pulse generator connect a crystal earpiece between the base of TR3 and negative line where a ticking should be heard if all is well.

By connecting a 10V meter (v.o.m.) across one of the output resistors R10, R13, etc., the action of the shift register can be checked by observing that the needle indicates about 8V at regular intervals with X and Y connected.

The tone generator can be tested by connecting a crystal earpiece across transistor TR11, and switch on. A lively oscillation should be heard.

COMPOSING MELODIES

If the shift register is extended to embrace eight or twelve notes there is no need to use all the outputs as the music will become uniform and dull. The melody will be more natural when some of the outputs are left out so achieving breaks or pauses.

In order to make certain tones last longer than others, the presets of two stages may be adjusted to produce the same tone giving the effect of sustain. The speed at which the shift register advances depends upon the value of C1. This may be altered to suit the melody being composed.

If C1 is reduced to around 1μ F a special effect is achieved which makes the Autotone suitable as an exciting doorbell alarm.

Electronic NOSE

Natural gas, petrol, butane, propane, alcohol and smoke, the Electronic Nose can "smell" them all. Its uses range from a simple breathalyzer to amuse your friends, to a detector for tracing lethal concentrations of odourless gas in the bilges of boats.

This easy to construct device uses a new gas detector and incorporates an audio oscillator for easy interpretation of readings. It also uses battery power for complete portability.

Twin Power Supply

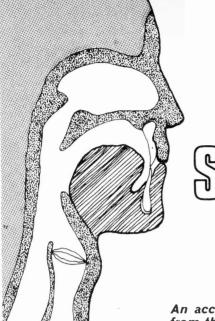
A high quality power supply unit is invaluable in the workshop for testing and evaluating equipment. This design features excellent regulation and extremely low ripple voltage. It also has a feature normally restricted to the most expensive commercial units, an accurate current limit facility, allowing current to be held constant at anything up to the maximum of one amp.

Lighting Controller Unit

A simple thyristor light control circuit provides a variety of switching actions to give effects with lighting circuits including anything from Xmas lights to discotheques. Simple in construction and operation.

ELECTRONICS

JULY ISSUE ON SALE JUNE 8, 1973



An account of the attempts to build a speaking machine from the eighteenth century to the present day.

By A. V. Flatman (North Staffordshire Polytechnic)

HE synthesis of speech is a subject that has preoccupied humanity for a long time. It has been legendary precept and mystery since the eighteenth century. Today, again, this question conserves a mysterious aspect, because speech is concerned with human activity.

Research contributions date back almost two centuries to the "talking machines" of Kempelen and Faber, who, as pioneers, established a great understanding in the acoustic structure of speech, a science that later became known as phonetics. It was unfortunate, however, that the early synthesisers were limited to a purely mechanical and inefficient construction.

The mid-twentieth century arrival of electronics proved to be the technology all researchers were waiting for. The tape recorder and oscilloscope made possible the acoustic spectrogram or "voice print" for analysis and synthesis of speech waveforms, whilst computer automation has effectively "untied" the operator's hands.

SPEECH ANALYSIS

Before the various techniques of synthesising speech are examined, we must first analyse its acoustic formation.

All acoustical signals can be represented in a sonogram or acoustic spectrogram; this is simply a graph of frequency plotted to a base of time, with acoustic intensity shown as line thickness. Fig. 1 gives examples of sonograms for elementary sounds.

When a complex sound enters a resonator, the harmonics near the resonant frequency will be amplified to add a certain "colour" to the sound. This frequency zone is known as a formant. Fig. 2 shows the effects of fixed and variable resonant frequencies on several types of acoustical signal.

English spoken in Britain can be broken down into approximately forty fundamental sound units called phonemes, the acoustic formation of which may be demonstrated with the aid of the human speech apparatus shown in Fig. 3.

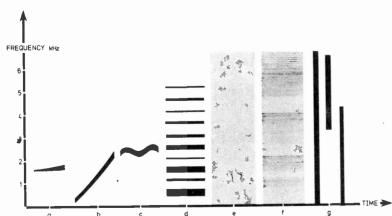


Fig. 1. Sound spectrogram for some elementary sounds

- (a) Simple sound of increasing amplitude
- (b) Simple sound of increasing frequency
- (c) Vibrato
- (d) Harmonic sound
- (e) White noise
- (f) Coloured noise
- (q) Noise impulses

The human speech system comprises several resonant cavities which have the ability of superimposing variable formants upon the complex sounds issued from the vocal cords. One particular vocal sound, with suitable formants to produce certain yields of frequency, will acoustically represent the vowels.

Consonants, on the other hand, are a little more complex in formation. Some consonants are purely breathing noises, from the roof (sound "CH"), teeth (sound "S") or in between the teeth and lips (sound "F"); whilst others may be explosive noises produced by sudden releases, from the roof (sound "K"), teeth (sound "T") or lips (sound "P").

Spoken messages may be analysed pictorially with the aid of the acoustic spectrograph. However there are two characteristics of speech which present difficulties. Firstly, speech has a continuous appearance and is rather difficult to segment into its various phonemes. Secondly, the useful or semantic information in speech is modified somewhat by a "tone of speech" reflecting the speaker's mood.

THE KRATZENSTEIN RESONATORS

One of the earliest efforts at speech synthesis was made by Kratzenstein at the Imperial Academy of St. Petersburg in 1779. This mechanical speaking machine comprised a set of acoustic resonators somewhat similar in size and construction to the human mouth. With a vibrating reed similar to that used in the harmonica, he mimicked the vocal chords by interrupting the airstream in each of the resonators to synthesise the five vowels with tolerable accuracy.

THE KEMPELEN SPEAKING MACHINE

Kempelen's speaking machine, which was made in Vienna in 1791, is shown in Fig. 4. One can observe the relative simplicity of the results of Kempelen's research into the problems of speech synthesis. After some practice, the operator could manipulate this machine to synthesise a somewhat limited vocabulary. Whistles are used to give the "S" and "CH" sounds and a reed to give the "R" sound, whilst the rubber mouthpiece is controlled alone to generate the vowels and in conjunction with the bellows to produce a variety of explosive consonants.

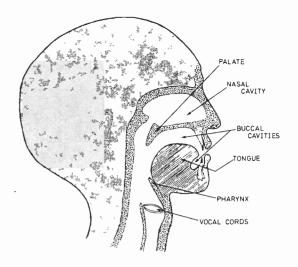


Fig. 3. Diagram of the human speech apparatus

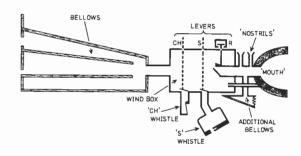


Fig. 4. Cross sectional diagram of Kempelen's speaking machine

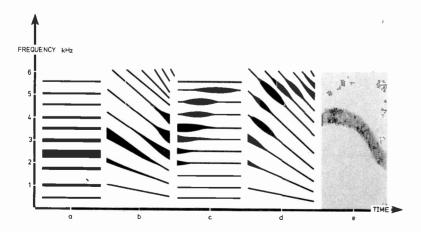


Fig. 2. Some different ways of representing formants

- (a) Fixed formant on fixed spectrum
- (b) Fixed formant on variable spectrum
- (c) Variable formant on fixed spectrum
- (d) Variable formant on variable spectrum
- (e) Variable formant on continuous spectrum (noise)

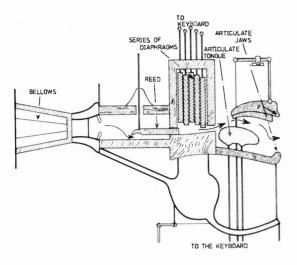


Fig. 5. Cross sectional diagram of Faber's speaking machine

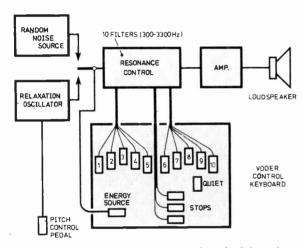


Fig. 6. Diagram showing the basic principles of the Voder

THE MACHINE OF FABER

Faber, professor of mathematics at Vienna, finished a speaking machine in 1835, which created a vast amount of interest and curiosity throughout Europe. Faber put into practice many of Kempelen's theoretical suggestions for improvement, which subsequently contributed to the success of the new machine. The strongest aspect of Faber's research was, however, his consideration of the operator in the use of a keyboard selection technique as shown in Fig. 5. For a century afterwards, many scientists attempted to expand upon the theory of Faber, unfortunately without much success.

THE VODER

The first electrical analogue speaking machine was built by Dudley in 1939. The Voder (VOice DemonstratOR), shown in Fig. 6, comprises ten filters whose passbands are uniformly distributed between 300 and 3,300Hz. The filters are stimulated with white noise or with a signal rich in harmonics. A keyboard controls the frequency and amplitude of the stimulus as well as the filter selection. Manual operation of the machine was again rather complex, and although the phonemes were relatively accurate in reproduction, they could not be linked efficiently to synthesise continuous speech.

Nevertheless, the Voder represents a new and promising approach in speech synthesis. We will examine in sequence the four types of synthesisers which are used at present: the Vocoder; the Analogue Synthesiser; the Playback; and the Units of Verbal Response.

ELECTRONIC SPEECH SYNTHESIS

Dudley's invention of the Voder marked the beginning of several decades of intense research on both sides of the Atlantic. The new understanding of information theory demonstrated that the telephone bandwidth (300 to 3,300Hz) was exceedingly large for transmitting the semantic information of

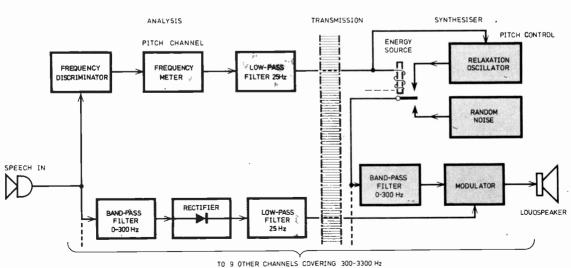


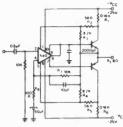
Fig. 7. Block diagram showing the principle of Dudley's Vocoder

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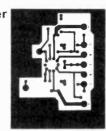
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> Practical Electronics June 1973

speech. The telephone allows a data transfer rate of 20,000 binary bits per second, whilst that of the semantic content of speech rarely exceeds 100 bits per second. Speech synthesis then discovered a new field of application in communication system bandwidth reduction, by suitable phoneme encoding.

THE VOCODER

Dudley expanded upon the principle of his Voder to develop a system which greatly reduced the bandwidth of a communication channel. The Vocoder (VOice CODER) analyses incoming speech by initially splitting it into ten frequency bands by suitable filtering and then obtains the modulation "envelopes" of each band by rectifying and low-pass filtering.

Fig. 7 shows how ten separate channels are used to link the analysed or "encoded" speech to the corresponding filters in a Voder. Incoming speech pitch is sensed in a similar way and an eleventh channel is used to select the appropriate energy

source in the Voder system.

The Vocoder transmission bandwidth is within 300Hz (11 channels each of 25Hz) and represents a great saving in this respect. Synthesis takes place in the established Voder system, which is now operated electronically to reproduce intelligible speech with tolerable continuity.

THE ANALOGUE SYNTHESISER

The principle of this machine consists more of simulating the function of the phonetic organs, rather than synthesis via analysis. The pharynx and buccal cavities of the human speech apparatus are firstly considered as the three damped resonators shown in Fix. 8; X, Y and Z respectively. Sound is emitted from the vocal cords into the first resonator X, then via the narrow passages to the resonators Y and Z, to result in the emergence of the combined sound, Un + Um.

Fig. 8 also shows the electrical analogue of the resonators X, Y and Z, where electrical properties of resonance, damping, etc., are matched to the properties of the acoustic resonators. The circuit will then act in a similar way to the phonetic organs if stimulated by the appropriate electrical signals, Ug.

Bell Telephone Laboratories have recently developed an extremely useful aid in the study of speech synthesis. It is a simulator in which the value of each component in the electrical analogue is controlled by a computer, at the same time a crosssection of the corresponding phonetic organ state is

displayed on a screen.

Programmes have subsequently been written to electrically simulate the phonetic organs with some degree of success. Unfortunately, the rate at which the electrical circuit will deal with successive phoneme simulation is not fast enough for continuous synthesis; however, this technique appears to be very interesting for the physiological study of phonetics.

THE PLAYBACK

Acoustic engineers have demonstrated that all information in speech may be detected on the acoustic spectrogram. Conversely, speech may be reconstructed by careful manipulation of a suitable acoustic spectrogram, as seen in the playback synthesiser. The Icophone, whose principle is shown in Fig. 9, is the most recent synthesiser to use the playback technique.

The principle of the Icophone is quite simple. Opaque zones of the acoustic spectrogram represent zero speech content and allow transmission of a narrow beam of light to the corresponding photoelectric cells. Each cell, in turn, represents a spectrographic frequency zone and controls the continuity between each of the 44 oscillators and the blender. Intonation of the synthesised speech is controlled by the adjustment of each oscillator output level and the switching thresholds.

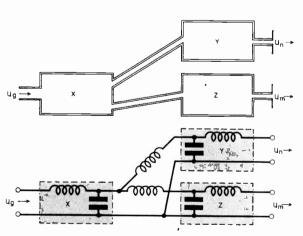
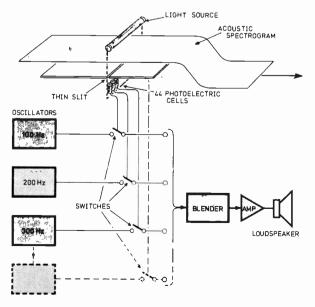


Fig. 8. Analogue synthesis of phonemes by simulation of three formants

Fig. 9. Diagram showing the principle of the Icophone



Accurate feeding speeds of the acoustic spectrogram result in intelligible synthesis of phonemes and prepared speech. The ultimate and most flexible system of "real time" synthesis from a library of prerecorded phonemes, however, proves to be somewhat inefficient due to the time taken in phoneme selection.

To overcome the inherent drawbacks of electromechanical synthesisers, digital computers have been used to synthesise speech by the playback principle. Semantic information of the basic library of 40 phonemes is digitised in the frequency, time and amplitude domains and held in a digital store (corestore, magnetic drum or disc).

The computer is then programmed to "empty" the contents of certain parts of the store sequentially with sub-millisecond speed, to synthesise a more efficient or continuous speech. Computer synthesis techniques have produced the most accurate subjective results to date, but unfortunately the amount of computing power required is both demanding and costly.

VERBAL RESPONSE UNITS

Certain large makers of calculating machines have commercialised some units of verbal response. These machines allow a computer to answer, in verbal form a question put by the user in coded form, using, for example, the ordinary telephone dial. This system has an immense future in some applications.

Words and phrases are prerecorded on tape and "linked" by computer programme to form a verbal output of a wide range of systems. Having a somewhat limited flexibility one would usually find this type of synthesiser performing specialised tasks of varying complexity, from the "speaking clock" to a verbal instruction generator in an aircraft flight simulator.

THE FUTURE

We have progressed from the relatively inarticulate, mechanical experiments of Kempelen and Faber to the "talking computers" of today. The recent appearance of electronics and the computer have somewhat altered the design philosophy of speech synthesisers. Dexterity, a prominent operator requirement of the past, has been replaced by the modern tool of computer programming.

Nevertheless there remains much room for improvement of today's synthesisers. Imagine the potential of an efficient, flexible, compact and inexpensive "text to speech converter"! It not only has applications in business and education, but think of the way in which it could make the lives of dumb or blind people that bit more bearable.

Speech recognition, an associated field of speech synthesis, is still relatively unconquered. The automatic recognition of speech is a much more difficult task due to the fact that a machine must cope with the enormous amount of variation in human speech, whereas synthesis of different pronunciations of words is normally not required. Obvious differences occur in accent, but considerable variety in pronunciation is present even with the same speaker. Further problems arise in recognition because the speech wave cannot easily be segmented into appropriate words or phonemes.



ELEMENTS OF LINEAR MICROCIRCUITS
F. D. Towers, M.B.E., M.A., B.Sc., M.I.E.R.E.
Published by Iliffe Books (Butterworth & Co. Ltd.)
108 pages, 6in × 8½in. Price £2.80

MAJOR growth area is represented by linear microcircuits. After a late start, following the digital devices that really established the microelectronics industry, the linear form of integrated circuit is now well established and the future holds promise of ever increasing varieties of circuit for all kinds of application.

Therefore, this book is welcome. It is based on a series of articles published in *Wireless World* and provides a compact, readable digest of the subject. Because the rate of development in this field is so rapid it cannot include the very latest developments; nevertheless it is a very useful reference book. It includes practical information in the form of lists of manufacturers, component coding methods, and advice concerning the handling and use of these devices.

A chapter is devoted to each of the major categories of device, e.g. a.f. amplifiers, operational amplifiers, r.f. & i.f. amplifiers, and voltage regulators. The circuit configurations of many typical commercial devices are illustrated and described in considerable detail.

D.D.R.

ELECTRONICS—A COURSE BOOK FOR STUDENTS G. H. Olsen, B.Sc., M.I.E.R.E., M.Inst.P. Published by The Butterworth Group 351 pages, 6in × 8½in. Price £2.60

WITH the continuing movement of electronics into almost every other walk of life it is becoming increasingly important for students of all disciplines to be aware of the vagaries of the art. Thus any volume which eases this understanding is to be welcomed and the present document is specifically designed with this end in mind.

In fact the book is a shortened version of the successful "Electronics: A General Introduction for the Non-Specialist" and it manages to take the reader through basics right up to the complexities of integrated circuit operation without any noticeable reference to mathematics and abstruse formulae.

Included, of course, are suitable references to matters of measurement and suitable indicators, power supplies and their construction, amplifiers and oscillators.

Wherever possible the author has illustrated his text with sufficient clarity for the competent to construct many of the items discussed. Indeed, much of the equipment discussed by way of application details is culled from component manufacturers' application notes.

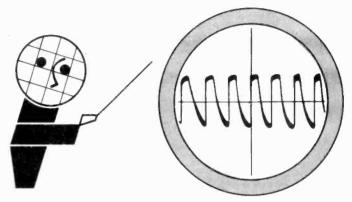
The author is principal lecturer in the Department of Physics and Physical Electronics at the Newcastle upon Tyne Polytechnic and he is concerned with the electronic content of C.N.A.A. degrees, organising post-graduate courses in electronics, and with consultancy in the field of electronic design.

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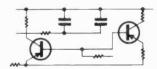
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1	2%	$\Omega MI - \Omega \Omega I$	E24	3-5p	3р
í	10%	$1\Omega = 3.9\Omega$	E12	i p	0.8p
I	5%	4-7Ω-IMΩ	EI2	1 p	0.8p
- 4	10%	ΙΩ-10Ω	E12	6р	5-5p

Quantity price applies for any selection. Ignore fractions on total order.

DEVELOPMENT PACK

0.5 watt 5% iskra resistors 5 off each value 4.7 Ω to IM Ω . E12 pack 32S resistors £2.40, E24 pack 650 resistors £4.70.

POTENTIOMETERS

Carbon track $5k\Omega$ to $2M\Omega$, log or linear (log $\frac{1}{4}W$, $\lim \frac{1}{2}W$). Single, 12p. Dual gang (stereo), 40p. Single D.P. switch 24p.

SKELETON PRESET POTENTIOMETERS

Linear: 100, 250, 500 Ω and decades to SM Ω . Horizontal or vertical P.C. mounting (0-1 matrix)

mounting (0.1 matrix).
Sub-miniature 0.1W, 5p each. Miniature 0.25W, 6p each.

T	R	Α	Ν	S	IS	Т	0	R	S	

AC107 15p AF125 20p BD132 75p OC28 50p 2N3702 13p AC126 12p AF125 20p BD133 75p OC35 50p 2N3703 12p AC127 12p AF126 20p BF115 25p OC35 50p 2N3703 12p AC128 12p AF139 32p BF173 20p OC44 12p 2N3705 12p AC131 12p AF188 32p BF173 20p OC70 12p 2N3706 11p AC132 12p AF188 40p BF178 32p OC70 12p 2N3707 12p AC187 21p BC108 40p BF178 32p OC70 12p 2N3707 12p AC187 21p BC108 9p BF180 32p OC71 12p 2N3708 10p AC188 22p BC108 9p BF180 32p OC72 12p 2N3709 11p AC188 22p BC108 9p BF180 32p OC72 12p 2N3709 11p AC188 22p BC108 9p BF180 32p OC82 12p 2N3709 11p AC188 21p BC108 9p BF180 32p OC82 12p 2N3710 11p AC187 31p BC108 31p BC148 13p BF195 15p OC82D 12p 2N3710 11p AC161 31p BC148 13p BF195 15p 2N2904 20p 2N4062 12p AC161 31p BC148 13p BF195 15p 2N2904 20p 40360 35p AC161 32p BC158 14p BF750 20p 2N2926R 9p 40361 35p AC161 32p BC158 14p BF750 20p 2N2926R 9p 40361 35p AC161 32p BC158 14p BF750 20p 2N2926G 9p 40362 40p AC161 32p BC158 14p BF752 20p AC161 32p BC158 14p BF752 20p BC158 40p BC158 14p BF752 20p BC158 40p BC15	IRAN	31310	K2							
ACI227 12p AFI27 20p BFI15 25p OC42 12p 2N3704 13p ACI28 12p AFI37 20p BFI15 20p OC44 12p 2N3705 12p ACI33 12p AFI38 32p BFI73 20p OC45 12p 2N3705 12p ACI33 12p AFI38 32p BFI78 32p OC70 12p 2N3707 12p ACI37 12p AFI81 40p BFI78 32p OC70 12p 2N3707 12p ACI37 22p BCI08 9p BFI80 32p OC71 12p 2N3708 10p ACI88 22p BCI08 9p BFI80 32p OC72 12p 2N3709 11p ACI88 22p BCI08 9p BFI80 32p OC82 12p 2N3709 11p ACI88 22p BCI08 9p BFI80 32p OC82 12p 2N3710 11p ACI49 45p BCI47 13p BFI80 32p OC82 12p 2N3710 11p ACI49 45p BCI47 13p BFI94 15p OC82D 12p 2N3710 11p ACI49 45p BCI47 13p BFI95 15p 2N2904 20p 2N4062 12p ACI41	ACI07	15p	AF12S	20p						
AC132 13p AF139 32p BF173 20p OC44 12p 2N3705 12p AC131 13p AF189 34p BF177 28p OC45 12p 2N3705 12p AC132 12p AF180 40p BF178 13p OC70 12p 2N3707 12p AC187 21p AF180 40p BF178 13p OC70 12p 2N3707 12p AC187 21p BC108 9p BF180 32p OC71 12p 2N3708 10p AC188 21p BC108 9p BF180 32p OC72 12p 2N3709 11p AC188 21p BC108 9p BF180 32p OC81 12p 2N3709 11p AC188 21p BC108 9p BF180 32p OC81 12p 2N3701 11p AC188 21p BC108 9p BF180 32p OC82 12p 2N3709 11p AC188 21p BC108 9p BF180 32p OC82 12p 2N3710 11p AC180 32p BC148 13p BF180 32p OC82 12p 2N3710 11p AC180 32p BC148 13p BF195 15p 2N2904 20p 2N3710 11p AC180 32p BC148 13p BF200 32p 2N29268 9p 40360 35p AC181 20p BC158 14p BF750 20p 2N29269 9p 40361 35p AC181 20p BC158 14p BF750 20p 2N29269 9p 40362 40p AF117 20p BC158 14p BF752 20p 2N29269 9p 40362 40p AF117 30p BC158 14p BF752 20p AF117 30p BC158 14p BF752 20p 2N29269 9p 2TX302 15p AF117 30p BC187 22p BU105 225p 2N39365 8p ZTX500 15p	ACI26	120	AFI26	20p	BD133	75p	OÇ35			
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ACI31 12p					BF173	20n	OC44	12p	2N3705	12p
ACI32 12p AFI80 40p BF178 32p OC70 12p 2N3707 12p ACI36 12p AFI80 40p BF178 32p OC71 12p 2N3707 12p ACI37 12p AFI81 40p BF179 32p OC71 12p 2N3708 10p ACI88 21p BC108 9p BF180 32p OC72 12p 2N3709 11p ACI88 21p BC108 9p BF180 32p OC81 12p 2N3701 11p ACI88 21p BC108 9p BF180 32p OC81 12p 2N3710 11p ACI88 21p BC109 3p BF181 32p OC81 12p 2N3710 11p ACI88 21p BC109 3p BF194 15p 2N2904 20p 2N3711 11p ACI88 21p BC109 32p 3N2904 20p 2N3711 11p ACI88 21p BC109 32p 32p 32p 32p 40360 35p ACI88 21p BC109 32p 32p 32p 40360 35p ACI88 21p BC109 32p 32p 32p 40360 35p ACI88 21p BC109 BC158 14p BF750 20p 2N2926R 9p 40361 32p ACI88 21p BC109 BC158 14p BF750 20p 2N2926G 9p 40408 40p ACI88 21p ACI88 21p BC109 BC159 14p BC158 21p BC109 BC159 14p BC158 21p BC109 2N2926G 3p ACI88 22p BC109 BC159 14p BC159 21p ACI88 22p BC109 2N2926G 3p ZTXS00 15p ZTXS00 15p ZTXS00 15p ACI88 22p							OC45		2N3706	Hp
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Of 124 asp 00131 100 0023 100 2110000 100					OC26	45n	2N3055	60p	ZTX502	20p
	AT 127	h	00131							

ZENER DIODES400mW S% 3·3V to 30V, 12p. | WIRE WOUND POTS, 3W, 10, 25, 50 Ω and decades to 100k Ω, 35p.

