#### **INSTRUCTION MANUAL**

### **MODEL 193 20 MHz SWEEP/MODULATION GENERATOR**

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#### **SAFETY**

This instrument is wired for earth grounding via the facility power wiring. Do not bypass earth grounding with two wire extension cords, plug adapters, etc.

BEFORE PLUGGING IN the instrument, comply with installation instructions.

MAINTENANCE may require power on with the instrument covers removed. This should be done only by qualified personnel aware of the electrical hazards.

The instrument power receptacle is connected to the instrument safety earth terminal with a green/yellow wire. Do not alter this connection. (Reference:  $\bigoplus$  or  $\widehat{\Lambda}$  stamped inside the rear panel near the safety earth terminal.)

WARNING notes call attention to possible injury or death hazards in subsequent operations.

CAUTION notes call attention to possible equipment damage in subsequent operations.



## SECTION GENERAL DESCRIPTION

#### 1.1 THE MODEL 193

The Wavetek Model 193, a 20 MHz Sweep/Modulation Generator, is a precision source of sine  $\wedge$ . triangle  $\wedge$ , square  $\square$ , AM sine  $\bigotimes$  and dc. All waveforms are front panel variable from 0.002 Hz to 20 MHz and can be internally and externally modulated. Outputs can be continuous, triggered or gated by an external trigger signal or a front panel manual trigger switch. The main generator can be swept, frequency modulated, 0 to 100% amplitude modulated and suppressed carrier modulated using the auxiliary generator or an external source. The auxiliary generator, in addition to being an internal modulation generator, can be used as an independent function generator. Both generators have symmetry control of waveforms and can be frequency modulated by an external signal. Each modulation parameter has a static setup feature to allow its exact selection. Amplitude of the waveforms is variable from 30 Vp-p (15 Vp-p into  $50\Omega$ ) down to 1.5 mVp-p. DC reference of the waveform can be offset positively or negatively. Maximum 150 mA peak current can be continuously varied over an 80 dB range. A sync output provides a TTL level into  $50\Omega$ .

#### 1.2 SPECIFICATIONS

#### 1.2.1 Main Generator

#### **Waveforms**

Selectable sine  $\wedge$  , triangle  $\wedge$  , square  $\square$  , AM sine  $\mbox{\em $\omega$}$  and dc.

#### Symmetry

Symmetry of all waveform outputs is continuously adjustable from 1:19 to 19:1\*. Varying symmetry provides variable duty-cycle pulses, sawtooth ramps and distorted sine waves.

#### **Operational Modes**

Continuous: Generator runs continuously at selected frequency.

Triggered: Generator is quiescent until triggered by external signal or manual trigger, then generates one complete waveform cycle at selected frequency.

Gated: As triggered mode, except output continues for duration of gate signal. Last waveform started is completed.

#### Frequency Range

0.002 Hz to 20 MHz in nine overlapping decade ranges with 1% of full scale vernier.

#### **Function Output**

Waveforms variable to 30 Vp-p (15 Vp-p into  $50\Omega$ ). Waveforms may be attenuated continuously to 80 dB.  $50\Omega$  source impedance.

#### DC Output and DC Offset

Adjustable between  $\pm 15$  Vdc ( $\pm 7.5$  Vdc into  $50\Omega$ ) with signal peak plus dc offset limited to  $\pm 15$  Vdc ( $\pm 7.5$  Vdc into  $50\Omega$ ). DC offset and waveform attenuated proportionately 10 dB/step to 70 dB.

#### **Sync Output**

TTL level pulse into  $50\Omega$ . Duty cycle varies with SYM control.  $50\Omega$  source impedance.

#### **GCV Output**

0 to +5V open circuit voltage level proportional to main generator frequency.  $600\Omega$  source impedance.

#### **AM—Amplitude Modulation**

Either an internal signal (AUX GEN), or external signal (AM IN), or internal plus external signals (AUX GEN MODE—AM, main generator FUNC—AM) amplitude modulates the main generator's sine wave in either the 0 to 100% AM or suppressed carrier mode.

AM Input: 5 Vp-p gives 100% modulation 10 Vp-p gives suppressed carrier operation.  $600\Omega$  input impedance.

Carrier Level (0 to 100% AM): Adjustable 10 to 50% of full amplitude at function output.

Carrier Null (Suppressed Carrier AM): Adjustable  $\pm 2\%$  of full amplitude at function output.

#### VCG—Voltage Controlled Generator

Up to 1000: 1 frequency change with external 0 to  $\pm\,5\text{V}$  signal. Upper frequency limited to maximum of range.

Slew Rate: 2% of range per  $\mu$ s.

Linearity:  $\pm 0.5\%$  thru X100K range;  $\pm .5\%$  on X1M and X10M.

Impedance:  $10 \text{ k}\Omega$ .

#### **Trigger Input**

Input Range: 1 Vp-p to ± 10 V.

Trigger Level Adj: -5V to +5V.

Impedance:  $1.5k\Omega$  shunted by 1.5 pF.

Pulse Width: 25 ns minimum.

Repetition Rate:

 $\begin{array}{lll} \text{Input} & \text{Max Rep Rate} \\ \pm 1\text{V} & \text{1 MHz} \\ \pm 2.5\text{V} & \text{10 MHz} \end{array}$ 

#### **Frequency Precision**

#### **Dial Accuracy**

 $\pm$  3% of full scale from X .1 Hz to X 1 MHz range.  $\pm$  5% of full scale on X 10M range.

#### 1.2.2 Amplitude Precision

#### Sine Frequency Response

 $< \pm 0.2$  dB all ranges thru X 100K.

 $< \pm 0.5$  dB on X 1M range.

 $< \pm 1.5$  dB on X 10M range.

#### **Step Attenuator Accuracy**

± 0.3 dB with 10, 20 and 40 dB.

± 0.6 dB with 30, 50 and 60 dB.

±0.9 dB with 70 dB setting.

#### 1.2.3 Waveform Characteristics

#### **Sine Distortion**

<0.5% on X 100, X 1K and X 10K. <1.0% on X 0.1 to X 100.

All harmonics 30 dB below fundamental on X 100K, X 1M range, and 25 dB below fundamental on X 10M range.

#### **Square Wave**

Rise/Fall Time: <15 ns (10% to 90%).

Total Aberrations: 5% of full amplitude for each waveform peak.

#### **Time Symmetry**

Square wave variation from 0.1 to 2 on dial less than: ±1% to 200 kHz. ±10% to 20 MHz.

#### **Triangle Linearity**

>99% for 0.002 Hz to 200 kHz.

#### 1.2.4 Auxiliary Generator

#### **Waveforms**

Selectable  ${}^{\textstyle \frown}$  ,  ${}^{\textstyle \frown}$  ,  ${}^{\textstyle \frown}$  , and  ${}^{\textstyle \frown}$  . Symmetry of  ${}^{\textstyle \frown}$  and  ${}^{\textstyle \frown}$  adjustable 1:19 thru 19:1\*.

#### **Frequency Range**

0.1Hz to 100 kHz in 4 ranges.\*

#### **Auxiliary Output**

Waveforms selectable and variable to 10 Vp-p (5V p-p into  $600\Omega$ ).

#### **Auxiliary Sync Output**

Rear panel BNC. TTL level pulse coincident with AUX GEN output. Duty cycle varies with symmetry control.

#### **AUX VCG Input**

Rear Panel BNC. Up to 33:1 frequency change with external  $\pm$  5V signal. Upper frequency limited to maximum of selected range. 11 k $\Omega$  input impedance.

#### 1.2.5 Modes of Operation

Internal auxiliary generator used as the modulation source.

#### FM

Two setup and one operate mode.

Set-Freq/Aux-Gen-Off: Disables auxiliary generator for main generator adjustment.

Set  $\Delta F$ : Allows vernier setup of peak deviation. Vernier range is up to  $\pm 10\%$  of main generator range.

FM: Operate mode.

#### **Sweep**

Two setup and one operate mode.

Set Start: Allows setup of main generator start frequency.

Set Width: Allows vernier setup of main generator stop frequency.

Sweep: Operate mode.

#### AM

Two setup and one operate mode for double sideband or suppressed carrier operation.

Set Carrier: Allows setup of carrier amplitude.

Set  $\Delta M$ : Allows the setup of the modulation level.

AM: Operate mode. 0 to 100 % AM or suppressed carrier modulation of main generator by auxiliary generator plus external signal if present at AM IN.

Aux Out Only: Disconnects the modulator for independent operation of auxiliary generator.

#### 1.2.6 General

#### Stability

Amplitude, frequency and dc offset at FUNC output after 2 hour warmup:  $\pm\,0.05\,\%$  for 10 minutes.  $\pm\,0.25\,\%$  for 24 hours.

#### **Environmental**

Specifications apply at 25°C  $\pm$ 5°C. Operates 0°C to  $\pm$ 50°C.

#### **Dimensions**

28.6 cm (11 $\frac{1}{1}$  in.) wide; 13.3 cm (5 $\frac{1}{1}$  in.) high; 28.6 cm (11 $\frac{1}{1}$  in.) deep.

#### Weight

4.6 kg (10 lb) net; 6.4 kg (14 lb) shipping.

#### Power

100/120/220/240V ( +5% -10%), 48 to 66 Hz, ≤70 VA. VA.

NOTE: All specifications apply from 0.1 to 2.0 on frequency dial, when FUNC OUT amplitude is max and 50  $\Omega$  terminated, and with SYM control OFF.

SYMMETRY and VERNIER controls affect frequency calibration. Maximum possible asymmetry is a function of frequency setting.

\*NOTE: When SYMMETRY control is used, indicated frequency is divided by approximately 10.

## SECTION 2 INITIAL PREPARATION

#### 2.1 MECHANICAL INSTALLATION

After unpacking the instrument, visually inspect all external parts for possible damage to connectors, surface areas, etc. If damage is discovered, file a claim with the carrier who transported the unit. The shipping container and packing material should be saved in case reshipment is required.

#### 2.2 ELECTRICAL INSTALLATION

#### 2.2.1 Power Connection

#### **WARNING**

To preclude injury or death due to shock, the third wire earth ground must be continuous to the facility power outlet. Before connecting to the facility power outlet, examine extension cords, autotransformers, etc., between the instrument and the facility power outlet for a continuous earth ground path. The earth ground path can be identified at the plug on the instrument power cord; of the three terminals, the earth ground terminal is the nonmatching shape, usually cylindrical.

#### **CAUTION**

To prevent damage to the instrument, check for proper match of line and instrument voltage and proper fuse type and rating.

#### NOTE

Unless otherwise specified at the time of purchase, this instrument was shipped from the factory with the power transformer connected for operation on a 120 Vac line supply and with a ¾ amp fuse.

Conversion to other input voltages requires a change in rear panel fuse holder voltage card position and fuse (figure 2-1) according to the following procedure.

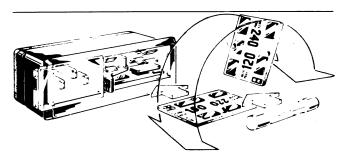


Figure 2-1. Voltage Selector and Fuse

- 1. Disconnect the power cord at the instrument, open fuse holder cover door and rotate fuse-pull to left to remove the fuse.
- Remove the small printed circuit board and select operating voltage by orienting the printed circuit board to position the desired voltage to the top left side. Push the board firmly into its module slot.
- 3. Rotate the fuse-pull back into the normal position and insert the correct fuse into the fuse holder. Close the cover door.
- 4. Connect the ac line cord to the mating connector at the rear of the unit and the power source.

<b>Card Position</b>	Input Vac	Fuse
100	90 to 105	³/₄ amp
120	108 to 126	3/4 amp
220	198 to 231	3/8 amp
240	216 to 252	³/ <sub>8</sub> amp
	l	

#### 2.2.2 Signal Connections

Use RG58U  $50\Omega$  coaxial cables equipped with BNC connectors to distribute signals when connecting this instrument to associated equipment.

#### 2.3 ELECTRICAL ACCEPTANCE CHECKOUT

This checkout procedure verifies the generator operation. If a malfunction is found, refer to the Warranty in

the front of this manual. A dual trace, 150 MHz bandwidth oscilloscope with X10 time base magnification, a 50 $\Omega$  load, a coaxial tee and three 50 $\Omega$  cables are required to perform this checkout.

Set up as in figure 2-2 and preset the generator front panel controls as follows; then perform the steps in table 2-1.

Control	Position
FREQ/START FREQ	1.0
FREQ MULT	1K
VERNIER/SYM FREG	Q CAL (cw)
SYM Off	(extended)
MODE	CONT

TRIG LEVEL	
FUNCTION	
DC OFFSET (On/Off)	OFF (Extended)
DC OFFSET (Variable Control)	CCW
OUTPUT ATTEN 40, 20, 10	All extended
AMPLITUDE	MAX (cw)
SUPPRESSED CARRIER	AM (extended)
AM CARRIER LEVEL/NULL	ccw`
AUX GEN MODE	AUX OUT ONLY
WIDTH/ΔF/ΔM	<b>cw</b>
AUX GEN FUNC	
AUX GEN SYMMETRY	<b>cw</b>
AUX GEN FREQ	100-3K
AUX GEN VERNIER	MAX (cw)
POWER	

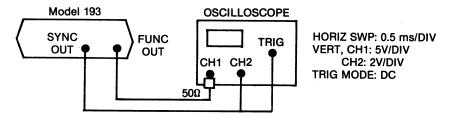


Figure 2-2. Main Generator Check Setup

Table 2-1. Checkout Procedure

Step	Control	Position/Operation	Observation
1	Oscilloscope	Trig level and slope, both positive.	CH2: Square wave that begins on positive going edge. CH1: 15 Vp-p sine wave.
2	Dial and VERNIER/SYM	Rotate dial full cw, vernier full ccw. Then the opposite. Return dial to 1.0, vernier to CAL.	CH2: Square wave remains in sync for all dial positions. Range is greater than 2 Hz to 2000 Hz (1000:1).
3	FREQ MULT	Rotate to all positions. Return to 1K position.	Frequency is 1 × each range position.
4	AMPLITUDE	Set to 6 Vp-p on scope.	CH1: Amplitude decreases to approximately 6 Vp-p.
5	DC OFFSET	Depress DC OFFSET switch, then rotate DC OFFSET Control cw. (Extended DC OFFSET upon completion.)	Full ccw gives negative offset. Clipping occurs when the offset plus waveform peak amplitude exceeds approximately $\pm 7.5 \text{V}$ into $50 \Omega$ . Initially the negative peak is clipped, but as the dc offset is rotated cw the clipping of the negative peak disappears and eventually the positive peak begins to clip.

**Table 2-1, Checkout Procedure (cont)** 

Step	Control	Position/Operation	Observation
6	AMPLITUDE	Rotate cw.	Waveform returns to 15 Vp-p.
7	OUTPUT ATTN 10, 20, 40	Depress buttons in various combinations. Then release all buttons.	Output level varies from 15 Vp-p (0 dB) to 4.7 mV (70 dB).
8	FUNC (Main Generator)	Select DC, $\wedge$ , $\wedge$ , $\square$ . Reset to $\wedge$ .	Observe 0 Vdc level; $\wedge$ , $\wedge$ and $\square$ are 15 Vp-p. Note phase relationships; $\square$ in phase with $\wedge$ and $\wedge$ .
9	SYM, VERNIER/SYM.	Depress SYM switch and rotate VERNIER/SYM control ccw. Extend SYM, return VERNIER/SYM to CAL.	Frequency decreases to approximately 100 Hz. ccw of the 12 o'clock position gives 1:19; cw gives 19:1 (askewed sine wave and variable duty cycle pulses can be observed for $\sim$ and $\square$ .)
10	MODE and FUNC (Main Generator)	Select GATE. Select $\land$ , $\land$ , $\ \square$ . Return to $\ \land$ .	A dc level near zero volts (except \( \backslut \) function; quiescent level is at negative peak value).
11	MANUAL TRIGGER	Press, hold and release.	A burst of $ {}^{ \! \! \! \! \! \! \! \! \! \! \! \! \! \! \! \! \! \! \!$
Set u	o trigger source (200 H	z triangle 10Vp-p), scope, and Mod	el 193 as shown in figure 2-3.
12	TRIG LEVEL	Rotate throughout its range. Return to 10 o'clock.	CH1: The number of waveform cycles in each gated ''burst'' varies with the trigger level. Notice relationships between Channels 1 and 2 waveforms as the TRIGGER LEVEL is rotated.
13	MODE	Select TRIG	CH1: A single triggered $\sim$ recurring at the 200 Hz trigger rate.
Retur	n all controls to the init	ial setup. Set up scope and Model	193 as shown in figure 2-4.
14	AUX GEN FUNC	Rotate to $\wedge$ , $\wedge$ , $\square$ .	$\sim$ , $\sim$ , $\square$ are approximately 10 Vp-p, 3 kHz.
15	Width/ΔF/ΔM	Rotate ccw. Return to cw position.	CH1:AUX OUT level varies from approximately 10 Vp-p to 0 Vp-p.
16	AUX GEN FUNC and SYMMETRY	Rotate to $\nearrow$ 1, $\mathbin{ ightharpoonup}$ 1. Vary symmetry control. Return to $\mathbin{^{\checkmark}}$ 2.	✓ and ∟ are approximately 10 Vp-p, 300 Hz. Symmetry varies from 19:1 to 1:19.
17	AUX GEN FREQ and VERNIER	Rotate VERNIER in all FREQ range positions. Return FREQ range to 100-3K and VERNIER to cw.	Frequency varies in each range to the range limits as marked.
	MODEL 193  SIG 500 OUT  FUNCT GENER/	TRIG (REAR)  OI (REAR)  AUX OUT	NOTE: UT Verify scope is properly triggered PANEL) (on positive going edge).  MODEL 193 OSCILLOSCOPE TRIG CH1 CH2 NO LOAD
	Figure 2-3. Chec	kout Setup	Figure 2-4. Auxiliary Generator Check

Table 2-1. Checkout Procedure (cont)

Step	Control	Position/Operation	Observation		
Set u	Set up the Model 193 and scope as shown in figure 2-5.				
18	AUX GEN MODE	Set to SET FREQ/AUX GEN OFF.	FM center frequency of 1kHz on CH1.		
19	AUX GEN FREQ Range, VERNIER	FREQ range to .1-3 and VERNIER to 12 o'clock.	(This sets the rate of modulation.)		
20	AUX GEN MODE and WIDTH/ΔF/ΔM	Select SET $\Delta F$ and vary WIDTH/ $\Delta F/\Delta M$ for 1.1 kHz on scope.	Approximately 1.1 kHz on CH1. This will be maximum peak frequency of modulated signal.		
21	AUX GEN MODE	Select FM.	Both channels: CH1 frequency varies with instantaneous amplitude of AUX OUT waveform on CH2.		
22	AUX GEN FUNC and SYMMETRY	Select ✓ and full ccw rotation.	(Ramp's instantaneous amplitude controls sweep: A slow sweep up and fast return to start frequency.)		
23	AUX GEN MODE	Select SET START	(Ramp is held at its start voltage.)		
24	FREQ/START FREQ Dial and VERNIER/SYM	Set dial to .02 and VERNIER full ccw.	CH1: Approximately 2 Hz. (start frequency).		
25	AUX GEN MODE and WIDTH/ΔF/ΔM	Select SET WIDTH and rotate WIDTH/ΔF/ΔM cw.	CH1: Approximately 2 kHz. (Max sweep frequency for range).		
26	AUX MODE	Select SWEEP	Sweep of main generator from 2 Hz to 2 kHz. (If AUX GEN symmetry is rotated cw, the main generator sweeps from 2 kHz to 2 Hz.		

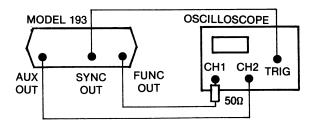


Figure 2-5. FM and Sweep Check Setup

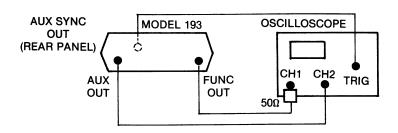
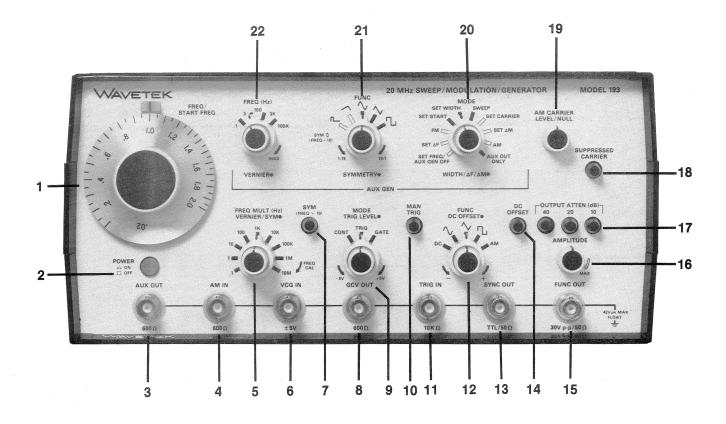
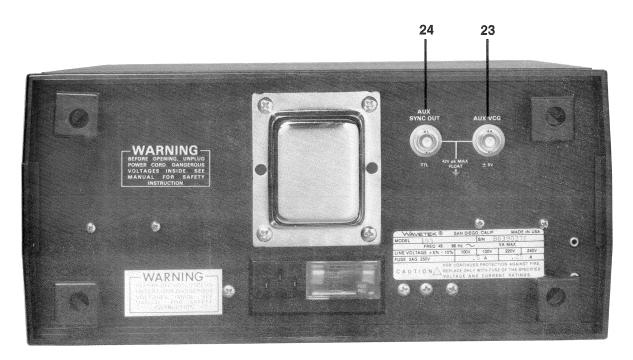


Figure 2-6. AM Check Setup

Table 2-1. Checkout Procedure (cont)

	Control Position/Operation Observation				
Step	Control	Position/Operation	Observation		
	Set up the Model 193	and scope as shown in figure 2-6.			
27	AUX GEN MODE FREQ MULT (Main Gen) FREQ/START FREQ Dial FUNC (Main Gen)	Select SET CARRIER 10K 1.0 AM	CH1: When AM is selected, the unmodulated carrier level at the FUNC OUT decreases to one-half the normal amplitude to prevent clipping of AM peaks. 7.5 Vp-p unmodulated carrier shown on CH1. (Note: AM CARRIER LEVEL/NULL varies carrier level.		
28	AUX GEN MODE AUX GEN FUNC AUX GEN FREQ AUX GEN FREQ VERNIER	Δ M √ 100l3K 9 o'clock			
29	WIDTH/ΔF/ΔM	Adjust for 4 Vp-p sine wave.	CH1: 4 Vp-p sine wave.		
30	AUX GEN MODE	AM	CH1: Modulation envelope varies with instantaneous amplitude of CH2 waveform.  Approximately 30% modulation.		
31	SUPPRESSED CARRIER	Depress	Suppressed carrier signal on CH1. (Adjusting AM CARRIER LEVEL/NULL controls the balance of the modulation envelope peaks.)		





REAR PANEL

Figure 3-1. Controls and Connectors.

# SECTION OPERATION

#### 3.1 CONTROLS AND CONNECTORS

The generator front panel controls and connectors are shown in figure 3-1 and keyed to the following descriptions.

- FREQ/START FREQ Dial Settings under the dial index mark summed with 6 and multiplied by 5 determine the main generator signal frequency at FUNC OUT 15.
- 2 POWER Pushbutton Depressed is power on, extended is power off.
- **3 AUX OUT Connector** This BNC is the auxiliary generator waveform output. The FUNC control **21** of the AUX GEN selects the waveform and the WIDTH/ $\Delta$ F/ $\Delta$ M control **20** varies the waveform level from 0 to 10 Vp-p (5 Vp-p into 600Ω). Source impedance is 600Ω.
- **4 AM IN Connector** This BNC receives the external amplitude modulation or suppressed carrier signal. When FUNC switch **12** selects AM, a 5 Vp-p signal gives 100% modulation for normal AM (SUPPRESSED CARRIER **18** extended) or a 10 Vp-p signal gives suppressed carrier operation (SUPPRESSED CARRIER **18** pressed). Source impedance is 600Ω.
- 5 FREQ MULT Control Outer coax control selects one of nine frequency multipliers for dial 1 setting.
  - VERNIER/SYM Control When SYM 7 is off (extended) this inner coax control is a fine adjustment of the dial 1 setting. When SYM 7 switch is on (depressed) this control varies the symmetry of the main generator waveforms (normally 50% duty cycle). Symmetry range is 19:1 to 1:19 (half cycle to half cycle ratio). When SYM is used, the main generator frequency is divided by 10. Extending SYM switch ensures 1:1 (50%) symmetry.
- 6 VCG IN Connector This BNC accepts ac or dc voltages to proportionately control main generator frequency within the range determined by the FREQ MULT 5. Positive voltages increase the frequency set by the dial 1;

negative voltages decrease the frequency. The VCG IN will not drive the generator frequency beyond the normal limits of a range. (Upper limit:  $2 \times FREQ$  MULT setting. Lower limit: 1/1000 upper limit). Input impedance is  $5 \text{ k}\Omega$ .

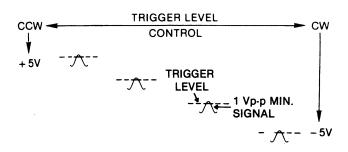
- 7 SYM Pushbutton This switch, when depressed, allows main generator waveform symmetry to be varied 19:1 to 1:19 range by the VERNIER/SYM control 5 (as a result, the main generator frequency is divided by 10). When extended, the switch allows the main generator to produce normal (50% duty cycle) waveforms.
- **8** GCV OUT Connector This BNC provides do excursions of 0 to +5V (open circuit) that represents the main generator output frequency in the range selected by FREQ MULT **5**. Source impedance is  $600\Omega$ .
- **9** MODE Control This outer coax control selects one of the three main generator operating modes:

**CONT** — Continuous output at FUNC OUT **15** and SYNC OUT **13** connectors.

TRIG — A dc level output until the generator is triggered by the MAN TRIG 10 or with a signal at the TRIG IN connector 11. When triggered, the generator output is one cycle of waveform followed by a dc level.

**GATE** — As for TRIG except the output is continuous for the duration of the manual or external trigger signal. The last waveform cycle started is always completed.

TRIG LEVEL Control — This inner coax control is a continuously variable adjustment of the trigger circuitry firing point. When full ccw, a positive going signal at approximately +5V is required for triggering (see figure 3-2). In the full cw position, a positive going signal at approximately –5V or more positive voltage is required for triggering. In the gated modes, the generator will run continuously when the control is cw of 12 o'clock.



Trigger signal must be a positive going signal exceeding the TRIGGER LEVEL setting.

Figure 3-2. Minimum Trigger Signal

- MAN TRIG Pushbutton Triggers or gates the output signals when main generator mode is TRIG or GATE 9. In trigger mode, one waveform cycle is output when the button is pushed. In gated mode, waveform cycles are continuously output as long as the button is held in.
- 11 TRIG IN Connector This BNC receives the external trigger and gate signals. These signals are applied to the trigger and gate circuit when the MODE switch 9 is in the TRIG or GATE positions. Refer to Section 1, Trigger (and Gate). The TRIG LEVEL control 9 varies the firing point of the TRIG IN signal.
- **12 FUNCTION Selector** Outer coaxial knob selects one of the three main generator waveforms (sine, triangle, square), dc, and AM.

DC OFFSET Control — Inner coaxial control offsets the main generator output waveform vertically from its normal position and, when FUNCTION (outer coaxial switch) 12 is in the DC position, controls polarity and voltage of dc output. DC output range is  $0 \pm 10$  Vdc ( $\pm 5$  Vdc into  $50\Omega$ ). DC OFFSET switch 14 must be depressed to enable this DC OFFSET control. Extending the DC OFFSET switch ensures zero volt offset. DC offset and waveform are attenuated by the OUTPUT ATTEN control 17, but dc offset is not attenuated by the AMPLITUDE control 16. Waveform peak voltage plus dc offset is limited to  $\pm 15$  Vdc ( $\pm 7.5$  Vdc into  $50\Omega$ ). See figure 3-3.

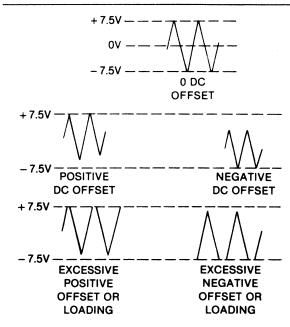


Figure 3-3. DC OFFSET Control

- 13 SYNC OUT Connector The sync signal from this BNC is a TTL level into 50Ω, synchronous with FUNC OUT 15 signal. In □, duty cycle varies with waveform symmetry. Source impedance is 50Ω.
- **14** DC OFFSET Pushbutton Depressed button activates dc offset **12**. Extended button ensures zero dc offset.
- **15 FUNC OUT Connector** This BNC is the waveform, or dc, output of the main generator (alone or modulated). Maximum output is 30 Vp-p (15 Vp-p into 50Ω). Source impedance is 50Ω.
- 16 AMPLITUDE Control Continuously varies waveform amplitude within each OUTPUT ATTEN 17 range. Full ccw rotation reduces waveform amplitudes by greater than 10 dB. DC and dc offset voltages are not affected by this control.
- 17 OUTPUT ATTEN Pushbuttons Select the attenuation range of the FUNC OUT 15 signal. The AMPLITUDE control 16 allows continuous waveform (but not dc) level variations within each attenuator range. Each of the three buttons may be used individually for 40, 20 or 10 dB steps of attenuation, or pressed in combinations for up to 70 dB of attenuation. The attenuator attenuates both the waveform and dc offset.
- **18** SUPPRESSED CARRIER Pushbutton When pressed, selects suppressed carrier operation and if extended, selects AM operation.

19 AM CARRIER LEVEL/NULL Control — In the AM function (FUNC 12) with the MODE switch 20 set to SET CARRIER, this control sets the peak-to-peak level of the carrier (main generator sine wave) at FUNC OUT 15.

If SUPPRESSED CARRIER **18** is pressed, the AM CARRIER LEVEL/NULL control balances the peaks of the suppressed carrier signal at FUNC OUT **15**.

20 MODE Switch — Outer coax control selects one of three basic operating modes of the auxiliary generator: frequency modulation, frequency sweep, or amplitude modulation. Additional settings allow static setup of these modes. The following sections describe each switch setting.

SET FREQ/AUX GEN OFF — Auxiliary generator is turned off for main generator only operation, or to allow center frequency (f<sub>o</sub>) adjustment for FM operation. The dial **1** and FREQ MULT **5** sets the center frequency (f<sub>o</sub>) for frequency modulation.

**SET**  $\Delta F$  — In preparation for FM operation, FUNC OUT **15** frequency is held at the maximum frequency deviation ( $f_{\circ} + \Delta f$ ). The WIDTH/ $\Delta F$ /  $\Delta M$  control **20** sets the maximum frequency. (Refer to previous paragraph for  $f_{\circ}$  selection.)

**FM** — Allows the auxiliary generator to frequency modulate the main generator creating an FM signal at FUNC OUT **15**.

**SET START** — In preparation for sweep operation, FUNC OUT **15** is held at initial sweep frequency. Dial **1** and FREQ MULT **5** sets the initial frequency. FUNC **21** sets the waveform that controls the main generator frequency.

**SET WIDTH** — In preparation for sweep operation, FUNC OUT **15** frequency is held at the final sweep frequency. The WIDTH/ $\Delta$ F/ $\Delta$ M control **20** sets the final sweep frequency.

**SWEEP** — Allows the auxiliary generator to sweep the main generator creating a frequency swept signal at FUNC OUT **15**.

**SET CARRIER** — In preparation for AM operation, allows setting of unmodulated output carrier level (main generator output) at FUNC OUT **15** by the AM CARRIER LEVEL/NULL control **19** and AMPLITUDE control **16**.

**SET**  $\Delta M$  — In preparation for AM operation, holds FUNC OUT **15** at the valley of the modulation signal from the auxiliary generator to allow level adjustment by the WIDTH/ $\Delta F/\Delta M$  control **20**.

**AM** — Allows the auxiliary generator to AM or suppress carrier modulate the main generator for AM or suppressed carrier signal at FUNC OUT **15**.

**AUX OUT ONLY** — Disconnects the auxiliary generator from the main generator. Allows the auxiliary generator to be used as an independent function generator whose output is AUX OUT **3**.

WIDTH/ $\Delta$ F/ $\Delta$ M — Inner coax control serves several functions dependent upon the outer coax MODE switch setting. With MODE in  $\Delta$ F position, WIDTH/ $\Delta$ F/ $\Delta$ M sets the maximum positive frequency deviation. With MODE SET WIDTH, WIDTH/ $\Delta$ F/ $\Delta$ M sets the final sweep frequency. With MODE in SET  $\Delta$ M, WIDTH/ $\Delta$ F/ $\Delta$ M sets the valley of the FUNC OUT **15** modulation signal. Finally, with MODE in AUX OUT ONLY, WIDTH/ $\Delta$ F/ $\Delta$ M controls the signal amplitude at AUX OUT **3**.

21 FUNC Control — This outer coax control selects one of five auxiliary generator waveforms: sine, triangle, square, sawtooth and pulse. Symmetry of the sawtooth and pulse is adjustable between 1:19 and 19:1 by the inner coax SYMMETRY control. When sawtooth and pulse are selected, the auxiliary generator frequency is divided by 10.

**SYMMETRY** — When outer coax FUNC control is set to sawtooth or pulse, this inner coax control varies the symmetry of sawtooth and pulse over a 1:19 to 19:1 range.

**22** FREQ Control — Outer coax control selects one of four auxiliary generator frequency ranges; e.g., 3-100 Hz.

**VERNIER** — Inner coax control provides continuous frequency control of the auxiliary generator frequency range selected by outer coax FREQ control.

- 23 AUX VCG Connector This BNC allows up to a 33:1 auxiliary generator frequency change with a  $\pm 5V$  signal. Upper frequency is limited to the maximum of the selected range. Input impedance is 11 k $\Omega$ .
- 24 AUX SYNC OUT Connector This BNC provides a TTL level pulse which leads the AUX OUT 3 sine wave by 90°, lags the triangle wave by 90° and is 180° out of phase with the square wave. AUX SYNC OUT duty cycle varies with SYMMETRY control 21.

#### 3.2 OPERATION

Perform the initial checkout in Section 2 for the feel of the instrument. Any questions concerning individual controls and connectors may be answered in paragraph 3.1.

#### 3.2.1 Signal Termination

Proper signal termination, or loading, of the generator connectors is necessary for its specified operation. For example, figure 3-4 shows proper termination of the FUNC OUT connector. Placing the  $50\Omega$  terminator, or  $50\Omega$  resistance, in parallel with a higher impedance, matches the receiving instrument input impedance to the coax characteristic and generator output impedance, thereby minimizing signal reflection or power loss on the line due to impedance mismatch.

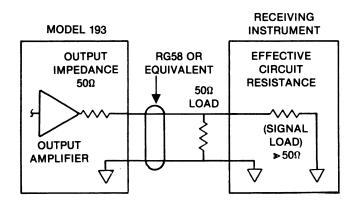


Figure 3-4. Signal Termination

The input and output impedances of the generator connectors are:

Connectors	Impedance
FUNC OUT	$\dots \dots 50 \Omega$
SYNC OUT	$\dots \dots \dots 50 \mathbf{\Omega}$
GCV OUT	$\dots\dots\dots$
AUX OUT	$0000 \dots$
AUX SYNC OUT	TTL
TRIG IN	1.5k $\Omega$
VCG IN	10k $\Omega$
AM IN	$\dots \dots $
AUX VCG IN	$\dots\dots 11 k\Omega$

#### 3.2.2 Manual Main Generator Basic Operation

The following steps demonstrate setting up the main generator for continuous operation. (Bold numbers are keys to figure 3-1. Refer to paragraph 3.1 for further explanations of controls and connectors.) Outputs are shown in figure 3-5.

Step	Control/Connector	Setting
1	AUX GEN MODE <b>20</b>	Select AUX GEN OFF.
2	FUNC OUT 15	Connect to under test (refer to paragraph 3.2.1).
3	MODE 9	Select CONT.
4	SYM <b>7</b>	Extended.
5	FREQ MULT 5	Set to desired range of frequency
6	FREQ/START FREQ Dial <b>1</b>	Set to desired frequency within the range.
7	FUNCTION 12	Select desired waveform.
8	DC OFFSET 14, Inner Control12	Set as desired. Limit offset to prevent clipping (See figure 3-3).
9	OUTPUT ATTEN 17	Select for desired attenuator range.

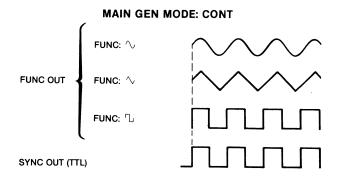
As an actual exercise in operation, perform steps 1 through 13 in table 2-1.

## 3.2.3 Voltage Controlled Main Generator Operation

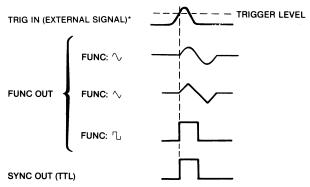
Operation of the main generator as a voltage controlled function generator (VCG) is as for a manually controlled function generator, only the frequency within particular ranges is additionally controlled by an external voltage ( $\pm$  5V excursions) injected at the VCG IN connector. Perform the steps given in paragraph 3.2.2, only set FREQ/START FREQ to determine a reference from which the frequency is to be voltage controlled:

- For frequency control with positive dc inputs at VCG IN, set the dial for a lower frequency limit.
- 2. For frequency control with negative dc inputs at VCG IN, set the dial for an upper frequency limit.
- For modulation with an ac input at VCG IN, set the dial at the desired center frequency. When applying VCG voltage, do not exceed the maximum dial range of the selected frequency range.

Figure 3-6 is a nomograph with examples of dial and voltage effects. Example 1 shows that with OV VCG input, frequency is determined by the dial setting, 1.0 in this example. Example 2 shows that with a positive

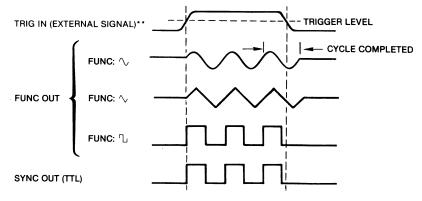


#### MAIN GEN MODE: TRIG



\*TRIGGERS ONE FUNCTION GENERATOR CYCLE FOR EACH POSITIVE GOING TRANSITION.

#### MAIN GEN MODE: GATE



\*\*GATES FUNCTION GENERATOR ON WHEN RISING EDGE EXCEEDS TRIGGER LEVEL AND OFF WHEN FALLING EDGE GOES BELOW TRIGGER LEVEL. LAST CYCLE STARTED IS COMPLETED.

Figure 3-5. Main Generator Basic Operation

VCG input, output frequency is increased. Example 3 shows that with a negative VCG input, output frequency is decreased. (Note that the Output Frequency Factor column value must be multiplied by a frequency range multiplier to give the actual output frequency.)

#### NOTE

Nonlinear operation may result when the VCG input voltage is excessive; that is when the attempted generator frequency exceeds the range limits. The upper limit is 2 times the multiplier setting, and the lower limit is 1/1000th of the upper limit.

FREQ/ START FREQ DIAL	VCG IN (VOLTS)	OUTPUT FREQUENCY
DIAL	(VOL13)	FACTOR*
2.0 🕂	-5 T	⊤ .002
1.8	-4+	E32
1.6	-4 - -3 - EXAMP	+ .4
1.4	-2-4	+ .6
1.2	-1+	8. +
1.0	0 + EXAM	IPLE 1 1.0
.8 + ``	+1+	+ 1.2
.6+	+2+	+ 1.4
.4 +	+2 + +3 + EXAM, +4 +	9, + 1.6
.2 +	+4 +	1.8
.002 ⊥	+5 丄	`

\*Must be multiplied by FREQ MULT switch setting

Figure 3-6. VCG Voltage-to-Frequency Nomograph

The up to 1000:1 VCG sweep of the generator frequencies available in each range results from a 5V excursion at the VCG IN connector. With the frequency dial set to 2.0 and the main generator VERNIER fully ccw, excursion between -5V and 0V at VCG IN provide the up to 1000:1 sweep within the set frequency range.

#### 3.2.4 Manual Auxiliary Generator Operation

The auxiliary generator, when used as a separate function generator, provides frequency and function selection in a continuous mode. (Bold numbers are keyed to figure 3-1. Refer to paragraph 3.1 for further explanations of controls and connectors.) The following steps demonstrate setting up of the auxiliary generator for continuous operation.

Step	Control/Connector	Setting
1	AUX GEN MODE 20	Set to AUX OUT only.
2	AUX OUT 3	Connect to circuit under test, source impedance is $600\Omega$ .
3	FREQ 22	Select desired frequency range.
4	VERNIER <b>22</b>	Set to desired frequency within the range.
5	FUNC <b>21</b>	Select desired wave- form.
6	WIDTH/ΔF/Δ <b>M 20</b>	Select desired amplitude at AUX OUT.
NOTE		

AUX SYNC OUT Connector—This rear panel BNC provides a TTL level pulse which lags the AUX OUT 3 triangle signal by 90°.

As an actual exercise in operation, perform steps 14 through 17 of table 2-1.

## 3.2.5 Voltage Controlled Auxiliary Generator Operation

Operation of the auxiliary generator in Voltage Controlled Generator (VCG) mode is very similar to VCG operation of the main generator (paragraph 3.2.3) except there is a maximum of 33:1 frequency change with a 0 to  $\pm$  5V change at the AUX VCG input BNC (a 1000:1 change is maximum for the main generator).

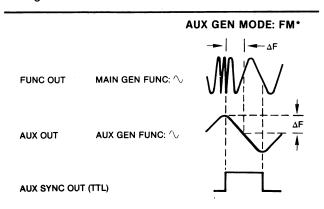
#### 3.2.6 FM Operation

In FM operation, the instantaneous frequency of the output signal varies with the instantaneous amplitude of the modulating signal (Figure 3-7). Both main and auxiliary generators may be frequency modulated by external signals at their VCG inputs (ref: paragraphs 3.2.3 and 3.2.5): Internal operation is the main generator modulated by the auxiliary generator. Internal FM operation is the same as a manually controlled main generator operation (ref: paragraph 3.2.2) except that, in addition, the main generator is frequency modulated by the auxiliary generator. The following steps demonstrate setting up the unique controls for FM modulation. (Bold numbers are keyed to figure 3-1. Refer to paragraph 3.1 for further explanations of controls and connectors.)

## Step Control/Connector Setting 1 AUX GEN Mode 20 Select SET FREQ/AUX GEN OFF.

2	FREQ/START FREQ Dial <b>1</b>	Set for FM center frequency.
	FREQ MULT <b>5</b> and VERNIER/SYM <b>5</b>	
3	AUX GEN MODE <b>20</b> WIDTH/ΔF/ΔM <b>20</b>	Select SET $\Delta F$ . Set for peak frequency deviation.
4	AUX GEN	
	MODE <b>20</b>	Select FM.
5	FUNC 21	Select modulation function.
6	AUX GEN FREQ, AUX GEN VERNIER <b>22</b>	Set modulation rate.

As an actual exercise in operation, perform steps 18 through 21 of table 2-1.



\*Similar waveforms apply when external signals are used to AM, FM and sweep the main generator. AUX OUT and AUX SYNC OUT signals are applicable to AUX GEN use only.

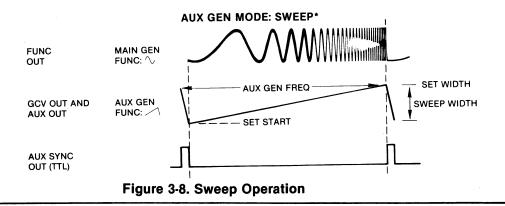
Figure 3-7. FM Operation

#### 3.2.7 Sweep Operation

Both the main generator and auxiliary generator may be swept with a ramp voltage applied to their respective VCG inputs (ref: paragraphs 3.2.3 and 3.2.5). Internal sweep is the main generator being swept by the auxiliary generator. Internal sweep operation is the same as manually controlled main generator operation (ref: paragraph 3.2.2), except that, in addition, the auxiliary generator sweeps the main generator frequency between two preset frequency limits (Figure 3-8). The following steps demonstrate setting up the controls for frequency sweep of the main generator. (Bold numbers are keyed to figure 3-1. Refer to paragraph 3.1 for further explanations of controls and connectors.)

Step	Control/Connector	Setting
1	AUX GEN MODE <b>20</b>	Select SET START.
2	START FREQ Dial 1, FREQ MULT 5 and VERNIER/SYM 5	Select start frequency.
3	AUX GEN MODE <b>20</b>	Select SET WIDTH.
	WIDTH/ΔF/ΔM <b>20</b>	Set maximum sweep (stop) frequency.
4	AUX GEN MODE <b>20</b>	Select SWEEP
5	FUNC 21	Select modulation function.
6	AUX GEN FREQ, AUX GEN/ VERNIER <b>22</b>	Set sweep rate.

As an actual exercise in operation, perform steps 22 through 26 of table 2-1.



#### 3.2.8 AM Operation

20 AUX GEN

MODE 20

In amplitude modulation, the instantaneous amplitude of the output signal varies with the instantaneous amplitude of the modulation signal figure 3-9. The main generator may be amplitude modulated by an extenal signal (with MODE set to AUX GEN OFF) applied to the AM IN BNC connector as well as internally modulated by the auxiliary generator. For internal AM operation, set up the generator as for continuous operation (refer to paragraph 3.2.2). The following steps demonstrate setting up the additional controls for AM Modulation. (Bold numbers are keyed to figure 3-1. Refer to paragraph 3.1 for explanations of controls and connectors.)

Step	Control/Connector	Setting
1	Main Generator FUNC <b>12</b>	Select AM. FUNC OUT decreases by 50% to prevent waveform clipping when amplitude modulating.
2	AUX GEN MODE <b>20</b>	Select SET CARRIER.
3	AM CARRIER LEVEL/NULL <b>19</b>	Set unmodulated output carrier level.
4	AUX GEN MODE <b>20</b>	Select ΔM.
5	AUX GEN MODE WIDTH/ΔF/ΔM	Select the valley of the modulation signal.

Select AM.

7 AUX GEN Sets amplitude modula-FREQ/VERNIER tion rate
22
8 AUX GEN Selects modulation func-FUNC 21 tion

As an actual exercise in operation, perform steps 27 through 30 of table 2-1.

#### 3.2.9 Suppressed Carrier Operation

Operation as a suppressed carrier modulated generator is the same as for AM operation (ref: paragraph 3.2.8), except the main generator carrier is suppressed (figure 3-9). The following steps demonstrate setting up the controls for suppressed carrier modulation. (Bold numbers are keyed to figure 3-1. Refer to paragraph 3.1 for explanations of controls and connectors.)

Step	Control/Connector	Setting
1	SUPPRESSED CARRIER 18	Depressed. Selects suppressed carrier operation.
2	AM CARRIER LEVEL/NULL <b>19</b>	Adjusts for equal balance of modulation envelope peaks. Balanced peaks ensures greatest carrier suppression.

As an actual exercise in operation, perform steps 27 through 31 of table 2-1.

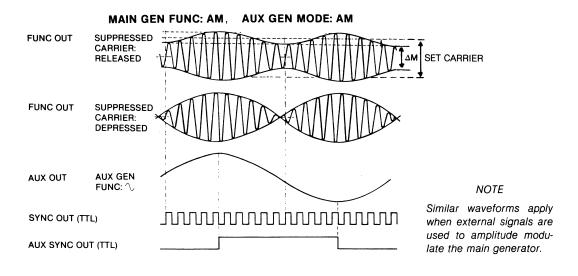


Figure 3-9. AM Operation

## SECTION 4 CIRCUIT DESCRIPTION

#### 4.1 INTRODUCTION

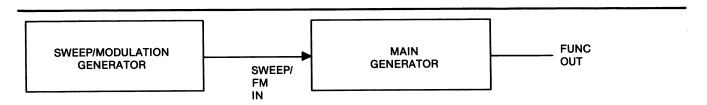
The Model 193 consists of two completely seperate, independent function generators: the main generator and sweep/modulation generator. This section describes the functional relationship between the two generators, and the function and relationship of the major circuit element of each generator.

While each generator can operate independently, they can be interconnected for frequency sweep, frequency modulation (FM), doubleside band amplitude modulation, and suppressed carrier amplitude modulation (refer to figure 4-1). For sweep and FM, the sweep/modulation generator varies the frequency of the main generator. In AM, both suppressed carrier and double side band, the main generator supplies a carrier signal that mixes, on the sweep/modulation generator board, with the modulating signal, created by the sweep/modulation generator, to produce the amplitude modulated signal which feeds back to the main generator output amplifier.

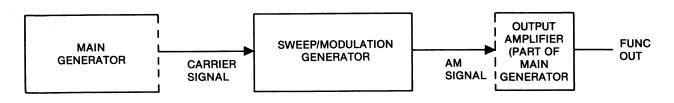
Figure 4-2 illustrates the Main Generator circuit elements and their functional relationships, which paragraphs 4-2 and 4-4 describe. Figure 4-3 shows the Sweep/Modulation Generator circuit elements and their functional relationships, which paragraph 4-3 and 4-5 describe.

## 4.2 MAIN GENERATOR BLOCK DIAGRAM ANALYSIS

As shown in figure 4-2, the VCG (Voltage Controlled Generator) sums the voltage inputs from the FREQ/START FREQ Dial, VCG IN, sweep/FM input and frequency vernier to provide a voltage control signal for the positive and negative current sources and the GCV amplifier. The positive and negative current sources generate precision currents, linearly related to the output of the VCG summing amplifier, which pass through the current switch to the timing capacitors. Additional linear currents are generated for loop dc delay compensation and the trigger baseline com-



#### SWEEP/FREQUENCY MODULATION OPERATION



AMPLITUDE MODULATION/SUPPRESSED CARRIER OPERATION

Figure 4-1. Model 193 Overall Block Diagram

pensation. The GCV amplifier provides a dc voltage proportional to the main generator frequency.

The current switch, controlled by the hysteresis output, causes either the positive current source or the negative current source to charge the timing capacitor selected by the frequency multiplier. When the positive current source is switched in, the charge on the timing capacitor will rise linearly producing the positive-going triangle slope. Likewise the negative current source produces the negative going triangle slope.

The triangle buffer amplifier is a unity gain amplifier whose output is fed to the hysteresis switch, sine converter and function switch. The hysteresis switch operates as a "window" comparator with limit points set to the triangle peaks. When the positive going ramp reaches + 1.0V, the hysteresis switch toggles to a low state causing the current switch to connect the negative current source. This causes the timing capacitor voltage to linearly ramp to -1.0V. As the timing capacitor voltage reaches -1.0V, the hysteresis switch toggles to a high state, switching in the positive current source. The generator loop continues to oscillate producing simultaneous triangle and square waves, at a frequency determined by the frequency multiplier and the magnitude of the timing current controlled by the sum of the dial setting, the VCG input, and the vernier.

Depressing the SYM button produces an unsymmetrical waveform and a division of the frequency by a factor of 10. The VERNIER/SYM control creates an imbalance in the current sources and therefore an imbalance in the waveform symmetry up to a ratio of 19:1. The result is variable duty cycle pulse, variable askewed sine wave and variable "sawtooth" triangle waves.

On the X1M and X10M ranges the dc loop delay compensation circuit compensates for delays in the generator loop. This circuit causes the hysteresis switch trip points to switch earlier in the cycle, and prevents the timing capacitors from charging beyond  $\pm 1.0V$ . The switch points are adjusted in proportion to the charging current, thus ensuring a constant amplitude as frequency is varied.

The capacitance multiplier is an active circuit which simulates capacitors up to 10,000 times larger than the timing capacitor, thus allowing very long charging times using physically small capacitors. This circuit is used in the four lowest frequency ranges.

The sine converter accepts a  $\pm 1.0$  volt triangle signal from the triangle buffer and converts it to a sine wave

current. The output is fed via the function switch to the preamplifier.

The trigger circuit allows precise single or multiple (gated) cycles at the output in response to external trigger signals or manual trigger operation. The trigger circuit operates by holding the timing capacitor at 0 volts, via the loop stop signal, on the positive going triangle ramp, until a trigger signal occurs. In the TRIG mode a single cycle is produced for each trigger signal above the variable trigger level threshold. In the GATED mode continuous cycles are generated for the time period at which the external signal is above the trigger level threshold plus the time for completion of the last partial cycle. The RUN signal causes the SYNC output to stay in the low state when the generator is quiescent. The TRG RST signal resets the trigger circuit and generator to the quiescent state on every generator cycle to arm it for the next trigger input. The trigger baseline compensation, LOOP STOP, circuit holds the generator output at zero volts, LOOP STOP, (within specified limits) during the quiescent intervals at any position (value) of the frequency dial, FREQ MULT, VCG IN, or VERNIER.

The sync circuit accepts the square wave signal from the hysteresis switch or zero crossing detector and converts it to a true  $50\Omega$  TTL level output. In square wave function (SYNC SELECT enabled) the sync is in phase with the output, but in triangle or sine functions (SYNC SELECT disabled), the zero crossing detector causes the sync output to be in phase withthe zero crossing of the output waveform.

When square is selected by the function switch, the square shaper accepts the signal from the hysteresis switch and converts it to a clean, fast square wave current which drives the preamplifier. In sine, triangle or DC functions, the square shaper input and output are disabled so as not to interfere with the selected waveform.

The preamplifier is fed from both the function switch and the square shaper. In all functions except AM, the preamplifier voltage output drives the output amplifier via the amplitude control. But when the AM function is selected, the voltage output (carrier signal) drives the amplitude modulator on the Sweep/Mod board.

The output amplifier receives input signals from a section of the function switch via the amplitude control and drives the output attenuator. DC offset is achieved by offsetting the output amplifier.

The output attenuator, fed directly from the output amplifier, provides up to 70 dB of attenuation to the selected waveform or DC offset. This signal is connected directly to the FUNC OUT BNC.

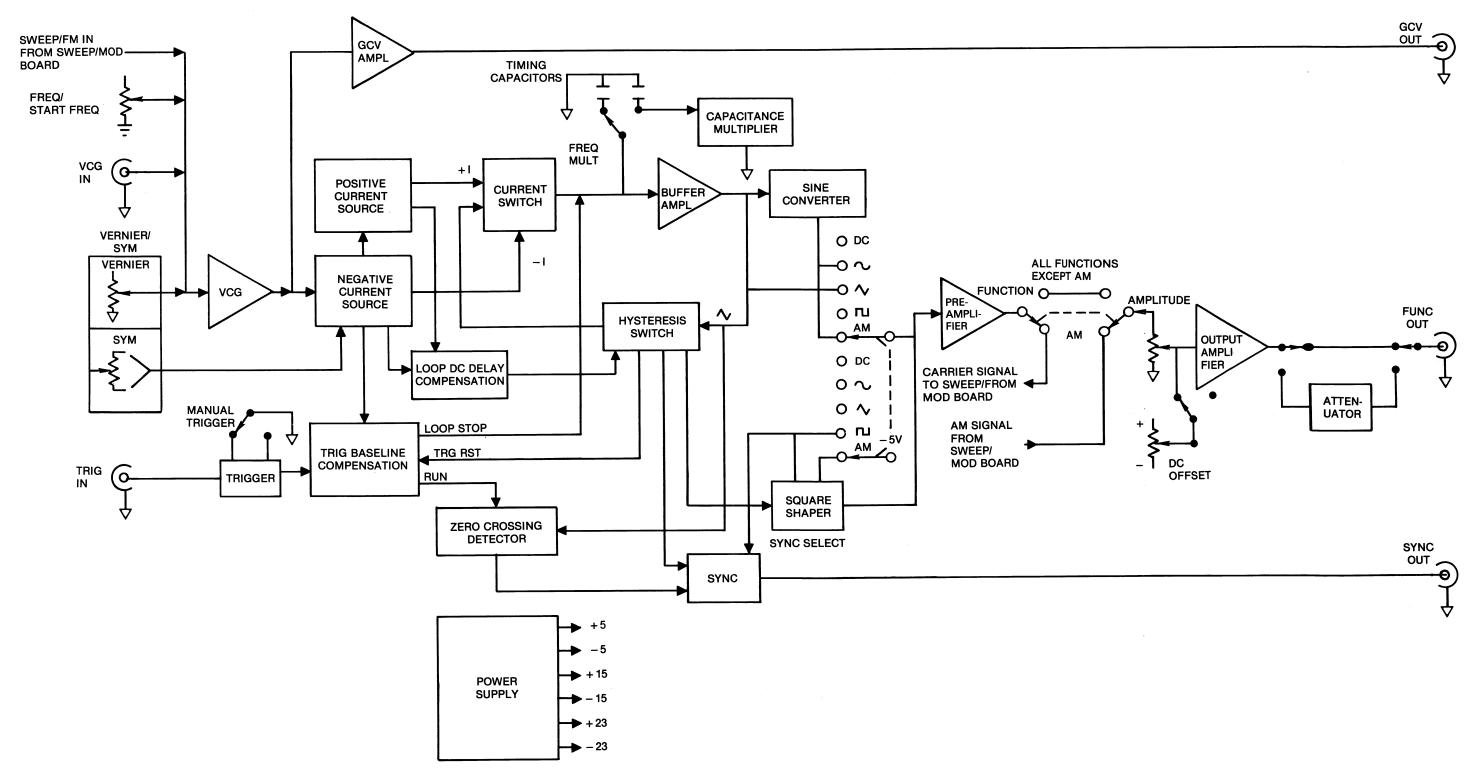


Figure 4-2. Main Generator Block Diagram

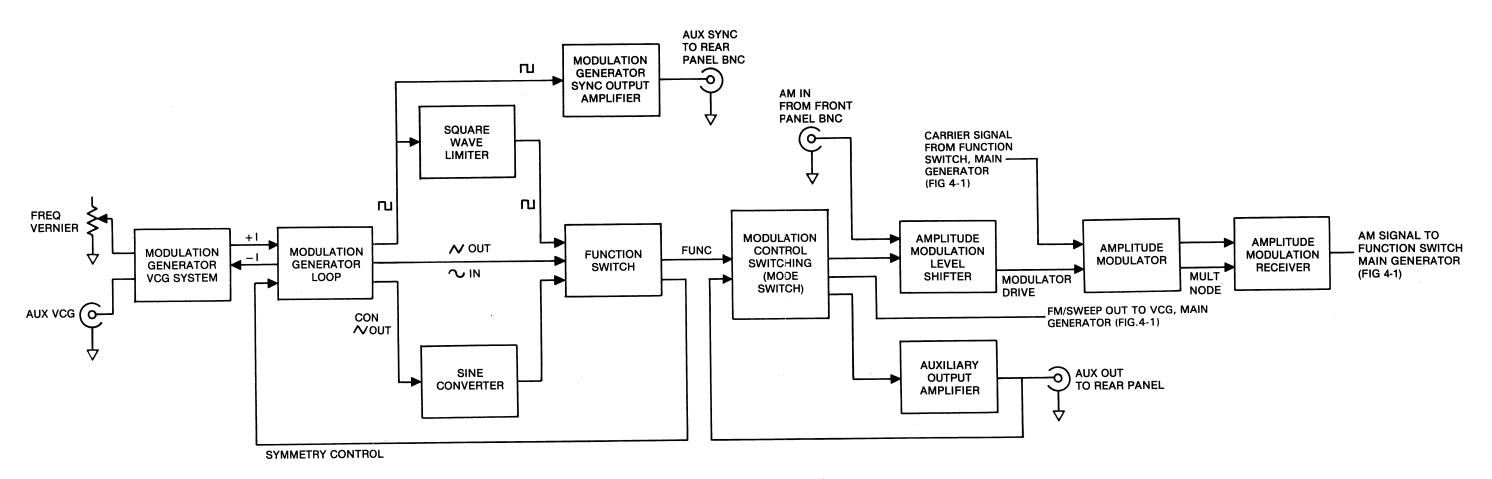


Figure 4-3. Sweep/Modulation Generator Block Diagram

### 4.3 SWEEP/MOD GENERATOR BLOCK DIAGRAM ANALYSIS

As shown in figure 4-3 the modulation generator VCG (voltage controlled generator) system sums voltage inputs from the FREQ VERNIER and AUX VCG input to provide a voltage control signal for the positive and negative current sources. The positive and negative current sources generate precision currents, linearly related to the VCG system output, which controls the modulation generator loop.