DIODES				
RECTIFI BY 127 BZY 10 BZY 13 IN 4001 IN 4004 IN 4007	1A AA AA 1A 1A	12p 25p 20p 7p 8p 10p	SIGNAL OA85 OA90 OA91 OA202 IN4148 BA114	7 p 5 p 7 p 5 p 8 p

BRUSHED ALUMINIUM PANELS $12in \times 6in = 25p$; $12in \times 2\frac{1}{2}in = 10p$; $9in \times 2in = 7p$

SLIDER POTENTIOMETERS

SLIDER POTENTIOMETERS
86mm × 9mm × 16mm, length of track 59mm.
SINGLE 10K, 25K, 100K log. or lin. 40p.
DUAL GANG, 10K + 10K etc. log. or lin. 60p.
KNOB FOR ABOVE 12p.
FRONT PANEL 65p.
18 Gauge panel 12in × 4in with slots cut for use with slider pots. Grey or matt black finish complete with fixings for 4 pots.

THERMISTORS VA1055S VA1066S VA1077 (RS3 15p 15p 15p £1-35

THYRISTORS

2N5060 50V 0.8A 30p, 2N5064200V 0.8A 47p, CRS1/40 400V 1A 25p, 106F 50V 4A 40p, 106D 400V 4A 55p.

MULLARD POLYESTER CAPACITORS C296 SERIES 400V: $0.001\mu\text{F}$, $0.0015\mu\text{F}$, $0.0022\mu\text{F}$, $0.0022\mu\text{F}$, $0.0033\mu\text{F}$, $0.0047\mu\text{F}$, 24p, $0.0068\mu\text{F}$, $0.01\mu\text{F}$, $0.015\mu\text{F}$, $0.022\mu\text{F}$, $0.033\mu\text{F}$, 19, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 0.14π , 49, $0.15\mu\text{F}$, 69, $0.22\mu\text{F}$, 74p, $0.33\mu\text{F}$, 19, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 0.14π , 49, $0.015\mu\text{F}$, 34p, $0.015\mu\text{F}$, $0.022\mu\text{F}$, $0.033\mu\text{F}$, $0.015\mu\text{F}$, 44p, $0.022\mu\text{F}$, $0.033\mu\text{F}$, $0.015\mu\text{F}$, 44p, $0.022\mu\text{F}$, 13p, $0.016\mu\text{F}$, $0.016\mu\text{F}$, $0.016\mu\text{F}$, $0.022\mu\text{F}$, 39D, $0.033\mu\text{F}$, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 34p, $0.015\mu\text{F}$, $0.012\mu\text{F}$, $0.022\mu\text{F}$, 39D, $0.033\mu\text{F}$, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 34p, $0.15\mu\text{F}$, $0.022\mu\text{F}$, 39D, $0.033\mu\text{F}$, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 34p, $0.15\mu\text{F}$, $0.022\mu\text{F}$, 39D, $0.033\mu\text{F}$, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 34p, $0.15\mu\text{F}$, $0.022\mu\text{F}$, 39D, $0.033\mu\text{F}$, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 34p, $0.15\mu\text{F}$, $0.022\mu\text{F}$, 39D, $0.047\mu\text{F}$, $0.068\mu\text{F}$, 34p, $0.082\mu\text{F}$, 34p, $34\text{$

MYLAR FiLM CAPACITORS 100V 0.001μF, 0.002μF, 0.005μF, 0.01μF, 0.02μF, $2\frac{1}{3}$ p. 0.04μF, 0.05μF, 0.068μF, 0.1μF, $3\frac{1}{3}$ p.

CERAMIC DISC CAPACITORS 100pF to 10,000pF, 2p each.

20

ECTROLYTIC CAPACITORS-MULLARD 015/6/7 RANGE REPLACES C426,

ELECTROLYTIC CAPACITORS—FIGURAND 015/6/1 NANGE REPLACES C426. C457 RANGES. (1,1F,19) 1-0/63, 15/63, 2-2/63, 3-3/63, 4-7/63, 6-8/40, 10/25, 10/63, 15/16, 15/40, 15/63, 22/10, 22/25, 22/63, 33/6-3, 33/40, 47/4, 47/10, 47/25, 47/40, 47/63, 68/6-3, 68/16, 100/4, 100/10, 100/25, 100/40, 150/6-3, 150/16, 150/25, 220/4, 220/16, 230/4, 330/4, 330/4, 470/6-3, 59 each. 68/63, 150/40, 220/20,25, 330/16, 470/10, 680/6-3, 1000/4, 9p. 100/63, 150/63, 220/40, 470/25, 680/16, 1,000/10, 1,500/6-3, 12p. 220/63, 470/40, 680/25, 1,000/16, 1,500/10, 2,200/6-3, 15p. 330/63, 680/40, 1,000/25, 1,500/16, 2,200/10, 3,300/6-3, 4,700/4, 18p.

SOLID TANT	TALUM B	EAD CAPACIT	ORS			10
)·IμF 35∨	2-2µF	35∨	22μF 33μF	16V	
0.	22µF 35V	4.7µF	35∨	33µF	107	
	47µF 35∨	6.8µF	25∨	47µF		
	.0µF 35∨	10µF	25 V	100µF	3∨	

ı	VEROBOARD		JACK PLUGS AN	D SC	CKETS	
	0-1 2½ × 3½ 22c 2½ × 5 24c 3½ × 3½ 24c 3½ × 3½ 24c 3½ × 5 27c 17 × 2½ 175; 17 × 3½ (plain) — 17 × 3½ (plain) — 2½ × 5 (plain) — 2½ × 5 (plain) — 2½ × 3½ (plain) — 2½ × 3½ (plain) —	0 16p 0 24p 0 24p 0 27p 0 57 1p 0 78p 82p 60p 42p 12p	Standard screened Standard insulated Stereo screened Standard socket Stereo socket D.I.N. PLUGS AN 2 pin, 3 pin, 5 pin 18 Plug 12p. Socket 1 4 way screened cable 6 way screened cable	15p 18p D SC 0°, 5 Sp. , 15p	pin 240°, 6 pin /metre.	8p 8p 63p 8p 8p
	Pin insertion tool 52; Spot face cutter 42; Pkt. 50 pins 20;	p 42p	BATTERY ELIMIT 9V mains power supp			£1·50 ttery.

LARGE (CAN) ELECTROLYTICS 1600µF 64V 74p 2500µF 64V 80p 2500µF 40V 74p 2800µF 100V £2:60 2500µF 50V 58p 3200µF 16V 50p 4500µF 4500µF 5000µF

 $\begin{array}{cccc} \textbf{HIGH VOLTAGE TUBULAR CAPACITORS-} \\ & 0.01 \mu \textbf{F} & \textbf{10p} & 0.047 \mu \textbf{F} & \textbf{13p} \\ & 0.022 \mu \textbf{F} & \textbf{12p} & 0.1 \mu \textbf{F} & \textbf{13p} \end{array}$ 1,000 VOLT 0·22μF 0·47μF

POLYSTYRENE CAPACITORS 160V 2½% 10pF to 1,000pF E12 Series Values 4p each

SMOKE AND COMBUSTIBLE GAS DETECTOR—GDI
The GDI is the world's first semiconductor that can convert a concentration of gas
or smoke into an electrical signal. The sensor decreases its electrical resistance when
it absorbs deoxidizing or combustible gases such as hydrogen, carbon monoxide,
methane, propane, alcohol, North Sea gas, as well as carbon-dust containing air or
smoke. This decrease is usually large enough to be utilized without amplification.
Full details and circuits are supplied with each detector.
Detector GDI, £2. Kit of parts for detectors including GDI and P.C. board but
excluding case. Mains operated detector £3.0. 12 or 24V battery operated audible
alarm £7.30. As above for PPS battery, £6.40.

97p
Draw the planned circuit onto a copper laminate board with the P.C. Pen, allow to
dry, and immerse the board in the etchant. On removal the circuit remains in high
relief.

ITT/TEXAS IC's NOW ΙN STOCK LARGE RANGE

PRICES ARE CALCULATED ON TOTAL NUMBER ORDERED REGARDLESS OF MIX 25-99 12-24 100 + 135 250 95 105 170 100 100 112 185 185 36 85 130 220 89 95 160 86 96 105 180 164 150 170 40 95 140 280 7450 7451 7453 7454 7460 7470 7472 7472 7474 7475 7480 7481 7482 7484 7485 7490 7493 7493 7495 7496 74121 74141 74150 74151 74153 74154 74155 74156 74190 74191 74192 74196 74197 14 14 14 14 14 25 50 32 90 110 95 110 240 37 60 85 65 65 85 7400 7401 7402 7403 7404 7405 7406 7407 7410 7410 7411 7416 7420 7421 7421 7421 7421 7432 7440 7441 7442 16 16 16 18 18 14 14 14 16 16 16 40 44 27 27 26 39 14 27 23 16 70 115 150 330 13 13 14 14 14 35 38 23 33 13 23 22 34 13 23 24 28 14 65 65 115 iin 120 200 150 130 110 180 120 136 190 190 190 180 190 20 50 56 36 18 23 36 45 18 36 32 30 40 40 55 40 100 125 100 120 28 38 36 52 36 95 115 96 97 28 49 30 85 50 30 16 21 30 28 43 16 30 29 18 36 18 75 195 195 200 200 200 200 105 85 92 105 230 33 52 79 60 52 80 90 85 230 250 45 75 100 75 75 95 105 245 42 67 92 70 68 90 100 95 240 LINEAR IC's 709 741 741 723 747 748 DIL socket, 14 pin DIL 8 pin DIL 14 pin DIL 14 pin DIL 14 pin DIL 8 pin DIL 20 40 20 80 34p 28p 95p 85p 35p 80 250 14 pin and 16 pin

RINGS OF SATURN

Once again a theory has been challenged by a practical test. The original theoretical study of Saturn's rings was made by Clark-Maxwell. He gave as his opinion that these rings must be dis-continuous and made up of discrete particles.

This view was accepted by astronomers and until recently the generally recognised explanation was that the particles were ice crystals or accretions of dust. This appeared to be a satisfactory conclusion since the thickness of the rings is around 1.0km, although the diameter reaches some 85,000km.

Now as a result of studies made at Goldstone, California, two researchers R. Golstein and G. Norris, have suggested that the rings consist of boulders approximately 1.0m

in diameter.

They used the large Goldstone radio dish and a power of 400kW to direct a beam to the planet 700 million miles away. The beam, using a wavelength of 12:5 centimetres, calculated boulder size from the reflected signal.

This new finding represents a considerable hazard to spacecraft and it may well be that there are scattered lumps of rocks at higher and lower latitudes. The discovery prompts the thought that perhaps the rings are after all the debris of a former satellite.

The approach of the present programmed probes will be close enough to check the conclusions that have been made and possibly even photograph the rocks.

ASTEROID BELT

The safe journey of Pioneer 10 through the asteroid belt has caused a sigh of relief from the scientists planning the missions. The knowledge that the density of the belt and the size of the particles which differs considerably from the former conjectures will enable plans for future probes to go forward with confidence.

The size of the particles in the belt were between 100 micrometres and 1.0mm. Nothing larger than 1.0mm was encountered all the way out from the earth's orbit. Even those of 1.0mm size were rare. It seems, therefore, that the belt is composed of the 17,000 to 20,000 identified asteroids and very fine dust. During the whole period of the transit through the belt Pioneer 10 did not have one of the known asteroids in view.

COMPUTER INTEGRITY

A short time ago the possibility of a tenth planet was raised again. Observers have been unable to locate it and computation did not supply sufficient information for a search.

A new attempt at the three body problem was tackled again recently



and out of this exercise came the conclusion that residual errors, which are unavoidable in a computer analysis of data, will always predict a planet or body. It seems therefore that this situation rebuts computer integrity for problems such as this.

TECHNICAL EUTHANASIA

At the press of a switch Sir Brian Flowers FRS, Chairman of the Science Research Council put an end of the life of Atlas 1. This computer, the most advanced design in the world when it was installed. had come to the end of a useful life because it was old and its transistors germanium. The availability of spares and the high cost of maintenance made the euthanasia decision imperitive.

The computer which dealt with the pilot data from weather satellites, retrieval of data from UK 3, and other space projects was designed by a team at Manchester University under Dr (now Prof.) T. Milburn. It contained a supervisory system which has only quite recently been emulated.

The work will now be taken over by a 1906A and a 370/195 which is already dealing with UK 4 and Essa 8. These two computers are, of course, much faster than Atlas 1. The main activities of Atlas have been concerned with theoretical chemistry and crystallography. There is no doubt that the decision to set up the enterprise of the Atlas Laboratory by the Research Council was a wise one and the original investment justified by results.

ORANGE SOIL OF THE MOON

The excitement of the discovery of the orange soil on the moon, thought to be iron oxide, has now been rather dashed by the findings of the investigators. It seems that

the soil is made up of grains of

glass.

It is thought that the soil has only been exposed to cosmic rays on the surface for about 8 to 10 million years. When the material was formed, which would be about 3,700 million years ago, it was buried deeply and not exposed to cosmic rays. It was thrown up later by meteoric impact and not by volcanic action of the Shorty crater even though it is of recent forma-

The orange soil is the finest grained that has been found on the moon and consists of coloured glass in droplets and fragments. Chemically the samples from the Apollo 17 mission are the same as those from the Apollo 11 mission which was several hundred miles away. There is, however, a difference in that the Apollo 11 samples were not so rich in zinc.

LUNAN PROBE

In the March issue of Spacewatch mention was made of the work of D. A. Lunan regarding the possibility of a probe being situated in the orbit of the Moon. This was suggested to have originated from Epsilon Boötes a star in the constellation of that name and at a distance from the Earth of approximately 105 light years. The whole matter has caused some stir and the paper was given before a large audience in Caxton Hall in March this year.

There has been a further development in this matter and America has now taken an interest in the project. Originally a project named OSMAR was initiated by Frank Drake and Carl Sagan in America. This was subsequently abandoned but a new project for which a very large Radio System is being set up is to be undertaken. This involves a system of aerials in a line 25

miles long.

A more modest project is being set up in this country based on Lunan's suggestions. It is called GOLDE (Ground Observations of Long Delayed Echoes) and is to be operated under the aegis of the British Interplanetary Society. This is a project in which the help of

amateurs is required.

The principle of operation is to send a strong signal to the place where calculations show that the probe should be and time the echoes which may come back, EMI have loaned a special set of equipment to help the project, which will be in operation for a long time. From America has come an offer to parcipate in the programme. This is from Stanford University where a large dish is available.

The British team is lead by D. A. Lunan and A. T. Lawton who would welcome amateurs who are interested to take part in this ex-

periment.



By Prof. G. D. Sims, O.B.E., Ph.D. (Head of Dept. of Electronics, Southampton University)

AN APPRECIATION OF THE CHANGES ALREADY BROUGHT ABOUT AND AN EVALUATION OF FUTURE PROSPECTS

1973 marks the 25th anniversary of the announcement of the transistor and few in 1948 would have forecast the immeasurable impact which it was to make on the evolution of our society. Since that time "electronics" has seen the appearance of a continuous stream of new devices and "systems" for communications, data processing and control, and our pattern of living has been affected by these developments in a multitude of ways.

FROM SMALL BEGINNINGS TO A SMALLER FUTURE!

Looking back, it was clear that once the junction transistor had been produced, the day of the thermionic valve was limited, though the uptake of the new technology by U.K. industry was relatively slow. Certainly, at that time, both the educational and practical problems associated with its use were formidable, but since, there has been a steady adaptation to solid state devices in almost all low-power applications.

The new "active devices" were no longer necessarily expensive and as the new "systems" emerged, they grew in complexity, bringing new problems associated with "statistical" device and wiring failures. A solution had to be found and once planar technology had been mastered, the ultimate emergence of the microcircuit was inevitable: this, in its turn, brought still lower cost, greater reliability and, as a bonus, yet smaller size. We were now able to buy a complete amplifier, in a single can, while even more significantly logic families and complete integrated digital sub-systems, such as counters and shift registers were also available.

SYSTEMS

At this point, the revolution in basic electronic techniques took a new turn, for now quite massive systems could be envisaged, of which the control systems for the Apollo missions were perhaps the most challenging and spectacular.

In parallel with these developments new information storage techniques were emerging and "memory", whether in semiconductor or magnetic form, is now another readily purchaseable item.

Memory, together with faster digital techniques, now enables us to envisage future "super-systems"

capable of working at data processing rates which were quite inconceivable in 1948. Whereas at that time a microsecond pulse was short, now nanosecond pulses are everyday things and picosecond pulses are beginning to be considered for the communication systems of the future.

These have all been direct effects of the development of the transistor and have often diverted attention from some equally important side-effects, not the least of which was the awakening of interest in the properties of "pure" materials generally.

MATERIALS

The development of semiconductor materials, which first took us from germanium to silicon, soon turned to other materials such as gallium arsenide. This has since given us power devices at microwave frequencies and possibly, of even greater importance, useful infra-red solid state lasers. Subsequently, gallium phosphide has arrived and using much the same pn-junction techniques as those employed at low frequencies, we now have compact alpha numeric indicators in this material and its derivatives.

We came to realise, too, that glass could be an important material and "Ovonic" storage devices using switching properties in glass offer great potential, even though, at the moment, they remain little understood and unreliable. We had already accepted that ceramics had their uses, too, whether for substrate materials or in more esoteric combinations such as with ferroelectric glasses, whose interesting optical and piezo electric properties suggest many potential applications.

It would be no exaggeration to say that all of the major device developments which have taken place since 1948 have arisen either from the utilisation of new materials or from improvements in materials technology which enabled us to use already known materials properties more efficiently than before.

DESIGN CONSIDERATIONS

In order to exploit our devices, in the new systems, it became necessary to marshal a whole new battery of design techniques: the simplest of these were concerned with the design of logical systems, and switching theory has now become an important tool in the hands of almost all electronic engineers. Further, as systems have grown in capability, the designer has

had to find other means for carrying out the kind of routine calculations which hitherto had been his main occupations.

If a computer aided design programme could provide details of how to design a filter, or a feedback circuit, it was clearly sensible to use it rather than to waste valuable professional time on "chores".

With the bigger systems, it is seldom practicable to "bread-board" everything in the initial design stages. Thus simulation techniques, in which systems are modelled on computers, have also developed apace.

Finally, and perhaps most important, is the growing pre-occupation with systems reliability, and design for maintenance ("terotechnology"). These considerations have to be taken into account at the outset of the design process and wherever possible systematic calculations must be performed to put realistic limits on component life.

Long component life is an expensive commodity and whilst it is essential that the system is *reliable enough*, it is seldom that excessive reliability can be justified. Perfectionism is all very well but it costs money!

THE YEARS TO COME

What then of the future?

At the system level we have already spoken of the "super-system" which will provide a communicating power, exceeding by orders of magnitude that are available at the moment. This, using faster digital techniques and optical fibre waveguides, will open up the almost unlimited bandwidths available at optical frequencies and will allow many new possible applications of video systems, such as information retrieval from the local data bank or library and generally "instant-optical communication".

At the device level it is not easy at the present time to envisage a replacement for the silicon integrated circuit, though as technologies evolve, we are already seeing a noticeable movement from bi-polar techniques to MOS processes, particularly in relation to the new forms of solid state memory.

VACUUM TUBE SURVIVORS

In the area of general electronics, the cathode ray tube and imaging tubes remain apart, as almost the sole survivors of the vacuum tube era—but for how long?

We are already seeing important developments by way of self-scanned solid state silicon arrays which may soon challenge vacuum image tubes for some applications: new ideas are continually emerging in the use of new materials for image storage: new techniques also appear for writing, both with light and electron beams, either on photochromatic materials or thin films, for data recording purposes.

At the moment, notwithstanding, the cathode ray tube as a "picture tube" appears to be secure, for as yet nothing fast enough can challenge it. The same remains broadly true of "power" tubes where as yet solid state devices have made a limited impact—though undoubtedly their time will come, too!

THE ENGINEER AND TECHNICIAN

Implicit in our discussion above is the suggestion that the roles of the engineer and the technician in electronics have also changed. The subjects of fundamental importance to either remain much the same,

but the achievable level of complexity is now so great that many engineers must be less-concerned with circuit detail than with the properties and specification of the overall system and its organisation. The technician at the same time has a task which, although it may in some respects be simplified through the microcircuit, is in other ways more complicated, requiring a greater overall understanding of electronics as a whole.

EDUCATION

For those in education, the scene has been changing continually and the challenge is unrelenting. Graduate courses remain at three-years and, somehow or other, competently educated engineers must be produced in that time. As always, the education has to be mainly concerned with fundamentals and with developing, within the student, the ability to continually self-update his knowledge, in his subsequent professional life in Industry. At the same time, however, he requires an awareness of the attitudes and practices of the contemporary world and the task of balancing these needs is a delicate one.

THE VITAL AREAS

The areas above all in which more effort is needed, if the student is to cope with the changes which the future will bring, are probably the two mentioned above, viz. "Materials" and "Systems".

For the man who is going to be concerned primarily with devices, materials have assumed an importance many times greater than was appreciated 25 years ago: while for the designer of "capital goods", the systems considerations which could be overlooked when the radio set represented the ultimate challenge in complexity, are now at the fore-front!

Circuit theory, of course, remains as important as ever, but, to it, we have had to add logic design and some indication of how we can check our ideas and translate them into practical terms through CAD and simulation.

Thus the education sector has had to learn and adapt rapidly, too, and it is vital that it should continue to do so, for the industry is critically dependent on an assured inflow of high quality personnel.

PROSPECT

Despite the recession, the half-million strong electronics industry of this country has been remarkably successful and will be of key importance both nationally and to the EEC, in the years to come.

Just as the development, successively, of the telephone, radio, television, and the computer have all marked stages in a revolution which has been social as well as technical, the super-systems of the future will change our habits still more. It has been, above all, the invention of the transistor which has accounted for the continually accelerating pace of this social revolution, which began, so innocently, with the invention of the thermionic valve at the turn of the century!

THIS month the module containing the Sample and Hold circuitry and Noise Generator is described together with application details.

SAMPLE AND HOLD

The Sample and Hold shown in block form in Fig. 5.1 is another programming device which is capable of producing formalised staircase waveforms (Fig. 5.2) or random staircase waveforms. In both cases the "rise" of each step in the staircase is dependent upon the amplitude of the voltage being sampled while the "tread" of each step is governed by the sampling rate set by the clock.

The clock itself consists of two separate circuit forms as shown in the theoretical circuit diagram Fig. 5.3. The first is an astable multivibrator (ICI) in which the rate of oscillation is variable between 0.25Hz and 50Hz. The circuit switches alternately between its positive and negative saturation levels so that, for 15 volt supply rails, the output voltage swing is of the order of 28 volts.

The second circuit form is a monostable multivibrator (IC2) which is triggered by the negative going steps of the astable squarewave thus producing a pulse train of the same frequency. This circuit also switches between its positive and negative saturation levels. In the stable condition the output is positive. The introduction of a negative trigger pulse at the input, C3, causes the output to switch to its negative saturation level and to remain there for a period determined by the components R12 and C4.

PRECISION GATE

Clock pulses from the monostable are used to trigger an analogue gate (IC3) which has two summing inputs. One is coupled internally to a d.c. source and the other directly to an external sample socket. The gate will only recognise negative inputs.