The modulation generator loop produces triangles and square waves whose frequency is related to the magnitude of the current from the modulation generator VCG system and the FREQ (Hz) range selected.

The sine converter accepts the 5 Vp-p triangle from the modulation generator loop and convertes it to a 5 Vp-p sine wave. The sine converter output is routed to the function switch. The square wave limiter receives a square wave input directly from the modulation generator loop and shapes the square wave to drive the function switch.

The function switch selects the sweep/modulation generator's output waveform, either a sine, triangle or square wave. Also, the function switch enables the SYMMETRY control when pulse or sawtooth function is selected.

The modulation control switching (sweep/generator's MODE switch) selects how the sweep/modulation generator will modulate the main generator. When internal sweep or FM is selected, the switching circuit routes the modulating waveform to the VCG of the generator board. Or, if AM is selected, the circuit routes the modulating waveform to the amplitude modulation level shifter.

The auxiliary output amplifier performs two functions. First, it provides a method of controlling the level (WIDTH/ $\Delta$ F/ $\Delta$ M) of the modulating signal, which is routed back through the modulation control switching to the amplitude modulation level shifter. Second, it supplies an output signal of the same frequency and function as the modulating signal to the AUX OUT. The Output level is proportional to the modulating signal.

The amplitude modulation level shifter sums inputs from the modulation control switching (internal AM) and AM IN (front panel). It shifts the dc level of the waveform and drives the amplitude modulator. The level shifter output level depends upon the suppressed carrier switch position: double side band up to 2.5 Vp-p (max.) offset between -2.2 and -3.7V; suppressed carrier up to 5Vp-p (max.) offset between -2.2 and -2.5V.

The amplitude modulator then mixes the waveform from the level shifter (Modulator Drive) and a sine wave (Carrier Signal from the main generator). The differential output from the modulator drives the amplitude modulation receiver, which converts the differential input to a single ended output. This signal is routed back to the output amplifier at the main generator.

The modulation generator sync output amplifier converts the square wave from the modulation generator loop to a TTL square wave (0 to +5V open circuit) that drives the AUX SYNC BNC.

## 4.4 DETAILED MAIN GENERATOR BOARD CIRCUIT DESCRIPTION

#### 4.4.1 VCG and Current Sources

Refer to the Generator Board Schematic sheet 4. The VCG IN (J7), Sweep/FM input (from Sweep/Mod board), and FREQ VERNIER (R88) are summed with the FREQ/START FREQ dial potentiometer (R56) at the summing node, U14 pin 6 of the VCG amplifier. Full scale on the dial causes a -5 volt control signal at the dial buffer output U14 pin 7. Rotating the dial to minimum, plus turning the FREQ VERNIER ccw produces -5 mV at U14 pin 7. The output of the buffer drives both the GCV buffer and current sources. The GCV output at U14 pin 1 is +5.0 volts at full scale.

The voltage output from U14 pin 7 is present at U13 pin 1. The output of U13 at pin 12 is fed through level shifting transistor Q14 to U8 pin 6. The collector current at pin 7 flows from ground through R81 and R80. As the voltage at U14 pin 7 varies, amplifier U13 and transistor Q14 adjust the base drive of U8 pin 6, and hence the collector current, until the voltage at U13 pin 2 equals the voltage at U13 pin 1. Because U8 is an array of matched transistors with the bases connected together, and all emitter resistors are equal with VERNIER selected, all collector currents are also equal.

The positive current source is controlled by a current control signal at U8 pin 1, which is held at 0 volts by the servo action of U13 pins 6, 7 and 10, level shifting transistor Q15 and U7 pins 6 and 7. The current "I" in R84 must flow through R93, and because these resistors are both  $1k\Omega$ , an equal but opposite base control voltage is present on U7 pin 6 compared to U8 pin 6. Because the transitors in U7 are matched and their bases are at the same point, a positive current "I" flows in R97 and hence the positive current source. A small amount of adjustable balance is provided by R95 and R94 to enable the positive and negative currents to be set for correct symmetry.

On the 1M and 10M ranges, the timing current is increased by approximately 25%, allowing the use of larger timing capacitors and hence, minimizing the effect of any stray capacitance. On the higher ranges, the parallel resistance across R83 (at ISCAL) is greater than the resistance on the lower ranges. This would decrease the current through U8 pin 8 were it not for the servo loop action of U13 pin 12, Q14 and U8 pins 6, 7 and 8. For any VCG setting at U14 pin 7 and U13 pin 1, no matter which range is selected, this servo loop maintains the voltage at U13 pin 2 equal to pin 1. Because the voltage at U13 pin 2 remains constant from range to range, the voltage across, and therefore the current through R80 and R81 also remains constant. This current also flows through U8 pins 7 and 8. To enable this current to remain constant, the servo loop drives the base voltage at U8 pin 6 in a positive direction. Because all of the bases in U8 are at the same point, the current relative to the lower ranges increases in R84 through R87 and also in the collectors of U8 pins 1, 14, 2, and 9.

Variable symmetry is controlled by R88 which doubles as the frequency vernier. With VERNIER selected, R88 functions as a frequency vernier with one end of the control connected to ground and the other connected to the - 15 volt supply. The wiper supplies current to the summing node U14 pin 6. Additionally, one end of 1kΩ resistors R84 through R87 are all connected to the - 15 volt supply. For any given dial setting, the current through each of the four resistors is "I". With SYM selected, R88 functions as a variable symmetry vernier with the wiper connected to the - 15 volt supply. One end of this vernier supplies current to R84 and R85, while the other end supplies current to R86 and R87. With the vernier centered, each leg is approximately  $5000\Omega$  and reduces the current through each of these 4 resistors to 1/10 I, dividing the generator frequency by 10. As the symmetry control is varied, emitter resistance in the positive and negative current sources change unequally, hence the current sources are unbalanced and the timing for the positive waveform is varied in respect to the negative waveform, resulting in variable symmetry.

Loop delay dc compensation currents (+ ICMP and - ICMP), are supplied by Q16 and U8 pin 9 and track the timing currents.

A current (ITRGBL), is supplied by U8 pin 14 to the trig baseline circuit to compensate for variations in dial settings when the generator is in a quiescent trigger or gated mode.

#### 4.4.2 Current Switch

Refer to sheet 3. The current switch is driven by the

square wave signal (ISWCTRL) from the hysteresis switch. Level shifting transistor Q10 provides a control signal for the diode bridge CR8, CR9, CR30 and CR31. When the control signal is +1.8 volts, CR30 is reversed biased, allowing CR8 to conduct current from the positive current source to the timing capacitor selected by SW9-D. This produces a positive going ramp. CR31 is also turned on, which reverse biases CR9 and prevents current sinking from the timing capacitor to the negative current source. When the control signal is -1.8 volts, both CR30 and CR9 are forward biased, while CR31 and CR8 are reversed biased. At this time, current from the negative current source sinks from the timing capacitor, producing a negative going ramp.

#### 4.4.3 Triangle Buffer Amplifier

Refer again to sheet 3 of the schematic. The signal on the selected timing capacitor is present at both the gate of Q11, and at U9 pin 2. These devices provide a very high input impedance for the signal to avoid leakage which would otherwise cause poor triangle linearity. The output current of Q11 controls the base drive to emitter follower Q13 and hence the output voltage on the emitter. This voltage is sensed at U9 pin 3, causing U9 to adjust the base voltage of Q12 until the differential input of U9 is zero. The low impedance source output voltage at the emitter follower Q13 now follows the high impedance input signal at the gate of Q11 with a circuit gain of unity.

#### 4.4.4 Hysteresis Switch

Refer to sheet 2. U10 pin 5 is the input to the positive peak comparator, while pin 10 is the input to the negative peak comparator. A level shifted triangle signal of -0.9 volts to -2.8 volts is present at pins 5 and 10 of U10. Assume a positive going ramp. R18 and R19 set the reference voltage on U10 pin 4 at -0.9 volts. When the voltage on pin 5 exceeds the reference voltage on pin 4, the positive comparator changes state and the voltage on pin 3 pulses from an ECL low (-1.8V) to an ECL high (-0.8V). This signal is connected to clear direct (pin 4) of D flip flop U5. The output of U5 pin 2 goes low, while U5 pin 3 goes high. These outputs toggle the differential pair Q7 and Q8 so that Q7 is on and Q8 is off. This causes the current switch control signal (ISWCTRL) to go low, which connects the negative current source to the timing capacitor, and causes the triangle to begin to ramp negative. The negative peak comparator functions in an identical manner to the positive comparator except that the reference voltage at U10 pin 9 is -2.8 volts. At the negative triangle peak, U10 pin 6 pulses high, causing a set direct at U5 pin 5, toggling the current

switch signal (ISWCTRL) high and producing a positive going ramp. In addition to being used to store the first peak comparison pulse from U10 pins 3 and 6, U5 also ignores "chatter" from both positive and negative comparators.

#### 4.4.5 Loop DC Delay Compensation

The circuit is also located on sheet 2 of the schematic diagram. The purpose of this circuit is to adjust the reference voltages on the comparators in the two highest frequency ranges so that the triangle peaks do not increase in amplitude due to loop delay. Q2 functions as a variable positive current source controlled by the range switch and the main current source. As the generator frequency is increased, the base voltage of Q2 progressively moves negative causing positive current through R15 and increasing the reference voltage on U10 pin 9 in a positive direction. This causes the negative peak to switch earlier in time, compensating for the loop delay and maintaining constant triangle amplitude and correct frequency tracking.

The positive peak comparator reference is changed in an identical way, except that the voltage on U10 pin 4 becomes more negative with increased frequency. Q4 is a variable negative current source. Q1 and Q3 function as temperature compensating diodes.

#### 4.4.6 Capacitance Multiplier

Refer to schematic diagram sheet 5. The capacitance multiplier is a precision current splitter which shunts up to 99.990% of the VCG current away from the integrating capacitor (C57) to produce the 100 through 0.1 frequency ranges. Timing current is divided between C57 and R114, then again between R113 and the selected timing resistor (R110 through R112 or R108).

The signal at U11 pins 2, 6, and 7 is a  $\pm 1.0$  volt triangle. U11 (pins 6, 7, and 10) is a non-inverting amplifier with a gain of 8. The waveform at U11 pin 1 is a  $\pm 1.0$  volt triangle with 0.5 volt spikes at each peak. At any given moment, the junction of R103 and C55 (differentiator circuit input) has 8 times the voltage as the junction of R104 and C55. This voltage difference causes a constant current to charge C55 through R104 and the selected timing resistor. Thus a frequency dependent charging current flows into the summing node of U11 pin 1, producing an inverted square wave component at the differentiator output U11 pin 12 sinking or sourcing current from the main current sources and limiting the amount of current available to charge C57. The ± 1.0 triangle at U11 pin 2 provides the triangle portion of the waveform at U11 pin 12. Since the triangle slopes on U11 pins 1 and 12

are identical, only the square wave component of the waveform at U11 pin 12 is across the timing resistor. The amount of current supplied to charge C55 is therefore this voltage divided by the range resistor value. As the range resistor is increased, the feedback for U11 between pins 1 and 12 is also increased, causing less current to charge C55 and increasing the amount of current being shunted to U11 pin 12 by a factor of 10 for each lower frequency range.

#### 4.4.7 Sine Converter

Refer to sheet 6 of the schematic. The sine converter converts the buffered  $\pm 1.0$  volt peak triangle to a sinusoidal current of 2mA peak. The input triangle voltage (TRIBUFC) passes through a voltage divider network to the input of the diode at pins 1, 4 and 6. As this signal progressively increases, the diode between pins 1 and 9 is progressively reversed biased, sinking less current and causing the diode between pins 2 and 5 to pass increasingly more current in a sinusoidal manner to IFUNC. This produces the positive half of the sine wave at the output of the preamplifier. At the same time, the diode between pins 2 and 8 is progressively reversed biased. This slows and eventually prevents current from flowing from the negative portion of the sine converter.

When the input waveform moves negatively, the diode between pins 2 and 5 is reversed biased and the diode between pins 2 and 8 progressively conducts, producing the negative half of the sine wave.

R159 sets the input amplitude for correct biasing of the sine conversion diodes, while R165 adjusts the input signal offset. Thermister R161 adjusts the input voltage to compensate for the diode voltage change with temperature. The network consisting of R166, R167 and C102 provides a signal (SINCMP) to the non-inverting input of the preamplifier to compensate for the effects of diode capacitance which would otherwise distort the sinewave peaks at high frequencies.

#### 4.4.8 Trigger Circuit

Refer to sheet 5. The trigger input at J8 is added to the voltage from the trigger level control R119 and compared at U12 pin 5 with a reference at U12 pin 4. When the signal at U12 pin 5 exceeds pin 4 by a few millivolts, U12 pin 3 goes high. R120 and C60 ensure a noise free pulse at U12 pin 3 which is one of two wire ORed inputs to U4 pin 7. The second input originates from the MAN TRIG switch circuit. When this switch is depressed, R115 pulls U12 pin 10 low. Pin 10 is compared to the Vbb reference voltage at pin 9, latching pin 6 high and preventing false triggering due to switch contact bounce. Pin 13 connected to pin 6, is

referenced to pin 12, causing pin 15 to also go high. When either U12 pin 3 or pin 15 go high, U4 pin 3 goes low because these outputs are wire ORed to U4 pin 7. U4 pin 3 is connected to pins 4 and 10. Because pin 10 was previously high, U4 pin 14 was low causing a low at U4 pin 5. The trigger pulse low at U4 pin 4 causes a 10 ns ECL high pulse at U4 pin 2. At the same time, U4 pin 14 goes high and after the time delay set by R126 and C62, U4 pin 5 also goes high. This causes U4 pin 2 to return low.

In the gate mode CR14 holds U4 pin 11 high, forcing pin 14 low. The length of the control pulse at U4 pin 2 is now equal to the period during which U4 pin 7 is held high. In the continuous mode, U4 pin 2 is held high by CR16 regardless of any input trigger signals.

#### 4.4.9 Trigger Baseline

Refer to sheet 5. In the trigger mode, with no trigger inputs, U5 pin 12 is held low. On the next positive going triangle, the trigger reset (TRIG RST) signal at U5 pin 11 causes U5 pin 14 to go high. This turns Q18 off and Q17 on, which turns off Q19. The Q19 emitter voltage is pulled down by the negative current sources Q20 and Q21, causing CR19 to conduct. Because the anode is at ground and CR18 is matched to CR19, the voltage at the anode of CR18 is also zero. This causes the triangle on the positive going ramp to stop at exactly zero volts. When a trigger signal occurs, U5 pin 12 goes high for about 10ns, causing pin 15 to also go high. This turns on Q18 and turns off Q17, which turns on Q19, causing the emitter to rise to about 1.7 volts. This reverse biases CR18 and CR19 causing the generator to run for exactly one cycle. In the gate mode, U5 pin 12 is held high for the duration of the input signal causing the generator to run for this interval plus the time required to complete a partial cycle.

In the trigger or gated mode, quiescent state, positive charging current I flows in CR18. As the VCG current is varied, I also varies, causing the voltage across CR18 to vary. To prevent this from causing a baseline shift, current (I) must also flow in the reference diode CR19. A negative current source (ITRGBL) is connected to the bases of Q20 and Q21. Negative current (-I) flows through the collector of Q20 and R133. Because of the configuration of Q20 and Q21, and because R133 and R134 are both  $1k\Omega$ , an equal amount of current - I also flows through the collector of Q21 and R134, causing -21 to flow at the junction of R133 and R134. Half of this current (-I) flows through CR19, while the remaining current flows through CR18. Therefore, the anode of CR18 is held at zero volts regardless of the VCG summing node current.

The RUN signal is used to hold the sync output low during guiescent periods.

#### 4.4.10 Sync

Refer to schematic sheet 2. The SYNC OUT amplifier is driven from the signal at U6 pin 10 in the triangle and sine functions, and from U6 pin 7 when the function switch is in the square function. These two inputs are wire ORed at U6 pin 13.

In the triangle and sine functions, SYNC SEL allows R23 to pull CR4 high causing a low at U6 pin 2. This enables the signal from the zero crossing detector output (U10 pin 15), and disables the hysteresis switch input at U6 pin 7. When the positive going ramp crosses 0 volts at the zero crossing detector input U10 pin 13, U10 pin 15 and U6 pin 10 go high. This causes a low at U6 pins 14 and 13. U6 pin 9 goes low and pin 15 goes high turning on Q5 and turning off Q6. This results in a high at SYNC OUT. As the triangle at U10 pin 13 crosses 0 volts in a negative direction, pin 15 goes low, causing Q6 to be turned on, producing a low at SYNC OUT. Therefore the SYNC OUT always toggles when the triangle crosses 0 volts.

When the square wave function is selected, CR4 pulls U6 pins 4 and 6 low. U6 pins 2 and 11 now go high, disabling the zero crossing detector input from pin 10, and enabling the square wave input from U6 pin 7. U5 pin 3 now drives the SYNC OUT connector in a similar manner as U10 pin 15. The SYNC OUT is in phase with the square wave output.

R26, a 49.9 $\Omega$  resistor sets the 50 $\Omega$  output impedance.

#### 4.4.11 Square Shaper

The square shaper schematic is located on sheet 6. In square function, CR20 pulls U4 pin 13 low, enabling the hysteresis switch input (HYS) at U4 pin 12. A low at U4 pin 12 causes a low at U4 pin 15 and a high at pin 9. Q22 turns on while Q23 is turned off, producing a +1.2 volt high at the bases of the current switch control transistors Q24 and Q25. Transistor Q25 is on, reverse biasing CR23. Transistor Q24 is off allowing positive current to flow through R147, CR22, R154 and into the preamplifier node via R152.

When  $\overline{\text{HYS}}$  toggles high, Q23 turns on forcing the bases of Q24 and Q25 to -1.2 volts. Q24 turns on and Q25 turns off, allowing negative current to flow through R157, CR23, R154 and the the amplifier node via R152.

R152 and R154 form a current divider to obtain a 2mA full scale current into the preamplifier. Overshoot caused by diode capacitance is reduced by R153 and C73. The output of the square shaper is disabled in all

other functions by turning on Q26 and CR24 which reverse bias CR22 and CR23 and prevents current from flowing through R152.

#### 4.4.12 Preamplifier

Refer to sheet 7 of the schematic circuitry. For all functions, full scale output voltage is produced when 2mA is injected into the input summing node U1 pin 8. Transistor array U1 forms a cascaded differential stage. Transistor Q27 is a fixed current source. Q28 and Q29 form a high gain voltage follower. DC negative feedback is applied through R195 to U1 pin 8. The closed loop voltage gain of the amplifier is determined by the ratio of R195 to the input resistors, R152 for square wave and R176 for triangle. The sine converter output supplies the correct current directly from U3 pin 2 to U1 pin 8. The servo action of the preamplifier holds this point at 0 volts, therefore no voltage can be measured. U1 pin 4 is the non-inverting input and is used both to adjust the offset to 0 volts at TP2 using R185 and to inject the sine converter compensation signal (SINCMP) described under paragraph 4.3.7. Sine Converter. High frequency compensation is provided by R182, C81, C86 and C153. Zener diode CR29 provides increased collector voltages for U1 pins 11 and 12 and also allows these two points to be relatively close in voltage.

#### 4.4.13 Output Amplifier

The output amplifier consists of an ac coupled amplifier for signals above about 16 kHz, and a dc coupled amplifier for signals below about 16 kHz, and to maintain zero dc output offset within specified limits. Refer to the simplified output amplifier schematic, figure 4-4.

Assume zero input voltage at the junction of R203 and R218. The output at R222 and R224 is maintained at 0 volts by dc amplifier U2. U2 pin 3 is connected to a 0 volt reference. If the output drifts away from 0 volts, this will be sensed at U2 pin 2 through R256, R257 and R254. Amplifier U2 will sense a difference between its inputs and produce an output voltage which adjusts the bias in the ac coupled amplifier to return the output to 0 volts. Because R218 and R223 form half of a balanced bridge, and R253, R256 and R257 form the second half, the amplifier node at the junction of R218 and R223 will be held at 0 volts as U2 has returned the junction of R253 and R256 to 0 volts.

A dc input of +1 volt at R218 and R253 is sensed as a positive increase at U2 pin 2, causing U2 pin 6 to go negative. The ac amplifier output goes negative in response to the dc control input. This continues until the output becomes sufficiently negative to sink all the input current, and return U2 pin 2 to 0 volts. The bridge circuit causes the ac amplifier node to be 0 volts. If the input is +1 volt and the node at the junction of R218 and R223 is 0 volts then the input current is 8.26 mA which flows through R223 to produce an output voltage of -16.52V. Therefore the amplifier voltage gain is 16.52 (R223/R218).

Above about 16 kHz, the ac amplifier controls the summing node directly, sinking or sourcing current through R223 by adjusting the output voltage to hold the node at 0 volts. The ac amplifier gain is also R223/R218 = 16.25. This is divided by 2 at the output terminal, due to the  $50\Omega$  source impedance resistors R222 and R224, providing the output is also terminated into  $50\Omega$ .

Refer to sheet 7. The top half of the circuit amplifies

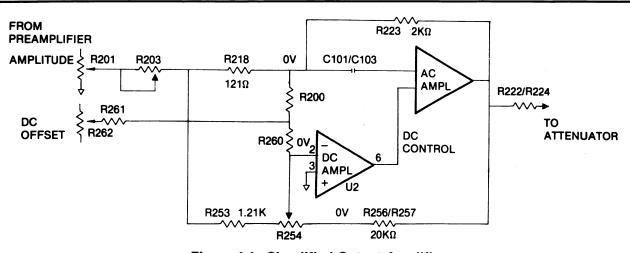


Figure 4-4. Simplified Output Amplifier

the positive portion of the signal, and the bottom half amplifies the negative part. Q30 and Q31 form an ac gain stage. An emitter follower stage is formed by Q32 and Q33, to provide a low impedance drive to the second voltage gain stage Q36 and Q39. This stage drives the parallel output emitter followers Q37 and Q38 on the positive side, and Q40 and Q41 on the negative. Diodes CR23 and CR26 set thermally stable bias for the output transistors. Networks R211, R212, C94 and C93 bypass emitter resistor R208, while R245, R246, C107 and C106 bypass R244. As frequency is increased, these components decrease the local negative feedback in the driver stage, increasing the high frequency gain. Voltage regulators VR5 and VR6 have external current limiting circuitry set to limit at about 220 mA to prevent damage in the event of a shorted transistor. When the offset button is depressed, offset current is injected directly into both nodes in proportion to the feedback resistor values. The amplifier responds exactly as described above for a dc input.

#### 4.4.14 Output Attenuator

Also refer to sheet 7. Each attenuator button selects an independent voltage divider, which has  $50\Omega$  input and output impedances to correctly load the amplifier and to provide a constant  $50\Omega$  impedance at the FUNC OUT terminal.

The 10dB attenuator has a 3.16/1 voltage division ratio. The 20dB attenuator has a 10/1 voltage division ratio, and the 40dB stage has a 100/1 ratio. These ratios multiply in voltage. For example if the 20dB and 40dB buttons are depressed, the voltage division ratio is 1000/1. The attenuators add algebraically in dB, therefore any attenuation from 10 to 70dB may be selected in 10dB steps.

### 4.5 SWEEP MODULATION BOARD DETAILED CIRCUIT DESCRIPTION

#### 4.5.1 Modulation Generator VCG System

Refer to Sweep/Mod board schematic sheet 1. The modulation generator VCG system consists of an input buffer, the positive current source, and negative current source.

#### 4.5.1.1 Input Buffer

The input buffer U1 pin 7 sums inputs from the AUX VCG (J16) and FREQ VERNIER. The buffer drives both positive current source inverter U1 pin 2 and the negative current source U1 pin 9. With FREQ VERNIER full cw and no AUX VCG input, the buffer supplies -0.546V, and -0.014V with FREQ VERNIER full ccw.

#### 4.5.1.2 Positive Current Source

U1 pin 1 is an amplifier with a gain of -1 that drives the positive current source U1 pin 13.

The positive current source produces a current that charges the modulation generator loop timing capacitor. The positive current source receives its input at U1 pin 13. The output U1 pin 14 is dc coupled by a zener diode CR1 and resistor R13, which shifts the dc bias, to the two bases U2 pins 11 and 14. The base bias U1 pins 11 and 14 controls the emitter currents U1 pins 10 and 13. Emitter resistors R14 and R16 are matched and supplied by the same + 15V supply, thus providing equal emitter currents and, consequently, equal collector currents. The collector current U2 pin 12 flows through resistor R17 and the function switch. to the negative current source. The collector U2 pin 12 connects to U1 pin 12, this provides the feedback to stabilize the positive current source. The collector current at U2 pin 15, which equals the collector current U2 pin 12, flows to the current switch in the modulation generator loop (see paragraph 4.5.2.1). The RC network R12 and C1, across U1 pins 12 and 13, provide loop stabilization.

#### 4.5.1.3 Negative Current Source

The negative source is similar to the positive source, except it supplies the current that discharges the modulation generator loop's timing capacitor. The negative current source receives its input at U1 pin 9 from the input buffer U1 pin 7. The amplifier output U1 pin 8 drives the two transistor bases U2 pins 5 and 8 via the dc bias shifter CR2 and R23. As with the positive current source, the emitter resistors R24 and R26 are matched and biased to the -15V supply; therefore, the collector currents U2 pins 6 and 9 are equal. U2 pin 9 drives the current switch in the modulation generator loop (refer to paragraph 4.5.2). The RC network, C2 and R22, across U1 pins 9 and 10 provide loop stabilization.

#### 4.5.1.4 Symmetry Control

When the function switch is set to a symmetrical waveform: sine, triangle or square, the symmetry control R19 is shorted out. This allows the current to flow directly from the positive to the negative current source.

But if the ramp or pulse is selected, SW2-B connects the wiper of the SYMMETRY potentiometer to circuit ground, and SW2-C opens the short across R19. This increases the total resistance in the string R17, R19, R21 by ten times, thus reducing the current and frequency to 1/10th. As the SYMMETRY potentiometer is rotated from one end to the other, one current

increases as the other decreases; thus changing the slope of each half cycle over a 19:1 range without changing the frequency.

### 4.5.2 Modulation Generator Loop

Refer to Sweep/Mod board schematic sheet 1. The modulation generator loop consist of the current switch, timing capacitors and triangle buffer, and hysteresis switch.

#### 4.5.2.1 Current Switch

The current switch selects which current source will charge or discharge the timing capacitor. The hysteresis switch U13 pin 6 controls the current switch.

When the hysteresis output is high (+3V), diode U3 pins 4 and 9 conducts current, Q2 is reverse biased, and the positive current source charges the timing capacitor. In addition, transistor Q1 is forward biased, thus reverse biasing diode U3 pins 3 and 4, this sinks the negative current source to the +15V supply.

When the hysteresis output is low (-3V), diode U3 pins 3 and 4 is forward biased, this reverse biases transistor Q1, the current now flows from the timing capacitor through the diode U3 pin 3 and 4 to the negative current source. Meanwhile, transistor Q2 conducts current, which reverse biases diode U3 pins 4 and 9, thus the -15V supply sinks the current from the positive current source.

### 4.5.2.2 Timing Capacitor and Triangle Buffer

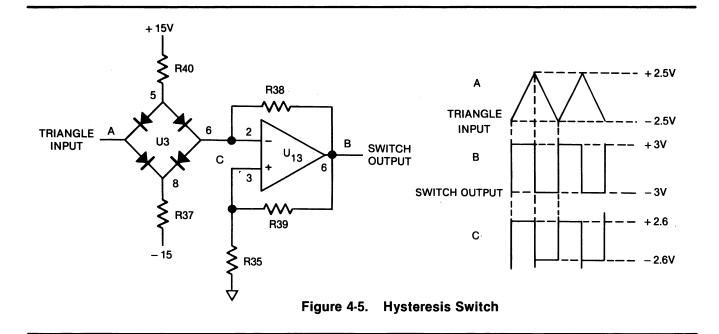
The timing capacitor C5 and C6 together with the currents from the current sources determine the frequency of the modulation generator loop. On two ranges 3-100 and 3K-100K, the current sources directly charge and discharge the timing capacitors. But, on the .1-3 and 100-3K ranges, resistors R46 and R47 acts as a capacitance multiplier. R46 connects between the current switch and the timing capacitor, while R47 provides a shunt to 97% of the current, thus reducing the current at the timing capacitor to 3% of the current from the current sources.

The triangle buffer U12 pin 6, a unity gain non inverting amplifier, follows the voltage across the timing capacitor.