If the trigger input to the gate is left open circuit or is grounded the gate behaves like a unity gain inverter but only for negative going signals. A positive input signal would tend to swing the output of the gate negative, a tendency which is prevented by the bounding action of diodes D6 and D7.

Under normal conditions overriding control of the state of the gate depends upon the polarity of the signal presented at the trigger input. As has been explained, the gate is closed, that is, passing inverted negative signals, when the trigger input is open circuit or grounded. The same situation exists when the polarity of the trigger signal is negative.

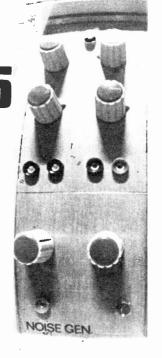
When the trigger signal is positive however the gate opens and its output is zero provided that the potential of the sample input is less than the potential of the trigger signal. This latter situation exists for all conditions of internal sampling where the maximum d.c. level attainable is determined by the divider R26-VR4, and for all conditions of external sampling from a single programming source.

When the d.c. and external sampling sources are combined, providing a single programming source only is used, the maximum sampling potential will never exceed 11.5V as compared to a trigger potential of 14V, and thus the closed period of the gate will never exceed 560 microseconds.

In circumstances where two or more programming sources are combined either with, or without, amplification, the situation can exist where the sample potential exceeds the trigger potential. Under these conditions the closed period of the gate is solely dependent upon the period of "high" sample potential.

PE Sound Synthesiser 5 SAMPLE-HOLD and NOISE GENERATOR

By G.D. SHAW





INTEGRATOR

Output from the gate is led to the programming input of an integrator/comparator (IC4/IC5) arrangement which is basically similar to the ramp generator described in last month's article. In this case, however, the time constant of the integrator is much shorter resulting in an extremely rapid integration.

This feature is necessary in order to provide the steep rise between steps if a crisp tone change is to

be achieved.

Further additions to the basic ramp generator circuit include an indicator lamp switched by TR1 which serves to indicate the "on" period of the staircase, a clamping diode D1 which serves to limit the integrator output to approximately 650 millivolts in the event that the integrator tends to go into positive saturation, and an input bias control provided by R1, VR1, R2. Input bias is required to compensate for integrator capacitor leakage during the periods of hold between samples, particularly when the sampling rate is low. It also serves to allow the Sample and Hold to be used as another Ramp Generator if required. This is achieved by disabling the trigger socket (JK1) by the insertion of an opencircuit jack plug and setting the ramp rate by adjustment of the d.c. and bias controls. Very high ramp rates may be achieved by this means.

CONSTRUCTION

The circuit board layout of the Sample and Hold is shown in Fig. 5.4. Layout is not critical although space problems may occur if relative component

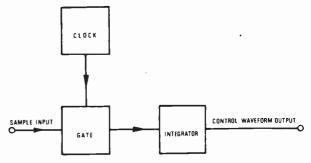


Fig. 5.1. Elements of Sample and Hold in block form

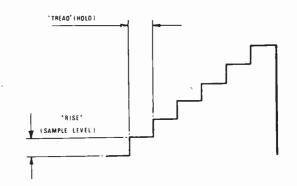


Fig. 5.2. Staircase waveform produced by Sample and Hold circuit. The rise of each step is dependent on the amplitude of the sampled voltage while the tread depends on the clock rate

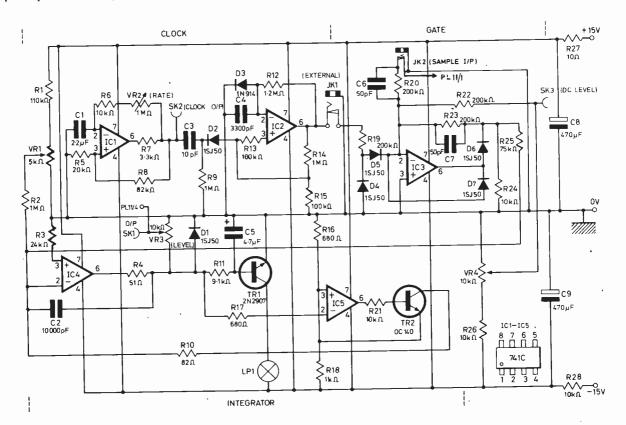


Fig. 5.3. Sample and Hold circuit

SAMPLE AND HOLD BOARD

COMPONENTS . . .

Resist	ors		
R1	110kΩ	R15	$100k\Omega$
R2	$1M\Omega$	R16	Ω 086
R3	24kΩ	R17	Ω 086
R4	51Ω	R18	1kΩ
R5	20 kΩ	R19	200k Ω
R6	$10k\Omega$	R20	200kΩ
R7	3-3kΩ	R21	10kΩ
R8	82 kΩ	R22	200kΩ
R9	1MΩ-	R23	200k Ω
R10	82Ω	R24	10kΩ
R11	9·1kΩ	R25	75k Ω
R12	1·2MΩ	R26	10kΩ
R13	100kΩ	R27	10Ω
R14	$1M\Omega$	R28	10Ω
All 5	% ⅓ watt carbon		

Potentiometers

 $\begin{array}{lll} \text{VR1} & \text{5k}\Omega \text{ lin. miniature moulded} \\ \text{VR2} & \text{1M}\Omega \text{ lin. miniature moulded} \\ \text{VR3-VR4} & \text{10k}\Omega \text{ lin. miniature moulded (2 off)} \\ \end{array}$

Transistors

TR1 2N2907 TR2 OC140

Capacitors

Č1	2⋅ 2 μF	35V Tantalum
C2	10,000pF	polystyrene
C3	10pF	polystyrene
C4	3,300pF	polystyrene
C5	4·7μF	40V tantalum
C6	50pF	polystyrene
C7	50pF	polystyrene
C8-C9	470µF	25V elect. (2 off)

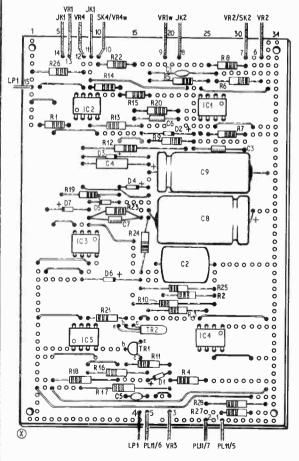
Diodes

D1	I S J50	
D2	ISJ50	
D3	IN914	
D4-E	7 ISJ50 (3 off	١

Integrated Circuit IC1-IC5 741C (5 off)

Miscellaneous

LP1-28V sub-miniature indicator lamp SK1-SK3 2mm miniature sockets (3 off) JK1, JK2 3·5mm jack socket (2 off) 0·1in matrix Veroboards as required



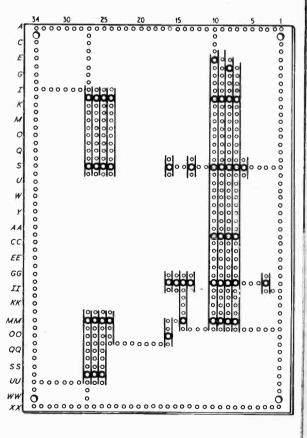


Fig. 5.4. Component layout and wiring of Sample and Hold Board

sizing differs appreciably from those shown. The power supply decoupling electrolytics present the biggest hazard as far as cramping of the board is concerned and these components should not be larger than 32mm in length by 16mm in diameter. R.S. Components Tube type are suitable and comply with the dimensions given.

EXPERIMENTAL CIRCUITS

The Sample and Hold provides the principal means by which the Synthesiser offers its most fascinating feature, that of "playing" by itself. Coupling the output to a v.c.o. and careful adjustment of the sample sources can provide a range of repetitive tone sequences the repeat period of which may be varied from a few seconds to several minutes.

Whether the sequence is truly repetitive or entirely random depends very largely on the choice of sample source. An article in the February issue of Hi-Fi News reviewed a recently issued record in which the "music" had been derived from computer stored data relating to changes in the Earth's magnetic field

measured at a series of selected points.

In a similar manner existing data sources may be used to provide sample information. Crystal clocks, binary counters, ring counters, old non-erased computer tape and other digital data sources of various kinds, signal generators—even legitimate recorded

music may be pressed into service, amplified, attenu-

ated and blended together in various ways to serve the cause of random programming.

The discerning constructor will have noted that whatever the source of sample information the overall effect of the Sample and Hold is to turn this into a voltage which is progressively increasing, in steps, to a predetermined level at which point it returns to zero only to commence climbing once again. In practice the feature of a regular return to zero of the Sample and Hold output does not become obtrusive except at relatively slow sampling rates when the sample voltage shows very little variation between successive samples.

NOISE GENERATOR

White noise, defined in some circles as unwanted sound, is a very useful addition to the aural facilities provided by the synthesiser. It is a known fact that a great many sounds otherwise considered to be "pure" actually contain a relatively high noise content. The edge-tone in a wind organ is a typical example.

For imitative synthesis the addition of noise in greater or lesser degree is essential if the greatest approach to realism is to be achieved. This factor applies particularly to the synthesis of naturally occurring sounds such as rainfall, surf on the beach, storms, etc., and also to certain man-made sounds such as gunfire, explosions, train whistles, steam

engines and so on.

Fig. 5.5 shows the theoretical circuit of the noise generator. The circuit is really quite simple. R5-R9; C3-C4 and D1 represent the noise generation section. The noise diode D1 is a specially selected noisy Zener marketed only by Semitron Ltd., and in the circuit configuration shown provides an output of about 75mV.

The noise bandwidth and level may be adjusted to a certain extent by varying the values of C4 and R6 respectively although it will be found, in practice, that the values shown are suitable for most purposes. C3 serves to decouple the noise diode from the inverting amplifier based around IC1.

Cost reduction can be achieved by omitting the offset adjustment preset VR3 and substituting a capacitor between the values of 0.01 and $0.1\mu F$ in the output of the operational amplifier as shown dotted.

LOW-PASS FILTER

R1, VR2 and C2 serve as a simple yet severe lowpass filter in order to provide a degree of control over the colouration of the noise. With VR2 at its minimum setting the output of the noise generator is reduced to a rough triangular waveform with a frequency in the region of 6kHz. Under these con-

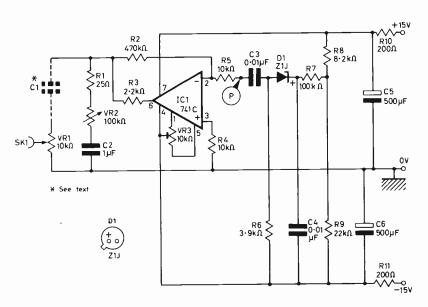


Fig. 5.5. Circuit diagram of Noise Generator

ditions the loading on the output of the operational amplifier is quite heavy and R3 is therefore included to limit the output current drain. Power supply decoupling is essential if noise is to be prevented from leaking back to the power distribution busbars.

In the circuit shown the current requirements are 2.5mA per rail and the addition of 200 ohm decoupling resistors will therefore result in a voltage drop of about 0.5V. If noise leak-through continues to be a problem these latter resistors may be increased in value quite considerably although if values over 500 ohms are used some adjustment of the values of R8 and R9 may be necessary to maintain their junction voltage at about +20V with respect to the negative rail.

Fig. 5.6 shows the circuit board layout of the noise generator. A piece of Veroboard or similar of 17 × 34 ways is suitable. Note that screened leads are used to connect this board's outputs with its associated components on the front panel. These leads should go direct to their respective components and not be bound into the wiring harness.

CONSTRUCTION

In general the construction of this module should follow the pattern adopted with those already described. The wiring harness for the Sample and Hold should pass out at the top of the front panel and down the length of the circuit board support plate to join the circuit board which is mounted adjacent to the McMurdo plug. The noise generator

circuit board should be mounted over the lower pair of supports adjacent to the base of the front panel. Details of the front panel layout and module wiring are given in Fig. 5.7.

SETTING-UP

It is recommended that setting-up and testing be established as a continuing process during construction of this module. With the noise generator, for example, it is suggested that the noise generating section consisting of R5-R9; C3-C4 and D1 be built first and tested by making temporary connections to the power supply rails and observing the output by connecting an oscilloscope between point "P" and the negative rail.

The expected output of the noise amplifier can be calculated at this stage by measuring the noise output of the diode with respect to signal ground and multiplying this by the gain of the amplifier.

In the prototype the total noise output was 3.5V maximum which is more than adequate. If the performance of this stage is satisfactory construction of the amplifying section can go forward being similarly tested on completion and before mounting the finished board into the module.

There is no actual "setting-up" to be done with the Sample and Hold and the purpose of testing is merely to establish that the circuit performs within the previously described limits.

|||||C3

Fig. 5.6. Board layout and wiring for Noise Generator

OF BOARD

COMPONENTS . . .

Resis	stors				
R1	25Ω	R5	10 kΩ	R9	$22k\Omega$
R2	470k Ω	R6	3.9kΩ	R10	200Ω
R3	2 ⋅2kΩ	R7	100k Ω	R11	200Ω
R4	10k Ω	R8	8-2kΩ		

All ½ watt 5% carbon

Capacitors

NOISE GENERATOR BOARD

C1*	See text
C2	1μF polyester
Ca	0.01 E

C4 0.01μF

C5 500μF elect. 16V C6 500μF elect. 16V

Diode

D1 Z1J Zener (Semitron Ltd)

Potentiometers

VR1 $10 \mathrm{k}\Omega$ lin. miniature moulded VR2 $100 \mathrm{k}\Omega$ lin. miniature moulded VR3 $10 \mathrm{k}\Omega$ lin. horizontal carbon preset

Integrated Circuit

IC1 741C

Miscellaneous

SK1 2mm miniature socket 0.1in Veroboard as required

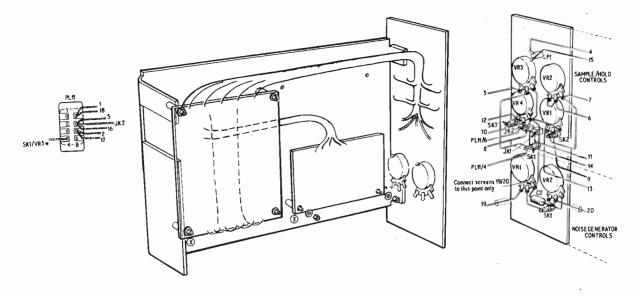


Fig. 5.7. Front panel component assembly with details of mounting and disposition of board on the module support plate. The small ringed X on the board edges indicates orientation. (See board assembly figures) For direct programming from the v.c.o. module connect SK11/4 to SK8/2

USING THE SAMPLE AND HOLD

It is best to confine initial experiments with the Sample and Hold to the formation of relatively simple staircase patterns, derived from the sampling of fixed d.c. voltages, in order to become familiarised with the effect of the adjustment of the various controls. Adjustment of the bias control, for example, can be quite critical when slow sample rates are being used and it is helpful to observe the output waveform on the oscilloscope so that drift between successive samples can be more easily balanced out.

When the Sample and Hold is programming a v.c.o. changes in "tread" voltage can be clearly discerned by ear but this becomes progressively more difficult as the sampling rate is increased. Note that it is difficult, if not impossible, to eliminate drift on the first step of a multi-step staircase due to the low charge on the integrating capacitor. This is not necessarily a disadvantage since it is possible to programme out the first step of the staircase by means of the envelope shaper which is to be described in a future article.

Progression to the sampling of varying voltages is the next logical step. Fig. 5.8 shows the effect of sampling a negative ramp having a period of about 0·1Hz. Note how the "rise" between "treads" increases in proportion to the increase in the ramp voltage. Variation in the ramp level by means of the input amplifier can cause remarkable changes in the output rhythm. A low ramp level and rapid sampling rate gives rise to an arpeggio-like sound in which the separation between the first few "treads" is barely discernible.

A high ramp level, on the other hand, causes the output of the integrator to reach its reset point fairly rapidly but since the sampling is continuing on an ever increasing ramp level the next staircase will have fewer steps and reach its reset point even more quickly. If the second reset is still well within the ramp period the third staircase will demonstrate even fewer steps while the fourth and subsequent staircases may consist of only one step, i.e. a square wave.

Variation of the sample voltage (ramp level) and sampling rate can ring the changes over a very wide range and produce some very interesting results.

SAMPLING A POSITIVE RAMP

Fig. 5.9 illustrates the effect of sampling a positive going ramp. In this case an initial condition of sampling fixed d.c. should be set and the ramp level adjusted so as not to exceed the d.c. sample voltage.

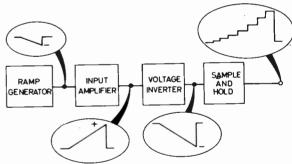


Fig. 5.8. Variable voltage programming. Note how the rise on consecutive steps at the output increase in proportion to the ramp level. Here d.c. level is zero

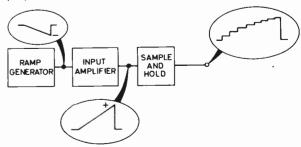
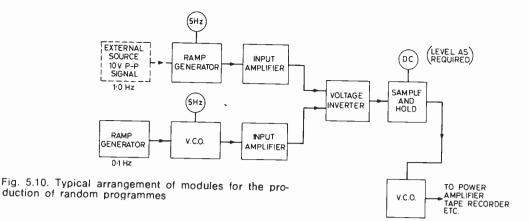


Fig. 5.9. Variable voltage programming. Here consecutive rises on the staircase output decrease with increase in ramp level. D.C. level is equal to or greater than the ramp level



The effect of these settings is shown, i.e. a relatively large "rise" on the first few samples which gradually decreases as the ramp level rises. In other words a reversal of the situation illustrated in Fig. 5.8.

FURTHER EXPERIMENTS

Further experiments may be carried out in which the sample voltage is derived from two or more sources simultaneously and a typical arrangement is shown in Fig. 5.10.

The sample sources used need not, of course, be centred within the Synthesiser itself. Almost any

device producing a varying output voltage may be used providing that the voltage amplitude concerned is compatible with the devices used in the synthesiser.

The output from a pick-up cartridge, tape recorder or radio can be amplified to a suitable level and used as sample material. Music which has wide and fairly rapid changes in dynamic range gives the best results.

Next month: Some general views on the establishment of an experimental sound studio and the construction of the Tone Control module will be described.

SOUND'73

Sound 73 International, organised by the Association of Public Address Engineers and held again this year at the Bloomsbury Centre, London, ran from 13 to 15 March under what can only be described as "Luxury" conditions with deep pile carpets, a warm inviting atmosphere and the sound of happy music welling up around one.

In addition to the main exhibition, the event included lectures on each day, a number of social activities, and public demonstrations of some of the available

equipment.

It is as well that the environment was conducive to communications since the sheer quantity of microphones, loudspeakers, amplifiers, Discos and so on to be seen must have led many into a feeling of confusion.

To add to this, the growing availability of semiconductor equipments and the adoption of current stylings have certainly led to a degree of similarity of gear

from stand to stand.

On the lecture front the subjects covered included microphones and their circuitry; limiters and compressors, their application and use; the industrial design of PA equipment; and finally the marketing of this equipment. Of course the event is basically directed to the professional sound engineer and the manufacturer but for all that there is invariably something of interest for the casual visitor at such an event.

FROM THE PAST

An interesting contrast with the past was provided in the presence of a public address caravan once the property of Cecil Clarabut of Bedford. The van, equipped with horn speakers, a rack of amplifiers and associated microphones and 78RPM turntables, started operating in 1927 and although the current amplifiers and other items to be seen in the van were installed in the mid 1940's the contents are no less interesting for that.

Apparently the outfit was used as recently as June of 1958 and since then it has been in store. Now it has been restored and Mr. Clarabut has donated it to the

Association for Museum purposes.

Returning to a more modern theme, two points come to mind from the show. In the first place the fairly universal adoption of slider controls in tone and volume circuits. This has tended to give much of the equipment similarity of appearance which is emphasised with items like a Disco unit where the layout can be little else but symetrical if it is to work successfully.

A second point is the tendency to place wattage ratings at 100 or 200 nowadays. Apart from the obvious dangers to the listener's hearing when faced with power at this level coming from one speaker, there seems to be the question of this type of high level being fashionable rather than necessary. One or two of the loud-speaker suppliers made this point with surprise but no doubt at the same time with a degree of pleasure at increased sales.

Whilst much of the equipment was of the type one might see in any Discotheque or audio supply house, one or two items caught the eye. For example there was a portable speaker unit intended for indoor or

outdoor use mainly in PA applications.

Called the Electrovoice Sonocaster portable extension speaker, the unit uses an 8-inch transducer, can handle up to 30 watts peak and, whilst limited in frequency response to 70Hz to 13kHz, gives a very good showing in comparison to a £400 unit from the same source, Goulton Europe Ltd. Priced at between £11 and £16 (figure not yet finalised) this plastic-housed unit will no doubt collect its fair share of interest.

The latest news from the APAE is that John Robins, MD of SNS Communications Ltd., has been elected President. This year's President elect is Keith Monks of Fleet and John Weed of Uxbridge is again Treasurer.

PATENTS BEVIEW...

MAGIC WIPERS

Motorists who like the idea of triggering their windscreen wipers as if by magic should read BP 1 287 752 from Joseph Lucas Industries.

The Lucas circuit Fig. 1 shows a vehicle battery supplying power to an oscillator coupled to a transmitting aerial. Usually the aerial will be in the form of a wire loop built into the driver's seat belt. The receiver aerial, in the form of an electrode built into the vehicle dashboard, is connected to the gate of the field effect transistor TR1.

The circuit is physically and electrically constructed such that the coupling between aerials is insufficient to cause TRI to conduct. To operate the load (e.g. the windscreen wipers) the driver touches the receiver aerial to increase capacitive coupling between the two aerials. Transistor TRI conducts, base current reaches transistor TR2 and thereby also transistor TR3. Relay RLA then latches, due to the rectified current flow through the transistor, and "Hey Presto!" the wipers start up.

Because a self-latching relay is used it will hold the wipers working until next time the aerial is touched. A non-latching relay could of course be used for on/off touch control.

ULTRASONIC GUIDE STICKS FOR THE BLIND

Geoffrey Mowat in BP 1 284 027 provides interesting details for the construction of a walking stick for the blind with ultrasonic guide capabilities. In his patent Mowat shows a transmitter formed from an oscillator, TR7, and amplifier, TR8, see Fig. 2. These produce ultrasonic electrical oscillations which are converted into ultrasonic sound waves by transducer X2.

No details are given of this fairly routine arrangement, but it is suggested that initially oscillation trains should be transmitted 8 times per second. The transducer X2 is directional in that it transmits only over a beamed path.

For reception a transducer X1 converts received ultrasonic soundwave pulses into electrical pulses and applies them to the receiver amplifier, TR1, 2 and 3. The amplified echo pulses are

BP 1 287 752

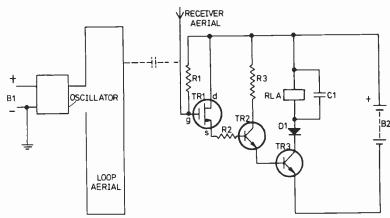


Fig. 1

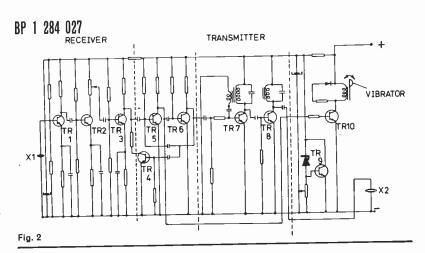
then fed to an astable multivibrator (TR5, 6) which, when no reflected signal is received, is free running.

When an object is close enough to the transducers to cause the latter to receive a reflection, the multivibrator will be triggered from its original to its alternative state for a predetermined length of time, and in which state it is insensitive to any subsequent pulses. When the multivibrator reverts to its original state it will cause the transmitter to emit a train of pulses again. Simultaneously an indicator amplifier TR10 will activate a vibrator which the subject can feel.

Thus when the transducer X1 receives a reflected pulse, the multivibrator will "flip" over and control a fixed cycle of events. As

the transmitted pulse will be received back in a shorter time as the reflecting object qets closer, the repetition rate felt at the vibrator will vary with the distance of the reflecting object, i.e. the vibrator will vibrate at a frequency which increases as the object gets closer. And, of course, because the multivibrator is triggered by the first received pulse, the unit as a whole will respond only to the nearest object in its path. Transistor TR4 attenuates direct path (non reflected) pulses.

The inventor claims that with the arrangement fitted in a walking stick, the system will allow ready distinction between objects such as a telegraph pole and a building behind it.





Timer C

By J.B. DANCE M.Sc.

SIGNETICS International Corporation has recently introduced a new economical integrated circuit, the 555, which can be employed in simple timing circuits for an extremely wide range of applications and is equally suitable for use by both the professional equipment designer and the amateur enthusiast.

The 555 devices can be employed to provide accurate time delays from microseconds to hours. The time delay is almost independent of the power supply voltage. The device can also be employed as an astable oscillator for pulse-width modulation as one of a series of timers or a frequency divider.

THE INTEGRATED CIRCUIT

The integrated circuit and its mode of operation will be described in some detail so that readers may gain an understanding of the circuit and thus be able to devise their own applications. Full constructional details of a general purpose timer appear on page 486 of this issue and show how this monolithic integrated circuit can be employed for automatically timing the exposure in photographic enlarging or as an industrial timer.

TYPES

The 555 timer is available in two types of package. An eight lead dual-in-line encapsulation with a silicon moulded body material is used for the NE555V, and the NE555T has a circular TO-99 case with eight leads. The connections to both types are shown in Fig. 2. The electrical characteristics of the two types are identical.

The SE555T is a close tolerance version of the 555 device and is available only in the circular TO-99 package at present. The connections are as in Fig. 1. The SE device can operate over the temperature range -55°C to +125°C, whilst the NE types can operate only over the narrower temperature range from 0°C to 70°C. The SE555T is considerably more expensive than the NE types and it generally provides a somewhat smaller drift of the time delay with the supply voltage and with temperature.

The writer feels that the dual-in-line construction of the NE555V is rather more convenient to use than the TO-5 encapsulation and it is the cheapest type of 555.

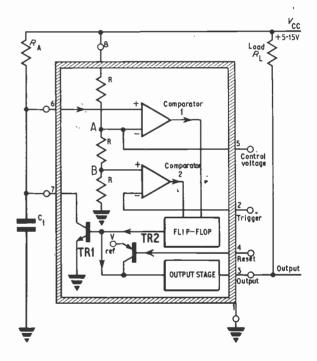


Fig. 1. General schematic of the 555 timer chip showing external circuit connections for monostable operation

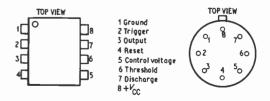


Fig. 2. Pin number connections for 8-pin DIL and TO99 can

The NE555V has therefore been used in the applications which will be described, but the NE555T or the SE555T are equally suitable.

MONOSTABLE OPERATION

In the monostable mode the timer is triggered by an input pulse or by the operation of a switch. This causes the output voltage to change for a pre-determined time (the delay) after which this voltage returns to its former value. The delay is determined by the product of the values of a capacitor and resistor connected externally to the integrated circuit.

The integrated circuit is shown connected as a monostable in Fig. 1 with the operational functions of the timer shown as blocks.

Initially the external capacitor, C_1 , is kept in a discharged state by the transistor TR1 inside the timer. This transistor is held in a fully conducting state by the bias applied to its base by the flip-flop stage.

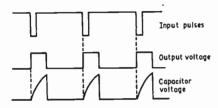


Fig. 3. Waveforms associated with the operation of the circuit of Fig. 1

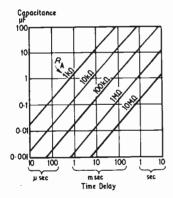


Fig. 4. Time delays obtainable with various values of C, and R,

The point B is held at a potential of $V_{ee}/3$ by the potential divider containing three resistors of equal value R. When a negative-going trigger pulse causes the potential of pin 2 to fall below the value $V_{cc}/3$, comparator 2 causes the flip-flop to be switched. results in the output voltage rising and in TR1 being cut off.

This sequence of operations may generally be started by momentarily connecting pin 2 of the timer to ground instead of applying a negative-going pulse to it. However, triggering will then normally occur at the instant the connection with ground is broken and this may introduce an appreciable error if the delay period is short.

WAVEFORMS

The voltage across C_1 now increases exponentially with a time constant $R_A C_1$ as current flows to it via R_A ; the waveforms are shown in Fig. 3. When the voltage across C_1 becomes equal to that at point A (that is, to $2V_{\rm cc}/3$), comparator 1 of Fig. 1 resets the flip-flop. The output from the latter causes the voltage at the output (pin 3) to return rapidly to its quiescent low value. In addition, TR1 is biased so that it conducts and the capacitor C_1 is rapidly discharged.

SUPPLY VOLTAGE

If the value of the supply voltage, V_{ee} , is increased, the potential of point A and the rate of charge of the capacitor C_1 at any given point in the charging cycle are both increased in proportion. The time at which the two inputs to comparator 1 become equal (that is, the time at which the end of the delay occurs) is therefore almost unaffected by the value of V_{ee} .

RESET

Once the circuit has been triggered by a negative going pulse to pin 2, it will remain in this state until the pre-set time has elapsed, no matter whether it is triggered again during this time or not. If, however, a negative going pulse is applied to the reset terminal, pin 4, before the circuit has returned to its quiescent state, the capacitor C_1 will be discharged and the circuit will be reset.

The output is in its low voltage state during the time the reset pulse is applied to pin 4. Resetting can also be effected by momentarily connecting pin 4 to ground. The reset current is about 100μ A.

In applications where the reset terminal will not be used, it is advisable to connect it to the positive supply line to avoid any possibility of undesired resetting.

TIME DELAY

The time, t sec, for which the output is in its high voltage state is given by the equation:

Equation 1 $t = 1.1 R_{\rm A} C_{\rm 1}$ where R_A is expressed in ohms and C_1 in farads.

Equation 1 may be deduced in the following way. When a capacitor C_1 charges from a voltage V_{cc} through a resistor R_A , the voltage V across the capacitor after a time t is given by the equation:

$$V = V_{-1}(1 - e^{-t/R_AC_1})$$

 $V = V_{cc}(1 - e^{-t/R_AC_1})$ In order to find the delay, t. we put $V = 2V_{cc}/3$, since this is the voltage at which the flip-flop is reset.

$$2V_{cc}/3 = V_{cc}(1 - e^{-t/R_AC_1})$$

$$e^{-t/R_AC_1} = 1/3$$

$$t/R_AC_1 = -\log_e(1/3) = \log_e 3 = 1 \cdot 1.$$

If one requires the output to remain in its high voltage state for one millisecond (10-3s), one may choose a reasonable value of C_1 , say $0.1\mu\text{F}$ (=10⁻⁷F), and calculate R_A using equation 1 to see if a reasonable value is obtained:

 $R_{\rm A} = t/1 \cdot 1C_1 = 10^{-3}/(1 \cdot 1 \times 10^{-7}) \approx 9 \cdot 1 \, k \, \Omega$. Similarly, if $R_{\rm A} = 100 \, k \, \Omega$ and $C_1 = 100 \, \mu \, \text{F}$, $t = 1 \cdot 1 \times 10^5 \times 10^{-4} = 11$ seconds. The chart of Fig. 4 shows the time delays obtainable for values of C_1 from $0.001\mu\text{F}$ to $100\mu\text{F}$ and for values of R_A from $1\text{k}\Omega$ to $10M\Omega$.

SERVICE CONTRACTOR OF CONTRACT

It should be noted that reasonable values of R_A and C_1 must be selected. A typical current of $0 \cdot 1 \mu A$ (maximum $0 \cdot 25 \mu A$) flows through R_A to pin 6 of the 555 timer. If R_A is $20 \text{M}\Omega$, this current alone will produce a voltage drop of up to 5V (typically 2V). Thus this value of R_A is about the maximum which should be employed.

If R_A is chosen as $10M\Omega$ and C_1 as $10,000\mu$ F, t can be calculated as about 30 hours. However, the leakage current passed by the capacitor might well be so large that a much longer delay occurs before the voltage across the capacitor reaches a value of $2V_{cc}/3$.

Apart from the fact that capacitor leakage current can affect the calculated time values, one should remember that the values marked on many capacitors are only very approximate. This is especially true in the case of electrolytic and Hi-K ceramic capacitors. The actual value may exceed the marked value by as much as 50 per cent.

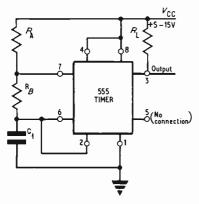


Fig. 5. Circuit of the 555 connected for astable operation

CONTROL VOLTAGE

In the previous discussion it has been assumed that pin 5 of the timer circuit has been left unconnected. However, the time delay may be changed over a range of about 10:1 by applying various control voltages to this pin. The impedance at pin 5 is a few thousand ohms. The time delay will still be independent of the value of $V_{\rm cc}$ if the control voltage is derived from $V_{\rm cc}$ by means of a potential divider so that it is proportional to $V_{\rm cc}$.

OUTPUT

If one merely requires an output pulse at the end of the delay time, one may connect the output pin 3 to a resistor (perhaps $4.7k\Omega$) which is returned to the positive supply line, as in Fig. 1. The output pulses are taken directly from pin 3.

When the output voltage is high, it is near to that of $V_{\rm cc}$. When it is low it is only very slightly greater than the ground potential of the negative supply line. If the output voltage is low and the current passing to pin 3 is 10mA or less, the output voltage will normally be within 0.2V of the ground potential. At output

currents of 100mA, the output voltage is usually within 2V of the potential of one of the supply lines.

An output current of up to 200mA can be obtained in either the high or low voltage states.

The output pulses rise and fall rapidly, typical rise and fall times being 0·1µs. The output pulses can be used to drive a high power transistor (such as the common npn type 2N3055 or the somewhat smaller npn types 2N3054 or RCA 40250). The high power transistor can then control a high current.

RELAY OPERATION

If the delay time exceeds about 0·1 sec, a small relay connected between pin 3 of the 555 and one of the supply lines can be operated directly from the timer circuit. The relay coil should be designed so that it operates from a voltage approximately equal to $V_{\rm cc}$ at a current which does not exceed 200mA.

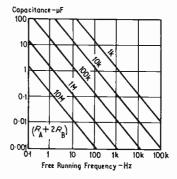


Fig. 6. Variation of frequency with component values in the astable mode

It is necessary to connect a diode across the relay so that the back e.m.f. generated when the current ceases to flow through the inductive relay coil is shorted out by the diode. This prevents possible damage to the timer circuit by the fairly high reverse transient voltage which appears across the relay coil. The diode must be connected so that it is reverse biased when the relay conducts. A fast switching diode is ideal.

TRIGGERING

The writer has found that the use of an inductive load (such as a relay) can cause the timing circuit to automatically re-trigger itself at the end of the delay time. This occurs so rapidly that the output voltage appears to stay quite constant and the relay remains closed. In the general purpose timer to be discussed, this problem is avoided by connecting pin 2 of the 555 through a resistor to the $+V_{\rm cc}$ line.

The triggering action in the 555 is extremely sensitive. If one touches pin 2 with one's finger, triggering will occur. Even moving one's hand near to a wire connected to pin 2 is adequate to trigger the circuit by a capacitive effect. The trigger current required is only about $0.5\mu A$ for a period of $0.1\mu s$.

D+ N'ESSSNOT=4.5v->16v lut SE has of to ISv. (perhaps 18)

POWER REQUIREMENTS

The integrated circuit itself requires a current of about 3mA (maximum 6mA) from a 5V supply, increasing to about 10mA (maximum 15mA) from a 15V supply. Any current passed by the load is additional to the current shown. The absolute maximum permissible power dissipation is 0.6W, but this should not be reached if the correct voltage is applied even if the maximum permissible output current of 200mA is passed.

The NE555V and NE555T should be operated from a suited developerating of the control of the cont

Variations of the delay time with the supply voltage are typically 0.01 per cent per volt and with temperature 0.005 per cent, per deg C.

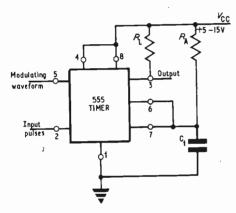


Fig. 7. Using the 555 for pulse width modulation

ASTABLE OPERATION

. If a 555 timer circuit is connected as in Fig. 5, it will "free run" and operate as an astable multivibrator. The trigger pin 2 is connected to pin 6 so that when C_1 discharges the resultant negative going pulse is used to trigger a new cycle automatically.

The current required to charge the capacitor C_1 flows through R_A and R_B in series. However, when TR1 (see Fig. 1) conducts, a current from C_1 flows through R_B into pin 7. Thus the charging time is proportional to $(R_A + R_B)C_1$, but the discharging time is proportional to R_BC_1 . The charging time cannot be made smaller than the discharging time.

TIMING

In the astable mode of operation, the capacitor C_1 charges and discharges repeatedly between the potentials $V_{\rm cc}/3$ and $2V_{\rm cc}/3$ provided that pin 5 is left unconnected. The charge and discharge times (and therefore the frequency of operation) are almost independent of the supply voltage, $V_{\rm cc}$. The output

voltage is high during the charging time.

Charging time
$$= t_c = 0.693(R_A + R_B)C_1$$
 Equation 2
Discharging time $= t_d = 0.693R_BC_1$ Equation 3
Total period $= t_c + t_d = 0.693(R_A + 2R_B)C_1$ Equation 4
Frequency $= 1/t = 1.44/(R_A + 2R_B)C_1$ Equation 5

The variation of frequency with component values is shown in Fig. 6.

Equation 2 can be deduced using the equation $V = V_{\rm cc}(1 - {\rm e}^{-t/RC})$ for a capacitor C charging through a resistor R from a voltage source $V_{\rm cc}$. One requires the time for V to increase from $V_{\rm cc}/3$ to $2V_{\rm cc}/3$ when charging through a resistor of value $R = R_{\rm A} + R_{\rm B}$. The discharging time is the time for V to decrease from $2V_{\rm cc}/3$ to $V_{\rm cc}/3$ when a resistor $R_{\rm B}$ is connected across the capacitor.

The charging and discharging times can be altered by the application of a control voltage to pin 5. Both of these times are affected, since an alteration of the potential of point A in Fig. 1 will affect the potential of point B.

OTHER APPLICATIONS

The astable mode of operation can be employed when a series of operations must be repeated at preset intervals many times. For example, one application occurs when one wishes to have the windscreen wipers of a car make single sweeps with a certain delay between successive sweeps. The circuit continues to operate with this delay until the timing is adjusted or until the current is switched off. Another application using the 555 in the astable mode involves the periodic switching of lights.

If a series of regularly spaced pulses is fed into pin 2 of Fig. 7, the mark/space ratio of the output from pin 3 is dependent on the instantaneous modulating voltage applied to pin 5. As the voltage at pin 5 increases, the time for which the output remains in its high voltage state increases until it becomes so long that alternate input pulses produce no output.

A number of the 555 timers can be connected in the monostable mode so that the output of the first triggers the second and the output of the second triggers the third, etc.

The operation of the first timer is started by connecting pin 2 momentarily to earth or by applying a negative pulse to it. The first 555 returns to its quiescent state after 1 sec (see equation 1) and triggers the second timer. This circuit produces an output pulse perhaps 100 sec later (owing to the higher values of R_3 and C_3) and triggers the third 555 circuit.

The output pulses from each of the timer circuits can be used to carry out any desired operation at the preset times after the first timer is started. If desired, the output from the last circuit may be used to trigger the first circuit so that output pulses continue to be generated at set intervals indefinitely.

On page 486 a general purpose timer constructional article discusses the use of the 555 in practical circuitry suitable for use in, say, a darkroom timer.

POINTS ARISING

NOVEL BATTERY ELIMINATOR (April 1973)

To prevent the possibility of any electrical accidents the cover and base of this unit should be made up of 18 s.w.g. aluminium. Alternatively an Ever-Ready 3-way 13A adaptor can be used.

For 6V use a 6.2V 400mW Zener diode can be

substituted such as the BZY88.

CAMERA SHUTTER TESTER (August 1972)

In Fig. 4; page 643, there should be a break in the copper strip between the end of R1 and the link wire which is close to VR1. The presence of this

incorrect link will probably damage the integrated circuit. Fig. 3, page 642, the reference to the integrated circuit as 7410PA should in fact read 741 operational amplifier.

GEMINI TUNER (April 1972)

The Gemini Tuner described over a year ago, in April 1972 has raised considerable interest and several hundred tuner heads have been supplied. Unfortunately these are made in Japan for an American company (thus keeping the cost down) who has now withdrawn the line because of shortage of some components and a changing product line. Both the importers and Practical Electronics have made extensive investigations to discover an alternative source or replacement product. Neither

LOGIC TUTOR EXPERIMENTS...



 $\label{eq:substitute} \textbf{S} \begin{tabular}{l} \textbf{OMETIMES} \b$

DEMONSTRATING IT

A simple application of inversion can be demonstrated on the Logic Tutor. Normally a level I will cause the lamps to be illuminated. Let's say we want a level I to extinguish a bulb; the first thing that has to be done is convert the I level to a 0. We have no simple transistor stages available to us on the tutor but we can simulate inversion by using one of the 2-input NAND gates. All you have to do is short both inputs together (next month's description of the NAND gate will make the reason for this clearer).

Connect the shorted inputs to one of the toggle switch outputs and also connect this node to one of the indicator lamps: also connect the output of the NAND to another lamp (Fig. 2.2). Set the switch to I and LPI will indicate level I at the input but LP2 will be extinguished indicating 0. Apply a 0 on the input and you will see a I appear at the output.

The symbol for invert is shown in Fig. 2.3 and we say that the output Q is \bar{A} (said NOT A). The truth table is very straightforward—whatever A is, Q is the opposite.

Invert is often used in logic systems but its occurence is sometimes disguised by the fact that it is mainly used in conjunction with AND gates to give us the NOT AND or NAND function that will be explained fully in the next issue.

by M. Hughes

In Part I last month the figures given should be transposed.

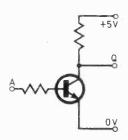


Fig. 2.1. A grounded emitter stage acts as a logic inverter

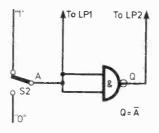


Fig. 2.2. The NAND gate with logic inputs shorted demonstrates the invert function

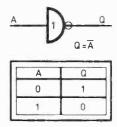


Fig. 2.3. The symbol and truth table for the invert function

Electronic Music Production

Alan Douglas

The author describes the tremendous potential of electronic music which is free from the limitations of conventional instruments and notation. The three principal methods of synthesizing music electronically are discussed, with examples of the latest available equipment. Excerpts from the writings of the world's leading researchers and other previously unpublished material are also included. This practical guide will be welcomed by all working in this field.

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PRACTICAL ELECTRONICS "SCORPIO" ELECTRONIC IGNITION SYSTEM



This Capacitor-Discharge Electronic Ignition system was described in the November and December issues of Practical Electronics. It is suitable for incorporating in any 12V ignition system in cars, boats, go-karts, etc., of either pos. or neg. earth and up to six cylinders. The original coil, plugs, points and contact-breaker capacitor fitted in the vehicle are used. No extra or special components are required. Helps to promote easier starting (even under sub-zero conditions), improved acceleration, better high-speed performance, quicker engine warm-up and improved fuel economy. Eliminates excessive contact-breaker point burning and the need to adjust point and spark-plug gaps with precision.
Construction of the unit can easily be completed in an evening and installation should take no longer than half an hour. A complete complement of components is supplied with each kit together with ready-drilled roller-tinned professional quality fibre-glass printed-circuit board, custom-wound transformer and fully-machined die-cast case. All components are available separately. Case size 7½in × 4½in × 2in approx.

Case size /210 A 7310 A 2000 approx.
Complete assembly and wiring manual 25p, refundable on purchase of kit. Price: £10-50 plus 50p P. & P.

PSYCHODELIC LIGHTING UNIT Mk. 3



This unit represents a natural progression from our phenomenally successful Mk. I and 2 Units As before the drive voltage is derived directly from the amplifier cusps or across the speakers. The unit converts the audio frequency signals into a three-coloured light display; the colour depending on the frequency of the signal and the intensity on the loudness of the audio source.

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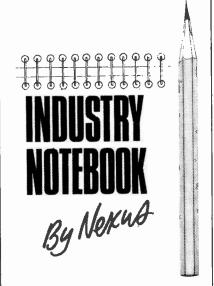
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FAMINE AHEAD?

It seems to be a flaw in our human character, perhaps even an immutable law, that we are unable to govern our affairs smoothly and sensibly. After the glut comes the famine—at least in professional components. Delivery times are lengthening all the while and prices are hardening and this trend is noticeable throughout Europe and the United States. The trend is that much more irritating because it is in such contrast to even a few months ago.

The Paris Components Show, still the world's biggest and best, showed a buoyancy in April unmatched since the boom year of 1969. And the London show which opens on May 21 will, according to pre-show gossip, be a much happier business event than the last, two years ago.

Some reports give lead times for deliveries of up to six months for capacitors, five months for potentiometers, up to four months for transistors.

Much of the trouble in Europe stems from the big upsurge in colour TV demand which has created an almost unprecedented call on components. And component manufacturers who burnt their commercial fingers by overexpanding production facilities and then had to cut back with heavy losses are quite naturally cautious about expanding again to meet what might only be a temporary boom period.

Component distributors in Britain who belong to the Association of Franchised Distributors of Electronic Components and therefore subscribe to an ethical standard in

their businesses, are almost as embarrassed by the present upsurge in business as when it was bad. The more sensitive of them are tormenting themselves on the ethics of switch selling (i.e. the supply of an equivalent component from a different manufacturer from that actually ordered) and suchlike niceties. Another of their problems is the so-called wheelerdealer who scours the world for job lots and then unloads them on a component-hungry market. The maxim is let the buyer beware and to trade only with reputable suppliers.

UNION CARBIDE EXPANSION

The solid tantalum capacitor was a hard-to-make sophisticated component which sold almost exclusively to the exotic space and defence market where price is secondary to performance and reliability. I was somewhat surprised, therefore, to discover on a visit to the Union Carbide plant at Aycliffe, Co. Durham, that the big expansion just completed is largely the consequence of tantalums having penetrated the consumer electronics market. Here is another company that has benefited from colour TV.

They have spent £250,000 beefing up production with new machinery and processes, much of it developed at the US production base at Greenville, South Carolina. The Aycliffe plant now turns out 2.5 million capacitors a month, a threefold expansion over one year ago. Marketing manager, Andy Thomson, was in jubilant mood explaining that the big takeoff was in low-cost dipped resin types which had opened up an entirely new and still expanding market.

The Aycliffe plant opened in 1952 with 17 employees and now has 230. Watch out for further developments. In the pipeline is a new range of monolithic ceramic capacitors which is forecast to make an important contribution to Union Carbide's European operations. Ceramic and tantalum chips will be included to meet the demands of the fast-growing hybrid market.

Jogging along in the background is a little-publicised activity—the manufacture of barium getters for TV tubes and radio valves which Aycliffe workers can turn out at the rate of 120 million a year, phew!

STILL DROPPING

The huge mark-ups of some manufacturers of pocket calculators are slowly being eroded as more types come on the market,

competition increases, production problems are ironed out and the cost of expensive tooling is amortised. This is one area where prices are not hardening.

Nearly 40 models are currently on sale in London shops, most of them carrying big discounts from the "recommended" price. One British manufacturer has chopped his "recommended" price by £20, and it is generally agreed that the simpler types will ultimately become available for a little over £20.

Shrewd buyers are holding off waiting until the market bottoms and some even shrewder people have been working on how to solve complex calculations on the simpler machines which are basically only four functions, add, subtract, multiply and divide.

Square roots, cube roots and other calculations are possible by using a number of discrete stages according to extensive correspondence in the US journal "Electronics" which has printed a number of ingenious methods of expanding calculator capability without paying for it.

POCKET BREATH ANALYSER

The annual Physics Exhibition in London is not normally the best place to look for good commercial ideas. But one item I spotted this year has good commercial prospects if the price is right.

It is a pocket breath alcohol meter which uses a fuel cell developed at the University of Wales Institute of Science and Technology, Cardiff. Breathing into the cell develops a potential which is amplified and displayed on a voltmeter. Accuracy is claimed to be within 5 per cent and sensitivity is such that 0.005mg of alcohol per litre of air can be detected. Size is a modest 6 × 10 × 2cm and weight is only 60g—a truly pocket sized instrument.

I also noted that Standard Communications Laboratories are gamely pressing on with optical waveguide communications. A liquid core fibre using tetrachlorethylene as the transmission medium is said to have sufficiently low attenuation for practical use and another solution is to use a single glass fibre waveguide instead of the conventional fibre optic bundle.