### 4.5.2.3 Hysteresis Switch

Refer to sweep/mod board schematic sheet 1 and figure 4-5.

The hysteresis switch U3 and U13 operates as a comparator. When the triangle input reaches the positive limit (+2.5V) the switch output toggles to -3V. And, when the triangle reaches the negative limit (-2.5V) the switch output toggles to +3V. The hysteresis switch output drives the current switch, square wave limiter, and modulation generator sync output amplifier. Figure 4-5 shows a simplified schematic and timing diagram.



During the triangle's positive-going ramp, the hysteresis switch output is +3V. Current flows from the +15V supply, through R40 and U3 pins 5 and 2 to the triangle buffer output U12 pin 6. At the same time, current flows from the hysteresis switch output U13 pin 6 through R38, U3 pins 6 and 8, and R37 to the -15V supply. Diodes U3 pins 5, 6 and 2, 8 are reverse biased. Resistor R38 provides negative feedback, while resistors R39 and R35 provide positive feedback.

As the triangle reaches the positive limit +2.5V, the diodes U3 pins 2, 8 and 5, 6 begin conduction. This also decreases the current through diodes U3 pins 2, 5 and 6, 8. This causes a positive increase in the voltage level at U13 pin 2. As U13 pin 2 becomes slightly more positive, with respect to pin 3, U13 pin 6 will instantaneously toggle to -3V.

During the triangles negative-going ramp, the current through diodes U3 pins 2, 8 and 5, 6 increases, while the current through diodes U3 pins 2, 5 and 6, 8 decreases and, eventually, cutoff.

As the triangle reaches the negative limit, -2.5V, the diodes U3 pins 6, 8 and 2, 5 begin conduction. This decreases the current through U3 pins 5, 6 and 2, 8; thus forcing U3 pin 2 slightly more negative, with respect to pin 3, which causes the output pin 6 to toggle to +3V.

## 4.5.3 Square Wave Limiter

Refer to Sweep/Mod board schematic sheet 1. The square wave limiter, which receives its input from the hysteresis switch, limits the square wave peaks to  $\pm 2.8V$  and routes the square wave to the function switch.

When the hysteresis switch output U13 pin 6 goes high (+3V), diodes CR4 and CR7 are foward biased, and diodes CR5 and CR6 are reverse biased. This allows current flow from the +15V supply, through R29, CR4 and R27.

The diodes CR5 and CR6 are forward biased when the hysteresis switch output is low (-3V), also diodes CR4 and CR7 are reverse biased. Now current flows from circuit ground through R27, CR6, and R28 to -15V supply.

#### 4.5.4 Sine Converter

Refer to Sweep/Mod board schematic sheet 2. The sine converter circuit converts a buffered  $\pm 2.5$ V triangle to a  $\pm 2.5$ V sine waveform. The sine converter uses the nonlinear characteristics of the diode to convert a linear triangle slope to a sinusoidal curve. The sine converter consists of a diode network, buffer

amplifier, and temperature compensated  $\pm 10V$  supplies. The triangle from the modulation generator loop is reduced in amplitude by resistors R41 and R42 and feeds the diode network U4 pins 3, 9 and 1.

The following section illustrates the operation of the diode network when shaping a sine wave.

At the negative peak of the input triangle the diodes U4 pins 3, 4 and 5, 6 are forward biased, which reverse biases U4 pins 2, 5. Also, diodes U4 pins 1, 9 and 6, 8 are cut off, which forward biases U4 pins 2, 8. When the input triangle slope increases from negative to positive peak, diodes U4 pins 1, 9 and 6, 8 begin to conduct, which decreases the current through U4 pins 2, 8. At the same time, diodes U4 pins 3, 4 and 5, 6 decrease conduction, which causes diode U4 pins 2, 5 to begin conduction. When the triangle reaches the positive peak, diode U4 pins 2, 5 is forward biased and U4 pins 2, 8 is cutoff. The current from the diode network U4 pin 2 drives the sine buffer U6.

Shaping the negative-going slope of the sine wave is similar to the positive-going slope. Except, when the input triangle ramps positive to negative, diodes U4 pins 1, 9 and 6, 8 decrease conduction and diodes U4 pins 3, 4 and 5, 6 increase conduction. This, eventually, forward biases diode U4 pin 2, 8 and reverse biases diode U4 pins 2, 5.

The sine buffer U6 pin 6 sums the currents supplied by the diode network to produce a 5Vp-p sine wave that drives the function switch.

The bias for the diode network is supplied by temperature compensated  $\pm\,10\text{V}$  supplies. The  $+\,10\text{V}$  supply is amplifier, U5 pin 1, with a gain of  $-\,2/3$ , which is referenced to the  $-\,15\text{V}$  supply. The  $-\,10\text{V}$  supply is amplifier U5 pin 7, with a gain of  $-\,1$ , which is referenced to the  $+\,10\text{V}$  supply. A temperature sensing device, RT1, adjusts the gain of the  $+\,10\text{V}$  supply, which gives the  $\pm\,10\text{V}$  supplies a negative temperature coefficient.

### 4.5.5 Function Switch

Refer to Sweep/Mod board schematic sheet 1. The function switch selects the function (waveform) of the sweep/mod generator. Wafer SW2-A routes the selected function, (sine, triangle or square) to SW3-D of the Modulation Contol switching. Wafers SW2-B and C control the current source when a pulse or sawtooth function is selected.

### 4.5.6 Modulation Control Switching

Refer to sweep/mod board schematic sheet 3. The modulation control switching, which selects the sweep/modulation generator operating mode, con-

sists of the MODE and SUPPRESSED CARRIER switches.

The MODE switch is a four wafer switch. Wafer SW3-A in AM and SET  $\Delta M$  routes either a modulating signal (AM) or a dc level (SET  $\Delta$ M) from the auxiliary output amplifier to the AM level shifter. Wafer SW3-B, when MODE is set to SWEEP, SET WIDTH, SET START, FM, SET  $\Delta F$ , and SET FREQ, connects either a modulating signal (SWEEP and FM) or a dc level (SET WIDTH, SET START, SET  $\Delta F$ , and SET FREQ) from the auxiliary output amplifier to the SWEEP/FM IN at the main generator VCG. Wafer SW3-C controls bias amplifiers U10 pins 1 and 7, part of the auxiliary output amplifier; table 4-1 shows the bias amplifiers output levels relative to MODE switch settings. Wafer SW3-D controls the input conditions, via the WIDTH/ΔF/ΔM control, for the auxiliary output amplifier: for AUX OUT ONLY and AM SW3-D routes the waveform (FUNC) to one end (E5) of the WIDTH/ $\Delta$ F/ $\Delta$ M control; for SET  $\Delta M$ , SET WIDTH, and SET  $\Delta F$  SW3-D grounds one end (E5) of the WIDTH/ΔF/ΔM control; and for SET CAR-RIER. SET START, and SET  $\Delta$ F SW3-D shorts out the WIDTH/ΔF/ΔM control.

The SUPPRESSED CARRIER controls the bias and gain of the amplitude modulation level shifter. When SUPPRESSED CARRIER is pressed, one section of SW4 shorts out R118, this increases the gain of the amplitude modulation level shifter, and removes R126, which was in parallel with R86. Also, the other section now supplies a  $\pm$ 15V bias when SET  $\Delta M$  is selected.

## 4.5.7 Amplitude Modulation Level Shifter

The amplitude modulation level shifter sums inputs from the modulation control switching (dc reference levels or ac waveforms), suppressed carrier switch SW4 (dc reference level), and the AM IN BNC J14 (an external signal). This circuit shifts the dc level, depending upon the position of the SUPPRESSED CARRIER switch, that drives the Amplitude Modulator.

### 4.5.8 Auxiliary Output Amplifier

Refer to sweep/mod board schematic sheet 3. The auxiliary output amplifier consists of a single amplifier U9. The amplifier input signal originates at the sine converter, square wave limiter, or modulation generator loop (triangle) and is selected by the function switch to be routed through the Mode switch of the modulation control switching to the WIDTH/ $\Delta$ F/ $\Delta$ M control. The WIDTH/ $\Delta$ F/ $\Delta$ M control varies the level of the auxiliary output signal. The output from U10 pin 7 provides dc offset to the inverting input of the auxiliary

Table 4-1. Bias Amplifier Levels Vs Modes

	Suppressed	Bias Amp	lifier Level
Mode Switch	Carrier Switch	U10 Pin 1	U10 Pin 7
SET FREQ AUX GEN OFF	No affect	+ 5V	+ 10V
SET $\Delta F$	No affect	- 2.5V	- 5V
FM	No affect	0V	OV
SET START	No affect	oV	0V
SET WIDTH	No affect	- 4.8V	- 9.6V
SWEEP	No affect	- 2.5V	- 5V
SET CARRIER	No affect	0V	0V
SET ΔM	Off On	+ 2.5V - 2.5V	+ 5V - 5V
AM	No affect	ov	0V
AUX OUT ONLY	No affect	ov	0V

output amplifier U9 pin 2; the mode switch SW3-C controls the dc offset. Refer to the table on the schematic sheet 3. The auxiliary output signal feeds both the AUX OUT BNC and MODE switch SW3-A and B for sweep, frequency or amplitude modulations.

If the MODE switch is set to SET FREQ, SET START, or SET CARRIER, both ends of the WIDTH/ $\Delta$ F/ $\Delta$ M are shorted together, R108 has no affect. This biases the Auxiliary output amplifiers output to a level dependent upon the Mode SW3-C selected: 0V for SET CARRIER and SET START, and +5V for SET FREQ. This allows setting of front panel controls for the generator board.

If the mode switch is set to SET  $\Delta$ F, SET WIDTH, or SET  $\Delta$ M, a dc level, referenced to ground is connected to the WIDTH/ $\Delta$ F/ $\Delta$ M control. This level represents a dc equivalent of the modulating signal and allows static set up conditions of the main generator. The maximum dc level applied to the WIDTH/ $\Delta$ F/ $\Delta$ M control depends on the mode selected and SUPPRESSED CARRIER switch position, refer to table 4-1.

When AUX GEN OFF is selected, the output from U10 pin 7 (LOOP STOP) shuts off the current source in the modulation generator VCG system.

### 4.5.9 Amplitude Modulator

Refer to sweep/mod board schematic sheet 2. The

amplitude modulator U7 is a wide-band four-quadrant multiplier that is configured as an amplitude modulator. The amplitude modulator produces an output current that is the linear product of the CARRIER SIGNAL U7 pin 9 and MODULATOR DRIVE U7 pin 8.

The CARRIER SIGNAL, originating at the main generator preamplifier (FUNC:AM only), drives the non-inverting X input U7 pin 9 with a fixed  $\pm$  0.2V sine wave. Resistors R73, R74, R75, and R76 bias the inverting X input U7 pin 12 at 0Vdc, this provides the reference voltage for the CARRIER SIGNAL.

The MODULATION DRIVE signal from the amplitude modulation level shifter, controls the inverting Y input. This signal level depends upon the type of AM selected: double-sideband modulation -2.5 to -5.0V (max) and suppressed carrier modulation 0 to -5.0V (max). The MODULATION DRIVE signal is referenced to -2.5V at the non-inverting Y input U7 pin 4.

The modulator outputs U7 pins 2 and 14 each receive 4 ma current from the amplitude modulation receiver. The modulator differentially varies these currents up to  $\pm 2$  ma.

### 4.5.10 Amplitude Modulation Receiver

Refer to sweep/mod board schematic sheet 3 and figure 4-6. The amplitude modulation receiver converts the differential current from the amplitude modulator to a single-ended output signal, which drives the output amplifier of the main generator. The receiver consists of two amplifiers (a bridge amplifier and output amplifier) and a resistor bridge.

To best understand this circuit refer to the simplified schematic figure 4-6. Basically, resistors R92, R93, R101, and R103 form a resistor bridge. When the currents I6 and I7 from the modulator outputs U7 pins 14 and 2 are equal, the bridge is balanced and the output E2 is 0V.

The amplitude modulator differentially varies the currents I6 and I7. For example, when I7 increases and I6 decreases, both an equal but opposite amount, current I2 increases, while current I3 remains unchanged. This increases the voltage E1. As a result, current I4 increases, but the modulator demands less current I6, therefore, the current I5 flowing through R103 equals the difference between I6 and I7. Thus, E2 swings negative.

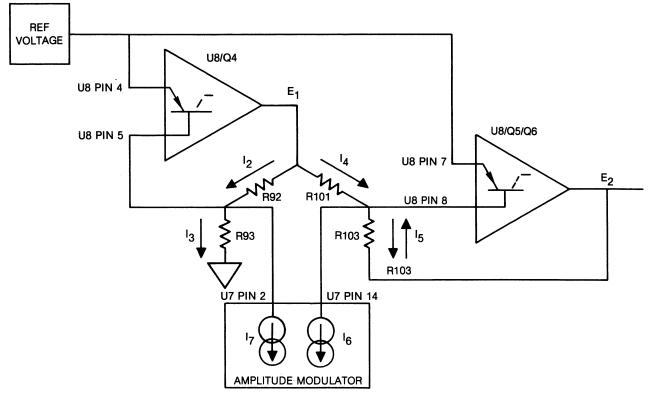


Figure 4-6. Amplitude Modulation Receiver

For the opposite polarity, the modulator demands less current I7 at U7 pin 2 and more current I6 at U7 pin 14. This decreased current I7 decreases the voltage E1, which decreases the current I4. But now, the modulator output U7 pin 14 demands more current I6 all of which cannot be supplied by current I4 and the difference in current must be supplied by I5. Output voltage E2 now swings positive.

The bridge amplifier (sections of U8 and Q4) supplies the drive voltage E1 for the resistor bridge. It supplies all the current for the R92, R93 legs of the bridge, and part of current for the R101, R103 leg of the bridge.

The output amplifier (sections of U8 and Q5/Q6) supplies the receiver output signal E2. This voltage E2 serves two functions; it supplies the AM signal which drives the main generator output amplifier, and provides a current source or sink for a portion of the resistor bridge.

Both amplifiers are tied to a common reference voltage. Resistors R95 and R100 bias transistors U8

pin 14 and U8 pin 11 (see sweep/mod schematic sheet 3) which serve as active loads for the input transistors of the amplifiers. These active loads set the emitter currents from U8 pins 4 and 7. The emitter currents collectively pass through R129 to ground. The voltage drop across R129 provides the reference voltage for the amplifiers.

## 4.5.11 Modulation Generator Sync Output Amplifier

Refer to sweep/mod board schematic sheet 1. The modulation generator sync output amplifier, driven by the hysteresis switch of the modulation generator loop, provides a TTL level output at AUX SYNC OUT J17. The amplifier Q3 is a transistor switch that is turned off and on by the driving square wave. This causes the collector of Q3 to switch between +5 and OV (no load).

### **5.1 FACTORY REPAIR**

Wavetek maintains a factory department for those cusstomers not possessing the necessary personnel or test equipment to maintain the instrument. If an instrument is returned to the factory for calibration or repair, a detailed description of the specific problem should be attached to minimize turnaround time.

### 5.2 REQUIRED TEST EQUIPMENT

Voltmeter	. Millivolt dc measurement
	(1% accuracy), true rms
Oscilloscope, Dual Chann	el100 MHz bandwidth
Counter	20 MHz (0.01% accuracy)
50Ω	± 1.0% accuracy, 2W
Distortion Analyzer	To 200 kHz
RG58U Coax Cable 3 f	length BNC male contacts
Spectrum Analyzer	To 20 MHz

### 5.3 COVER REMOVAL

### NOTE

Before removing the covers, disconnect the instrument from the ac power source.

Invert the instrument and remove the four screws in the bottom cover. Remove the bottom cover.

### NOTE

Remove the cover only when it is necessary to make adjustments or measurements.

### 5.4 CALIBRATION

After referring to the following preliminary data, perform calibration, as necessary, per table 5-1, 5-2. If performing partial calibration, check previous settings and adjustments for applicability. Figure 5-1 shows generator

board calibration points while figure 5-2 shows sweep/mod board calibration points.

### NOTE

The completion of the calibration procedure returns the instrument to correct alignment.

## CALIBRATION LIMITS AND TOLERANCES ARE NOT INSTRUMENT SPECIFICATIONS

Instrument specifications are given in Section 1 of this manual.

1. All measurements made at the FUNCTION OUT connector must be terminated into a  $50\Omega(\pm 1.0\%)$  load.

### **WARNING**

With the covers removed, dangerous voltage points may be exposed. Contact with any of these points could cause serious injury or death.

2. Start the calibration by removing covers as described in paragraph 5.3, connecting the unit to an ac source and setting these front panel switches as follows:

SYM	Off (extended)
TRIG LEVEL	12 o'clock
DC OFFSET	Off (extended)
OUTPUT ATTEN	0 dB (all extended)
AUX GEN MODE SE	T FREQ/AUX GEN OFF

3. Allow the unit to warm up at least 2 hours for final calibration. Keep the instrument covers on to maintain heat. Remove covers only to make adjustments or measurements.

 Table 5-1. Generator Board Calibration Procedure

 Note: Where there are no entries, open column indicates previous entry is applicable.

Step	Test	Freq/ Start Freq	Freq	Vern/ Sym	Mode	Func	Ampl	Test	Tester	Adjust	500 Load	Result	Remarks
1 (3)	±15V Balance	2.0	¥	Š	CONT	Šģ	»	Board + 15V	DVM	R3	o <sub>N</sub>	+15 ± .75 Vdc	Ref. gnd is TP7.
1 (2)	1	ł	ı	1	1	ı	1	Board - 15V	1	Verify	ı	- 15 ± .75Vdc See remarks	-15V =  +15V  (reading) $\pm 10$ mV , if not retouch R3.
2	+ 5V Supply	ı	ı	ı	ı	1		+ 5V	ı	ı	ı	+5 ± .25 Vdc	
က	- 5V Supply	ı	ı	١	١	ı	ı	- 5V	ı	ı	1	+5 ± .25 Vdc	
4	+ 23V Supply	1	1	ı	1	ı	1	FB1	1	1	1	+23 ± 1.15 Vdc	
. 0	- 23V Supply	١	1	١	ı	ı	1	FB2	1	ı	1	-23 ± 1.15 Vdc	
9	Power Ampl Zero	ı	ı	1		8	ccw	FUNC	1	R258	Yes	0 ± 20 mVdc	
7	Preamp Zero	1	ı	ı	l	1	Š	.	1	R185	1		
80	Top of Dial Symmetry	I	ı	l	1	Sqr	ı	·	Scope	R96	1	Asym < 1µs	Set for min asym. (Set by alternate triggering of scope ± slope.)
თ	VCG Null	.02	100K	1	I	1	1	ı	1	R65	1	See remarks	Set for min freq shift when VCG IN is grounded. Repeat steps 8 and 9 as necessary.
10	100:1 Symmetry	ı	1	ı	1	ı	1	1	•	R94	1	Asym <1μs	Set for min asym.
=	1000:1 Frequency	ı	ı	CCW	ı	١	1		Counter	R63		160 (+0, -20) Hz	
12	Triangle Offset	2.0	¥	o.	ı	Ē	١	1	DVM	R17	1	0 ± 20 mVdc	
13	Sine Distortion	1	I	1	1	Sine	ı	1	Dist. Analyzer	R159. R165	1	<.18%	
14	Triangle Trigger Baseline	ı	ı		TRIG	Ē	ı	1	DVM	R51	ı	0 ± 20 mV	
15 (1	15 (1) Dial Alignment	1	ı	I	CONT	Sqr	ı	SYNC	Counter	R81	1	2 kHz ± 10 Hz	
15 (2)		0.2	١	١	I	ı	1	ı	ı	Verify	1	200 ± 10 Hz	If satisfactory skip to step 16 (1).
15 (3)	- (8	See	1	1	1	ı	ı	ı	I	R81	1	2.088 kHz ± 10 Hz	Remove dial and set the shaft ccw.
15 (4)	- (1		1	1	I	I	ı		1	See remarks	I	200 ± 10 Hz	Replace dial, align to 0.2, tighten set screw and verify setting.
15 (5)	1	2.0	١	١	1	ı	I	ı	ı	R81	ı	2.00 ± 40 Hz	
16 (1)	1) X10M Frequency	1	10M	ı	I	ı	ı	ı		C37	ı	20 MHz ± 600 kHz	Optimize C66 value if setting is out of range for C37.
16 (2)	- (3	See	1	ı	ı	l	1	1	ı	Verify	1	Dial mark ± 600 kHz	Verify frequency at each major dial mark.
17 (1)	1) X1M Frequency	5:0	Ξ	cw	CONT	Sqr	1	SYNC	Counter	Verify	1	2 MHz ± 40 kHz	Trim C33 to Set 2MHz freq between 2.020 and 2.040 MHz.
18	X100K Frequency	1	10K	١	ı	1	I				I	200 ± 4.0 kHz	Optimize R39 value if necessary.
16	Capacity Mult	0.1	5	1		١	1	FUNC	Scope	R106	1	< 200µs	Set for min asym. (Very important for low freq sine dist.)
20 (1)		5.0		1	1	1	1	SYNC	Counter	R102	ı	199.5 ± .5 Hz	
21	Low Frequency Aberrations		¥	ŀ	١	ı	1	FUNC	Scope	R254	1	See remarks	Adjust the "Corner Shape" for just noticeable peaking.
22 (1)				1	1	Sine	1	1	DVM	R203	ı	5.35 Vrms ± .01V	Verify sine. tri and sqr ampl.
23	High Frequency Aberrations	ιvί	01 M	1	1	Sqr	1	1	Scope	R245 R211	ı	< 0.6 Vp-p	Worst case aberrations not to exceed 4% of full ampl for each peak.

Table 5-2. Sweep/Mod Board Calibration Procedure

NOTE: Replace the bottom cover, turn the instrument upright and remove the top cover. The following adjustments are located on the Sweep/mode Board. Set the AUX GEN controls as follows: WIDTH/ΔΕ/ΔΜ.......cw, SYMMETRY......cw.

-						A		Main C	Main Generator					
Step	Test	Freq (Hz)	Vernier	Vernier Aux Gen Func	Aux Gen Mode	Carrier Level/ Null	Suppressed Carrier	FUNC	Freq/Start Freq and Mult	Test Point	Adjust	Tester	Results	Remarks
-	Auxiliary Output Zero	100- 3 KHZ	οw	Tri	SET START	cw	Off (extended)	Sqr	2.0 × 1K	AUX OUT (No load)	R109	DVM (dc)	0 ± 5 mVdc	
2 (1)	Triangle Peak Adjustment	1			AUX OUT ONLY					1	R36		0 ± 5 mVdc	
2 (2)		1			1			1	ł		R34	DVM (see remarks)	2.88 ±.01 Vrms	True rms reading AC Voltmeter.
3 (1)	Top of Range Symmetry	3- 100 HZ		Sqr	1				1	1	R18	Scope and Counter	ASYM. <2% of period	Short either outside leg of the SYMMETRY pot (R19) to circuit ground before performing steps 3 and 4. Using scope, match the period of the positive half cycle to the negative half cycle while monitoring the
3 (2)	Frequency	-				İ				1	R20			frequency on the counter. Adjusting R18 and R20 affects frequency.
3(3)		1						l	1		Verify		105 ± 1 Hz	Repeat step 3 (1) and (2) if necessary.
4	Bottom of Range Symmetry Adjustment	3K- 100 KHZ	woo	Sqr		I		I		1	R10	Scope	ASYM <1%	Set for minimum asymmetry. Remove ground short from SYMMETRY pot (R19).
S.	Bottom of Range Frequency Adjustment	1				1	1		1	1	R2	Counter	2.7 ± .2 kHz	
ဖ	Capitance Mult. Balance	100- 3 KHZ			1	1	ı			1	R45	Scope	ASYM < .2% of period	Set for minimum asymmetry.
7	Sine	1	ð.	Sine			1	1			R58 and R66	Distortion Analyzer	Adjust for minimum distortion < .2% (typical)	
8 (1).	Sine Zero and						ļ				R62	DVM (dc)	0 ± 10 mVdc	
8(2)	Set	1			ı			1			R64 (	DVM (see remarks)	3.54 ±.03 Vrms	True rms reading AC voltmeter.
ō	Modulation	3- 100 HZ		Sqr		12 o'clock	On (depressed)	A	2.0 × 10K	FUNC OUT (500)	R83	Scope	Nuil (±50 mVdc)	Trigger scope from AUX SYNC OUT Set scope: vert 100 mV/div horz 2 ms/div
10	Carrier Zero				WA	See		1	I	1	R73		See remarks	Set scope to display 2 cycles. Merge both traces using CARRIER NULL and R73 so the peaks of both square waves coincide.

Table 5-2. Sweep/Mod Board Calibration Procedure (Continued)

						MA		Main G	Main Generator					
Step	Test	Freq (Hz)	Vernier	Aux Gen Aux Gen Func Mode	Aux Gen Mode	<u></u>	Suppressed Carrier	FUNC	Freq/Start Freq and Mult	Test	Adjust	Tester	Results	Remarks
11 (1)	Multiplier Zero	3- 100 HZ	cw	Sine	SET	See remarks	On (depressed)	Sine	2.0 × 1K	FUNC OUT (500)	<b>∀</b> Z	DVM (dc)	See remarks	Note vottmeter reading for reference in step 11 (2). CARRIER LEVEL/NULL remains unchanged from step 10.
11 (2)	1	l	ı	1	1	1		AM	1	1	R96			Adjust R96 for voltage referenced in step 11 (1) (±10 mV).
12 (1)	Carrier Amplitude		ı	1	1	1	1	Sine	1		Š Š	DVM (See	l	Using a true rms AC voltmeter, note reading for reference in step 12 (2).
12 (2)		1				1		AM	ı		R70		1	Adjust R70 for voltage referenced in step 11 (1) (±70 mVrms).
13 (1)	Sweep Width	1	l	1	1	ı	ı	Sine	2.0 x 100K	1	₹ Z	Counter		Note the frequency for reference in step 13 (2) ( $\pm$ 100 Hz).
13 (2)		l	l		SET WIDTH		ı		.02 × 100K	1	R120	1	ı	Set VERNIER/SYM ccw. Adjust R120 for frequency referenced in step 13 (1) (+200, -0 Hz).
14 (1)	Carrier Bandwidth	1		.1	SET AM	1	1	AM	1.0 × 1 MHZ	I	₹ 2	Scope	1	Set VERNIER/SYM cw. Use the scopes verifical controls to set a peak-to-peak reference level (6 cm. 30 minor divisions) for step 14 (2) and (3).
14 (2)	·	1		1	I	1	1	1	2.0 × 10 MHZ	1	C33		<14% deviation	Verify <14% deviation (±4.5 minor divisions) while rotating FREQ/STAPT FREQ dial between .1 and 2.0
14 (3)	1		1	1	1	1	1	1	1	1	1	ı		

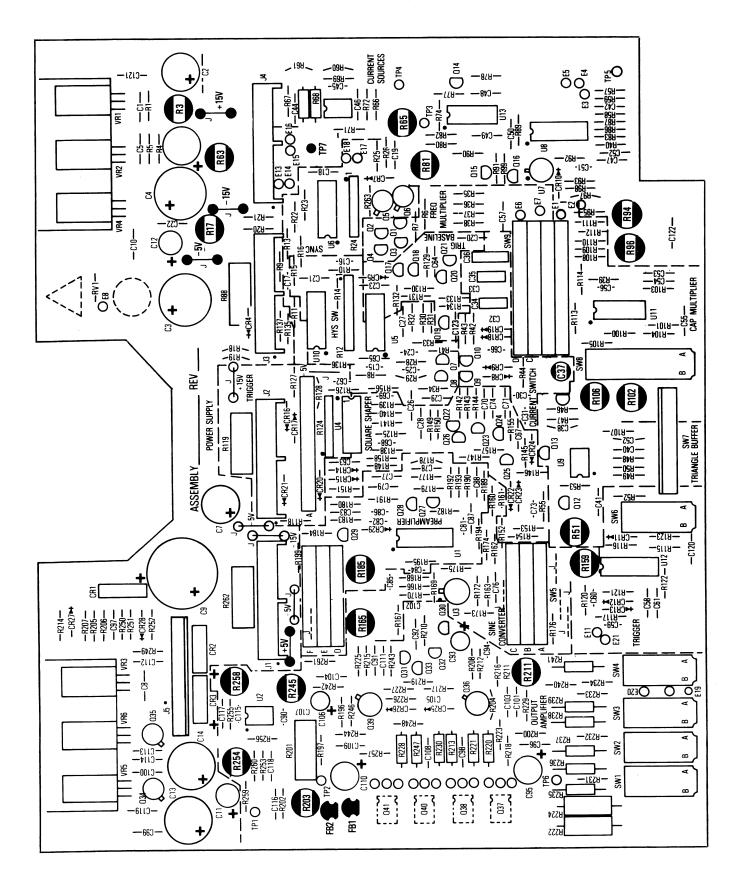


Figure 5-1. Generator Board Calibration Points

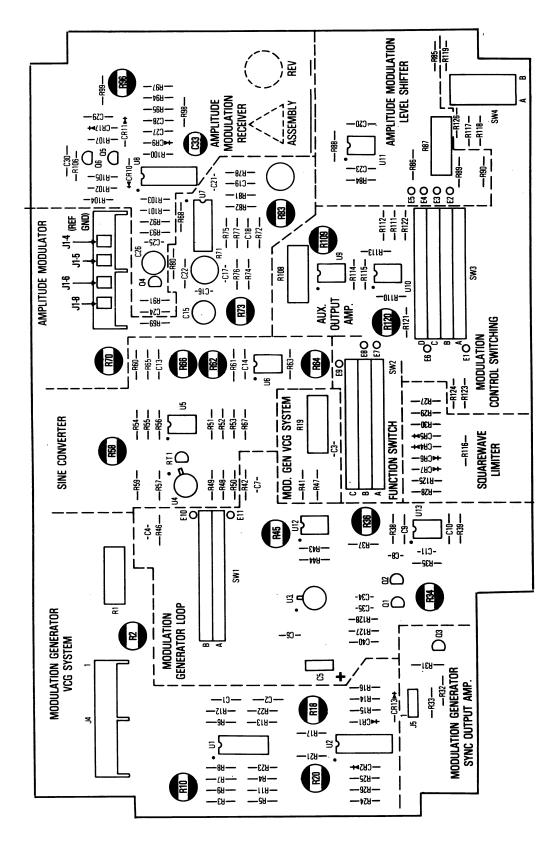


Figure 5-2. Sweep Mod Board Calibration Points

### **6.1 FACTORY REPAIR**

Wavetek maintains a factory repair department for those customers not possessing the necessary personnel or test equipment to maintain the instrument. If an instrument is returned to the factory for calibration or repair, a detailed description of the specific problem should be attached to minimize turnaround time.