The single fibre, say STL engineers, is better from all points of view from production which can be carefully monitored during extrusion through to more robust connection to the laser source and detection unit.

Wide Range

PULSE GENERATOR

By M.J.Trand

COMPACT and lightweight source of fast pulses is frequently required for testing. Integrated circuits can meet this need and the pulse generator described here uses only TTL i.c.s as active devices with the addition of a few extra resistors and capacitors. The unit may be powered by a standard mains supply, but a battery operated version can easily be adopted without any modifications to the circuitry.

INTEGRATED CIRCUITS

As mentioned, two 74 series i.c. chips produce the rectangular shaped pulses. Type SN7404, which is a sextuple single-input inverter, uses three of its inverters as part of the oscillator section. The second circuit block, type SN74121, is a monostable multivibrator connected to provide pulses of variable widths. These two integrated circuits can be obtained for well under £1 from many electronic components distributors.

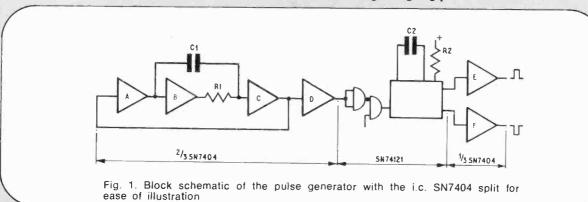
It should be mentioned at this stage that although several firms market i.c.s of the 74 series, similar devices such as the Mullard FJH241 and FJK101 will do equally well.

CIRCUIT

The block circuit diagram of the pulse generator is shown in Fig. 1. On switching on, the output of inverter A tends to go up to a logical 1 while the output of inverter B is pulled down to a logical 0, charging the capacitor C1. The speed of this action will be governed by the time constant C_1R_1 in the circuit. The output of inverter C will then gradually go to a 1 and also transmit the transient to the input of inverter A, whose output in turn follows to a 0, and so on, thus producing an oscillatory action. The fourth inverter D in the line acts simply as a buffer stage.

The pulses are next fed to a monostable block which will vary their duration according to the value of the external time constant applied by C2 and R2.

At the last stage pulses from the monostable complementary outputs are passed on to the remaining two inverters E and F, providing simultaneous positive and negative going pulses.



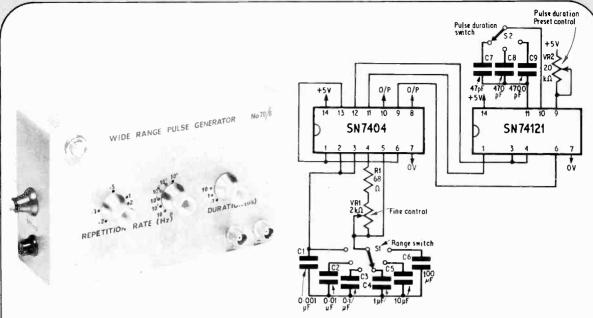
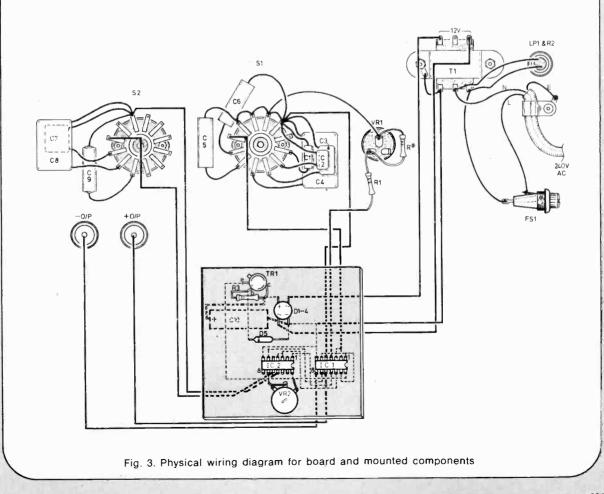


Fig. 2. Circuit diagram of the components external to the integrated circuits. VR1 is, in the final model, a $2 \cdot 2 k \Omega$ linear potentiometer with an $8 \cdot 2 k \Omega$ resistor connected from wiper to the end proximate R1



Components

GENERATOR

Resistors

Rf 68Ω (‡W)

Potentiometers

VR1 2kΩ Lin VR2 20kΩ Lin

Capacitors

C1 0.001 µF

C2 0-01 µF

C3 0-1 µF

C4 1.0µF

C5 10µF

C6 100µF C7 47pF

C8 470pF

C9 4700pF

Switches

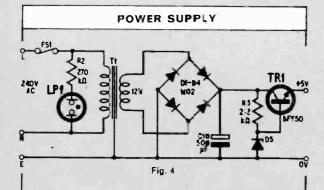
S1 Rotary, single pole, 6-way S2 Rotary, single pole, 3-way

Integrated circuits

IC1 SN7404 IC2 SN74121

Miscellaneous

Die-cast box and suitable knobs



Resistors

R2 270kΩ (‡W) R3 2·2kΩ (‡W)

Capacitors

C10 500 µF. 25V elect.

Diodes

D1-4 WO2 (Bridge)

D5 5.6V Zener

Transistors

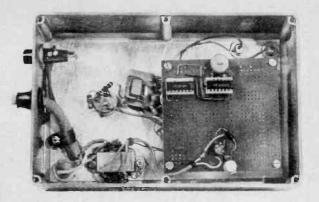
TRI BFY50

Transformer .

T1 12V output mains transformer

Miscellaneous

FS1 100mA fuse and panel-mounting holder LP1 Neon indicator and panel-mounting holder



CONSTRUCTION

A circuit diagram is shown in Fig. 2. and a wiring diagram in Fig. 3. In practice the two 14-pin i.c. holders are soldered on a small piece of perforated p.c.b. as shown, keeping wiring lengths small. The value of potentiometer VR1 should not exceed $2k\Omega$, as this would attenuate the feedback signal and stop oscillations.

A series resistor R1 of about 800 increases the stability of the oscillator when VR1 is at minimum. The value of capacitors C1 to C6 can be chosen from 0.001 µF to 100 µF giving a frequency range from 1Hz to above 1MHz in six steps.

To fit in with the frequency range chosen, C7 to C9 can have any reasonable value between 10pF and 10,000pF giving pulse durations from 0-lus to 10/4S.

If longer pulse durations are needed at lower frequencies then larger values of capacitor can be added. When electrolytic capacitors are used connect the negative terminal to pin 11 of the mono-

The switches \$1 and \$2 are ordinary single-pole rotary type.

SETTING UP

Calibration of the pulse generator, particularly VR1 the repetition rate fine control and VR2 the pulse duration preset control, can easily be done with a reference source such as the time base of commercial oscilloscope. This should be sufficiently accurate to enable a fairly precise marking of the front panel.

Pulses of 4V amplitude with a rise time of 40ns, and fall time of 20ns into about 150Ω are obtained provided stray capacitance is kept at a minimum. If necessary pulses with smaller amplitudes can be obtained using a set of plug-in attenuators.

POWER SUPPLY

A positive 5V stabilised d.c. supply is required. The diagram of such a circuit is shown in Fig. 4. The series transistor Zener diode (TR1-D5) arrangement provides sufficient stabilization against mains variations of ±10 per cent (nominal 230V a.c.). The d.c. current requirement is about 30mA.

If the instrument is to be battery operated it is recommended that the low power types of TTL be used (e.g. 74 L) in conjunction with a rechargeable battery-nickel cadmium for instance.

The new EMI 1515 stereo amplifier and smaller EMI SQ1500 quadraphonic decoder shown here with two LE2 15W loudspeakers forming a complete "add-on" package for converting an existing stereo system for SQ quadraphonic operation







The Saba 544G stereo reel-to-reel unit illustrates the modern styling current in today's equipment

PROBABLY the main theme of this year's Sonex exhibition, Sonex 73, held at the Excelsior Hotel, London Airport, was quadraphonic or 4-channel sound. Of course, much of the exhibition and the associated lectures was concerned with the more mundane things of life like the mechanics of disc reproduction, listening room acoustics, and so on.

But, as at any gathering of Hi-Fi enthusiasts, and Sonex was certainly that, the tendency is to discuss the latest development available on the market. Currently quadraphonics falls into this category, with all the usual arguments as to benefits of different systems and

methods.

Many of the exhibitors were prepared to comment on the validity of 4-channel sound but only a few offered equipment and, of course, in the small hotel rooms used by the exhibitors it was impossible to demonstrate the effects available decently.

In fact, this is one of the problems with using a hotel for an event of the nature of Sonex 73. Unlike RECMF or IEA or, for that matter any show held in a large hall-type space, not only is it difficult for the visitor to scan the field in general for his particular interest but hotel corridors have a habit of all looking the same.

Thus there is a tendency to miss items of interest and to get "lost" or disoriented. The only solution being the heavy reliance on the handbook. And we all know how impossible it is for all the information to be present

there.

Of course, the hotel environment does have advantages. As the sound-making equipment was housed in separate rooms it was possible for the exhibitor to give an almost private demonstration of his gear. The usual problem of interference from stand to stand was avoided.

Really Sonex is directed to the audiophile with the desire to purchase already-built equipment, with only a scattering of items such as Richard Allan speaker kits and Connoisseur turntable products for the more practically minded. However, it is always interesting to see how the professionals do things.

LOUDSPEAKERS

It is an accepted fact that the loudspeaker is one of the weakest links in the chain of sound reproduction and many of the exhibits showed just how far one can go with money and good intentions to put the orchestra in the sitting room. Anything up to several hundred pounds per unit can be spent. However, there must be a rational level between TV sound standards and using the dining room as a sound box for the rest of the house.

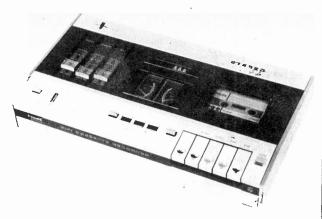
An interesting demonstration of loudspeaker performance was given by Acos using the Martin range from America. At the low price end was the Micro Dan, a America. At the low price end was the Micro Dan, a prototype baby unit which is expected to sell here for about £30 plus VAT per pair. The Mini Dan, already available here at £56.28 plus VAT per pair also gave a good performance for the price. But most impressive were a pair of Laboratory Mk. 2 units and although these sell at £57.95 each plus VAT the performance was perhaps more "musical" than anything else at the show.

Of course, these comments are governed by personal hearing to a degree but when it is remembered that the Laboratory Mk 2 units include three speakers and controls to set mid and high frequency balance, allowing "tuning" of the equipment to a room, a hearing test is in the end, the only valid method left.

In fact it was commented by someone at the show that if you provided an audience with "flat" response speakers then most people would not appreciate the

resulting sound at all.

Again on the subject of speakers, Richard Allan were displaying their well-known range of kits and assembled Hi-Fi units. Whilst they have not added to the kit range they were demonstrating an interesting professional/monitor unit, still in the prototype stages, called the Academy. This is physically a large item and not at all of the kittable type. It is expected to appear on



The Phillips N2510 stereo cassette unit, to DIN 45500 standards, is intended for use with Chromium Dioxide tapes

the market some time in July when the price will be settled. At the moment it is envisaged that £70 to £80 will be about right.

MADE OVERSEAS

Much of the equipment on show is made overseas. We have already mentioned American sourcing. From Norway comes the Tandberg line handled in the UK by Farnell-Tandberg Ltd. Included are a wide selection of tape and deck units, loudspeakers, separate amplifiers, tuner units and composites of various denominations. It is interesting to note the adoption by Tandberg, as almost everyone else, of a cassette deck which meets 45500 DIN requirements and includes a Dolby noise reduction system.

Japanese equipment was fairly in evidence and, as is to be expected, set a standard of appearance which some sources near home are having difficulty meeting. Again the emphasis was on 4-channel sound with specially designed controllers for remote handling of the output of each channel and special headphones designed for quadraphonics.

Of course, the Japanese exhibits included many examples of casette units, some claiming performances equal if not superior to reel-to-reel equipments.

SIDE SHOW

Perhaps one of the problems of an event of this type is the consistency of the exhibits. Whilst not quite an "If you've seen one you've seen them all" situation, there is certainly a tendency to feel this after a matter of hours looking at the exhibits.

Thus it was somewhat of a relief to vacate the Excelsior for a while and go to see a demonstration of quadraphonics put on by EMI at a separate hotel. Intended to launch two of their latest products for the Hi-Fi market, the EM 1515 stereo amplifier and the SQ 1500 quadraphonic decoder, this side-show (as it were) took the cake as a demonstration of the abilities of quadraphonics to enhance the realism of reproduced sound.

In addition to announcing and displaying the amplifier and decoder, EMI showed a new small eliptical speaker unit which is expected to come on the market soon at about £30/pair cased.

The amplifier is to sell at £46.50 including VAT and includes a claimed performance in excess of most equipments in this mid-price area. 15 watts per channel at 0.2 per cent distortion, and the ability to control power supply to ancillary equipment are some factors of interest.

The decoder is designed to give one the ability to handle the new SQ 4-channel records in conjunction with four channels of amplification. The 1515 is suited to this and EMI are to offer a kit including the decoder and one amplifier to those already possessing a stereo unit.

NEWS BRIEFS

Advanced Motorway Communication System

A new data transmission system to control the roadside telephones on the M2 has been ordered by the Northern Ireland Ministry of Development. Designed and manufactured in Britain by AP Electronics of Chiswick, the system will be capable of replacing a 185 core cable with just two wires.

The heart of the system is a complex logic network constructed from Motorola MCMOS units. With the new system the operator will be able to dial any of the roadside telephones so that emergency situations can be

handled before the less urgent breakdowns.

Underwater Detection of Aircraft Wreckage

Trials to improve the techniques for the detection of aircraft wreckage by sonar have just been carried out in Torbay. Designed to test the limitations of sonar detection, the trials covered a four-week period and the results should help in developing more effective systems.

A unique feature of the trials is the use of magnetic tape for the recording of the raw data required later for numerical analysis. More sensitive magnetic tape allows easier measurement of signal strength and hence easier differentiation between different types of wreckage.

Apart from the underwater detection of aircraft wreckage the techniques under trial could also be useful in such areas as location of ship wrecks, tracking of oil pipes and the assessment of inshore sediment accumulation.

Aircraft Tactical Simulator Takes Off

What is described as the world's most advanced simulator is now being "flown" by RAF crews after completion by the manufacturers, Marconi Space and Defence Systems. Built as the first of a number to be supplied to RAF Strike Command under a Ministry of Defence contract worth approximately £5 million, the simulator will reproduce for all 12 trainee crewmen, every operational facet of the world's most advanced anti-submarine aircraft, the Hawker-Siddeley NIMROD.

All operational equipment, including the radar, sonar, tactical navigation and weapon delivery systems, is fitted in a replica NIMROD, to reproduce actual missions, even

down to engine noise and low-level buffeting.

PO Contracts for 120 Mbit/s Digital Line System

As a first step to developing a digital trunk network and preparing for new facilities, the Post Office has placed contracts with STC, GEC and Plessey to develop digital transmission systems enabling pulse code modulation (PCM) to be used on Britain's trunk network.

The decision to develop a digital system for the UK trunk network stems from the results of feasibility studies carried out for the Post Office by GEC and Plessey in 1970-71. These studies confirmed that it is technically possible to introduce a digital system using the standard 1-2/4-4mm coaxial cable already in use for multichannel frequency division multiplex transmission.

multichannel frequency division multiplex transmission.

Under development contracts STC, GEC and Plessey have been commissioned to design, develop, manufacture and install systems transmitting information at a rate of 120 Mbit/s. Links between Guildford and Portsmouth, and Portsmouth and Southampton are to be set up, each system capable of transmitting up to 1,680 telephone conversations simultaneously.

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В	86	100	Sil. Diodes sub. min. IN914 and IN916 types	50p
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The ideal conditions for stability are as follows. The current through R3 should be equally divided between TR1 and TR2, thus implying that the base potential of TR2 should be the same as that of TR1, namely $V_{\rm tef}$. From this we can deduce that

$$V_{\rm ref} = V_0 \frac{R_5}{R_5 + R_6}$$

$$V_0 = V_{\rm ref} \, \frac{R_5 + R_6}{R_5}$$

From the last equation it can be seen that there are two alternative methods of varying V_0 ; both are shown in Fig. 1. Firstly we can vary the ratio of R_0 to R_0 by means of VR2. This suffers from the disadvantage that as V_0 changes so does the loop gain, invariably to the detriment of circuit performance.

The second alternative is to vary the value of $V_{\rm ref}$ by means of VR1. However, examination of the circuit discloses yet another disadvantage. For low values of $V_{\rm ref}$ the current through TR1 is relatively low but there must be a large voltage drop across R4 thus implying a high current through TR2. The converse applies for high values of $V_{\rm ref}$ and it should not be difficult to see that there is only one setting of VR1 that fulfills the fundamental condition for stability.

If, however, we replace R4 with another transistor stage TR5 (Fig. 2), which receives its base bias via the collector load resistor R8 of TR1 (Fig. 2) then this problem is resolved. If for all settings of VR1, the voltage across R8 is large compared with the emitter-base voltage of TR5 and the value of R8

is roughly the same as R4, then the current flowing through these two resistors will be substantially the same.

This circuit has the inherent advantage that TR2 has for its load the intrinsic collector resistance of another transistor which, having an extremely high incremental resistance, increases the loop gain to a value of several hundreds.

Since, at high frequency the base emitter junction of a transistor can be regarded as an RC network there will be a frequency at which this circuit will oscillate, although the frequency of oscillation is difficult to predict, depending largely on component layout and the types of transistors used. Instability can be prevented by making the output the dominant time constant by means of a capacitor, the minimum value of which in the prototype was found to be around 10µF.

This circuit has an output which is continuously variable from 6V to 18V at currents up to 1 amp with the output transistor suitably mounted on an adequate heat sink.

J. Davies, London W.14.

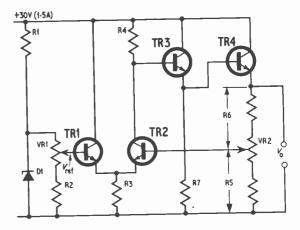


Fig. 1. Circuit for a conventional stabiliser

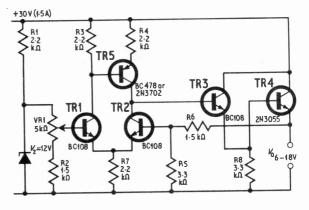


Fig. 2. Final circuit for a variable stabiliser for power supplies

BUILT the lamp strobe (Ingenuity Unlimited, April 1972) but found the multivibrator in my circuit was unstable and tended to lock at mains (50Hz) frequency. This was overcome by decoupling the supply and adding two more diodes to the bridge rectifier allowed pulsating d.c. to reach the gating transistor and hence trigger the thyristor. See Fig. 1.

The speed control circuit was also modified and now gives a range of about 2-12Hz, with 10µF timing capacitors, which gives a good strobe effect when using several low power lamps in parallel. Coloured 15W "pigmy" bulbs are very suitable for this application, giving a fair amount of light when mounted in simple reflectors.

A "one-shot" facility was added by switching out the multivibrator and arranging for a microswitch S2 to discharge a capacitor through the gate giving one bright light pulse from the lamp for every press of the micro-switch.

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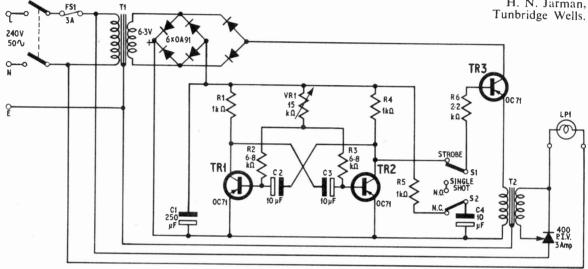


Fig. 1. Lamp strobe circuit diagram

TIMING CIRCUIT

HE main requirements of the timing circuit in Fig. I were that it should be cheap to build and run, while retaining reasonable accuracy.

In circuits of this kind it is usual to see a field effect transistor in the place of TR1, the main objection to a bipolar transistor being the low input resistance. In this circuit a bipolar transistor has been used for economy, the input resistance being increased by R2. Although this does not offer an input resistance comparable with that of a field effect transistor, the timing is accurate enough for many applications. The value of R2 should be found by experiment, but should not lie below $2M\Omega$.

When S1 is pressed TR1 will conduct, driving TR2 into saturation. This energises the relay which disconnects R1 and connects the emitter of TR2 to the power supply. C1 charges up through VR1 at a rate depending on its setting. When C1 has charged up enough, TR2 will no longer be able to hold the relay on, and the circuit will reset itself, C1 being discharged via R1.

The components specified will give a maximum timing period of a few minutes. If longer periods are required, the value of C1 can be increased.

The maximum current consumed by the prototype is about 25mA just after switching on, this value decreasing as the timing cycle progresses. The current consumed depends mainly on the relay chosen, one with a large coil resistance being most suitable.

The choice of power supply is left up to the constructor; the circuit can be run ecnomically from a small 9V battery if desired.

P. Chappell. Weston-Super-Mare

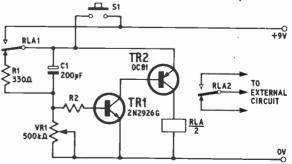


Fig. 1. Simple timing circuit

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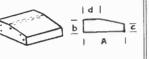




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301 301 301 301A 301A 301A 307 307 307 308 308A 709c 709c	DIL T099 8 PIN DIL DIL T099 8 PIN DIL DIL T099 T099 T099 DIL T099	50p 55p 46p 69p 69p 69p 69p 69p 645p 6 45p 8 40p 35p	723c 723c 741c 741c 741c 741c 747c 748c 748c 1437 1458 3046 7503	DIL TO99 8 PIN DIL 14 PIN DIL TO99 DIL TO99 DIL TO99 DIL DIL DIL	99p 95p 38p 39p 41p 46p 39p 41p 1-27p 1-27p 84p 1-27p	



Electrolytic Capacitors

	4 VOLI		16 VC) L I	40 VOLT	
	47μF 100μF 220μF 330μF 1000μF 4700μF	6 ½ P 6 ½ P 6 ½ P 6 ½ P 13 P 29 P	15μF 33μF 68μF 150μF 220μF 680μF 1000μF 1500μF 2000μF	6 1 p 6 1 p 6 1 p 8 p 9 p 17 p 17 p 25 p 43 p	47μF 100μF 150μF 220μF 470μF 680μF 1000μF 2200μF	6½p 9p 10p 11p 19p 25p 25p 44p
	6 3 VC) LT				
	33μF 68μF 150μF 470μF 680μF 1500μF 2200μF 3300μF	6½p 6½p 6½p 11p 13p 18p 18p 26p	25 VO 10μF 22μF 47μF 100μF 150μF 220μF 470μF 680μF	6 1 P 6 1 P 6 1 P 8 P 8 P 10 P 13 P 20 P	63 VC ΙμΓ 2·2μΓ 4·7μΓ	0LT 6½p 6½p 6½p
	10 VO	a T	1000μF	22p	6 · 8μF	4 3 P
	22μF 47μF 100μF 220μF 330μF	6½p 6½p 6½p 8p 10p	2200μF 5000μF	39p 68p	10μF 22μF 68μF 100μF 150μF	6 ½ p 6 ½ p 10 p 11 p 13 p
ı	470μF	10p	40 VO		220μ F	19p
ı	1000μF	lip	6 · 8μF	6 ½p	330μF	22p
ı	1500μF	20p	I5μF	6 ½ p	470μF	26p
8	2200μF	24p	33μF	6 ½ p	1000μF	44p

BARGAIN

Unmarked **Packs**

Pack of 25 1N4148 55p

Pack of 10 BC108 BC107 (Plastic can)

Pack of 10 Plastic BC109 55p

Pack of 10 BC169 55p (unmarked)

but tested 2N2646

(unmarked) 33p each Pack of 10 2N2926G

unbranded but tested Unmarked but fully tested 2N3055 33p 27p

10 plus FULLY

MARKED TYPES

AD161, AD162 M/P 59p 53p 10 plus

BC107-BC108 BC109 1-9 10-99 100 plus 9p 8p 61p

BC 182L:-3-4-212-4 1-9 10 plus 9p 8p

AC127 or AC128 1-9 13 10 plus 12 100 plus 11 12p

7FNFR DIODES

400 M/W 5% Miniature BZY 88 Range

All voltages 3.3-33 Volt 9p each

1 watt 5% All voltages -200 Volts 14p each

10 watt 5% All Voltages 7.5-100 Volts 51p each

Transistors

AC107 16p	BC138 36p	BF260 29p	OC44 141	Diodes &
AC126 14p	BC142 33p	BF329 18p	OC45 14	
AC127 13p	BC143 33p	BF330 18p	OC70 23	rectillers
AC128 13p	BC144 30p	BF390 37p	OC71 14	1N914 8p
AC142K 22p	BC145 26p	BFX84 28p	OC72 141	1N916 8p
AC141K 20p	BC147 9p	BFX85 35p	OC81 141	
AC176 15p	BC148 9p	BFX86 22p	OC83 22	1844 10p
AC187 13p	BC149 9p	BFX87 28p	OC84 28r	1N4007 22p
AC187K 20p	BC153 16p	BFX88 26p	TIP29A 53p	
AC188 13p	BC154 17p	BFY50 21p	TIP30A 64p	
AC188K 20p	BC157 13p	BFY51 17p	TIP31A 64p	18121 15p
ACY17 24p	BC158 12p	BFY52 17p	TIP32A 73p	
ACY 18 21p	BC159 14p	BFY64 89p	TIP33 £1.05	18131 11p
ACY19 25p	BC167 13p	BFY90 72p	TIP34A	18132 13p
ACY20 22p	BC168 11p	BSX 20 19p	£1·54	18920 8p
ACY21 23p	BC169 11p	C407 22p	TIP35A	18922 9p
ACY22 18p	BC177 15p	C426 33p	£2-53	18923 13p
ACY39 68p	BC179 15p	C428 31p	TIP36A	18940 6p
AD140 40p	BC182L 9p	C450 17p	£3·19	
AD142 44p	BC183L 9p	MP8111 35p	TIP41A 79p	
AD143 39p	BC184L 9p	MP8112 42p	TIP42A 91p	AAZ13 11p
AD149 38p	BC186 33p	MP8113 35p	2N706 13p	AAZ15 14p
AD150 60p	BC212L 11p	MP8121 33p	2N930 23p	AAZ17 14p
AD161 29p AD162 29p	BC213L 11p	MP8122 44p	2N1131 22p	
	BC214L 11p	MP8123 50p	2N1132 28p	
AD M/P 50p AP114 14p	BC258 9p	NKT21128p	2N1613 22p	BA115 8p
AF115 14p	BC259 9p BC267 14p	NKT21228p	2N1711 26p	
AF116 14p	BC268 15p	NKT214 25p	2N2904 40p	BA145 22p
AF117 14p	BC300 40p	NKT217 55p	2N2904A	BA154 14p BAX13 14p
AF118 92p	BC301 32p	NKT261 23p NKT271 20p	44p	
AF124 27p	BC302 30p	NKT274 20p	2N2905 46p 2N2924 18p	
AF139 39p	BC303 50p	NKT275 25p	2N2926 10p	
AF239 41p	BC304 40p	NKT403 71p	2N3053 26p	
AL100 77p	BCY70 17p	NKT405 83p	2N3054 55p	
AL102 66p	BCY71 37p	NKY603F	2N3054 55p 2N3055 52p	
AL103 55p	BCY72 17p	66p	2N3405 44p	BY127 19p
ASY26 31p	BD123 66p	NKT613G	2N3663 57p	BYX10 24p
ASY27 40p	BD130 50p	33p	2N3702 9p	OA5 19p
AU103 99p	BD131 68p	NKT674 26p	2N3703 9p	OA9 11p
AU110 £1.10	BD132 90p	NKT677G	2N3704 9p	OA10 24p
AU111 77p	BD135 42p	24p	2N3705 9p	OA47 9p
BC107 9p	BD136 50p	NKT713 32p	2N3706 9p	
BC108 9p	BD141 £1-87	NK T773 27p	2N3707 9p	OA70 8p
BC109 9p	BD142 50p	OC19 55p	2N3708 9p	OA73 11p
BC113 15 p	BF159 33p	OC20 55p	2N3709 9p	OA79 8p
BC116 16p	BF173 29p	OC23 33p	2N3710 9p	OA81 9p
BC125 16p	BF177 28p	OC25 28p	2N3711 9p	OA85 11p
BC126 25p	BF178 29p	OC28 33p	2N3794 17p	OA90 8p
BC132 16p	BF179 35p	OC29 33p	2N3819 28p	
BC134 16p	BF194 15p	OC35 38p	40361 50p	OA91 8p
BC135 16p BC137 16p	BF195 17p	OC36 38p	40362 50p	OA95 8p
BC137 16p	BF244 27p	OC41 14p	40636 50p	OA200 11p

MULLARD POLYESTER'S

MULLARD POLYESTER CAPACITORS C280 SERIES

250V P.C. mounting: 0.01μ F, 0.015μ F, 0.22μ F, 34p, 0.03μ F, 0.047μ F, 0.068μ F, 4p, 0.1μ F, 44p, 0.15μ F, 0.22μ F, 54p, 0.33μ F, 7p, 0.47μ F, 94p, 0.68μ F, 12p, 1.0μ F, 14p, 1.5μ F, 22p, 2.2μ F, 27p

MULLARD POLYESTER CAPACITORS C296 SERIES

400V:0-001μF, 0-0015μF, 0-0022μF, 0-0033μF, 0-0047μF, 21p, 0-0068μF, 0-01μF, 0-015μF 0-022μF, 0-033μF, 34p, 0-037μF, 0-068μF, 0-1μF, 44p, 0-15μF, 64p, 0-22μF, 84p, 0-33μF, 180V:0-01μF, 0-015μF, 0-22μF, 0-033μF, 0-047μF, 0-068μF, 8p, 0-1μF 34p, 0-15μF, 44p, 0-22μF, 54p, 0-33μF, 64p, 0-47μF, 84p, 0-68μF, 12p, 1-0μF, 144p,

VOLUME CONTROLS

Potentiometers	
Carbon track 500	Ω to 2.2M Ω
Log or Linear	
Single 13p. Dual	gang (stereo) 44p
Single type with	D.P. switch 13p extra.

SLIDE POTENTIOMETERS
88mm. TRACK
81NGLE GANGED, LOG or LIN 1k to 1M.
45p each
TWIN GANGED, LOG or LIN 1k to 500k.
86p each.

| 11p | S8mm, TRACK | | S8mm, TRACK | | S1mc, TRACK | S1mc, TRACK | | S1mc, TRACK | S1mc, | Matrix Matrix | Mat

SLIDE SWITCH SPST 11p each. D.P.D.T. 13p each.

MINIATURE NEON LAMPS 240V or 110V 1-4 5p, 5 plus 44p each.

MINITRON DIGITAL INDICATOR TYPE 3015F Reads 0-9 and decimals

(Data Sheet on request)
ONLY £1-50 50p OA200 11p Driven by 7447 33p £1.05

RECTIFIERS

P.I.V.	1 AMP	1.5 AMP
50	IN4001 4+p	PL4001 8p
100	IN4002 4 p	PL4002 9p
200	IN 4003 5 tp	PL4003 10p
400	IN 4004 6 p	PL4004 10p
600	IN4005 8p	PL4005 13p
800	IN4006 9p	PL4006 15p
1000	IN4007 10p	PL4007 20p

BRIDGE RECTIFIERS

	P.I.V.	1 AMP	2 AMP	5 AMP	10 AMP
	50	33p	53p	£1 76p	£2.20p
-	100	35p	57p		_
	200	37p	60p	£1.98p	£2.31p
	400	40p	64p	£2-15p	£2-42p
	600	44p	66p	£2-42p	£2.75p
	800	49p		_	_

PIV		50	100	200	300	400
	1 Amp 3 Amps	28 44	25 50	41 60	44	53 66
	7 Amps		96	1.01	-	1.2

RESISTORS

watt 5% carbon 1p each watt 10% carbon 1p each I watt 10% carbon 2p each Range 10 ohms to 4.7 meg. watt m/o 2% 3p each Range 10-1 meg ohms

AUTO-ELECTRIC CAR AERIAL

with dashboard control switch—fully extendable to 40in or fully retrac-table. Suitable for 12V positive or negative earth. Supplied complete with fitting instructions and ready wired dashboard switch. \$6.35 plus 25p post

RECORD PLAYBACK HEADS

Individual prices of these are—2 track record playback heads 50p each. 4 track record playback heads 72p each. Erase heads are also available separately—2 track 17p, 4 track 28p.

I R.P.H. MOTOR

Made by the famous Smith Company. 240V 50 cycle mains working. Ideal motor to drive clock mechanisms. Price 21-10 each or 10 for 40.00

MULTISPEED MOTOR

Six speeds are available: 500, 850 and 1,100 r.p.m. and 8,000, 12,000 and 15,000 r.p.m. Shaft 12,000 and 15,000 r.p.m. Shatt is \$1 in diameter and approx. I in long, 230/240V. Its speed may be further controlled with the use of our Thyristor controller. Very powerful and useful motor, size approx. 2in dia. x 5 in long. Price \$7p plus 23p post and inverse. insurance



MAINS OPERATED CONTACTOR

220/240V 50 cycle solenoid with laminated core so very silent in operation. Closes 4 circuits each rated at 10A. Extremely well made by a German Electrical Company. Overall size 2½ × 2 × 2in. 31-65 each.



TELESCOPIC AERIAL

for portable, car radio or transmitter. Chrome plated— sections, extends from 7½ to lole in bottom for 6BA screw. KNUCKLED MODEL FOR F.M. 55p

MAINS CLOCK & TIME SWITCH



Smith's main's driven clock with 15A programmable switch also notes showing how you can use this to wake up with music playing, kettle boiling or come home to a warm house, warn-off burglars, keeps pets warm-on halves your heating hills etc., etc. 42.20 +20p p. & p.

PP3 BATTERY CHARGER

Almost 3 times the life can be obtained from PP3 battery if you re-charge it from the mains—this ready to use is ready to use with instructions



IMMERSION HEATERS BY

REMPLOY Standard fitting for domestic water tanks, made by the famous Remploy



NEED A SPECIAL SWITCH?

00

Double Leaf Contact. Very slight pressure closes both contacts. 7p each, 10 for 63p. Plastic pushrod suitable suitable for operating. 6p each, 10 for 54p.

THERMOSTAT



sensor bulb connected by 33in of flexible tubing. On operation a 15A 250V switch is opened and in addition a plunger moves through approx. in. This could be used to open valve ventilator, etc. £1.65 plus 23p

p. & ins HIGH ACCURACY THERMOSTAT

Uses differential comparator I.C. with thermistor as probe. Designer claims temperature control to within 1/7th of a degree. Complete kit with power pack £6-25.

NUMICATOR TUBES



For digital instruments, counters, timers, clocks, etc. Hi-vac. XN. 3. Price £1.59 each.

12-WAY SUB-MINIATURE MULTI-CORE CABLE

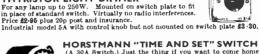
7.0076 copper cores each core. P.V.C. insulated and of different colour. P.V.C. covered overall and approx. 3/16in thick. Price 22p per yard.

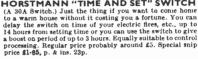
STANDARD WAFER SWITCHES

Standard size 1½ in wafer—silver-plated 5 amp contact, standard % in spindle 2 in long—with locking washer and nut

No. of Poles	2 way	3 way	4 way	5 way	6 way	8 way			12 way
1 pole	44p								
2 poles	44p		77p						
3 poles	44p	44p	44p	44p			77p	£1-04	
4 poles	44p	44p	44p	77p			77p	£1.32	£1.32
5 poles	44p	44p	77p	77p	£1.04	£1-04	£1-04		£1-60
6 poles	44p	77p	77p			£1.04			
7 poles	77p	77p	77p			£1.32			
8 poles	77p	77p	77p			£1-32			
9 poles	77p	77p	£1-04			£1-60			£2.70
10 poles	77p	77p	£1-04	£1-32	£1-60	£1.60	£1-60	£3-00	
11 poles	77p	£1-04				£1.87		£3.25	£3-25
12 poles	77p	£1.04	£1-04	£1-32	£1-87	£1.87	£1.87	£3.52	£3.52

THYRISTOR LIGHT DIMMER





13 AMP TWIN GANG SOCKETS

Offered at less than wholesale price your opportunity to replace those dangerous adaptors—brown bakelite flush mounting—standard fitting. Unswitched 22p each, separately switched 33p each. Less 10% ten or more +20p postage if order under £5.

-THIS MONTH'S SNIP-

KETTLE ELEMENTS

Made by the famous A.E.I. Co. Complete with washers and combined fixing ring and plug shroud. Normal 2 round pin and flat pin earth connection and overload reset push button. 2 Models—1jin (approx.) suitable for Swan and other similar models—1jin (approx.) suitable for G.E.C. Hotpoint, etc. All quick boil 2jkW elements at 240V. Price 21-38.

COMPUTER TAPE

2.400ft. of the Best Magnetic Tape money can buy. Some users claim good results with Video and sound. Iin, Jin or Jin wide. 21 plus 30p post. Spare spools and cassettes 50p.

Jin Scotch tape. Brand new. Suits most video recorders. lin Scotch to

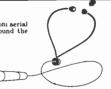


MULLARD UNILEX

MULLARD UNILEX
This D.I.Y. Stereo Amplifier is still available complete at \$7.70 for the four Mullard Modules, or Modules can be bought separately as follows: 4 watt amplifier module (2 required) Mullard Ref. No. E. P. 9000—\$1.60 each. Fre-amp module Mullard Ref. No. E.P. 9001, \$1.98 each. Power module—Mullard Ref. No. E.P. 9002, \$2.33 each. In addition and made to Mullard Specification we offer: Standard Control Unit with secutcheon, \$2.75, Knobs, 85p. Unit, \$2.54 with Set of 6.55p. Unit, \$2.54 with Set

RADIO STETHOSCOPE

EASION SIE! HOSCOPE
Easiest way to fault find—traces signal from aerial
to speaker—when signal stops you've found the
fault. Use it on Radio, TV,
amplifier, anything—complete kit comprises two special
transistors and all parts including probe tube and cluding probe tube and crystal earpiece. \$2.20—twin stethoset instead of earpiece 83p extra, post and ins. 20p.



MIGHTY MIDGET

Probably the tiniest possible radio, as described in Practical Wireless, January 73. All electronic parts £2.20 post paid.



DIGITAL COUNTER TIMER

Very stable and reliable crystal controlled circuit. Capable of work in excess of 15MHz. Construction simplified by use of 15 integrated circuits. Complete kit with case \$43.50 or construction data and price list 50p

INTEGRATED CIRCUIT BARGAIN

A parcel of integrated circuits made by the famous Plessey Company. A oncein-a-lifetime offer of Micro-electronic devices well below cost of manufacture. The parcel contains 5 ICs all new and perfect, first-grade device, definitely not sub-standard or seconds. 4 of the ICs are single silicon chip GP amplifiers. The 5th is a monolithic NPN matched pair. Regular price of parcel well over £5. Full circuit details of the ICs are included and in addition you will receive a list of many different ICs available at bargain prices £5p upwards with circuits and technical data of each. Complete parcel only £1 Post paid.

DON'T MISS THIS TERRIFIC BARGAIN.

TERMS:- 10% discount if ten of an item ordered, send postage where quoted-other items post free if order for items is over \$6 otherwise add 20p.



EXTRACTOR FAN

Cleans the air at the rate of 10,000 cubic ft per hour. Suitable for kitchens, bathrooms, factories, changing rooms, etc., it's so quiet it can hardly be heard. Compact rooms, etc., it's so quiet it can hardly be heard. Compact 5½in casing with 5½in fan blades. Kit comprises motor, fan blades, sheet steel casing, pull switch, mains connector, and fixing brackets, 12.75 plus 30p post and ins.

QUICK CUPPA

Mini Immersion Heater, 350W 200/240V. Boils full cup in about two minutes. Use any socket or lamp holder. Have at bedside for tea, baby's food, etc. £1.25, post and insurance 14p. 12V car model aleo available same price. Jug heater £1.75 plus p. & p. 14p.



MAINS TRANSISTOR POWER

PACK
Designed to operate transistor sets and amplifiers.
Adjustable output 6V, 9V, 12V for up to 500mA (class B working). Takes the place of any of the following batteries: PPI, PP3, PP4, PP6, PP7, PP9, and others. Kit comprises: mains transformer, rectifier, smoothing and load resistor condensers and instructions. Real snip at only \$1.10 plus 20p postage.



TREASURE TRACER Complete Kit (except wooden battens) to make the metal detector as the circuit in Practical Wireless, August issue. 23-30 plus 20p post and insurance.

WINDSCREEN WIPER CONTROL Beat dirty roads, drizzle, fog, etc. Kit of parts to make this useful accessory with circuit details.

12 VOLT I AM



Ferranti's latest device ZN414—gives results better than superhet. Supplied complete with technical notes and circuits. £1-35 each. 10 for £12.

HI-Q TUNER COMPONENTS

HI-Q TUNER COMPONENTS
Por experimenting with the ZN414
Kit No. 1—Pleasey Miniature Tuning Condenser
with built in LW awitch and 3in ferrite slab and
Litz wound MW coil. 72p.
Kit No. 2—Air spaced tuning condenser 6in ferrite
rod. litz wound MW and LW coils, 99p.
Kit No. 3—Air spaced TC with slow motion drive
Sin ferrite rod, with Litz wound LW and MW
coils, \$1-10.

coils, 21-10.

Kit No.4—Permeability tuner with fast and slow motion drive and LW loading coils, 50p.

DRY FILM LUBRICANT

DRY FILM LUBRICAN
In aerosol can for easy applicant into places where the
normal oil can cannot reach.
Home and everyday uses,
We have purchased a large
quantity of these from the
Liquidator and are able to
offer them to you for about
half of the original list price.
88p per (80x) can or 12 cans
for \$3.30, post paid. The
lubricant is I.C.I. fluon L169.



PHOTO TRANSISTOR
OCP70—deal for burglar
alarms and similar applications. Price 72p each.



A must for every busy man, gives almost instant heat also illumi-nated job. 100W \$2.50 plus post nated job. 10



20p post and insurance



20p post and insurance.

3 STAGE

Made originally for Radjomobile can be considered by the can b

ponents to make a complete compact receiver. Circuit supplied. Price 72p less 10% for 10.

J. BULL (ELECTRICAL) (Dept. P.E.), 7 Park Street, Croydon CRO IYD

(F2)

Now available from one company are kits for many of the articles published in the Electronics, Radio and TV Journals. Examples of our range are:

SCORPIO IGNITION SYSTEMS (P.E. Nov. 1971), £9-50. This kit includes all the parts for the assembly of this popular and reliable system. The hardware and the construction data are included.

DRILL SPEED CONTROLLER (E.E. Aug. 1972). Kit consists of resistors, rectifiers, thyristor and tag board as specified in the article, price £1.15. Kit with M.K. box, switch and plate included. £2:30.

ELECTRONIC PIANO (P.E. Sept. 1972). We can supply the various sections for this article in kit form:

Power Supply, £6.60, including all semiconductors, resistors. capacitors, transformer, heat sink and hardware less P.C.B.

Preamp and Tremelo, £3:20, complete with switches, hardware and electronic components, less P.C.B.

Main Amp., £4.20, complete with IC-12, electronic components and P.C.B.

13 Pitch Boards, £39.50. Kit contains all the resistors. capacitors and semiconductors as published, but not the P.C.B.'s or the inductor.





12 Lauderdale Road, London W.9. Telephone 01-286 0011 Telex 28479 **LIGHT DIMMER.** Kit contains all parts including circuit and construction data, 480 watts, fully suppressed. Price £2.10.

MUSIC MAKER ELECTRONIC ORGAN (P.W. Nov. 1972). Complete kit £4.40 or kit less resistors, P.C.B. and plastic box, £2-80. Stylus not supplied.

MIGHTY MIDGET (P.W. Jan. 1973). Kit contains all parts. less the battery, box and knob. £2.25.

TRIFFID RADIO (P.E. Feb. 1973). Kit includes all parts listed, less "tuning coil" and "miscellaneous components", £4.95. Ferrite rod and wire optional extra at 50p. P.C.B. optional extra at 65p.

We shall be offering kits for most articles published in the popular electronics magazines and will be pleased to quote prices.

Please note, prices shown do not allow for V.A.T. Please add 10% to your order.

Send for details of kits available (please enclose S.A.E.). All kits sent POST FREE within the limits of the U.K.

lease note. We reserve the right to withdraw kits without prior notification.



NEW COMPONENTS

Post and packaging free for orders over £1.50, include 10p P&P for each single pack under £1.50.

te pack under £1.50.

Mixed Resistors all types, 50p. 100 Mixed Modern and Miniature Resistors, 50pg 10mF 64V Electrolytic Caps, 5 for 25p; 12 for 50p, 640mF 16V Electrolytic Caps, 3 for 30p; 7 for 60p. 10mF 63V WIMA non-electrolytic, 15p each. 3200mF 10V, 15p; 3 for 30p; 5 for 60p.

SEMICONDUCTORS

Any 6 of the following, 50p, or 10p each. OC71, BFY50, 2N37 CV8615, BSY95A, NTG885, 2N930, OA81, 2 x 1N914. (OR P60 50p).

CONSTRUCTORS' ITEMS

Subminiature Omron 12V d.c. relays mounted on CCT board, 3 for 60p. P&P 9p; mounted on CCT board with components, 2 for 60p, P&P 7p GPO Relays, various 20002-700001, 30p each. Uniselectors, 10 pole, 25 way, 61 each, P&P 25p. Heavy Duty Foot Pedals, 50p, P&P 20p. 6V 5 digit High Speed Counters, £1-50, 15p P&P.

DICTATING MACHINES

£2-50 each—ideal for spares, motor, power pack, record/replay, electronics with mike £3-25.

TRANSFORMERS

Mains—13V 2-5A and 15V 0-75A £1-45, 20p P&P. Mains—13V 1A 12V 0-5A, £1-25, 20p P&P. Mains—13V 5A and 24V 2A, £2, 25p. P&P. Mains—24V 100 mA and 6V 100mA, 75p, 10p P&P. Power Pack—suitable to run Transistor Radio or Cassettes Recorder, 5-5V, 100mA, smoothed d.c. output £1-20, 20p P&P.

EX COMPUTER CIRCUIT BOARDS

IO Boards, 50p, 8p P&P; 25 Boards, £1, 18p P&P.
2 Boards with 2 power transistors, 1E 4 x OC28 type, 50p, 7p P&P

LAWBAK ELECTRONICS

(HI-FI AND COMPONENTS SPECIALISTS) 18 HIGH RD., SWAYTHLING, SOUTHAMPTON

> Telephone: Southampton 58479 No Half Day Closing

> Discount and Credit Terms Available

AUTOMATIC EMERGENCY SUPPLY

250v 50 Hz—150 watt Inverter. Full kit of parts excluding meter. Circuit as appeared in December P.W. Complete kit—£16:95 + 80p. P. & P.

OTHER INVERTERS AVAILABLE IN KIT FORM

150 Watt—£13-50 + 60p P. & P. 75 Watt—£7-80 + 60p P. & P. 300 Watt—£17-90 + 85p P. & P. 40 Watt—£5-20 + 40p P. & P. 25 Watt—£2-60 + 20p P. & P.

All above operate from 12v. battery and give 250v.-50Hz. Output. 24 volt types are also available, alternative outputs or taps can be supplied. Transformers and/or Transistors can be supplied separately.

SPECIAL OFFER

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10 TRANSISTORS. 9 TUNABLE WAVEBANDS, MW1, MW2, LW, SW1, SW2, SW3, TRAWLER BAND, VHF AND LOCAL STATIONS ALSO AIRCRAFT BAND.

Built-in ferrite rod aerial for MW/LW. Retractable, chrome plated 7 section telescopic aerial, can be angled and rotated for peak short wave and VHF listening. Push-pull output using 600mW transistors. Car Aerial and tape record sockets. 10 transistors plus 3 diodes. Fine tone moving coil speaker. Ganged tuning condenser with VHF section. Separate coil for Aircraft Band. Volume/on/off, wave change and tone controls. Attractive case in black with silver blocking. Size 9 in × 7 in × 4 in.

Easy to follow instructions and diagrams. Parts price list and easy build plans 30p (FREE with parts).