### 6.2 BEFORE YOU START

Since no troubleshooting guide can possibly cover all the potential problems, the aim of this guide is to give a methodology which, if applied consistently, will lead to the problem area. Therefore, it is necessary to familiarize yourself with the instrument by reviewing the functional description and the detailed circuit description in conjunction with the schematic. Successful troubleshooting depends upon understanding the circuit operation within each functional block as well as the block relationships.

For sweep/mod board problems, refer to paragraph 6.5. For all other problems, refer to paragraph 6.3.

## 6.3 GENERATOR BOARD TROUBLESHOOTING

### **WARNING**

With the covers removed, dangerous voltage points may be exposed. Contact with any of these points could cause serious injury or death.

Table 6-1 gives an index of common generator board symptoms. For each symptom a troubleshooting guide is referenced (Paragraphs 6.3.1 through 6.3.15) that, when correctly followed, will lead to a solution to the problem.

The troubleshooting guide is arranged in three (3) levels:

- 1. Identify improperly set controls.
- 2. Isolate the faulty functional blocks.
- 3. Identify the faulty circuit or component.

Individual component troubleshooting is given in paragraph 6.7, recommended test equipment is given in

paragraph 5.2 and circuit schematics are in the back of this manual.

In all problems:

- 1. Double check for proper control settings.
- 2. Calibrate or rule out calibration as a problem.
- 3. Inspect components, wiring and circuit boards for heat damage.
- 4. Recalibrate as necessary after circuit repair.

Find the instrument symptom in table 6-1 and proceed as directed to the proper troubleshooting paragraph. See paragraph 6.5 for sweep/mod board related problems.

Table 6-1. Generator Board Related Problems

Symptom	Paragraph
Fuse blows, no dial lamp.	6.3.1
Power supply >100 mVp-p ripple or out of specification.	6.3.2
Function out (all functions) distorted or missing.	6.3.3
Square output distorted or missing.	6.3.4
Sine wave output distorted or missing.	6.3.5
Triangle output distorted or missing.	6.3.6
Sync output distorted or missing (FUNC OUT normal).	6.3.7
Excessive high frequency sine or triangle roll off, excessive square wave overshoot and	6.3.8
rise/fall time.	
Low frequency square wave tilt.	6.3.9
Time symmetry cannot be adjusted within specification.	6.3.10
Frequency accuracy and FREQ/START FREQ dial response problems.	6.3.11
Trigger, gate and trigger baseline problems.	6.3.12
Voltage at VCG IN connector not changing frequency properly.	6.3.13
DC offset not functioning correctly.	6.3.14
Variable symmetry problems	6.3.15

### 6.3.1 Fuse Blows, No Dial Lamp

- 1. Fuse size incorrect for voltage setting.
- 2. Line voltage selector incorrectly positioned.

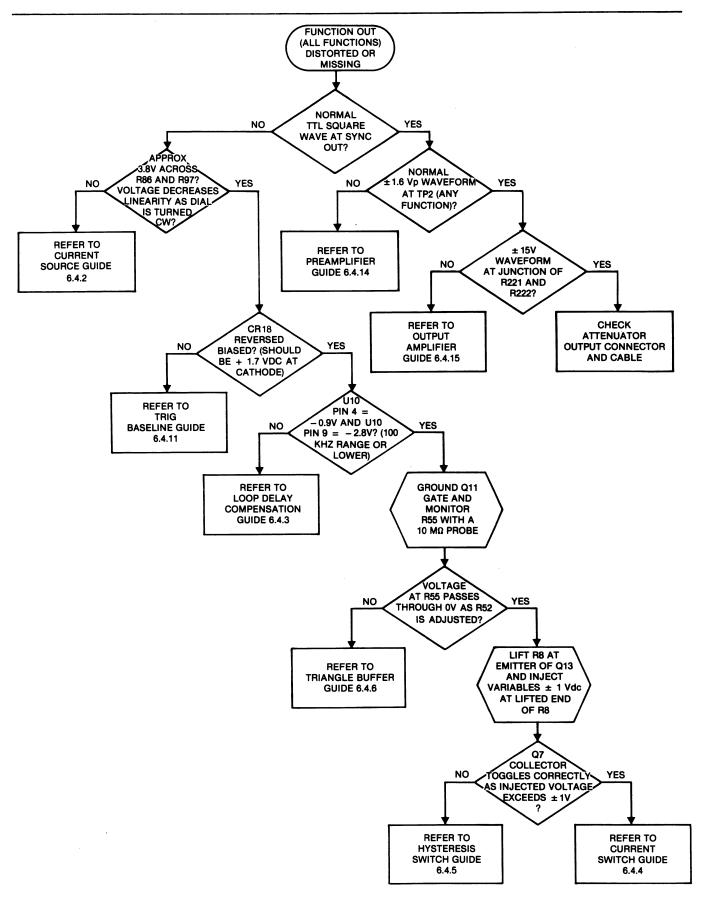


Figure 6-1. Function Output Troubleshooting

 Disconnect P5. If ac voltages are now correct, refer to the power supply guide, paragraph 6.4.1.
 If not, inspect the transformer and power receptacle.

## 6.3.2 Power Supply > 100 mVp-p Ripple or Out of Specification

- 1. Check line voltage selector for correct position.
- 2. If the supply is 0V, check for a short between the faulty supply and ground by lifting the jumpers at rear of the board.
- 3. Lift P5 from the board. If the voltages at P5 are not close to the values shown on the schematic table, inspect the transformer and power receptacle. If the voltages are normal, connect P5, then lift the jumpers (rear of board) for faulty supply. If the supplies are still bad, refer to paragraph 6.4.1. If not, the problem is caused by an excessive current drain by the generator circuits.

# 6.3.3 All Waveforms at FUNC OUT Distorted or Missing

Improperly set controls:

- OUTPUT ATTEN or AMPLITUDE controls incorrectly set too low for scope gain or voltmeter range.
- 2. FUNCTION switch incorrectly set to DC or AM.
- 3. MODE switch incorrectly set to TRIG or GATE.
- 4. SYM or DC OFFSET buttons depressed.

### Functional block isolation:

- 1. Verify power supply voltages are within  $\pm$  5% of nominal, with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- Check for a nonlinear triangle. If the triangle is nonlinear on only one range, check for a leaky capacitor on that range. If the triangle is nonlinear in more than one range, check for leaky capacitors or faulty active components in the frequency multiplier and triangle buffer circuits.
- 3. If the waveform is bad in one of the four lowest ranges (.1, 1, 10, 100), but the remaining ranges are normal, refer to the capacitance multiplier guide 6.4.9.
- If the waveform is bad only in 1M or 10M FREQ MULT positions, refer to paragraph 6.4.3. If the delay compensation circuit appears normal, refer to figure 6-1.
- 5. If square wave symmetry, measured at FUNC

- OUT, is out of specification and cannot be calibrated, refer to paragraph 6.3.10.
- 6. If none of the above conditions apply, refer to figure 6-1.

### 6.3.4 Square Wave Distorted or Missing

Improperly set controls:

- 1. SYM button depressed.
- 2. Excessive dc offset overdriving output amplifier.

### Functional block isolation:

- Verify power supply voltages are within ±5% of nominal, with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- 2. If the waveform is bad in one or more of the four lowest ranges (.1, 1, 10, 100), but the remaining ranges are normal, refer to paragraph 6.4.9.
- 3. If symmetry is not in specification and cannot be calibrated refer to paragraph 6.3.10.
- 4. If none of the above conditions apply, refer to figure 6-2.

### 6.3.5 Sine Wave Distorted or Missing

Improperly set controls:

- SYM button depressed
- 2. Excessive dc offset overdriving output amplifier.

### Functional block isolation:

- Verify power supply voltages are within ±5% of nominal, with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- Check the triangle for nonlinearity at FUNC OUT.
   If it is nonlinear, but only on one range, check for a leaky capacitor on that range. If the triangle is nonlinear on more than one range, check for a leaky capacitor or faulty active component in the frequency multiplier and triangle buffer circuits. (NOTE: Some nonlinearity above 200 kHz is normal and not specified.)
- 3. If the waveform is bad in one or more of the four lowest ranges (.1, 1, 10, 100), but the remaining ranges are normal, refer to paragraph 6.4.9.
- 4. Verify that square wave symmetry, at FUNC OUT, is in specification. If not and cannot be calibrated, refer to paragraph 6.3.10.
- 5. If none of the above conditions apply, refer to figure 6-3.

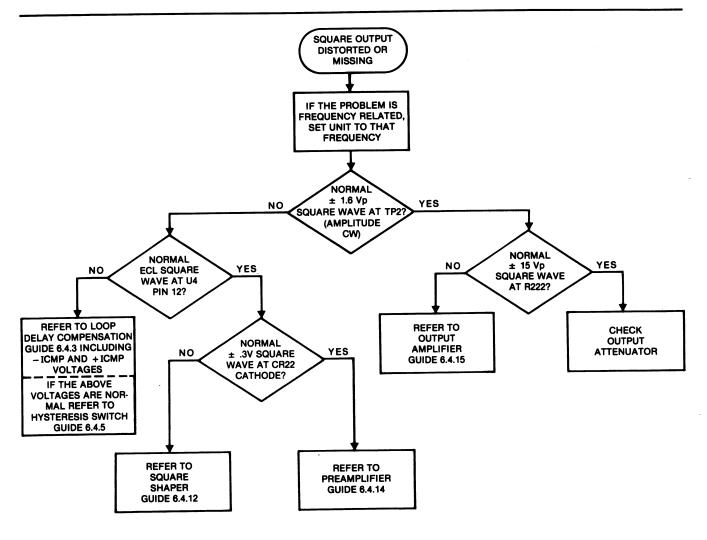


Figure 6-2. Square Output Troubleshooting

### 6.3.6 Triangle Distorted or Missing

Improperly set controls:

- 1. SYM button depressed.
- 2. Excessive dc offset overdriving output amplifer.

### Functional block isolation:

- 1. Verify power supply voltages are within  $\pm 5\%$  of nominal with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- 2. Check the triangle for nonlinearity at FUNC OUT. If it is nonlinear, but only on one range, check for a leaky capacitor on that range. If the triangle is nonlinear on more than one range, check for a leaky capacitor or faulty active component in the frequency multiplier and triangle buffer circuits.

- (NOTE: Some nonlinearity above 200 kHz is normal and not specified.)
- 3. If the waveform is bad in one or more of the four lowest ranges (.1, 1, 10, 100), but the remaining ranges are normal, refer to paragraph 6.4.9.
- 4. Verify square wave symmetry at FUNC OUT is in specification. If not and cannot be calibrated, refer to paragraph 6.3.10.
- 5. If none of the above conditions apply, refer to figure 6-4.

# 6.3.7 Sync Output Distorted or missing (FUNC OUT Normal)

Improperly set controls:

1. Because the FUNC OUT is normal, this cannot be caused by improperly set controls.

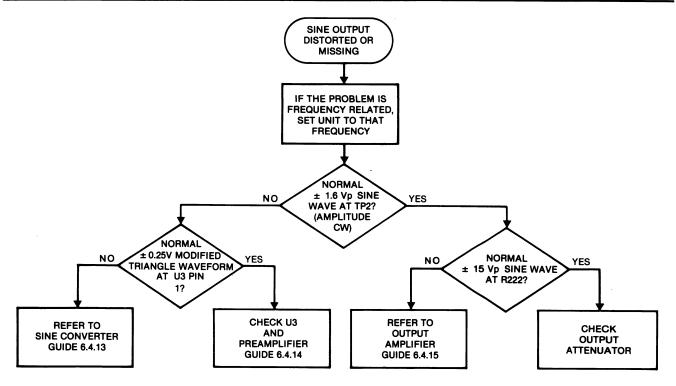


Figure 6-3. Sine Output Troubleshooting

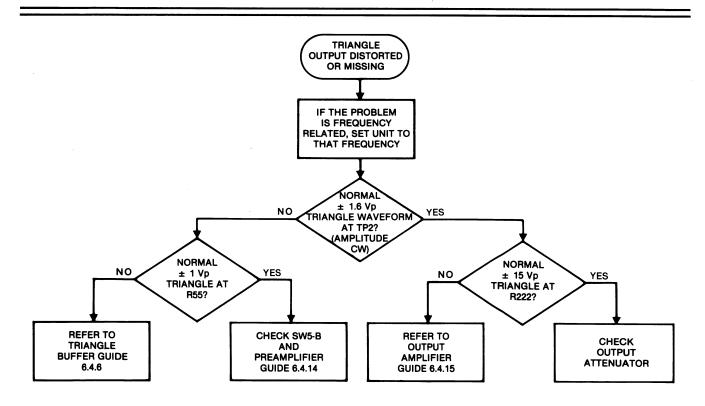


Figure 6-4. Triangle Output Troubleshooting

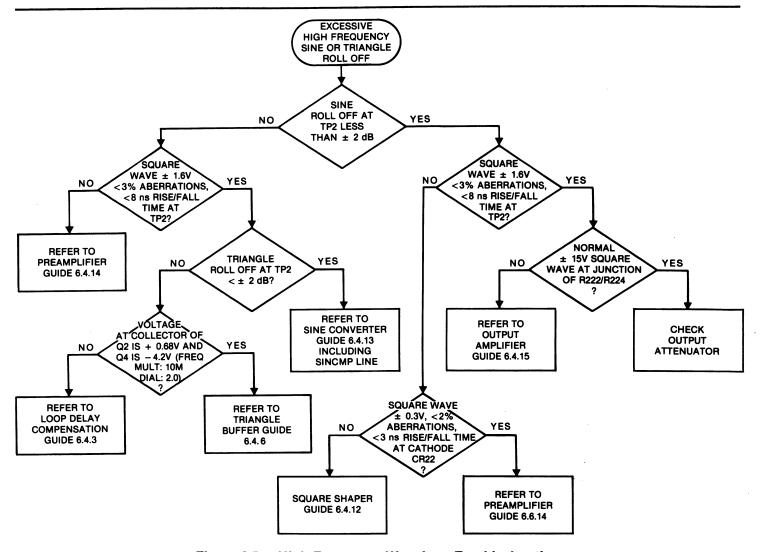


Figure 6-5. High Frequency Waveform Troubleshooting

Functional block isolation:

1. If there is no ECL square wave at U6 pin 10, refer to paragraph 6.4.7. If there is an ECL square wave, refer to paragraph 6.4.8.

# 6.3.8 Excessive High Frequency Sine or Triangle Roll Off

Improperly set controls:

- 1. Excessive dc offset overdriving output amplifier.
- 2. Verify  $50\Omega$  load on the cable at oscilloscope end.

### Functional block isolation:

 Verify power supply voltages are within ±5% of nominal with less than 100 mVp-p of ac ripple. If not refer to paragraph 6.4.1. 2. If none of the above conditions apply, refer to figure 6-5. Use a X10 probe with a very short ground lead and a spectrum analyzer, RF voltmeter or a 200 MHz bandwidth scope when performing sine or triangle roll-off tests.

## 6.3.9 Low Frequency Square Wave Tilt

Improperly set controls:

1. Scope improperly set to ac.

### Functional block isolation:

- 1. Verify power supply voltages are within  $\pm 5\%$  of nominal, and less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1
- 2. If none of the above conditions apply, refer to figure 6-6.

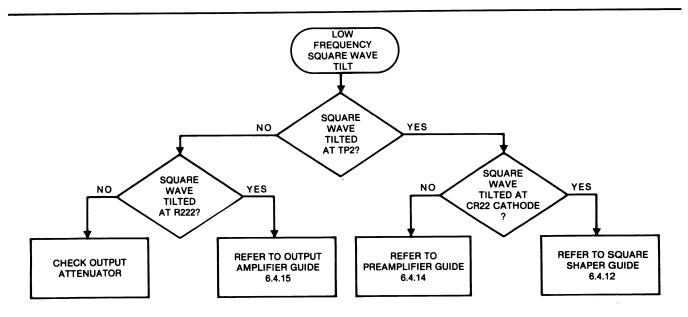


Figure 6-6. Low Frequency Square Wave Troubleshooting

# 6.3.10 Time Symmetry Cannot Be Adjusted To Within Specifications

Improperly set controls:

1. SYM button depressed.

Functional block isolation:

- Verify power supply voltages are within ±5% of nominal, with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- 2. If symmetry is out of specification in one of the four lowest ranges (.1, 1, 10, 100), but the remaining ranges are normal, refer to paragraph 6.4.9.
- 3. If symmetry is out of specification on FREQ MULT settings 1M or 10M only, refer to paragraph 6.4.3.
- If the voltages across R86 and R97 are not equal (typically 3.8V, Freq Dial: 2.0 Freq Mult: 100K or less), refer to paragraph 6.4.2

# 6.3.11 Frequency Accuracy and FREQ/START FREQ Dial Response Problems

Improperly set controls:

- 1. SYM button depressed.
- 2. External signal connected to VCG in BNC.
- 3. VERNIER not in FREQ CAL position.

### Functional block isolation:

1. Verify power supply voltages are within  $\pm 5\%$  of

- nominal with less than 100 mVp-p ac ripple. If not, refer to paragraph 6.4.1.
- 2. If the problem occurs in one of the four lowest frequency ranges (.1, 1, 10, 100), but the remaining ranges are normal, refer to paragraph 6.4.9.
- 3. If the frequency accuracy is out of specification on FREQ MULT settings 1M and 10M, refer to paragraph 6.4.3.
- 4. If the frequency is out of specification, but only on one range, check the range capacitor for that range.
- 5. If the problem occurs on the 1K, 10K, or 100K range, check the range capacitor.
- On the 1K range and frequency dial set at 2.0, check for 3.8V across R86 and R97. As the FRE-Q/START FREQ dial is rotated, this voltage should linearly track the dial settings within ±3% of full scale. If not, refer to paragraph 6.4.2.
- 7. If none of the above conditions apply, refer to figure 6-7.

## 6.3.12 Trigger, Gating and Trigger Baseline Problems

Improperly set controls:

- 1. MODE incorrectly set to CONT.
- 2. FUNCTION incorrectly set to DC.
- 3. DC OFFSET overdriving output amplifier.

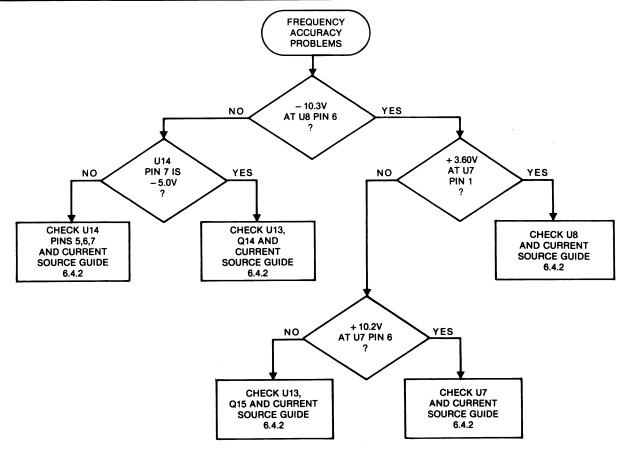


Figure 6-7. Frequency Accuracy Troubleshooting

### Functional block isolation:

- Verify power supply voltages are within ±5% of nominal, with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- If the trigger baseline cannot be calibrated within specification, set MODE to GATE and monitor the emitter of Q19. With TRIG IN disconnected, rotate the TRIG LEVEL ccw. The voltage should go about -0.7 Vdc. Rotating the TRIG LEVEL cw should change this voltage to about +1.8 Vdc. If these voltage readings are normal, check CR18 and CR19.
- 3. If none of the above conditions apply, refer to figure 6-8.
- 4. For high frequency (1M and 10M ranges) trigger or gate problems, set the controls as follows:

FREQ/START FREQ 2

2.0

FREQ MULT:

10M

SYM: MODE: OFF TRIG or GATE

(Depends on symptom-

GATE preferred)

TRIG LEVEL:

12 o'clock

Set the scope as follows:

Horizontal:

20 ns/div

Vertical:

1 V/div

Inject a 15 MHz 1 Vp-p trigger signal and refer to figure 6-9.

# 6.3.13 Voltage At VCG IN Connector Not Changing Frequency Properly

Improperly set controls:

- Excessive VCG IN voltage for dial setting (maximum input voltage is +5.0 Vdc with FREQ/START FREQ set at .02 and the Freq VERNIER turned ccw).
- 2. MODE (AUX GEN) incorrectly set to FM

Functional block isolation:

 Set FREQ/START FREQ to 2.0, FREQ MULT to 1K, and VCG IN with no input. Measure voltage across R86 and R97 (+3.8 Vdc). In addition, as the frequency dial is rotated, the voltage linearly

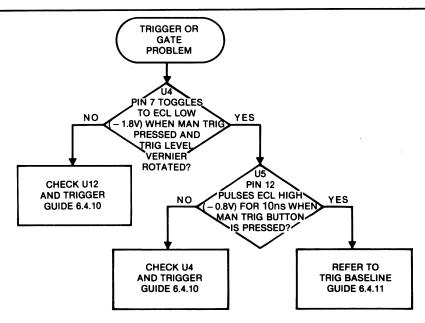


Figure 6-8. Trigger Gate Troubleshooting

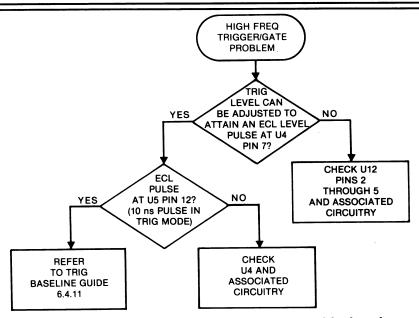


Figure 6-9. High Frequency Trigger/Gate Troubleshooting

tracks FREQ/START FREQ settings within  $\pm 3\%$  full scale. If it functions properly, check R67, R68, R69 and associated circuitry, but if not, refer to paragraph 6.3.11.

## 6.3.14 DC Offset Not Functioning Correctly

Improperly set controls:

- 1. Signal peak plus offset exceeding + or 7.5V (with a 50 $\Omega$  load), or  $\pm$  15V open circuit.
- Check OUTPUT ATTEN since this also attenuates output offset.

Functional block isolation:

- Verify power supply voltages are within ±5% of nominal with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- Take the following voltage measurements with the DC OFFSET button depressed and the DC OFFSET control rotated cw:
  - a) The junction of R260 and C116 should vary from +8.0V to -8.0V.
  - b) U2 pin 2 should hold at 0.0V.

- c) U2 pin 6 should vary from -1.0V to +1.0V.
   (Drifting of this voltage is typical because of constant compensation by U2 of variations in output transistor currents.)
- 3. If none of the above conditions apply, refer to paragraph 6.4.15.

### 6.3.15 Variable Symmetry Problems

Improperly set controls:

- 1. SYM button is incorrectly extended.
- 2. Note: When SYM is depressed the output frequency should be one-tenth the selected frequency.
- 3. DC offset is overdriving the output amplifier.

#### Functional block isolation:

- Verify power supply voltages within ±5% of nominal with less than 100 mVp-p of ac ripple. If not, refer to paragraph 6.4.1.
- When the voltage at the right leg of R88 (VERNIER/SYM CW) is -15V and when the voltage at the left leg of R88 (VERNIER/SYM) is -15V, refer to paragraph 6.4.2. If not, check R88 and SW8.

## 6.4 CIRCUIT GUIDES

Circuit guides provide listings of voltage levels, waveforms, and hints that, when used with the schematics, are helpful in isolating faulty circuits. Table 6-2 is an index of circuit guides.

Table 6-2. Circuit Guide Index

Circuit Guide	Paragraph
Power Supply	6.4.1
Current Source	6.4.2
Loop Delay Compensation	6.4.3
Current Switch	6.4.4
Hysteresis Switch	6.4.5
Triangle Buffer	6.4.6
Zero Crossing Detector	6.4.7
Sync	6.4.8
Capacitance Multiplier	6.4.9
Trigger	6.4.10
Trig Baseline	6.4.11
Square Shaper	6.4.12
Sine Converter	6.4.13
Preamplifier	6.4.14
Output Amplifier	6.4.15

### 6.4.1 Power Supply Guide

- 1. To determine a faulty power supply, check for the voltages given in table 6-3.
- 2. If the regulator input is bad, remove P5 and check for:
  - a. Shorted or open diodes (CR1, CR2, or CR3).
  - b. Shorted or open capacitors at the input of the regulator.
  - c. Short between the regulator metal mounting tab and chassis ground.
- 3. If the regulator input is good, check for:
  - a. Shorted or open capacitors at the output of the regulator.
  - b. Short between regulator metal mounting tab and chassis ground.
  - Excessive loading by main board circuits; to verify, lift jumper of the appropriate supply.
  - If all of the above conditions appear normal, replace the voltage regulator.

Table 6-3. Power Supply Checks

Tolerance	Ripple (p-p)	Ripple (p-p)
30 ± 1.5 Vdc (a)		
(b)	1.5 Vac	10 mV
(c)	1.5 Vac	10 mV
± 750 mV	1.5 Vac	10 mV
± 750 mV	1.5 Vac	10 mV
± 1.15 Vdc	1.5 Vac	10 mV
± 1.15 Vdc	1.5 Vac	10 mV
	(a) (b) (c) ± 750 mV ± 750 mV ± 1.15 Vdc ± 1.15 Vdc	(a) (b) 1.5 Vac (c) 1.5 Vac  ± 750 mV 1.5 Vac  ± 750 mV 1.5 Vac  ± 1.15 Vdc 1.5 Vac  ± 1.15 Vdc 1.5 Vac  n + 15V and - 15V supplies.

<sup>(</sup>c)  $-15V \text{ supply} = -|V_{+15} \pm .01V|$ .

### 6.4.2 Current Source Guide

**Top of Dial Check:** Set the controls as follows; then perform the checks in table 6-4.

Control	Setting
FREQ/START FREQ	2.0
FREQ MULT	1K
VERNIER	FREQ CAL
SYM	Off (extended)

VCG IN Disconnected MODE (AUX GEN) AUX OUT ONLY

**VCG Check:** Set the controls as follows; then perform the checks in table 6-5.

Control	Setting
FREQ START FREQ	0.2
FREQ MULT	1K
VERNIER	Full ccw
SYM	Off (extended)
VCG IN	+5.0 Vdc input
MODE (AUX GEN)	AUX OUT ONLY

Table 6-4. Current Source Check (Top of Dial)

Test Point	Desired Results
U14 pin 7	−5 ± .5 Vdc
U13 pins 1, 2	-5 ± .5 Vdc
Measure across R83	+3.8 ± .38 Vdc
U8 pin 6	$-10.3 \pm 1.03  \text{Vdc}$
Measure across R84 and R93	+3.8 ± .38 Vdc
U13 pin 6	0 ± .01 Vdc
U7 pin 6	+10.2 ± 1.02 Vdc
Measure across R86 and R97	+3.8 ± .38 Vdc

Table 6-5. Current Source (VCG IN)

Test Point	Desired Results
U7 pin 6	+14.38 ± 1.44 Vdc
U8 pin 6 (disconnect VCG IN)	-14.3 ± 1.43 Vdc

**10 MHz Range Check:** Set the controls as shown below, then perform the checks in table 6-6.

Control	Setting
FREQ/START STOP	2.0
FREQ MULT	10M
VERNIER	FREQ CAL
SYM	Off (extended)
VCG IN	Disconnected
MODE (AUX GEN)	AUX GEN ONLY

Table 6-6. Current Source Check (10 MHz Range)

Test Point	Desired Results
Measure across R99	+5.9 ± .59 Vdc
Measure across R83	+6.05 ± .61 Vdc
U8 pin 6	$-8.2 \pm .82  \text{Vdc}$

**Variable Symmetry Check:** Set the controls as shown then measure the voltage across resistors R84, R85, R86, R87, R93, and R97. The measured voltages should read  $\pm$  0.38  $\pm$  .04V.

Control	S	Settings
FREQ/START	STOP 2	2.0
VERNIER	. 1	2 o'clock position
VCG IN	[	Disconnected
FREQ MULT	1	K
SYM	C	On (depressed)

### 6.4.3 Loop Delay Compensation Guide

Set the controls as shown; then perform the checks in table 6-7.