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P.P. & INS. 52p (OVERSEAS P. & P. £1-05)



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TOTAL BUILDING COSTS **£7.68** P. P. & INS. 47p (OVERSEAS P. & P. £1.05)

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TOTAL BUILDING COSTS **£6·58** P. P. & INS. 47p (OVERSEAS P. & P. £1·05)

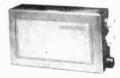




6 TUNABLE
WAVEBANDS:
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SWI, SW2.
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polished metal inserts. Size 9in x 5/in x 25/in approx.
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Darts).

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3 TUNABLE WAVE-BANDS: MW, LW, TRAWLER BAND WITH EXTENDED MW BAND FOR EASIER TUNING OF LUXEM-BOURG, ETC. 7 stages—5 transistors and 2 diodes, supersensitive ferrite rod aerial, fine tone inoving coil speaker. Attractive black and gold case. Size 5½m x 1½m x 3½m. Easy build plans and parts price list 10p (FREE with parts).

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5 TRANSISTORS AND 2 DIODES

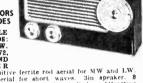
3 TUNABLE WAVE BANDS: MW, LW AND TRAWLER BAND, 7 stage—5 transistors and 2 diodes, ferrite rod aerial, tuning condenser, volume control, fine tone moving coil speaker. Attractive case with red speaker grille. Size 6jin *4jin *1jin. Easy build plans and parts price list 10p (FREE with parts).

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TRANS EIGHT

8 TRANSISTORS AND 3 DIODES

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1N85	0-88	A8Y27	0.82	BYZ11 BYZ12	0.32 0.80	OAZ223	0-45	Z8271 ZT21	0·18 0·25
1N253 1N256	0-50 0-50	ASY28 ASY29	0.25	BYZ13	0.25	OAZ224 OAZ241	0.45	ZT 43	0-25
1N645	0.25	A8Y36	0.25	BYZ15	1.00	OAZ242	0.28	ZTX 107	0-15
1N725A 1N914	0.20 0.07	A8Y50 A8Y51	0·17 0·40	BYZ16 BYZ88C3	0-82	OAZ244 OAZ246	0.22	ZTX 108	0.12
1N4007	0.20	A8Y53	0.20	DIMOGRA	0-15	OAZ290	0.88	ZTX300 ZTX304	0-12 0-25
18113 18130	0.15 0.18	ASY55 ASY62	0-20 0-25	CH11 CR81/05	0-85 0-25	OC16 OC16T	0.50	ZTX500	0.16
16131	0.18	49Y86	0-88	CR81/40	0.45	OC187 OC19	0-88 0-87	ZTX503	0.17
16202	0-28 0-22	A3Z21	0.42	('84B	2.50	OC20	0.85	ZTX531	0-25
2G371 2G381	0.25	ASZ23 AUY10	0-75 0-98	CS10B DD000	3·18 0·15	OC22 OC23	0.50	INTEGRA	TED
26414	0.80	AU101 BC107	1.50	DD003 DD006	0·15 0·18	0C24	0.80	CIRCUIT	8
2G417 2N404	0.20	BC108	0·10 0·10	DD006 DD007	0.40	OC25	0.87	7400 7401	0·20 0·20
2N697 2N698	0.15	BC109 BC113	0.10	DD00H GD3	0.88 0.83	OC26 OC28	0.80	7402	0.20
2N 706	0.10	BC115	0·15 0·20	(4D4		OC29	0.60	7403 7404	0-20 0-20
2N706A	0.12	BC116 BC116A	0.25	GD5	0.88	OC30	0.40	7404	0.20
2N708 2N709	0·15 0·63	BC118	0-80 0-25	GD8 GD12	0.25 0.05	OC35	0.50	7406	0.30 0.30
2N1091	0.88	BC121 BC122	0.20	GET102	0.80	OC36 OC41	0.60	7407 7408	0.20
2N1131 2N1132	0-25 0-25	BC122 BC125	0.20 0.68	GET103 GET113	0.22 0.20	0C41 0C42	0.80	7409	0.45
2N1302	0.18	BC126	0-65	GET114	0.15	OC43	0.40	7410 7411	0·20 0·23
2N1303 2N1304	0.18	BC140 BC147	0.55 0.15	GET115 GET116	0-45	OC44 OC44M	0·17 0·17	7412	0.42
2N1305	6.22 0.25	BC148	0.13	OET120 GET872	0.25	OC45	0.12	7418 7416	0-30 0-30
2N1306 2N1307	0.25	BC148 BC149 BC157	0-15 0-15	GET872	0-30 0-25	OC45M OC46	0·18 0·27	7417	0.30
2N1308	0.25	BC158	0.19	41ET880	0.87	OC57	0.60	7420 7422	0·20 0·48
2N2147 2N2148	0.75 0.60	BC160 BC169	0-68 0-13	GET881 GET882	0-25 0-25	OC59	0-60	7423	0.48
2N2160 2N2218	0.60	BCY31	0.85	GET885	0.25	OC66	0.50	7425 7427	0·48 0·42
2N2218 2N2219	0.20	BCY 32	0.55	GEX 44 GEX 45/1	0.08 0.10	0C70 0C71	0·12 0·12	7428	0-50
2N2369A	0·15 1·99	BCY33 BCY34	0-25 0-80	GEX 941	0.15	OC72	0.20	7430 7432	0.20 0.42
2N2444 2N2613	0.28	BUVSH	0.40	GJ3M GJ4M	0-25 0-88	OC73 OC74	0.80 0.80	7433	0.70
2N2646 2N2904	0-45 0-20	BCY39 BCY40 BCY42	1.00	GJ5M	0.25	0C75 0C76 0C77	0.25	7437 7438	0-65 0-85
2N2904A	0.25	BCY 42	0.50 0.25	(4J7M HG1005	0.87 0.50	0C76	0.25	7440	0.20
2N2906	0.20 0.23	BCY70	0.15	H8100A	0·20 0·25	OC78	0.20	7441 \N 7442	0-75 0-75
2N2907 2N2924	0.28	BCY70 BCY71 BCZ10	0.20 0.85	MAT100 MAT101	0.80	0C79 0C81	0.20	7450	0.20
2N2925 2N2926	0·15 0·10	BCZ11	0.50	MAT120	0.25 0.80	00811)	0.20	7451 7453	0-20 0-20
2N3054	0.50	BD121 BD123	0.86 0.80	MAT121 MJE520	0.87	OC81M OC81DM	0.20 0.18	7454	0.20
2N3055 2N3702	0.75	BD124 BDY11	0.75	MJE2955	1·87 0·87	OC81Z	0-40 0-25	7460 7470	0-20 0-80
2N 3705	0·10 0·10	BF115	1-62 0-25	MJE3055 NKT128	0.85	OC82 OC82D	0.20	7472	0.80
2N3706	0.28	BF117 BF167	0.50	NKT129	0.80	OC83	0.25	7473 7474	0-40 0-40
2N3707 2N3709	0·12 0·10	BF173	0.25 0.25	NKT211 NKT213	0.25 0.25	0C84 0CH4	0·25 0·38	7475	0-55
2N 3710 2N 3711	0·10 0·10	BF181	0.85	NKT214	0-15 0-37	0C114 0C122	0.80	7476 , 7480	0-45 0-80
2N3819	0.85	BF184 BF185	0.20	NKT216 NKT217	0.35	OC123 OC139	0.65 0.25	7482	0.87
2N5027 2N5088	0-58 0-88	BF194 BF195	0·17 0·15	NKT218 NKT218	1.18 0.88	OC140	0.85	7483 7484	1.00 0.90
28301	0.50	BF196	0.15	NKT222	0.20	OC141 OC169	0.80	4486	0.45
28304 28501	0.75 0.87	BF197 BF861	0·15 0·28	NKT224 NKT251	0.22		0.80	7490 7491AN	0.78 1.00
28702	0.62	BFS98	0.28	NKT271	0.25	OC171 OC200	0.40	7492	0.75
AA129 AAZ12	0.20 0.80	BFX12 BFX13	0.20 0.25	NKT272 NKT273	0.25 0.15	OC201	0.70 0.80	7493 7494	0.75 0.80
AAZ13	0.12	BFX29	0.25	NKT274	0-20	OC202 OC203	0.40	7495	0.80
AC107 AC126	0·87 0·20	BFX30 BFX35	0-25 0-98	NKT275	0.25 0.20		0.40	7496 7497	1·00 6·25
AC127	0.25	BFX63	0.50	NKT277 NKT278	0.25	OC205 OC206 OC207	0.90	74100	2.50
AC128 AC187	0.20 0.25	BFX84 BFX85	0.25 0.30	NKT301 NKT304	0-40 0-75	00207	0-90 0-20	74107 74110	0.50 0.80
AC188	0.25	BFX86	0.25	NKT403	0.75	OC460 OC470 OCP71	0.80	74111	1.45
ACY17 ACY18	0-80 0-25	BFX87 BFX88	0·25 0·20	NKT404 NKT678	0-55 0-80	OCP71 ORP12	0-97 0-50	74118 74119	1.00 1.90
ACY19	0.25	BFV10	1.00	NKT713	0.25	ORP60	0.40	74121	0.60
ACY20 ACY21	0.20 0.20	BFY11 BFY17	1·25 0·25	NKT773 NKT777	0·25 0·38	ORP61 S19T	0.42	74122 74123	1.85 2.70
ACY21 ACY22 ACY27	0.10	BVF18	0.25 0.25	078B	0.88	SAC40 SFT308	0.25	74141	1.00
ACY27 ACY28	0.25 0.17	BFY19 BFY24	0.45	OA6	0-12	ST722	0-88 0-88	74145 74150	1.50 3.85
ACY39	0.50	BFY44	1.00	OA47	0·10 0·10	8T7231	0.68	74151	1.10
ACY40 ACY41	0-15 0-15	BFY50 BFY51	0.22 0.20	OA70 OA71	0.10	8X68 8X631	0.20	74154 74155	2.00 1.55
ACY44	0.25	BFY51 BFY52	0.22	OA73	0.10	8X635	0-40	74156	1.55
AD140 AD149	0.50 0.50	BFY53 BFY64	0·17 0·42	OA74 OA79	0·10 0·10	9X640 8X641	0.50 0.55	74157 74170	1·80 4·10
AD161	0.87	BFY90	0.65	OA81	0.08	8X.642	0.60	74174	2.00
AD162 AF106	0.87 0.80	BSX 27 BSX 60	0-50 0-98	OTTO	0·12 0·15	8X644 8X645	0.75 0.75	74175 74176	1.35 1.80
AF114	0.25	BSX76	0.15	OA90	0.08	V15/30P V30/201P	0.50	74190	1.95
AF115 AF116	0-25 0-25	BSY26 BSY27	0·18 0·17	OA91 OA95	0-07 0-07		0.75	74191 74192	1.95 2.00
AF117	0.25	B8Y51 B8Y95A	0.50	OA200	0.07	V60/201P	0.75	74193	2.00
AF118 AF119	0.62 0.20	BSY95	0·12 0·12	OA202 OA210	0·10 0·25	XA102	0·10 0·18	74194 74195	2.50 1.85
AF124 AF125	0.25	BT102/50		OA211 OAZ200	0.30 0.55	XA151	0-15	74196	1.50
AF126	0.17	BTY42	0.92	OAZ201	0.50	XA152 XA161	0-15 0-25	14101	1.50 4.80
AF127 AF139	0·17 0·80	BTY79/1	0.00	OAZ202 OAZ203	0.42	XA161 XA162	0.25	74199	4.80
AF178	0.55	BT Y 79/4	00R	OAZ204	0.80	X B 101	0.48	Plug in se	nckets
AF179 AF180	0.65 0.52	BY 100	1-25	OAZ205	0.42	X B102	0.10	low pro	
AF181	0.42	BY126	0.15	OAZ207	0-47	X B103 X B113	0-25 0-12	14 min D1	IL
AF186 AFY19	0.40 1.18	BY 127 BY 182	0-17 0-85	OAZ208	0.82	XB121	0.48	16 pin D1	0-15 IL
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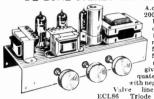


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QUALITY RECURD PLAYER AMPLIFIER ME. II A top-quality record player amplifier employing heavy duty double wound mains transformer, ECC83, ELS4, and rectifier, Separate Bass, Treble and Volume controls. Complete with output transformer matched for 3 ohm speaker, Size 7in. w. > 3 d. < 6 h. Ready built and tested. PRICE 24.40, P. & P. 40p. ALSO AVAILABLE mounted on board with output transformer and speaker, PRICE 25.85, P. & P. 50p.

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Beautifully made teak finish enclosure with most attractive Tygan-Vynair front. Size 16in high. V10lin wide V6in deep. Fitted with E.M. Ceramic Magnet 13in 8in bass unit, two H.F. tweeter units and crossover. Max. power handling 10W. Available 3, 8 or 15 ohm impedance.

Our Price £9.25 Carr. 70p

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Also available in 8 ohn with EMI 13in x 8in. bass speaker with parasitic tweeter. 27.15. Carr. 70p.

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A super quality gram amplifer using a double wound fully
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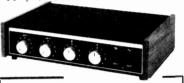
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123:10. Post Free. Note: The above a suitable for feeding two mono sources into inputs (e.g. mike, radio, tecin record decks, etc.) and will then provide mixing and fading facilities for medium powered Hi-Fi Disvotheque use, etc.



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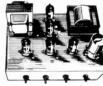
AMPLIFIER HA34 MK II
Designed for Hi-Fi reproduction of records. A.C. Mains operation. Ready built on plates heavy gauge metal chassis, size 7/3 m. 4 in. d. 4 lin. b. Incorporates ECC3. EL84, EZ80 valves. Heavy duty, double wound mains transformer and output transformer are output transformer are output giving bas and treble lift and cut. Negative feedback line. Output 4/3 watts. Front panel can be detached and leads extended for remote mounting of controls. Complete with knobs, valves, etc., wired and tested for only £5.50. P. x. P. 350.

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to follow each other Fully shrouded section wound output transformer to match 3-150 speaker and 2 independent volume controls, and separate bass and treble centrols are provided giving good lift and cut. Valve ince-up 2 ELP4s, ECV 83, EF86 and EZ86 rectifier. Simple instruction booklet 139 (Free with parts), All parts old spearately, ONLY \$8.80, P. & P. 55p. Also available ready built and tested \$12.10, P. & P. 60p.

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Sinclair Project 60

Now-the Z.50 Mk.2

with built-in automatic transient overload protection

When originally introduced, the Sinclair Z.50 proved how it was possible to design and produce a popularly priced modular power amplifier having characteristics to challenge the world's costliest amplifiers. Many thousands of Z.50's are now giving excellent service day in, day out. But we have also learned that constructors do not always use their Z.50's ideally. That is why we have introduced modifications whereby risk of damage through mis-use is greatly reduced and performance further enhanced. The Z.50 Mk 2 has improved thermal stability, more accurately regulated D.C. limiting to ensure more symetrical output voltage swing and clipping and still less distortion at lower power. Z.50 Mk 2 is compatible with all other Project 60 modules, and may be incorporated to advantage in existing systems. Eleven silicon epitaxial planar transistors are now used, two more than in the original Z.50'; circuitry has been re-designed, making this versatile high performance amplifier better than ever.



with free manual £5.48

Z.30 the power amplifier for quality and economy



with free manual

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The Z.30 provides excellent facilities for the constructor requiring a high fidelity audio system of less power than that available from Z.50's. Using a power supply of 35 volts, Z.30 will deliver 15 watts RMS into 8 ohms, or 20 watts RMS into 3 ohms using 30 volts. Total harmonic distortion is a fantastically low 0.02% at 15 watts into '8 ohms with signal to noise ratio better than 70 dB unweighted. Input sensitivity 250mV into 100K ohms. Size $80\times57\times13$ mm (3 $\frac{1}{8}\times2\frac{1}{8}\times\frac{1}{2})$ Z.30, Z.50 and Z.50 MK.2 modules are compatible and interchangeable

Guarantee

If, within 3 months of purchasing any product direct from Sinclair Radionics Ltd., you are dissatisfied with it, you money will be refunded at once. Many Sinclair appointed Stockists also offer this same guarantee in co-operation with Sinclair Radionics Ltd.

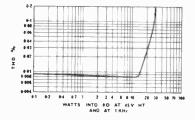
Each Project 60 module is tested before leaving our factory and is guaranteed to work perfectly. Should any defect arise in normal use, we will service it at once and without any charge to you, if it is returned within two years from the date of purchase. Outside this period of guarantee a small charge (typically £1.00) will be made. No charge is made for postage by surface mail. Air Mail is charged at cost.



Brilliant new technical specifications

Input impedance 100 K Ω Input (for 30w into 8 Ω) 400mV Signal to noise ratio, referred to full o/p at 30v HT 80dB or better Distortion 0.02% up to 20W at 8 Ω . See curve Frequency response 10Hz to more than 200 KHz \pm 1dB Max. supply voltage 45v (4 Ω to 8 Ω speakers) (50v 15 Ω speakers only)

Min. supply voltage 9vLoad impedance — minimum: 4Ω at 45v HT Load impedance — maximum: safe on open circuit



Typical Project 60 applications

System	The Units to use	together with	Units cost
Simple battery record player	Z.30	Crystal P.U., 12V battery volume control, etc.	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control, etc.	£9.45
12W. RMS continuous sine wave stereo amp. for average needs	2 x Z.30s, Stereo 60; PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers	2 x Z.30s, Stereo 60; PZ.6	High quality ceramic or magnetic P.Ü., F.M. Tuner, Tape Deck, etc.	£26.90
80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms)	2 x Z.50s, Stereo 60; PZ.8, mains transformer	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43

the world's most advanced high fidelity modules

Stereo 60 Pre-amp/control unit



Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout, a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly assystems.

SPECIFICATIONS—Input sensitivities: Radio — up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A. curve ±1dB:20 to 25,000 Hz. Ceramic p.u. — up to 3mV. Aux — up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE-12 to —12dB at 10KHz: BASS +12 to —12dB at 100Hz. Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm.

Built, tested and guaranteed.

£9.98

Project 60 Stereo F.M. Tuner



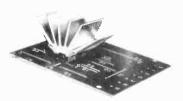
The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 pius 20 in I.C. Tuning range: 87.5 to 108MHz. Sensitivity: $7\mu V$ for lock-in over full deviation. Squelch level: Typically $20\mu V$. Signal to noise ratio: .65dB. Audio frequency response: 10Hz - 15KHz ($\pm 1dB$). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: $2\mu V$. Cross talk: 40dB. Output voltage: $2\times 150mV$ R.M.S. maximum Operating voltage: 25-30VDC. Indicators: Stereo on ; tuning. Size: $93\times 40\times 207mm$.

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£25

Super IC.12 Integrated circuit high fidelity amplifier



Having introduced Integrated Circuits to hi-fi constructors with the IC.10, the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the SuperIC.12, a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up. F. M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC. make possible, it is the equivalent of a 22 transactions.

sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC.12 is compatible with Project 60 modules which would be used with the Z.50 and Z.30 amplifiers. Complete with free manual and printed circuit board.

SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak). 6-80. Frequency Response: 5Hz to 100KHz±1dB. Total Harmonic Distortion: Less than 1%. (Typical 0-1%)- at all output powers and frequencies in the audio band (2BV). Load impedance: 3 to 15 ohms. Input Impedance: 250 Kohms nominal. Power Gain: 90dB (1,000,000,000 times) after feedback. Supply Voltege: 6 to 2BV. Quiescent current: 8mA at 2BV. Size: 22×45×28mm including pins and heatsink...

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The most reliable power supply unit ever made available to constructors. Brilliant circuitry makes failure from over load and even direct shorting of the output impossible. This is due to an ingenious re-entrant current limiting principle which, as far as we know has never before been available in any comparable unit outside the most expensive laboratory equipment. Ripple and residual noise have been reduced to the point of almost total elimination. This is, of course, the perfect unit for Project 60 assemblies, particularly where the new 2.50 MK.2 amplifiers are used. Nominal working voltage – 45.

PZ.8 Mk.3—£7.98 (Mains transformer, if required) £5 98 PZ.5 30v. unstabilised

(not suitable for Project 60 tuner) £4.98

PZ.6 35v. stabilised

(not suitable for IC. 12) £7.98

Project 605



Project 605 in one pack contains one PZ.5, two Z.30's, one Stereo 60 and one Masterlink, which has input sockets and output components grouped on a single module and all necessary leads cut to length and fitted with clips to plug straight on to the modules thus eliminating all

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p p p	p p p	8N74145 1 50 1 40 1 30
8N7400 0 20 0 18 0 16 8N7401 0 20 0 18 0 16	8N7450 0 20 0 18 0 18 8N7451 0 20 0 18 0 16	8N74145 1 50 1 40 1 30 8N74150 8 35 2 95 2 16
	8N7453 0-20 0-18 0-16	8N74151 1 10 0 95 0 90
8N7402 0 20 0 18 0 16 8N7403 0 20 0 18 0 16	8N7454 0 20 0 18 0 16	8N74153 1 35 1 27 1 20
8N7404 0 20 0 18 0 16	8N7460 0 20 0 18 0 16	8N74154 2:00 1:75 1:55
8N7405 0 20 0 18 0 18	8N7470 0 30 0 27 0 25	8N74155 1 55 1 47 1 35
8N7406 0 30 0 27 0 25	8N7472 0:30 0:27 0:25	8N74156 1 55 1 47 1 35
8N7407 0 30 0 27 0 25	BN7473 0-40 0-87 0-35	8N74157 1 80 1 70 1 50
8N7408 0 · 20 0 · 19 0 · 18	8N7474 0 40 0 87 0 85	8N74160 2 60 2 40 2 25
SN7409 0.45 0.42 0.35	SN7475 0-55 0-52 0-50	8N74161 2 60 2 40 2 25
8N7410 0 20 0 18 0 16	8N7476 0-45 0-42 0-39	SN74162 3 40 3 25 2 70
8N741I 0 23 0 22 0 20	8N7480 0 80 0 75 0 67	8N74163 8 40 3 25 2 70
8N7412 0-42 0-40 0-35	SN7481 1 25 1 15 1 10	8N74164 2 75 2 30 2 10
8N7413 0.30 0.27 0.25	8N7482 0-87 0-80 0-70	8N74165 4 00 3 50 8 00
8N7416 0.30 0.27 0.25	BN7483 1.00 0.90 0.85	SN74166 4 00 3 50 3 00
SN7417 0.30 0.27 0.25	8N7484 0-90 0-85 0-80	8N74167 6:25 5:60 5:10
8N7420 0.20 0.18 0.16	8N7486 0-45 0-41 0-38	8N74170 4 10 3 55 3 05
8N7422 0.48 0.44 0.40	8N7490 0 75 0 70 0 85	8N74174 2 00 1 75 1 30
8N7423 0.48 0.44 0.40	8N7491AN1-00 0-95 0-90	SN74175 1 35 1 27 1 15
8N7425 0 48 0 40 0 35	8N7492 0.75 0.70 0.65	8N74176 1 60 1 35 1 20
8N7427 0.42 0.39 0.35	8N7493 0.75 0.70 0.65	8N74177 1 60 1 35 1 20 8N74180 1 55 1 30 1 20
8N7428 0·50 0·45 0·42	8N7494 0.80 0.75 0.70	
8N7430 0 · 20 0 · 18 0 · 16 8N7432 0 · 42 0 · 39 0 · 35	8N7495 0 80 0 75 0 70 8N7496 1 00 0 97 0 95	8N74181 7 00 6 00 5 50 8N74182 2 00 1 80 1 80
	8N7496 1 00 0 97 0 95 8N7497 6 25 5 50 5 00	8N74184 2 40 2 00 1 80
8N7433 0-70 0-61 0-44 8N7437 0-65 0-60 0-50	8N74100 2 50 2 30 2 00	8N74185A 2 40 2 00 1 80
8N7438 0 · 65 0 · 60 0 · 50	8N74104 1 45 1 35 1 20	8N74190 1 95 1 85 1 75
8N7440 0 20 0 18 0 16	8N74105 1 45 1 35 1 20	8N74191 1 95 1 85 1 75
8N7441AN 0 75 0 72 0 70	8N74107 0 50 0 45 0 40	8N74192 2:00 1:90 1:80
8N7442 0.75 0.72 0.70	8N74110 0-80 0-70 0-60	8N74193 2:00 1:90 1:80
8N7443 1 00 0 95 0 90	8N74118 1 00 0 95 0 90	8N74194 2-50 2-25 1-90
	SN74119 1 90 1 78 1 65	8N74195 1 85 1 70 1 80
	8N74121 0-60 0-55 0-50	SN74196 1 50 1 40 1 30
8N7446 2:00 1:75 1:60	BN74122 1 35 1 25 1 10	SN74197 1 50 1 40 1 30
8N7447 1.75 1.80 1.45	SN74123 2 70 2 55 2 47	SN74198 4 60 3 70 3 35
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AAY42	15p	BC169C 12p	BY100 15p	OC45 15p	V405A 25p	2N3440 75p
AAZ13	10p	BC182 10p	BY126 15p	OC57 50p	ZTX 108 12p	2N3442 1.25
AC107	35p	BC214 15p	BY127 15p	OC71 15p	ZTX300 12p	2N3525 75p
AC126	25p	BCY32 75p	BYZ13 35p	OC72 25p	ZTX 301 15p	2N3614 59p
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AC128	25p	BCY39 1:00	GETHI 55p	OC81 25p	ZTX341 20p	2N3702 10p
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AC188	25p	BCY 55 2-50	LM309K	OC170 25p	2G301 30p	2N3714 1-80
ACY17	80p	BCY70 15p	(T03) 1-87	OC171 30p OC200 45p	2N404 20p 2N527 35p	2N3771 1.75 2N3773 2.00
ACY20	20p	BCY71 20p	MAT121 25p	OC200 45p OC201 75p	2N696 15p	2N3790 2:25
ACY21 ACY39	20p	BCY72 15p BCY87 2-99	MJE340 50p MJE370 70p	OC201 75p	2N697 15p	2N3819 35p
AU 139 AD 140	55p 50p	BCZ11 50p	MJE520 75p	OC202 50p	2N706 10p	2N3820 50p
AD149	50p	BD124 80p	MJE2955	OCP71 1-25	2N930 20p	2N3866 85p
AD143	35p	BD131 75p	1.10	ORP12 50p	2N987 45p	2N3903 15p
AD162	35p	BD132 80p	MJE3055	ORP60 40p	2N1131 25p	2N3906 12p
AF117	20p	BF115 25p	75p	P346A 20p	2N1132 25p	2N4061 12p
AF118	50p	BF167 25p	MPF105 40p	RASSIOAF	2N1302 18p	2N4062 12p
AF124	25p	BF173 25p	NKT214 20p	45p	2N1304 22p	2N4126 15p
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ASY27	30p	BF195 15p	NKT404 50p	TIL209 39p	2N1613 20p	28001 1:00
ASY28	25p	BF861 25p	OA5 50p	TIP29A 50p	2N1671 1-00	28012 10:00
BA102	30p	BFS98 25p	OA10 35p	T1P30A 60p	2N2147 75p	28018 6 25
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BC113	15p	BFY90 59p	OC28 65p	3:00	cols) 10p	40486 75p
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A.F. SIGNAL GENERATOR (Nov. 72)

S/c's, Rs, Cs, Pots, Sw's. PCB (2\frac{1}{2}\in \times 4\in) also holds Sw's, £3-15.