Controls	Settings
FREQ/START FREQ	2.0
VERNIER	FREQ CAL
VCG IN	Disconnected
FREQ MULT	10 <b>M</b>
SYM	Off (extended)

Table 6-7. Loop Delay Compensation Checks

Test Point	Desired Results
Q1 and Q2 emitters	+ 9.2 ± .92 Vdc
Q1 base and collector, Q2 base	+8.5 ± .9 Vdc
Q2 collector	+0.68 ± .07 Vdc
Q3 and Q4 emitter	-9.05 ± .91 Vdc
Q3 and Q4 bases, Q3 collector	-8.32 ± .83 Vdc
Q4 collector	-4.2 ± .42 Vdc
U10 pin 4	-1.6 ± .16 Vdc (+ Peak reference)
U10 pin 9	-2.17 ± .22 Vdc (- Peak reference)

## 6.4.4 Current Switch Guide

Set the controls as shown; then take waveform measurements. Refer to figure 6-10.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
SYM	Off (extended)
MODE	CONT

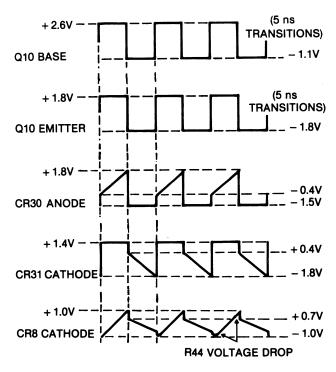


Figure 6-10. Current Switch Waveforms

## 6.4.5 Hysteresis Switch Guide

Set the controls as shown; then perform the checks in table 6-8, and take waveform measurements. Refer to figure 6-11.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
SYM	Off
MODE	CONT

Table 6-8. Hysteresis Switch Guide

Test Point	Desired Results
U10 pin 4	-0.9 ± .09 Vdc (+ Peak reference)
U10 pin 9	-2.8 ± .28 Vdc (- Peak reference)
Q7 and Q8 emitters	$-3.0 \pm .3 \mathrm{Vdc}$

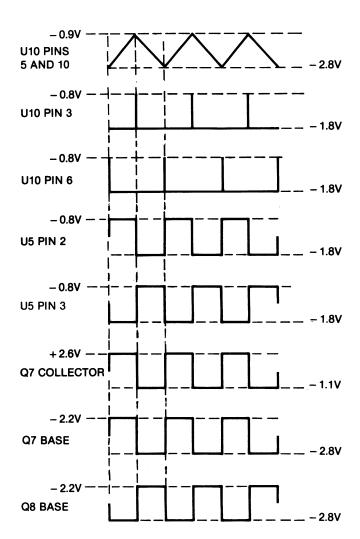


Figure 6-11. Hysteresis Switch Waveforms

## 6.4.6 Triangle Buffer Guide

Set the controls as shown; then perform the checks in table 6-9 and take waveform measurements. Refer to figure 6-12. If, after setting the controls, the generator loop does not run, lift R45 at E23 and inject a  $\pm$  1.0V triangle into R45.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
SYM	Off
MODE	CONT

Table 6-9. Triangle Buffer Checks

Test Point	Desired Results
Q11 drain	+6.5 ± .65 Vdc
Q12 emitter	$0.3 \pm .03$ Vp-p triangle, offset $-10 \pm 1$ Vdc
Q12 base	$0.3 \pm .03$ Vp-p triangle, offset $-9.3 \pm .9$ Vdc
Q13 collector	$+5.0 \pm .5 \mathrm{Vdc}$

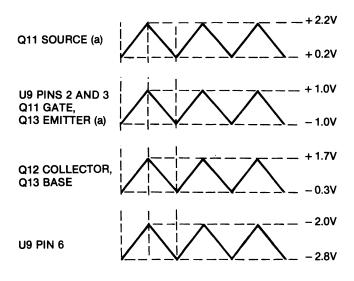


Figure 6-12. Triangle Buffer Waveforms

## 6.4.7 Zero Crossing Detector Guide

(a) Requires a X10 Probe (high impedance)

Set the controls as shown; then take waveform measurements. Refer to figure 6-13.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
SYM	Off
MODE	CONT

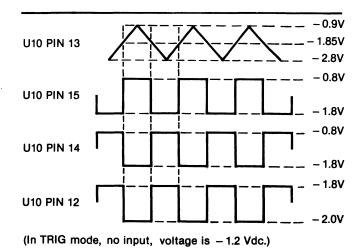
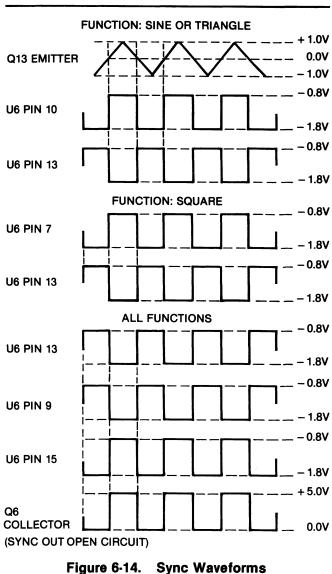


Figure 6-13. Zero Crossing Detector Waveforms



## 6.4.8 Sync Guide

Set the controls as shown then perform the checks in table 6-10 and take waveform measurements, see figure 6-14.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
SYM	Off
MODE	CONT

Table 6-10. Sync Check

	Desired Results	
Test Point	Function: Sine or Triangle Wave	Function: Square Wave
CR4 cathode	+1.2 ± .12 Vdc	-5 ± .5 Vdc
U6 pins 4 and 6	-1 ± .1 Vdc	-4.3 ± .43 Vdc
U6 pins 2 and 11	-1.8 ± .18 Vdc	-0.8 ± .08 Vdc
Q5/Q6 emitters	-1.6 ± .16 Vdc	-1.6 ± .16 Vdc

### 6.4.9 Capacitance Multiplier Guide

Set the controls as shown; then take the waveform measurements. Refer to figure 6-15.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	100
SYM	Off
MODE	CONT

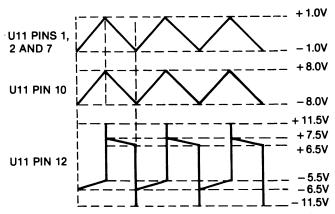


Figure 6-15. Capacitance Multiplier Waveforms

## 6.4.10 Trigger Guide

**TRIG or CONT Check:** Set the controls as shown; then take the waveform measurements. Refer to figure 6-16.

Controls	Settings
FREQ/START FREQ FREQ MULT MODE TRIG IN	2.0 1K TRIG or CONT ±1V 1 kHz Square wave

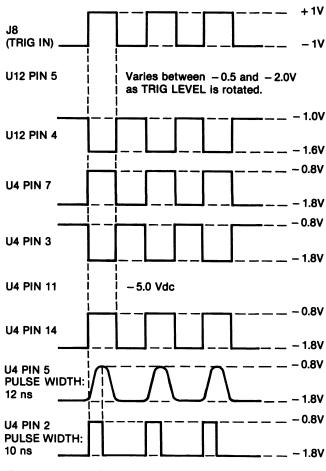
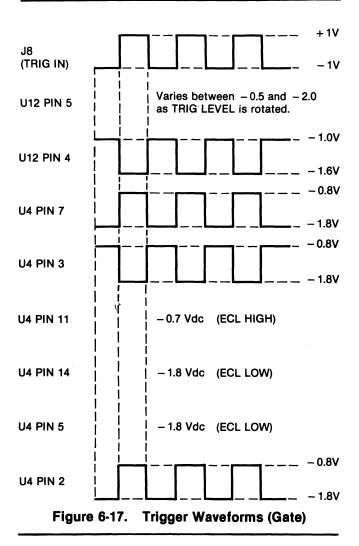


Figure 6-16. Trigger Waveforms (TRIG or CONT)

**GATE Checks:** Set the controls as shown; then take the waveform measurements. Refer to figure 6-17.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
MODE	GATE
TRIG IN	±1V 1 kHz Square wave



### 6.4.11 Trigger Baseline Guide

**Trigger or Gate Mode Problems:** Set the controls as shown; then take the waveform measurements. Refer to figure 6-18.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	10K
SYM	Off
MODE	TRIG or GATE
	(Depends on symp-
	tom—GATE pre-
	ferred
TRIG LEVEL	Approximately
	centered
TRIG IN	± 1V 10 kHz Square
	wave

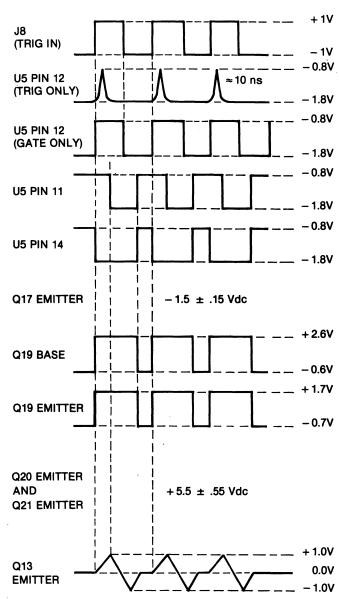


Figure 6-18. Trigger Baseline Waveforms

**Continuous Mode Problems:** Set the controls as shown; then take the waveform measurments. Refer to table 6-11.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	10K
SYM	Off
MODE	CONT

Table 6-11. Trigger Baseline Check (Continuous)

Test Point	Desired Results
U5 pin 12	$-0.8 \pm .08 \mathrm{Vdc}$
U5 pin 14	$-1.8 \pm .18 \text{ Vdc}$
Q17 emitter	-1.5 ± .15 Vdc
Q19 base	+2.6 ± .26 Vdc
Q19 emitter	+1.7 ± .17 Vdc
Q20 emitter	+5.5 ± .55 Vdc
Q21 emitter	+5.5 ± .55 Vdc
Q13 emitter	± 1.0V triangle

## 6.4.12 Square Shaper Guide

Set the controls as shown; then perform the checks in table 6-12 and take waveform measurements. Refer to figure 6-19.

Controls	Settings
FREQ/START FREQ FREQ MULT SYM MODE FUNCTION	2.0 1K Off CONT See Table 6-12 and Figure 6-19

Table 6-12. Square Shaper Checks

	Desired Results	
Test Point	Sine or Triangle	Square
Q22 emitter	$-3.0 \pm .3 \mathrm{Vdc}$	$-3.0 \pm .3 \text{ Vdc}$
U4 pin 13	$-0.8 \pm .08  \text{Vdc}$	$-4.3 \pm .43  \text{Vdc}$
Q26 base	$-0.8 \pm .08  \text{Vdc}$	-4.2 ± .42 Vdc
Q26 emitter	-1.6 ± .16 Vdc	$-4.0 \pm .4 \text{ Vdc}$
CR24 anode	+1.6 ± .16 Vdc	-1.5 ± .15 Vdc

### 6.4.13 Sine Converter Guide

Set the controls as shown; then perform the checks in table 6-13 and take waveform measurements. Refer to figure 6-20.

Controls	Settings
FREQ/START FREQ	2.0
FREQ MULT	1K
SYM	Off
MODE	CONT
FUNCTION	Sine

Table 6-13. Sine Converter Checks

Test Point	Desired Results
Junction R170 and R171	+14.8 ± 1.5 Vdc
Junction R173 and R174	- 14.8 ± 1.5 Vdc
U3 pin 2	0.0V (Full scale current = 2 mA

## 6.4.14 Preamplifier Guide

**DC Problems:** Set the FUNCTION control to DC; then perform the checks in table 6-14.

Table 6-14. Preamplifier Checks (DC)

Test Point	Desired Results
U1 pin 2	+14.86 ± 1.5 Vdc
U1 pin 3	$-0.7 \pm .07  \text{Vdc}$
U1 pin 13	-1.4 ± .14 Vdc
U1 pin 9	$-0.7 \pm .07  \text{Vdc}$
U1 pin 4	0.0 ± 10 mV
U1 pin 8	0.0 ± 10 mV
U1 pin 12	+5.8 ± .58 Vdc
U1 pin 11	+6.6 ± .66 Vdc
Q27 base	+9.6 ± .96 Vdc
Q27 emitter	+10.3 ± 1 Vdc
Q28 collector	+11.3 ± 1.1 Vdc

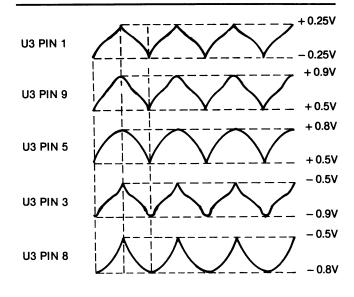


Figure 6-20. Sine Converter Waveforms

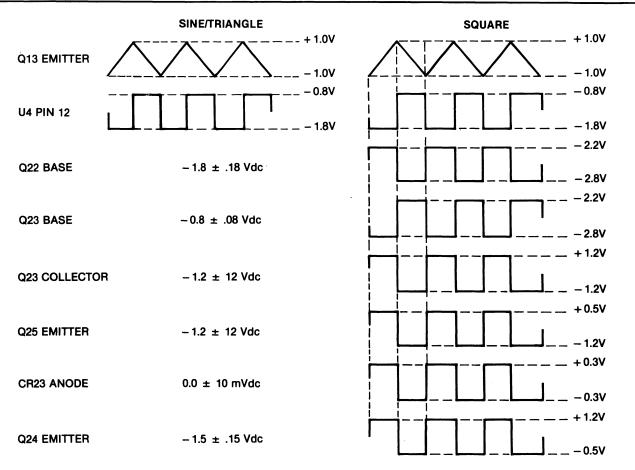


Figure 6-19. Square Shaper Waveforms

**Function Problems:** Set the controls as shown; then take the waveform measurements. Refer to figure 6-21.

2.0 K
Off CONT Square
+ 2.3V 0.9V

## 6.4.15 Output Amplifier Guide

Set the controls as shown; then take the waveform measurements. Refer to table 6-15.

Controls	Settings	
FUNCTION	DC	
DC OFFSET	Off	

Table 6-15. Output Amplifier Checks

st Point	Desired Results
Base	+11.7 ± 1.2 Vdc
Emitter	+11 ± 1.1 Vdc
Collector	+19 ± 1.9 Vdc
Collector	+22.8 ± 2.3 Vdc
Base	-12 ± 1.2 Vdc
Emitter	-11.3 ± 1.1 Vdc
Collector	-19 ± 1.9 Vdc
	Base Emitter Collector Collector Base Emitter

Table 6-15. Output Amplifier Checks (Continued)

140.00	or output / timpini	or oneone (oonunaoa)
Те	st Point	Desired Results
Q33	Collector	$-22.7 \pm 2.3  \text{Vdc}$
	Base	+18.3 ± 1.8 Vdc
Q36	Emitter	+19 ± 1.9 Vdc
	Collector	+0.7 ± .07 Vdc
Q37	Emitter	+0.05 ± .003 Vdc
Q01	Collector	+22.5 ± 2.3 Vdc
Q38	Emitter	+0.05 ± .005 Vdc
CR27	Cathode	+23 ± 2.3 Vdc
01127	Anode	+0.6 ± .06 Vdc
VR5	Input	+31 ± 3.1 Vdc
VIII	Output	+24 ± 2.4 Vdc
Q34	Collector	+22.8 ± 2.3 Vdc
Q39	Base	-18.3 ± 1.8 Vdc
	Emitter	-19 ± 1.9 Vdc
	Collector	$-0.05 \pm .005  \text{Vdc}$
Q40	Emitter	$-0.05 \pm .005  \text{Vdc}$
Q-10	Collector	$-21 \pm 2.1  \text{Vdc}$
Q41	Emitter	$-0.05 \pm .005  \text{Vdc}$
CR28	Anode	$-23 \pm 2.3  \text{Vdc}$
Q35	Base	$-0.6 \pm .06  \text{Vdc}$
	IN	$-31 \pm 3.1  \text{Vdc}$
VR6	ADJ	$-21.3 \pm 2.1 \text{ Vdc}$
	OUT	$-23.6 \pm 2.4  \text{Vdc}$
U2	Pin 2	0.0 ± 10 mVdc
	Pin 6	$-0.05 \pm .005  \text{Vdc}$

### 6.5 SWEEP/MOD BOARD TROUBLESHOOTING

### WARNING

With the covers removed, dangerous voltage points may be exposed. Contact with any of these points could cause serious injury or death.

Table 6-16 gives an index of common sweep/mod board symptoms. For each symptom a troubleshooting guide is referenced (Paragraphs 6.5.1 through 6.5.11) that, when correctly followed, will lead to a solution to the problem.

The troubleshooting guide is arranged in three (3) levels:

- 1. Identify improperly set controls.
- 2. Isolate the faulty functional blocks.
- 3. Identify the faulty circuit or component.

Individual component troubleshooting is given in paragraph 6.7, recommended test equipment is given in paragraph 5.2 and circuit schematics are in the back of this manual.

In all problems:

- 1. Double check for proper control settings.
- 2. Calibrate or rule out calibration as a problem.
- 3. Inspect components, wiring and circuit boards for heat damage.
- 4. Recalibrate as necessary after circuit repair.

Find the instrument symptom in table 6-16 and proceed as directed to the proper troubleshooting paragraph.

Table 6-16. Sweep/Mod Board Related Problems

Symptom	Paragraph
Amplitude Modulation At FUNC OUT Distorted or Missing	6.5.1
All Waveforms At AUX OUT Distorted or Missing	6.5.2
Output At AUX SYNC OUT Distorted or Missing	6.5.3
Loop Will Not Oscillate	6.5.4
Abnormal Waveforms (AUX OUT)	6.5.5
Waveform Symmetry Out of Specification	6.5.6
Incorrect Frequencies	6.5.7
Loop Oscillates In AUX GEN OFF MODE	6.5.8
AUX VCG Input Has No Effect	6.5.9
Variable Symmetry Bad	6.5.10
No Internal FM	6.5.11

# 6.5.1 Amplitude Modulation at FUNC OUT Distorted or Missing

Improperly set controls:

- 1. AUX GEN MODE and FUNC (Main Generator) switches must be set to AM.
- 2. WIDTH/ΔF/ΔM vernier must not be fully ccw.
- 3. AUX GEN and main generator frequency too low.
- 4. MODE (Main Generator) switch incorrectly set to TRIG or GATE.
- 5. Combination of OUTPUT ATTEN and AMPLITUDE controls excessively reducing signal.
- 6. DC OFFSET overdriving amplifier.

Functional block isolation:

If the amplitude modulated signal at the function out is distorted or missing, refer to figure 6-22.

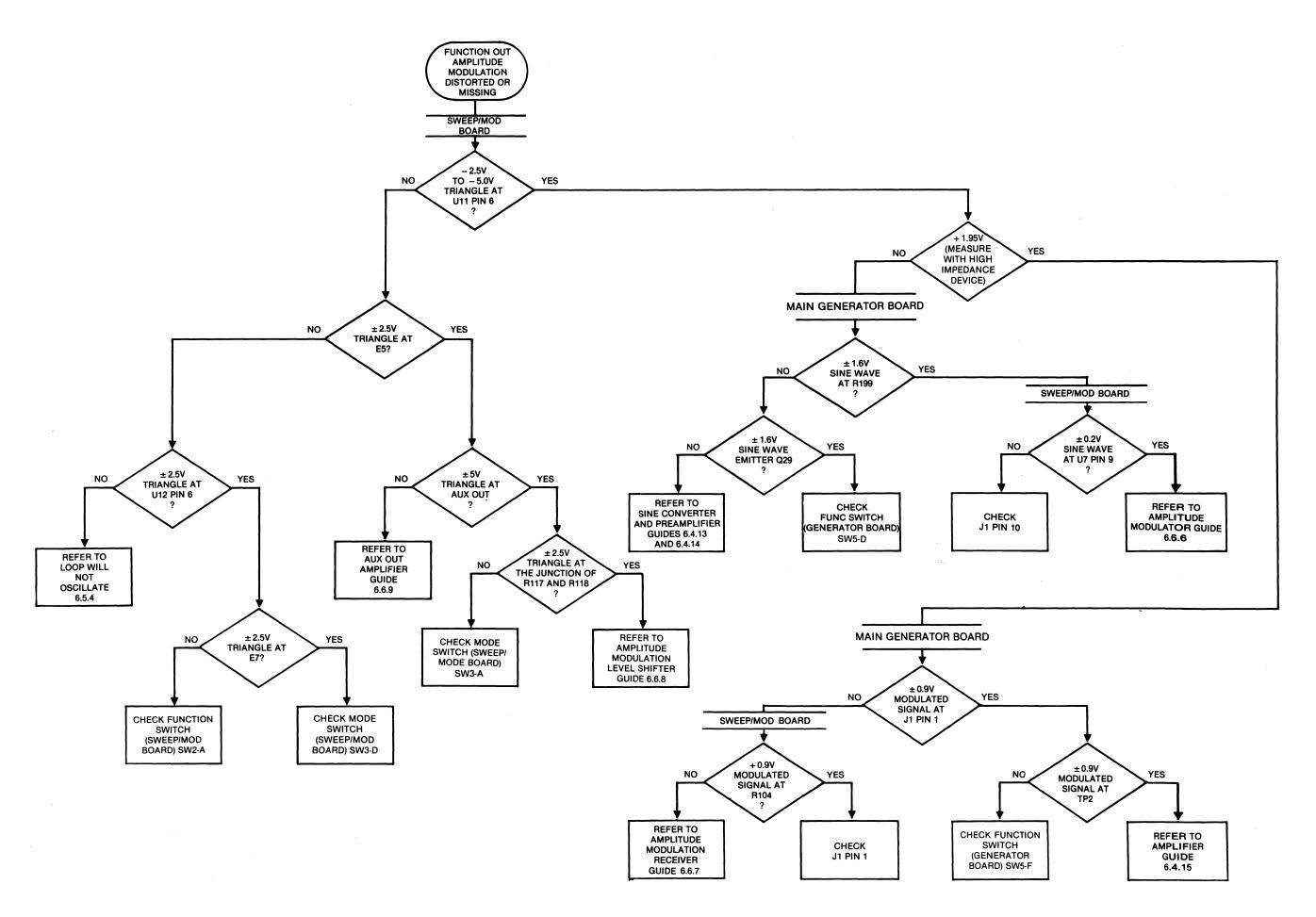
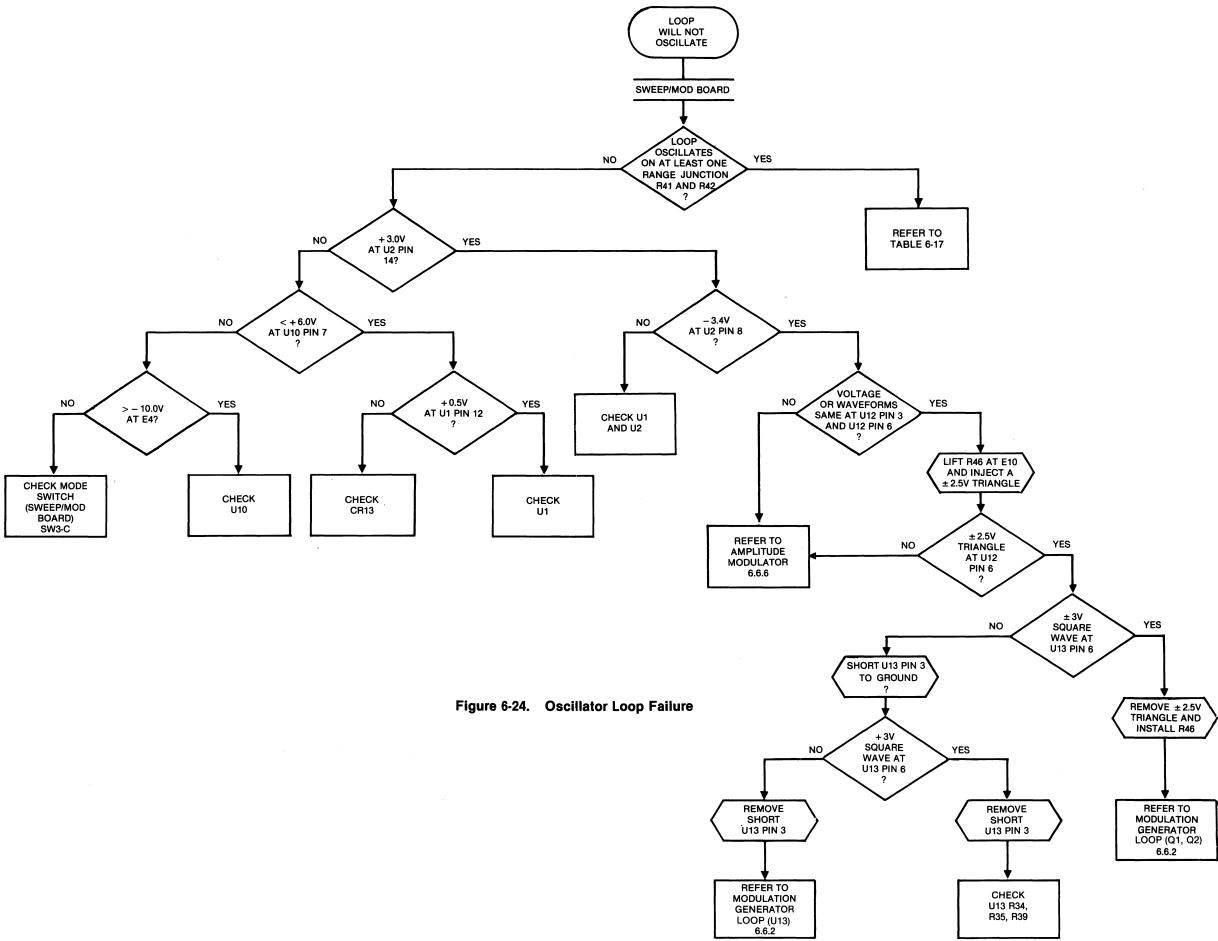


Figure 6-22. Amplitude Modulation Troubleshooting



# 6.5.2 All Waveforms At AUX OUT Distorted or Missing

Improperly set controls:

- AUX GEN frequency set too low.
- 2. AUX GEN MODE must be set to FM, SWEEP, AM, or AUX OUT ONLY.
- 3. WIDTH/ $\Delta$ F/ $\Delta$ M must not be fully ccw.

Functional block isolation:

If all waveforms at AUX OUT remains distorted or missing, refer to figure 6-23.

## 6.5.3 Output At AUX SYNC OUT Distorted or Missing

Improperly set controls:

- 1. AUX GEN frequency too low.
- AUX GEN MODE incorrectly set to SET FREQ or SET WIDTH.

Functional block isolation:

If a  $\pm 3.0$ V square wave is at U13 pin 2, refer to paragraph 6.6.4; if not, refer to paragraph 6.5.4.

### 6.5.4 Loop Will Not Oscillate

Improperly set controls:

- 1. AUX GEN frequency set too low.
- AUX GEN MODE incorrectly set to SET FREQ or SET WIDTH.

Functional block isolation:

If the loop still does not oscillate, refer to figure 6-24.

### 6.5.5 Abnormal Waveforms At AUX OUT

 Abnormal square wave, sine and triangle normal: If a ±2.8V square wave is present at the cathode of CR4, check function switch (Sweep/Mode board) SW2-A.

If not, refer to paragraph 6.6.3.

- Abnormal sine wave, square and triangle normal:
   If a ±2.5V sine wave is present at U6 pin 6, check the function switch (Sweep/Mod board) SW2-A.
   If not, refer to paragraph 6.6.5.
- Abnormal triangle, square and sine normal:
   Check function switch (Sweep/Mod board) SW2-A.

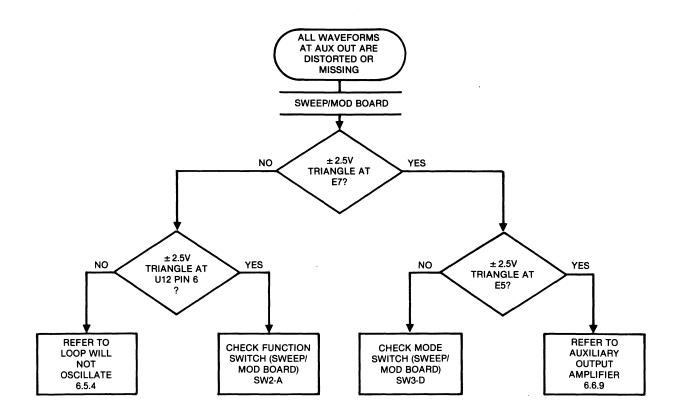


Figure 6-23. Distorted or Missing Waveforms at AUX OUT

Table 6-17. Oscillator Range Check

Inoperative Range	Check
3KI100K and 100I3K ranges only	C6 and SW1-A
3I100 and .1I3 ranges only	C5 and SW1-A
.1I3 and 100I3K range only	R46, R47 and SW1-B

## 6.5.6 Waveform Symmetry Out of Specification

Improperly set controls:

- AUX GEN FUNC switch incorrectly set to □ or
- 2. Signal incorrectly connected to AUX VCG input. Functional block isolation:
- Verify 0V across R19. If not, check AUX GEN FUNC switch SW2-C.
- 2. Refer to paragraph 6.6.1, especially U1 and U2.

### 6.5.7 Incorrect Frequencies

Improperly set controls:

- AUX GEN FUNC switch incorrectly set to ☐ or
- 2. Signal incorrectly connected to AUX VCG input.
- 3. WIDTH/ΔF/ΔM incorrectly set full ccw.

Functional block isolation:

If the frequencies are still incorrect, refer to table 6-18.

## 6.5.8 Loop Oscillates in AUX GEN OFF MODE

Improperly set controls:

AUX GEN MODE switch not set to AUX GEN OFF.

Functional block isolation:

If the Sweep/Mod generator continues to oscillate, refer to figure 6-25.

### 6.5.9 AUX VCG Input Has No Effect

Functional block isolation:

If the sweep/mod generator frequencies are correct

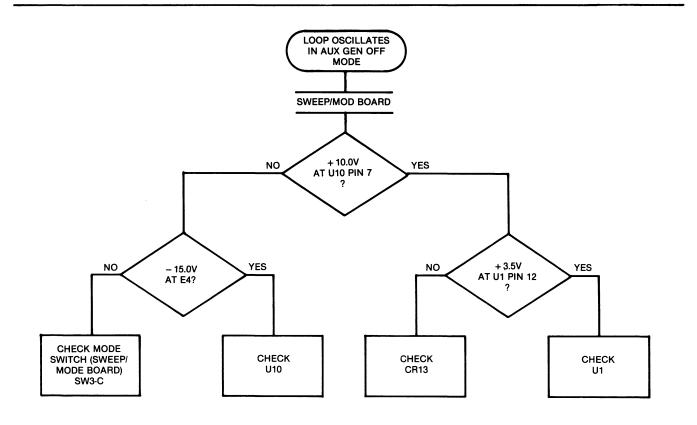


Figure 6-25. Loop Oscillates in AUX GEN OFF Mode

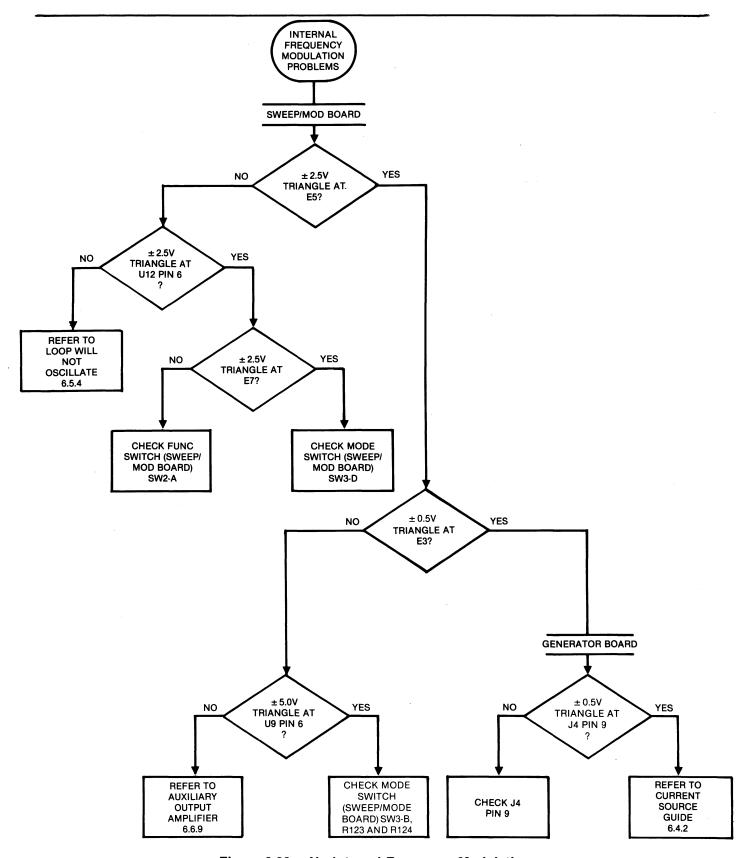


Figure 6-26. No Internal Frequency Modulation

(no AUX VCG input), check J16, J5 pin 1 and 2 and R5. If not, refer to paragraph 6.5.7.

Table 6-18. Incorrect Frequency Check

Incorrect Range	Check
3KI100K and 100I3K ranges only	C6 and SW1-A
3I100 and .1I3 ranges only	C5 and SW1-A
.1¶3 and 100¶3K range only	R46, R47 and SW1-B
All ranges	Modulation Generator VCG System 6.6.1 and SW1-B

# 6.5.10 Variable Symmetry (Sweep/Mod Board) Bad

Function block isolation:

If symmetry is normal in  ${\sim}$  ,  ${\sim}$  or  ${\sqcap}$  functions, check SW2-B, SW2-C and R19. If not, refer to paragraph 6.6.1.

## 6.5.11 No Internal Frequency Modulation

Improperly set controls:

- 1. FREQ/START FREQ set to 2.0.
- 2. WIDTH/ΔF/ΔM incorrectly set full ccw.

Functional block isolation:

If the sweep/mod board still does not frequency modulate the generator board, refer to figure 6-26.

### 6.6 SWEEP/MOD BOARD CIRCUIT GUIDES

Circuit guides provide listings of voltage levels, waveforms, and hints that, when used with the schematics, are helpful in isolating faulty circuits. Table 6-19 is an index of circuit guides.

Table 6-19. Circuit Guide Index

Circuit Guide	Paragraph
Modulation Generation VCG System	6.6.1
Modulation Generator Loop	6.6.2
Square Wave Limiter	6.6.3
Modulation Gen Sync Out Amp	6.6.4
Sine Converter	6.6.5
Amplitude Modulator	6.6.6
Amplitude Modulation Receiver	6.6.7
Amplitude Modulation Level Shifter	6.6.8
Auxiliary Output Amp	6.6.9

## 6.6.1 Modulation Generator VCG System Guide

Set the controls as shown below; then perform the checks in table 6-20.

Control	Settings	
AUX GEN FUNC AUX GEN MODE AUX VCG Input	${}^{}{\!$	

Table 6-20. VCG System Checks

	Desired Results	
Test Point	AUX GEN VERNIER	
	ccw	cw
U1 pin 7 and 10	-10 ± 1 mV	$-550 \pm 55 \text{ mV}$
U1 pin 1	+10 ± 1 mV	+550 ± 55 mV
U1 pin 14	+8 ± .8V	−8.9 ± .9V
U1 pin 12	0 ± 10 mV	+ 350 ± 50 mV
U2 pin 10 and 13	+4.5 ± .45V	+4.5 ± .45V
U2 pin 1	+14 ± 1.4V	+3.3 ± .3V
CR1 anode	+8.3 ± .8V	−3 ± .3V
U1 pin 8	−7.4 ± .7V	+8.7 ± .9V
U2 pin 4 and 7	-4.6 ± .46V	-4.6 ± .46V
U2 pin 5	-13.9 ± 1.4V	-3.4 ± .34V
CR2 cathode	-7.8 ± .8V	+3 ± .3V

### 6.6.2 Modulation Generator Loop Guide

Set the AUX GEN MODE switch to AM; then take the waveform measurements. Refer to figure 6-27.

## 6.6.3 Square Wave Limiter Guide

Set the AUX GEN MODE switch to AM; then take the waveform measurements. See figure 6-28.

## 6.6.4 Modulation Generator Sync Output Amplifier Guide

Set the AUX GEN Mode switch to AM; then take waveform measurements. Refer to figure 6-29.

## 6.6.5 Sine Converter Guide

Set the AUX GEN MODE switch to AM; then perform the checks in table 6-21 and take waveform measurements. Refer to figure 6-30.

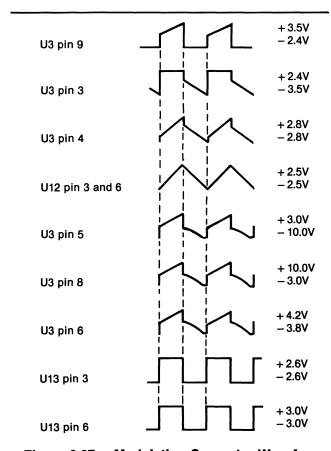


Figure 6-27. Modulation Generator Waveform

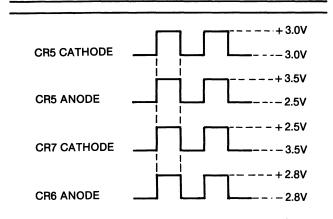


Figure 6-28. Square Wave Limiter Waveforms

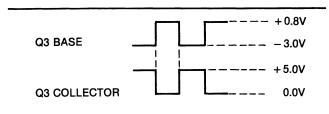


Figure 6-29. Mod Gen Sync Out Amp. Waveforms

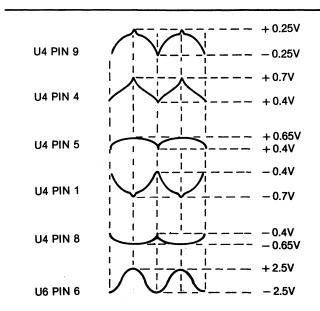


Figure 6-30. Sine Converter Waveforms

Table 6-21. Sine Converter Checks				
Test Point	Desired Results			
R57/RT1 junction	$-2.3 \pm .02  \text{Vdc}$			
U6 pins 2 and 3, U5 pins 2 and 6	0 ± 10 mVdc			

# 6.6.6 Amplitude Modulator Guide

Set the controls as shown; then perform the checks in table 6-22 and take the waveform measurements. Refer to figure 6-31.

Control	Setting
AUX GEN MODE	AM
WIDTH/ΔF/ΔM	Full cw
AM CARRIER LEVEL/NULL	Full cw
SUPPRESSED CARRIER	Off
FUNC (Main Generator)	AM

# 6.6.7 Amplitude Modulation Receiver Guide

Set the controls as shown; then perform the checks in table 6-23 and take waveform measurements. Refer to figure 6-32.

Control	Settings
AUX GEN MODE	AM
WIDTH/ΔF/ΔM	Full cw
AM CARRIER LEVEL/NULL	Full cw
SUPPRESSED CARRIER	Off
AUX GEN FUNC	Sine
FUNC (Main Generator)	AM

Table 6-22. Amplitude Modulator Checks

Test Point	Desired Results
U7 pin 1	+ 1.45 ± 0.14 Vdc
U7 pin 2	+1.95 ± 0.19 Vdc (measured with high impedance device)
U7 pin 3	- 13.0 ± 1.3 Vdc
U7 pin 4	-2.5 ± 0.25 Vdc
U7 pin 5	$-3.8 \pm 0.38  \text{Vdc}$
U7 pin 11	-1.35 ± 0.13 Vdc
U7 pin 12	0 ± 10 mVdc
U7 pin 13	-12 ± 1.2 Vdc
U7 pin 14	+1.95 ± 0.19 Vdc

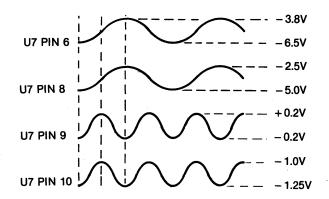
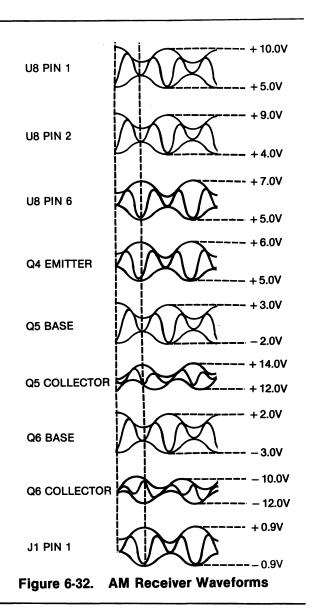


Figure 6-31. Amplitude Modulator Waveforms

Table 6-23. AM Receiver Guide

Test Point	Desired Results			
U8 pin 3	+14.7 ± 1.4 Vdc			
U8 pin 4	+1.25 ± .12 Vdc			
U8 pins 5 and 8	+2 ± .2 Vdc			
U8 pins 10 and 13	+14.1 ± 1.4 Vdc			
U8 pin 11	+13.3 ± 1.3 Vdc			
Q4 collector	+13.8 ± 1.3 Vdc			



# 6.6.8 Amplitude Modulation Level Shifter Guide

Set the controls as shown; then take the waveform measurements. Refer to figure 6-33.

Control	Setting
AUX GEN MODE	AM
WIDTH/AF/AM	Full cw
AM CARRIER LEVEL/NULL	Full cw
SUPPRESSED CARRIER	Off
AUX GEN FUNC	Triangle
FUNC (Main Generator)	AM

# 6.6.9 Auxiliary Output Amplifier Guide

Set the controls as shown; then take the waveform measurements. Refer to figure 6-34.

Control	Setting	
AUX GEN MODE	AM	
WIDTH/ΔF/ΔM	Full cw	
AM CARRIER LEVEL/NULL	Full ccw	
SUPPRESSED CARRIER	Off	
FUNC (Main Generator)	AM	
AUX GEN FUNC	Triangle	



Figure 6-33. AM Level Shifter Waveforms

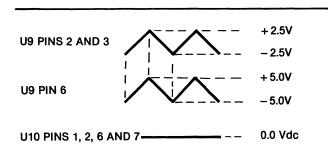


Figure 6-34. Auxiliary Output Amplifier Waveforms

# 6.7 TROUBLESHOOTING INDIVIDUAL COMPONENTS

#### 6.7.1 Transistor

- A transistor is defective if more than one volt is measured across its base-emitter junction in the forward direction.
- A transistor when used as a switch may have a few volts reverse bias voltage across base emitter junction.
- If the collector and emitter voltages are the same, but the base emitter voltage is less than 500 mV forward voltage (or reversed bias), the transistor is defective.
- A transistor is defective if its base current is larger than 10% of its emitter current (calculate currents from voltage across the base and emitter series resistors).
- In a transistor differential pair (common emitter stages), either their base voltages are the same in normal operating condition, or the one with less

forward voltage across its base emitter junction should be off (no collector current); otherwise one of the transistors is defective.

#### 6.7.2 Diode

A diode (except a zener) is defective if there is greater than one volt (typically 0.7 volt) forward voltage across it.

# 6.7.3 Operational Amplifier

- 1. The "+" and "-" inputs of an operational amplifier will have less than 15 mV voltage difference when operating under normal conditions.
- When the output of the amplifier is connected to the "-" input (voltage follower connection), the output should be the same voltage as the "+" input voltage; otherwise, the operational amplifier is defective.
- 3. If the output voltage stays at maximum positive, the "+" input voltage should be more positive than "-" input voltage, or vice versa; otherwise, the operational amplifier is defective.

#### 6.7.4 FET Transistor

- 1. No gate current should be drawn by the gate of an FET transistor. If so, the transistor is defective.
- The gate-to-source voltage is always reverse biased under a normal operating condition; e.g., the source voltage is more positive than the gate voltage for 2N5485, and the source voltage is more negative than gate voltage for a 2N5462. Otherwise, the FET is defective.
- 3. If the device supplying gate voltage to an FET saturates, the FET has too large a Vgs (pinch off) for the circuit and should be replaced.

#### 6.7.5 Capacitor

- 1. Shorted capacitors have 0V across their terminals.
- Opened capacitor can be located (but not always) by using a good capacitor connected in parallel with the capacitor under test and observing the resulting effect.
- Leaky capacitors will often have a decreased voltage across their terminals.

# 6.7.6 Digital ECL ICs

- 1. The device is operating correctly if the output high state is -0.81 to -0.96V and low state is -1.65 to -1.85V.
- 2. The input must show the same two levels as in step 1.

# SECTION PARTS AND SCHEMATICS

## 7.1 DRAWINGS

The following assembly drawings (with parts lists) and schematics are in the arrangement shown below.

## 7.2 ADDENDA

Under Wavetek's product improvement program, the latest electronic designs and circuits are incorporated into each Wavetek instrument as quickly as development and testing permit. Because of the time needed to compose and print instruction manuals, it is not always possible to include the most recent changes in the initial printing. Whenever this occurs, addendum pages are prepared to sumarize the changes made

and are inserted immediately inside the rear cover. If no such pages exist, the manual is correct as printed.

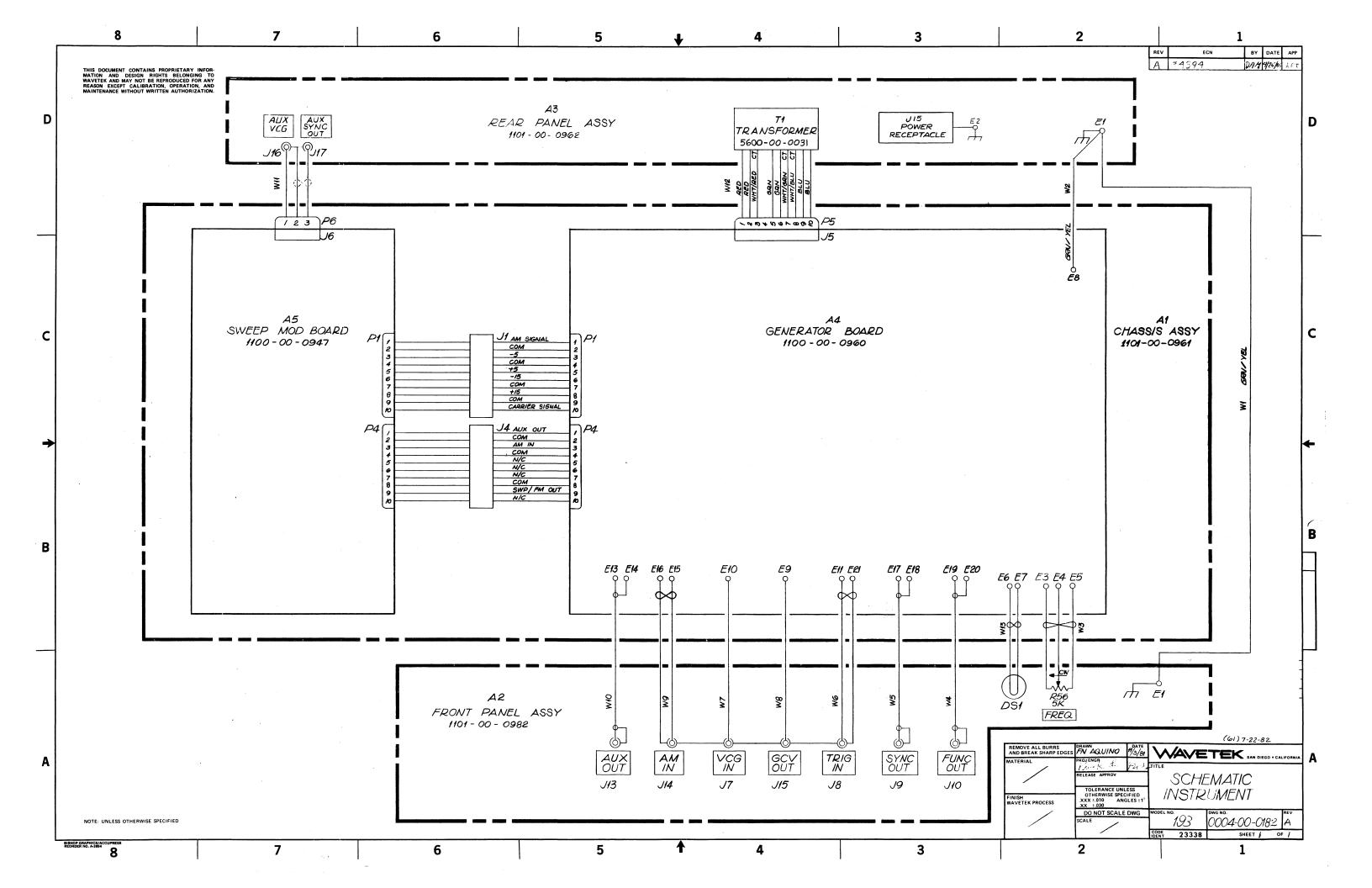
#### 7.3 ORDERING PARTS

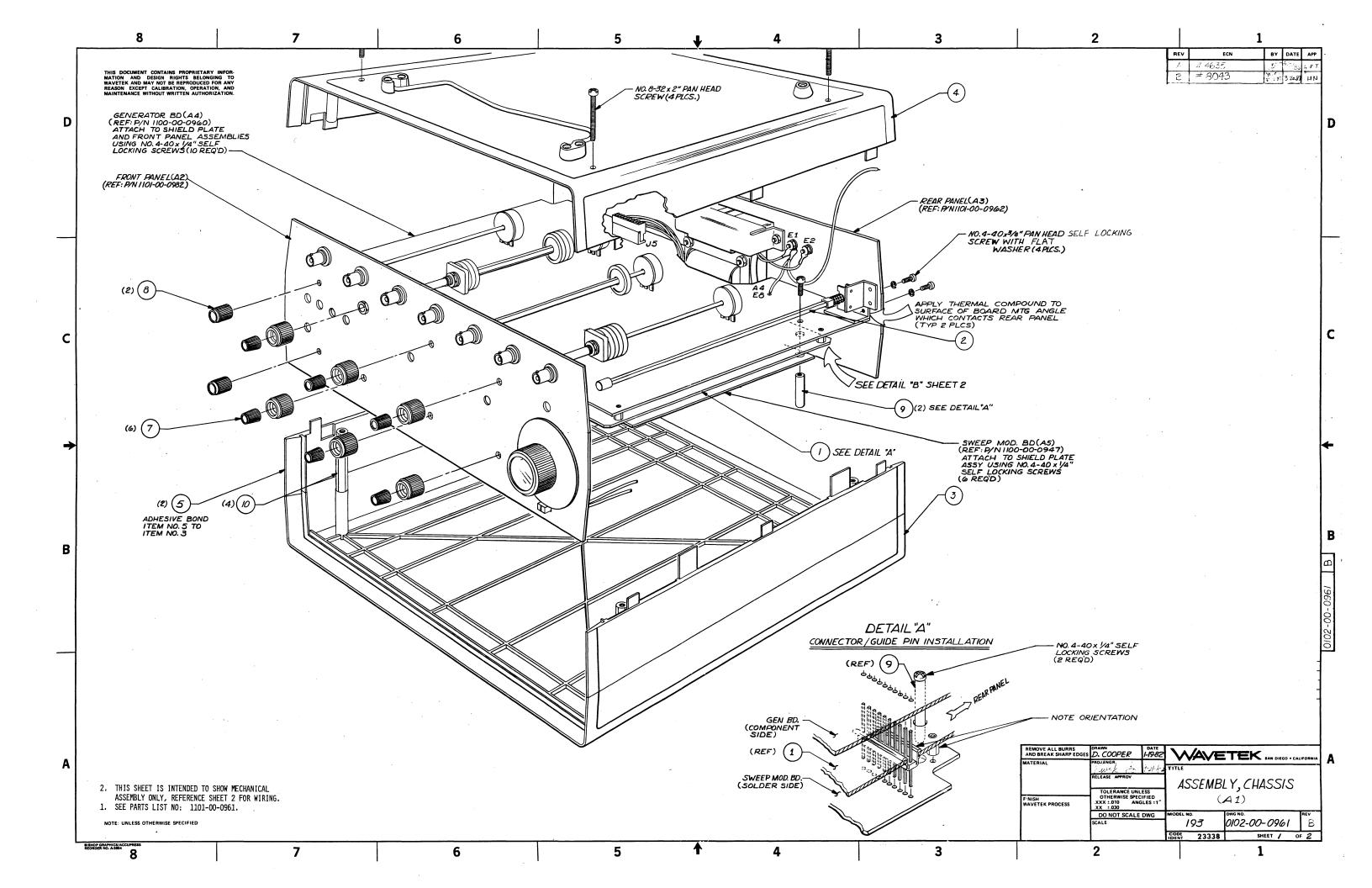
When ordering spare parts, please specify part number, circuit reference, board, serial number of unit, and, if applicable, the function performed.

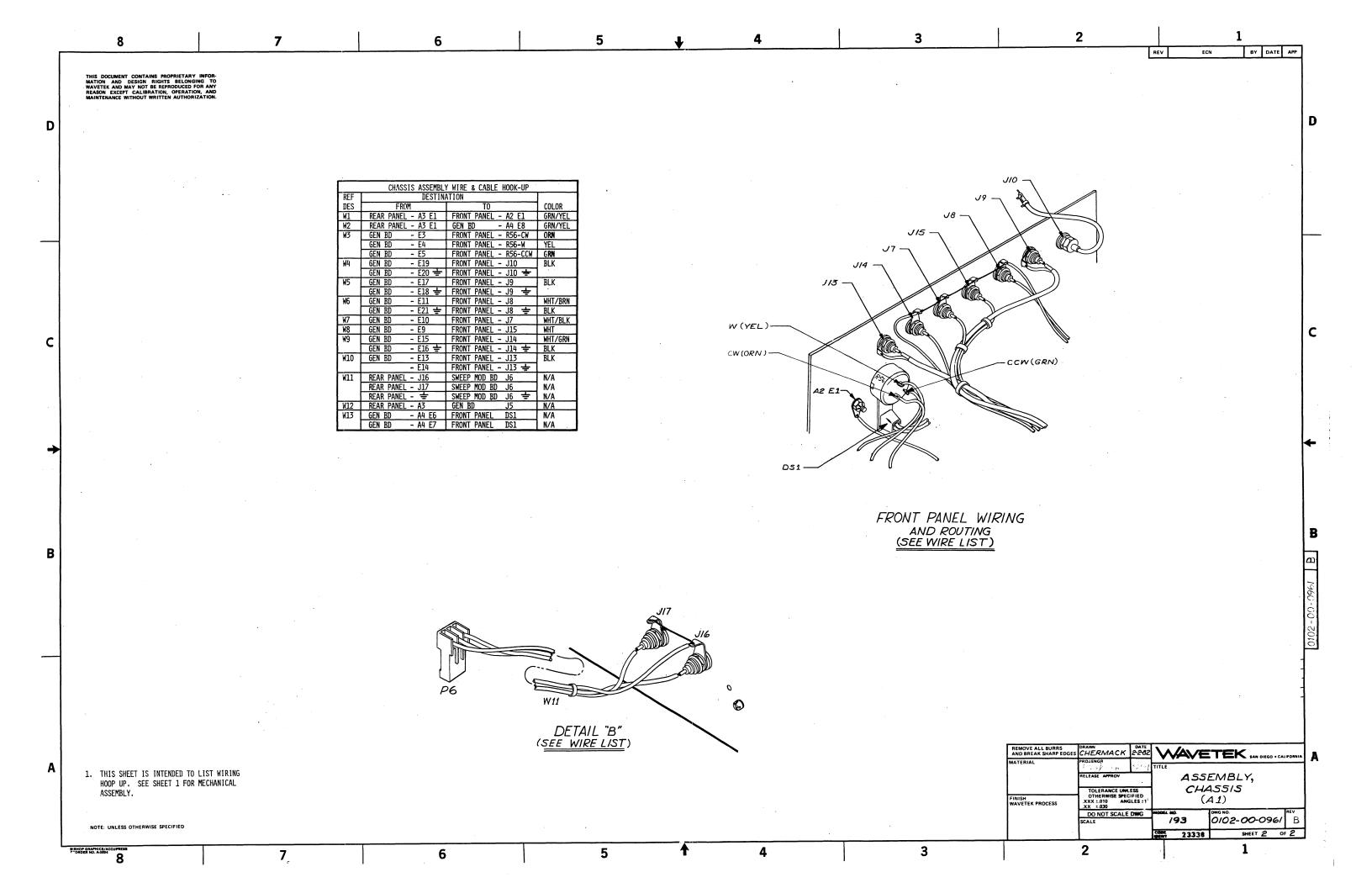
#### NOTE

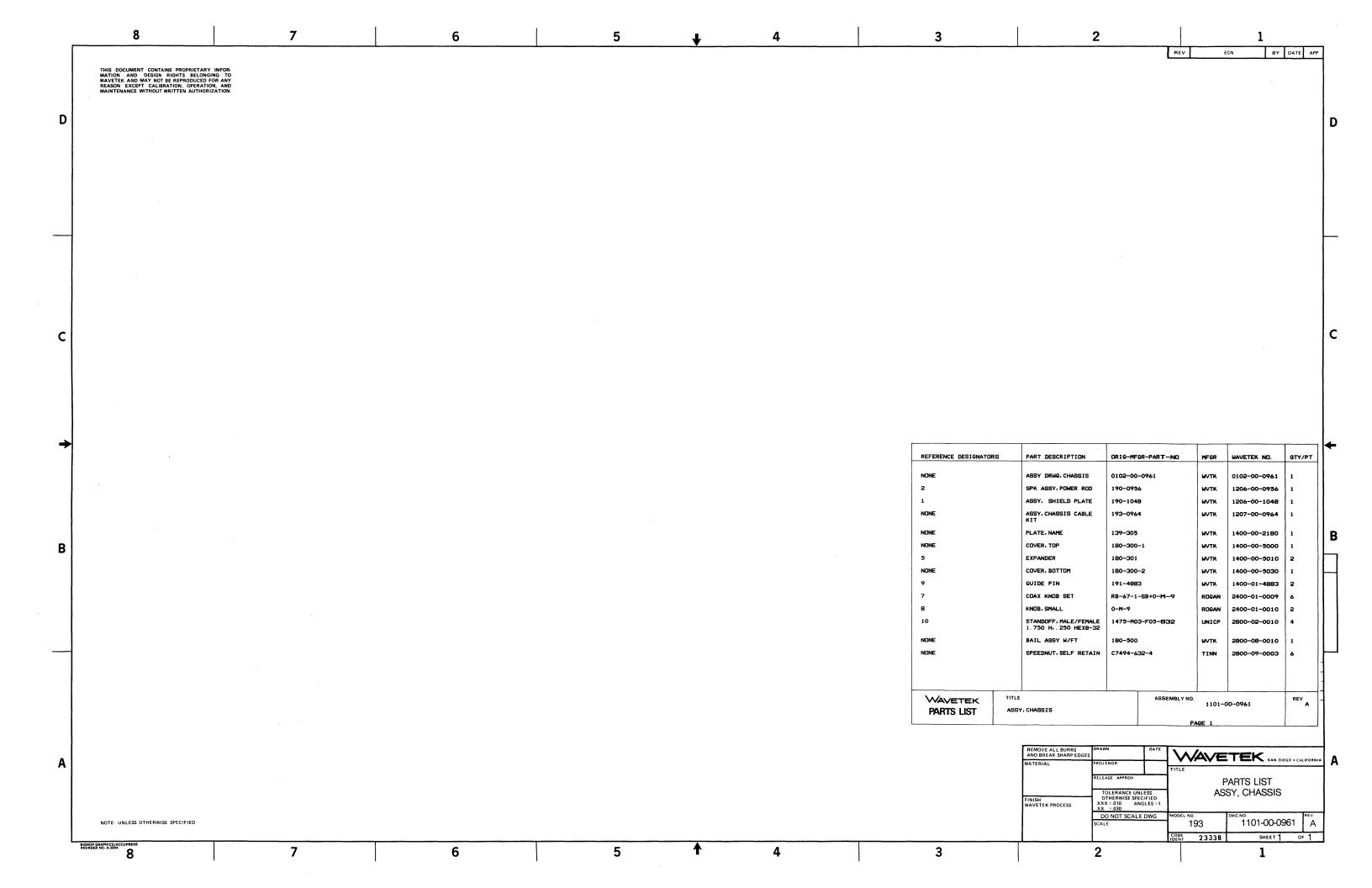
An assembly drawing number is not necessarily the assembly part number. However, the assembly parts list number is the assembly part number.