AUDIO MIXER (Jan. 72)
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BIOLOGICAL AMPLIFIER (Jan./Feb. 73)—Pre-amp Set S/c's, 1.c's Rs, Cs, Pots, PCB ($1\frac{1}{2}$ in \times $3\frac{3}{2}$ in), £3·70.

DOOR BELL YODELLER (Apr. 71)—S/c's, Rs, Cs, Pots, Transformer. Loudspeaker, PCB (3in × 3½in), £7-70.

ELECTRONIC PIANO (Sept. 72/Jan. 73)—Details in lists

GEMINI STEREO AMP (Nov. 70/Mar. 71) Stereo Sets and PCBs Pre-Amp—S/c/s, £1-85. Cs. Pots, Maka-Sw's—with ½W 2% M.O. Rs, £11-60 —with ½W 5% CF. Rs, £9-35.—with ½W 5% CF. Rs, £9-35.—with ½W 5% CF. Rs, £6-90. PCB (3½in × 10½in) for kits with M.O. or ½W CF. Rs. Holds pots and Maka-Sw's, £2-10. Main Amp—Rs, Cs. Pots, £5-40. PCB (3½in × Sin), £1-40. PSU—Rs, Cs. Pot, £3-70. PCB (2in × 4in), 75p.

GEMINI STEREO TUNER (Apr./Jun. 72) Rs, Cs, Pot, £3-80. PCB as published, £1-80.

HI-FI TAPE LINK (Mar./Apr. 73)
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LOGICAL RADIO CONTROL (Dec. 71/Jan. 72)-Details in lists

MODEL SERVO CONTROL (Feb./Mar. 72)-Details in lists.

MICROPHONE MIXER (Apr. 69)—S/c's, Rs, Cs, Pots, PCB (32in × 42in) also holds pots, £4-10,

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REVERBERATION UNIT (Practical Wireless Nov.-Dec. 72) Sicks, Rs, Cs, Slider Pots T/former, £6:80. PCB ($2\text{in} \times 11\frac{1}{2}\text{in}$) also holds sliders (compatible with publ. panel) £1:20.

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## 15 19 19 19 19 19 19 19		2N3403 2N3404		3N 199		AF240 AF279	0-54	BCY33	0-34	BFX44	0-33	
## Section	2G303 0.2	2N3405	0.27	3N141	0.58	AFY42	0-74	BCY38	0.40	BFX68	0.30	8N7403 0-20 8N7404 0-22 8N7420 0-20
12 12 12 12 12 12 12 12	2G309 0.3	2N3415 2N3416	0-15		0.92	AL102	0.75	BCY40	0-81	BFX85	0-29	53.777
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1.00 1.00	2N491 8.2	5 2N3705	0.10	10070	0.79	ASZ21	0.55	BCY71	0.22	BFY18	0.25	8N74123 1-35 8N74164 2-28 8N74193 1-90
1.00	2N696 0-1	5 2N3707	0.13	40251	0.81	BC107	0.14	BCY87	3-47	BFY20		SN74145 1-50 SN74167 6-25 SN74198 4-60
STORE A. 1. STORY CONTROLLED SET 1. S. 1.	2N698 0.2	5 2N3709	0.09	40310	0-59	BC109	0-14	BCY89	0.80		0.43	SN74151 1-10 SN74175 1-35
SILICON RECTIFIES 10	2N706 0-1	0 2N3711	0-90	40316	0.50	BC114		BCZ11	0.75	BFY50	0.16	
\$22711	2N708 0-1	8 2N3713	1.08	40322	0.92		0-15 0-18	BD121	0.75	BFY52	0.16	
### 1877 648 62377 638 6236 628 6213 628 628 6213 628 62	2N711 0-8	0 2N3715	1.28	40361	0.48	BC117	0.11	BD124	0-67	BFY56	0-34	
## 1971 0.45 0.5377 0.48 0.092 0.68 0.0137 0.03 0.0137 0.03 0.0137 0.03 0.0137 0.03 0.0137 0.03 0.0137 0.03 0.0137 0.03	2N718A 0-8	0 2N3773	2.35	43063	0.46		0.23	BD131	0.40	BFY75	0.40	1A 8p 9p 10p 11p 12p 15p 20p —
2879 2879 288 6660 284 6C18 288 6C18 6C1	2N721 0.5	5 2N3775	4-19 5-95	40394	0-50	BC125	0.15	BD135	0.43	BFY77	0.24	3A 15p 17p 20p 22\p 25p 27p 30p 35p 6A — 25p 30p 32\p 35p — —
Section Sect	2N916 0-1 2N918 0-8	7 2N 3778			0.38	BC132	0.30	BD137	0.55	BFY90	0.60	15A 36p 45p 48p 55p 65p 75p 87p
\$28500	2N919 0-2 2N929 0-1	4 2N3780	4-50	40409	0.52	BC135	0.11	BD139	0.71	BSX 19	0.13	
Silling 680 20078	2N1090 0.2	3 2N3782	8-37	40411	2.25	BC137	0.15	BDY10	1.25	BSX21	0-20	
THIS ISSUE. THIS	2N1131 0-8	0 2N3790	2.20	40438	1.44	BC140	0.34	BDY17	1.50	BSX28	0-25	NE555 TIMER I.C. 90p. AS DESCRIBED IN
251130	2N1302 0-1	6 2N3791	2.06	40468A	0.44	BC142	0.24	BDY19	1.97	BSX30	0-68	THIS ISSUE.
\$\frac{5}{2} \text{5}	2N1304 0-9	0 2N3794	0.10	40601	0-67	BC144	0-24	BDY38	0.80	BSX 60	0-54	
281189	2N1306 0-9	2 2N3820	0-47	40603	0.58	BC147		BDY62	0.75	BSX76	0-15	NE560 phase locked loop, £4.48. SCORPIO IGNITION KIT for improved performance £10 +
231483 0-86 283894 0-16 ACID 0-33 inclor 0-14 inclored 1-15 acid inclo	2N1308 0-2	5 2N3824	0.75	40636	1.10	BC149 BC153	0.18	BF117	0.43	ASX78	0.25	50p P. & P.
30 183	2N1483 0-1 2N1507 0-2	0 2N3854 4 2N3854A	0·16 0·16	AC107	0.16	BC157	0.14	BF121	0.25	BSY24	0-20	MONTHIV VEWS FRATIRE
28.1836 0-22 2.83858, 0-16 AC197 0-25 BC168 0-13 FF153 0-25 BS170 1-10 2.8385 0-15 AC197 0-25 BC168 0-13 FF153 0-15 BS170 1-10 2.8385 0-16 AC197 0-25 BC168 0-13 FF153 0-15 BS170 0-15 BS17	2N1631 0-3	5 2N3855A	0.16	AC117	0.20	BC159	ტ.14	BF125	0.25	BSY26	0.20	1. NEW OFFICE OPEN: 65 BATH STREET, GLASGOW.
2011 1.1 2.8838.	2N 1638 0-1	7 2N3856A	0.16	AC126	0.25	BC167B	0.11	BF152	0.20	BSY28	0.15	2. NEW OFFICE OPENING BRISTOL. WATCH THIS SPACE.
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2001 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	2N 1893 0-	4 2N3859A	0.16	AC142K	0.25	BC169C	0.13	BF160	0.23	BSY53	0.25	
2872192	2N2147 0-	0 2N3866	0.70	AC152V	0.17	BC171	0.11	BF163	0.20	BSY56	0.79	
287193 0-40 2873970 1-91 AC176 0-18 BE183 0-09 BE1717 0-95 BE785 0-90 BE1712 479 AC187 BE 287390 0-19 AC187 AC188 AC184 0-19 AC187 AC188 AC184 0-19 AC188 AC184 0-19 AC188 AC188 AC184 AC188 AC1	2N2192 0-	0 2N3877A	0.26	AC153K	0.25	BC182L	0.12	BF167	0.18	BSY78	0-40	
282194 0.27 2873900 0.28 27 2873900 0.28 27 2873900 0.28 287391 0.28 2873991 0.28 287391 0	2N2193 0-	0 2N3879	1.91			BC183L	0.09	BF177	0.25	BSY790	0.45	IN4007 20p AAZ15 12p BAY38 25p OA9 10p
2N2195 0-37 2N3903 0-27 CV18 0-15 BC187 0-25 BP181 0-32 C111 0-28 C112 CV18 0-15 BC187 0-35 BP181 0-32 C111 0-28 C112 CV18 0-38 C12 CV19 0-29 BC205 0-10 BP183 0-40 GET111 0-28 C112 CV18 0-38 C12 CV19 0-29 BC205 0-10 BP183 0-40 GET111 0-28 CV19 0-29 BC205 0-10 BP183 0-15 GET113 0-28	2N2194 0-	2N3900A	0.32	AC188K	0-26	BC184L	0.11	BF179	0.30	BU104	1-42	IS113 15p BA100 15p BY103 22p OA70 7p
2N2218A 0-80 2N9905 0-82 ACY219 0-80 BC295 0-10 BF183 0-40 GET111 0-45 [8113] 109 BA115 77 0-882 109 200 201 201 201 201 201 201 201 201 201	2N2195 0- 2N2195A 0-	37 2N3903 18 2N3904	0-17	ACY18	0.15	BC187	0.25	BF181	0.32	C111	0.53	18121 14p BA110 25p BY124 15p OA79 7p
5 NAPS20	2N2219 0-	37 2N 3906	0.22	ACY20	0.20	BC205	0-10	BF183	0-40	GET111 GET113	0.45 0.20	IS131 10p BA115 7p BY127 17p OA82 10p
5.00.221 A 0-38 5.00.4059 C-069 ACY30	2N2220 0-	20 2N4037	0.42	ACY22	0.13	BC207	0-10	BF185	0.17	GET115	0-50	18920 7p BA142 17p BYX10 22p OA91 7p
2N2929 0.40 1 2N4060 0.11 ACV40 0.17 BC212L 0.16 BF197 0.15 BF198 0.18 BF198 0.18 BF198 0.18 BF198 0.18 BF198 0.18 BF198 0.18 BF199 0.18 GET33 0.20 N2369 0.80 1.1 ACV40 0.17 BC212L 0.16 BF198 0.18 GET33 0.20 N2369 0.80 1.2 N4002 0.25 ACV44 0.31 BC214L 0.22 BF199 0.18 GET33 0.20 RET33 0.20 N2369 0.80 1.2 N4013 0.50 AD140 0.55 BC237 0.09 BF235 0.20 RET33 0.10 GET33 0.20 AD140 0.55 BF207 0.20 RET33 0.10 GET33 0.20 AD140 0.55 BF207 0.20 RET33 0.20 GET37 0.20 AD140 0.55 AD142 0.56 BC237 0.09 BF235 0.22 GET39 0.20 AD140 0.55 AD149 1.26 BC251 0.20 BF235 0.22 GET39 0.20 BF235 0.22 GET39 0.20 BF235 0.22 GET39 0.20 BF235 0.22 GET39 0.20 BF235 0.20 BF2	2N2221A 0.	38 2N4059	0.09	ACY30 .	0.42	BC209	0.10	BF195	0.15	GET120	0.25	18923 12p BA145 17p BYZ11 32p OA200 7p
2N2399 0.89 2N4902 0.28 ACV44 0.31 BC231 0.98 BC239 0.99 BC239 0.90 BC239 0.96 BC239 0.96 BC239 0.96 BC239 0.96 BC239 0.96 BC239 0.99 BC239 0.	2N2222A 0-	11 2N4061	0.11	ACY40	0.17	BC212L	0.16	BF198	0.15	GET536	0.20	BAX13 5p BYZ13 25p TIV307 50p
2N2466 0-80 0-80 2N4913 0-80 AD140 0-85 BC238 0-09 BF223 0-20 BC237 0-20 BC2	2N2369 0-	39 2N4302	0.32	ACY44 AD136V	0.96	BC236	0.16	BF200	0-35	GET873	0-12	
2N2712 0-12 0-12 0-14 0-17 0-17 0-17 0-17 0-17 0-17 0-17 0-17	2N2647 1.	2N4914	0.87	AD140	0-55	BC238	0-09	BF225J	0.19	GET883	0-20	MINITRON 3015F 7-SEGMENT Carbon:
28 28 28 28 28 28 28 28	2N2712 0.	12 2N4916	0.20	AD143	0.45			BF238	0.22	GET890	0.22	DRIVER 8N7447 21-30 Log. or Lin., with switch, 29p.
No.	2N2714 0-	17 2N4918	0-50	AD149V	1-26			BF245	0.33	T1P29A	0-49	Twin Ganged Stereo Pots, Log. or
Nation	2N2904A 0.	2N4920	0.60	AD161	0-49	BC258	0-09	BF247	0.49	TIP31A	0.62	DIODE. Made by TEXAS INST.
2N 2906 0-28 2N 4926 0-90 ADZ11 1-50 BC263 0-28 BF258 0-53 N 2900 PHOTORESISTORS N 2907 0-18 2N 4927 1-00 AP108 0-40 BC303 0-54 BF264 1-45 N 2902 0-12 2N 4930 2-20 AF116 0-25 BC300 0-54 BF264 1-45 N 2902 0-12 2N 4930 2-25 AF114 0-25 BC300 0-54 BF264 1-45 N 2902 0-12 2N 4930 2-70 AF115 0-24 BC304 0-43 BF271 0-21 N 2902 0-12 2N 4930 2-70 AF116 0-25 BC307 0-10 BF271 0-21 N 2902 0-12 2N 4930 2-70 AF116 0-25 BC307 0-10 BF271 0-21 N 2903 0-12 2N 4930 2-70 AF117 0-20 BC307 0-10 BF271 0-21 N 2903 0-10 2N 5175 0-28 AF118 0-50 BC307 0-10 BF273 0-25 N 2903 0-10 2N 5176 0-32 AF116 0-25 BC308 0-09 BF459 0-57 N 2903 0-10 2N 5176 0-32 AF126 0-20 BC3080 0-09 BF459 0-57 N 3005 0-40 2N 5195 1-46 AF127 0-20 BC3090 0-10 BF821A 2-30 N 3030 0-20 2N 5193 1-01 AF139 0-38 BC3090 0-10 BF821A 2-30 N 3030 0-20 2N 5193 1-01 AF139 0-38 BC3090 0-10 BF821A 2-30 N 3030 0-20 2N 5193 1-01 AF139 0-38 BC3090 0-10 BF821A 2-30 N 3030 0-20 2N 5193 1-01 AF139 0-38 BC3090 0-10 BF821A 2-30 N 3030 0-20 2N 5193 1-01 AF139 0-38 BC3090 0-10 BF821A 2-30 N 3030 0-20 2N 5193 1-01 AF179 0-25 BC337 0-24 BF861 0-25 N 3030 0-20 2N 5195 1-46 AF172 0-25 BC337 0-24 BF896 0-28 N 3030 0-12 2N 5459 0-43 AF180 0-50 BC338 0-19 BFW10 0-81 N 3030 0-12 2N 5459 0-33 AF186 0-40 BC337 0-19 BFW10 0-81 N 3030 0-12 2N 5459 0-33 AF186 0-50 BC338 0-19 BFW10 0-81 N 3030 0-12 2N 5459 0-33 AF186 0-50 BC338 0-19 BFW10 0-81 N 3030 0-12 2N 5459 0-33 AF186 0-50 BC338 0-19 BFW10 0-81 N 3030 0-12 2N 5459 0-33 AF186 0-50 BC338 0-19 BFW10 0-81 N 3	2N2905A 0-	2N4922	0.55	AD161	PR	BC261		BF255 BF257	3.47	TIP34A	1.51	PRESETS (CARBON)
2N2907A 0-25 2N4928 1-80 AF106 0-27 BC290 0-27 BF261 0-28 BC290 1-19240 0-79 T1P42A 0-79	2N2906A 0-	28 2N4926	0.90	ADZ11 ADZ12	1.50 1.75	BC263	0.28	BF259	0-48	TIP36A	3-70	0.2 Watt 6p OR
2N2924 0-12 2N4930 2-28 AF114 0-28 BC304 0-48 BF271 0-21 PR305 0-88 T1P3055 0-88 T	2N2907A 0- 2N2923 0-	25 2N4928 12 2N4929	1.80 2.23	AF106 AF109R	0-27 0-40	BC302	0.27	BF264	1-45	TIP42A	0.90	ENCAPSULATED FULL-WAVE
2N2926	2N2925 0-	12 2N4930 12 2N4931	2.70	AF115	0.24	BC304	0-43	BF271	0-21			SLIDE POTENTIOMETERS
Vellow 0-10 285175 0-32 AF118 0-50 BC387V1 0-10 BP274 0-53 AF118 0-50 BC388 0-09 BP458 0-65 DR387V1 0-10 BP274 0-53 AF118 0-50 BC387V1 0-10 BP274 0-53 AF118 0-50 BP274 0-53 AF118 0-50 BP274 0-50 BP27	Green 0.	10 2N5174	0.22	AF117	0.20	BC307A	0.10	BF273	0.25			2.5 watt 5% (up to 270 ohms SINGLE GANGED. LOG or LIN
28 3030	Orange 0.	10 2N5176	0.32	AF121	0-22	BC308	0-09	BF457	0.53	DETI	IDN	5 watt 5% (up to 8.2k Ω only), 9p. 10 watt 5% (up to 26k Ω only), TWIN GANGED. LOG or LIN
2N 3995	2N3054 3.	89 2N5190	0-92	AF125	0.20	BC308B	0.09	BF459	0.57	KEI	, IV 14	
2N3391 0-20 2N5194 1-10 AP170 0-25 BC313 0-30 BF861 0-28 POST 2N3391 0-20 2N5195 1-48 AP172 0-25 BC327 0-24 BF898 0-28 2N3392 0-13 2N5245 0-43 AP178 0-55 BC328 0-22 BFW10 0-61 2N3393 0-12 2N5457 0-33 AP180 0-50 BC338 0-19 BFW10 0-61 2N3394 0-17 2N5468 0-33 AP180 0-50 BC338 0-19 BFW10 0-35 BFW10 0-61 2N3402 0-12 2N5469 0-33 AP180 0-50 BC338 0-19 BFW10 0-35 BFW10 0-61 BFW10	2N 3055 0-	80 2N5192	1.24	AF127	0.20	BC309A	0.10	BF821A	2.30	0	F	Small value
2N3392 0-13 2N5245 0-43 AF179 0-55 BC238 0-22 BFW10 0-61 2N339 0-12 2N5457 0-35 AF179 0-65 BC338 0-19 BFW10 0-61 2N339 0-17 2N5458 0-33 AF180 0-50 BC338 0-19 BFW15 0-35 SERVICE 2N3302 0-12 2N5459 0-33 AF180 0-40 BC730 0-35 BFX13 0-23 SERVICE 2N3402 0-12 2N5459 0-33 AF180 0-40 BC730 0-35 BFX13 0-23 BFX13 0-2	2N3391 0-	20 2N5194	1.10	AF170	0.25	BC313	0.30	BF861	0.27	D∩	ST	
2N3394 0-17 2N5458 0-38 AF180 0-50 BC338 0-19 BFW15 0-35 SERVICE N3302 0-12 2N5459 0-38 AF180 0-40 BC338 0-19 BFW15 0-35 SERVICE MIGHT Part and Parking 18n per order. Europe 25p. Commonwealth (Air) 65p (MIN). Matching charge (audio	2N3392 0-	13 2N5245	0.43	AF178	0.55	BC328 BC337	0.22	BFW10 BFW11	0-61 0-61		<i>-</i> 1	
Post and Packing 13n per order. Europe 25n, Commonwealth (Air) 65p (MIN.). Matching charge (audio	2N 3394 0- 2N 3402 0-	17 2N5458 12 2N5459	0-33 0-33	AF180 AF186	0-50 0-40	BC338 BCY30	0.35	BFX13	0.23			
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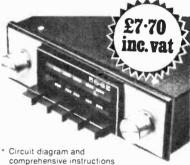
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071 15332	16	3300	2·4 amps	loz	15p
071 15472	16	4700	3.9 amps	loz	17p
071 15682	16	6800	5-8 amps	1 toz	22p
071 15103	16	10000	7-9 amps	2∮oz	27p
07! 18222	63	2200	5-8 amps	3oz	30p
072 15752	16	7500 + 7500	10.5 amps	3oz	37p
072 51 3	16	11000 + 11000	13-8 amps	410z	49p
071 16222	25	2200	2 2 amps	loz	15p
071 16472	25	4700	5.4 amps	l∮oz	22p
072 16502	25	5000 + 5000	9.6 amps	3½oz	37p
072 6752	25	7500 + 7500	12 6 amps	4 oz	49p
072 17342	40	3400 + 3400	9 I amps	3 ±oz	37p
072 17502 071 18681	40	5000 + 5000 680	12 0 amps	4½oz	49 p
072 18172	63		2 I amps	loz	15p
	63 .	1650 + 1650	7-8 amps	3oz	37p
106 and 10			_		
106 15103	16	10000	7 amps	2½oz	65p
106 16223	25	22000	17 amps	10oz	£1-12
106 17103	40	10000	12 amps	7±oz	94p
106 18153	63	15000	28 amps	I Boz	61.79
107 10222	100	2200	10 amps	5∮oz	74p
Type No.	Voltage	Capacitance	Weight		Price
102 15163	16	16000	8oz		20p
104 90003	20	39000	16oz		30p
102 16802	25	8000	7oz		25p
104 17562	40	5600	5oz		25p
104 90001	45	20000	l 6oz		50p
104 18332	63	3300	5oz		25 p
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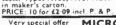
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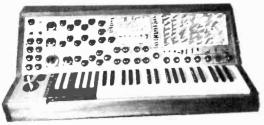
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