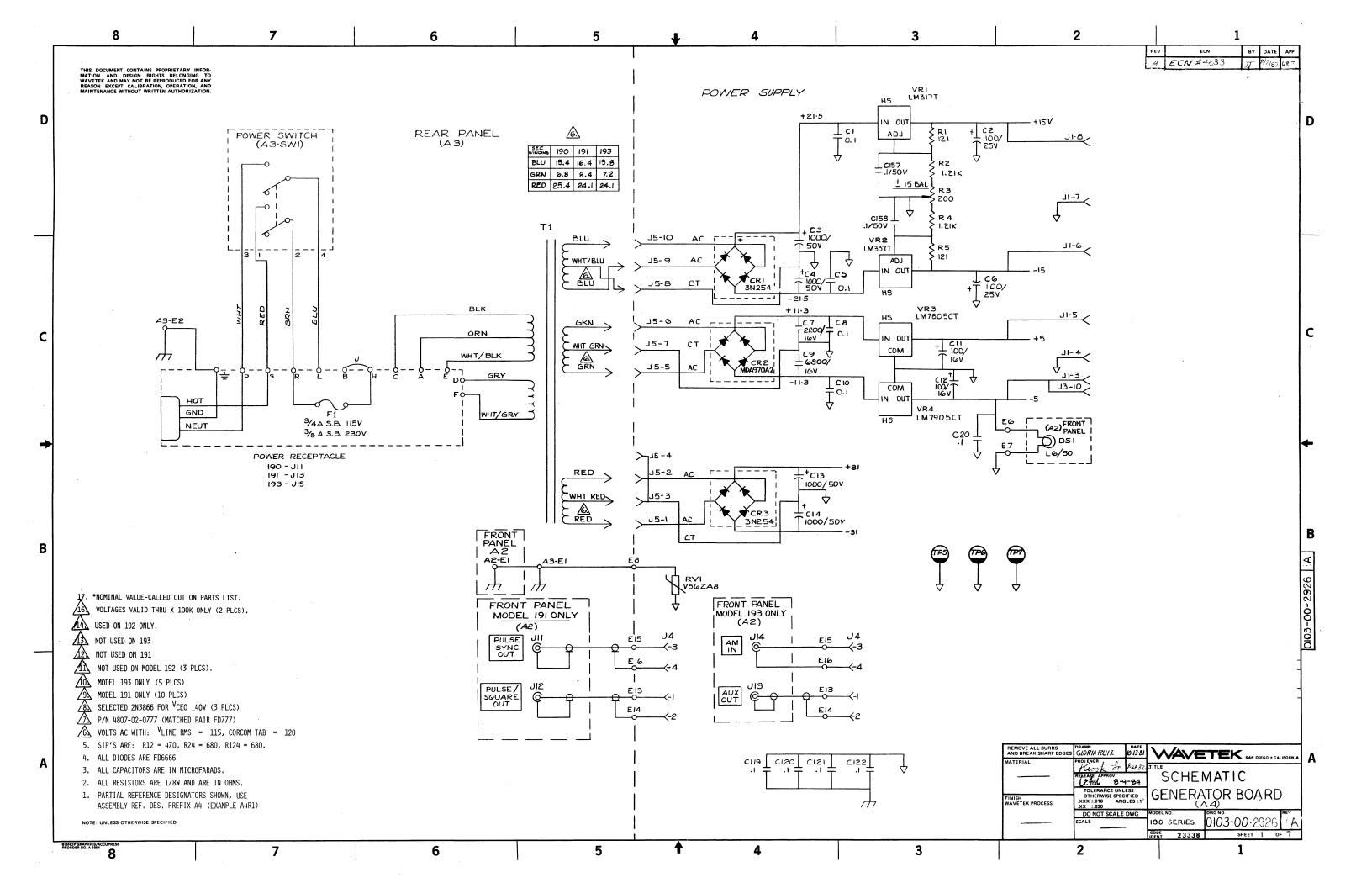
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Instrument Schematic	0004-00-0182
Chassis Assembly	0102-00-0961
Chassis Parts Lists	1101-00-0961
Generator Board Schematic	0103-00-2926
Generator Board Assembly	1100-00-0834
Generator Board Parts List	1100-00-0960
Generator Board Switch Detent	0102-00-0963
Generator Board Switch Parts List	1202-00-0963
Sweep/Mod Board Schematic	0103-00-0947
Sweep/Mod Board Assembly	1100-00-0947
Sweep/Mod Board Parts List	1100-00-0947
Sweep/Mod Switch Detent	0102-00-0965
Sweep/Mod Switch Parts List	1202-00-0965
Rear Panel Assembly	0102-00-0962
Rear Panel Parts List	1101-00-0962
Front Panel Assembly	0102-00-0982
Front Panel Parts List	1101-00-0982

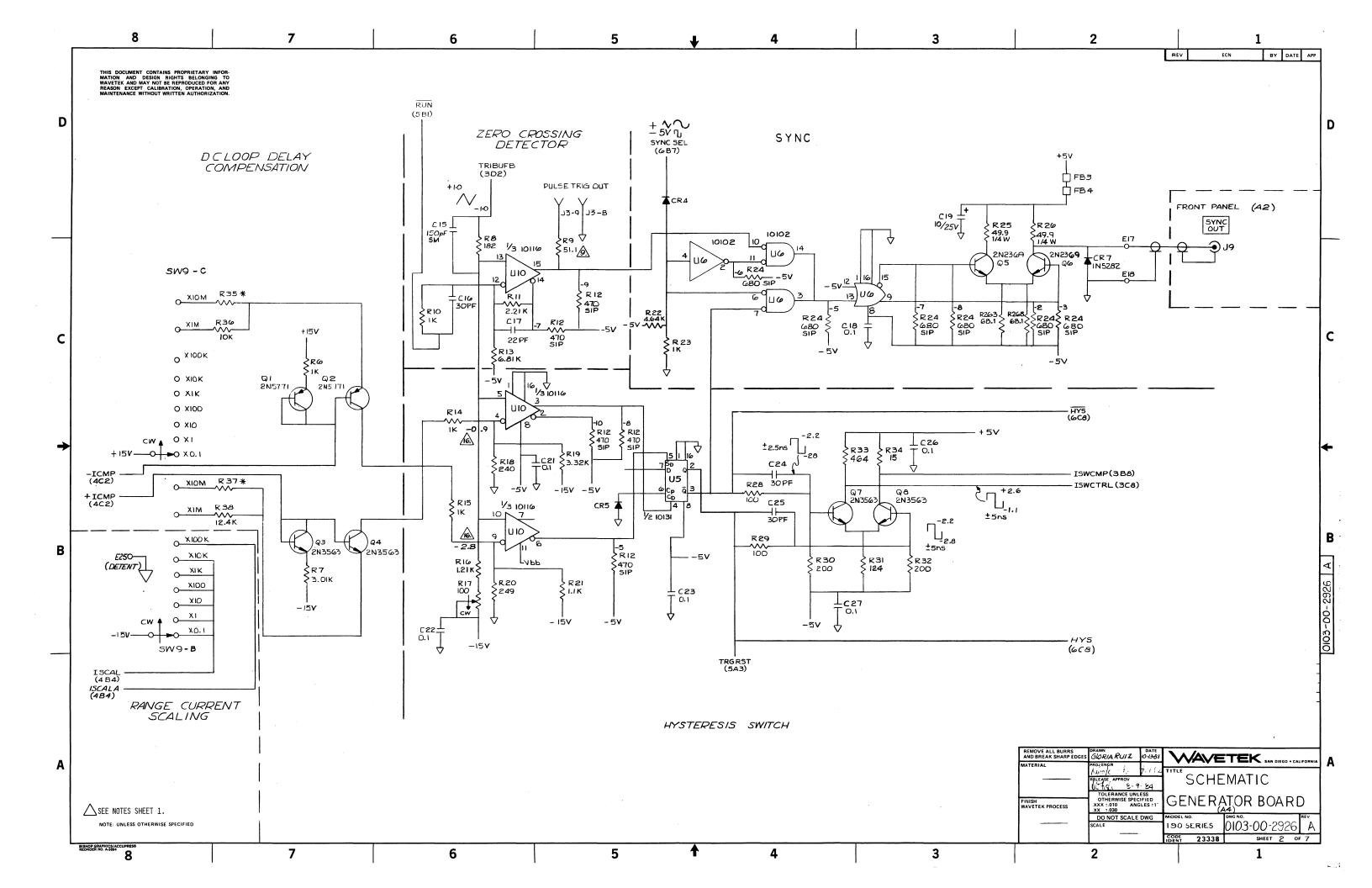


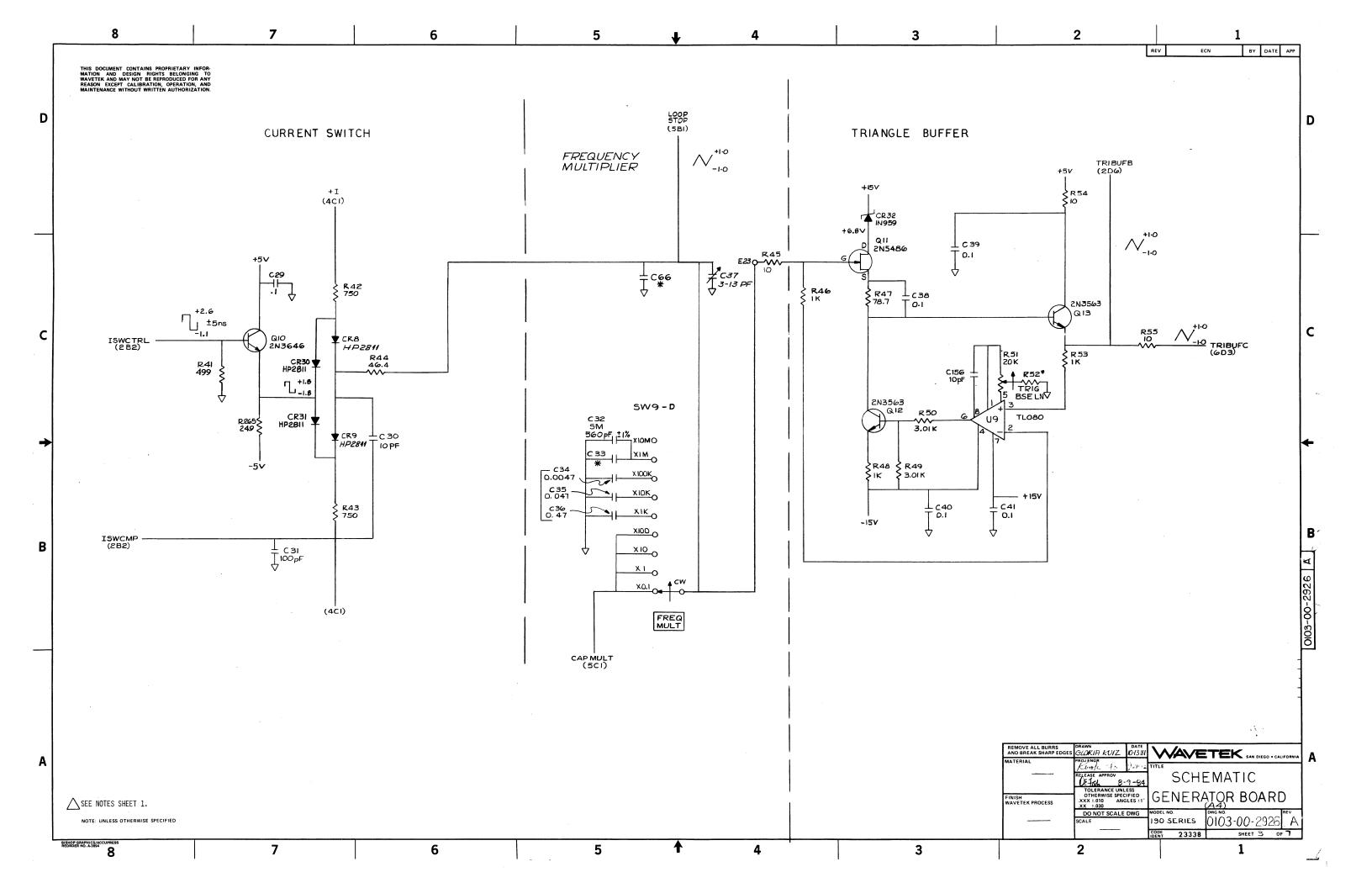


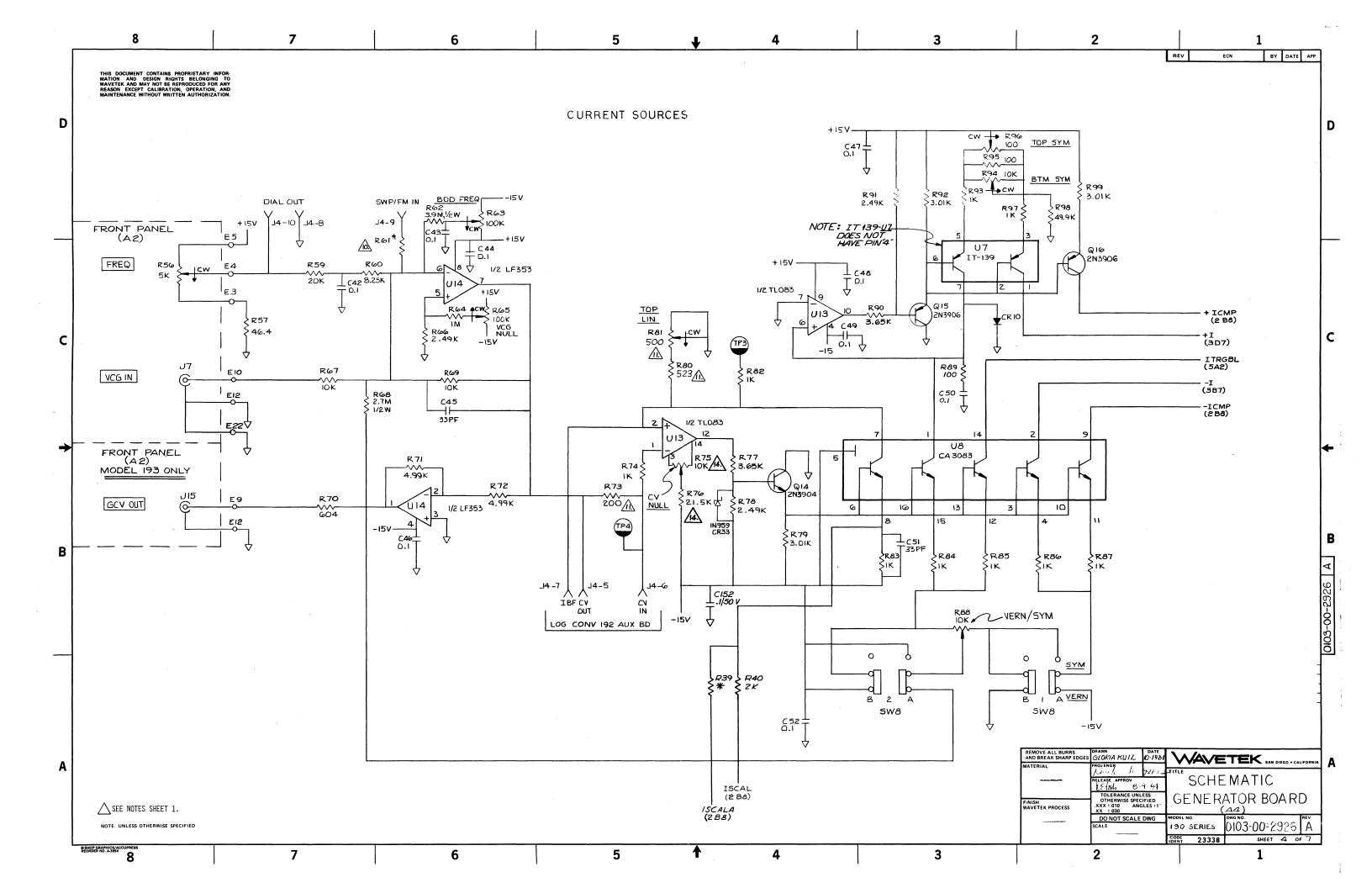


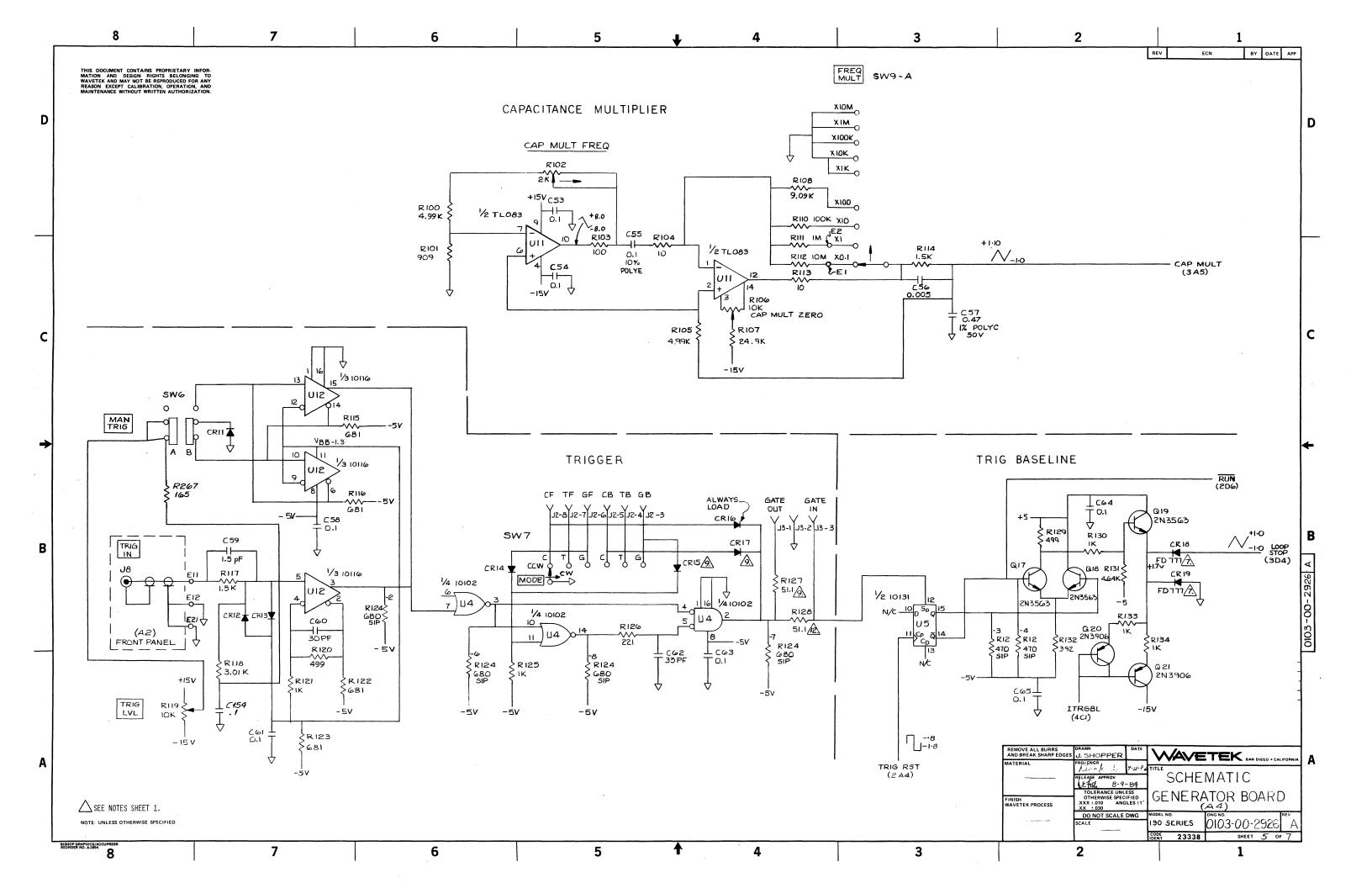


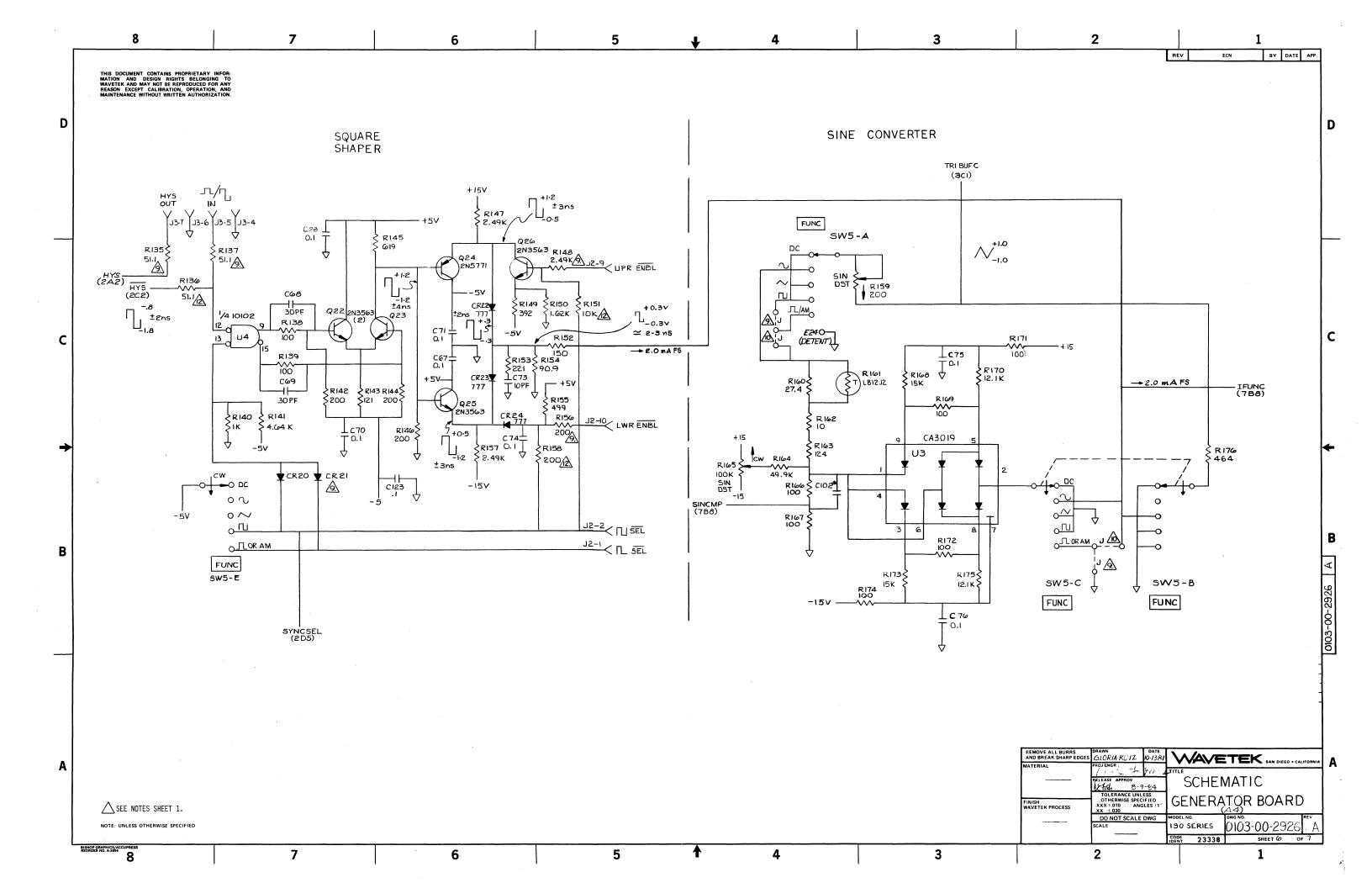


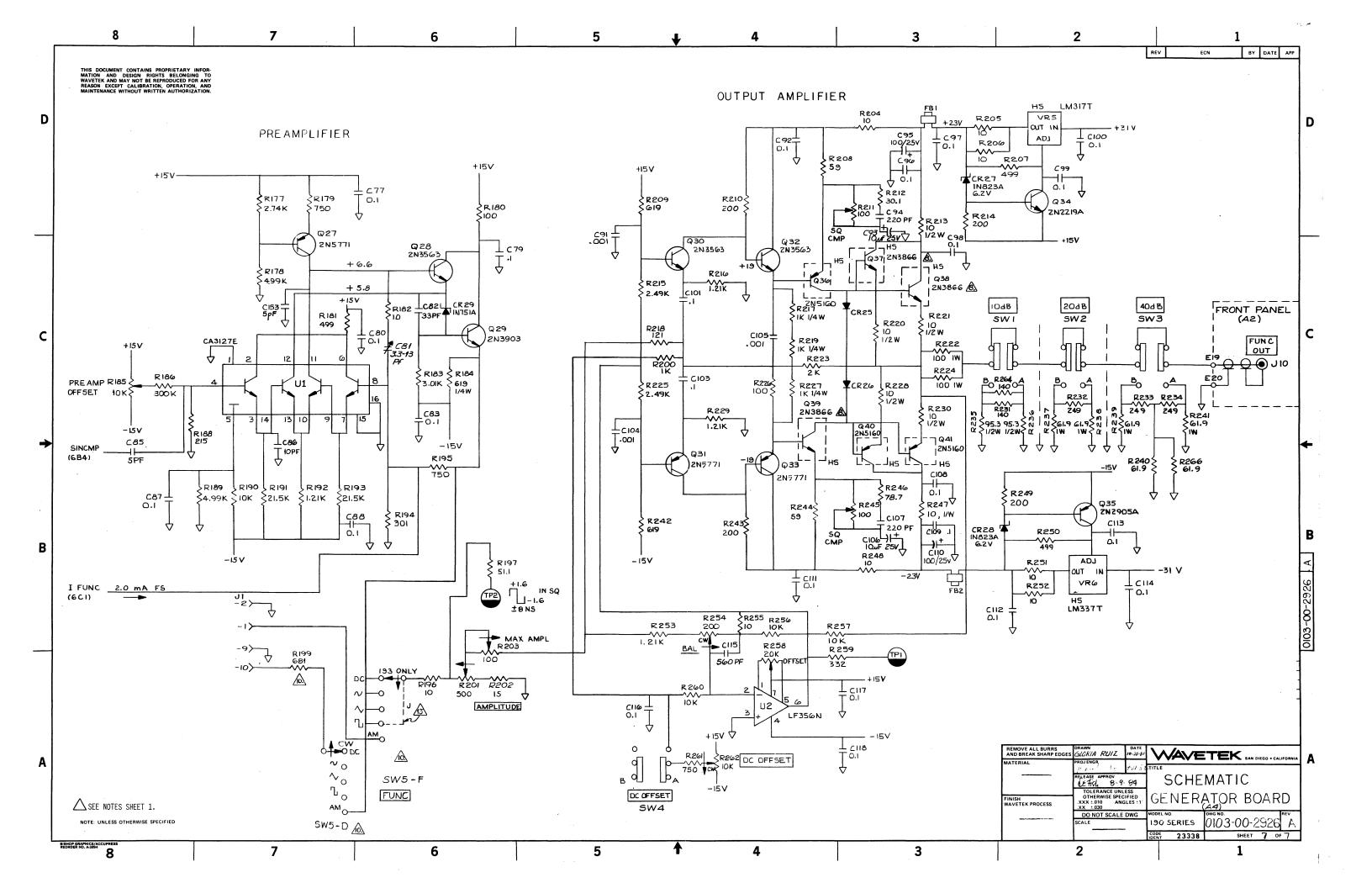


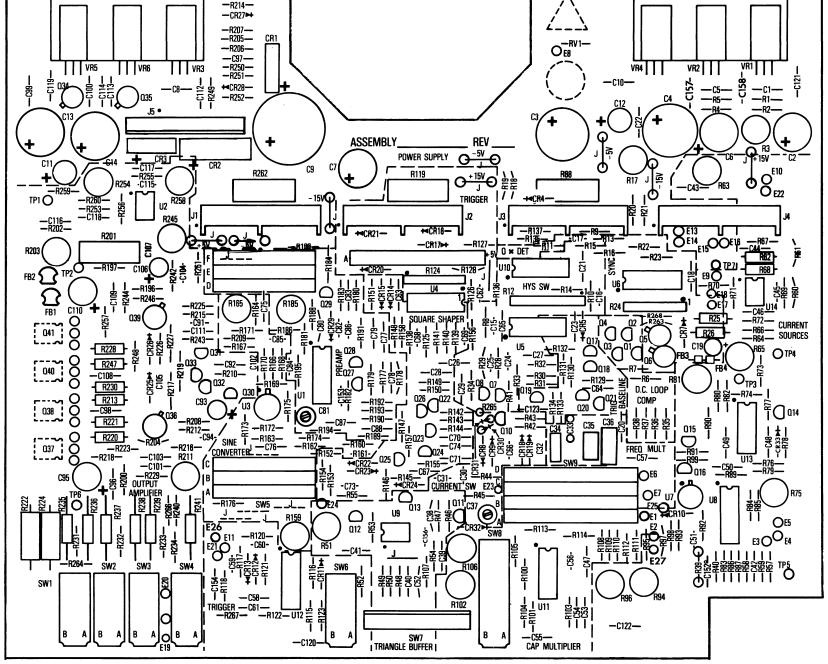










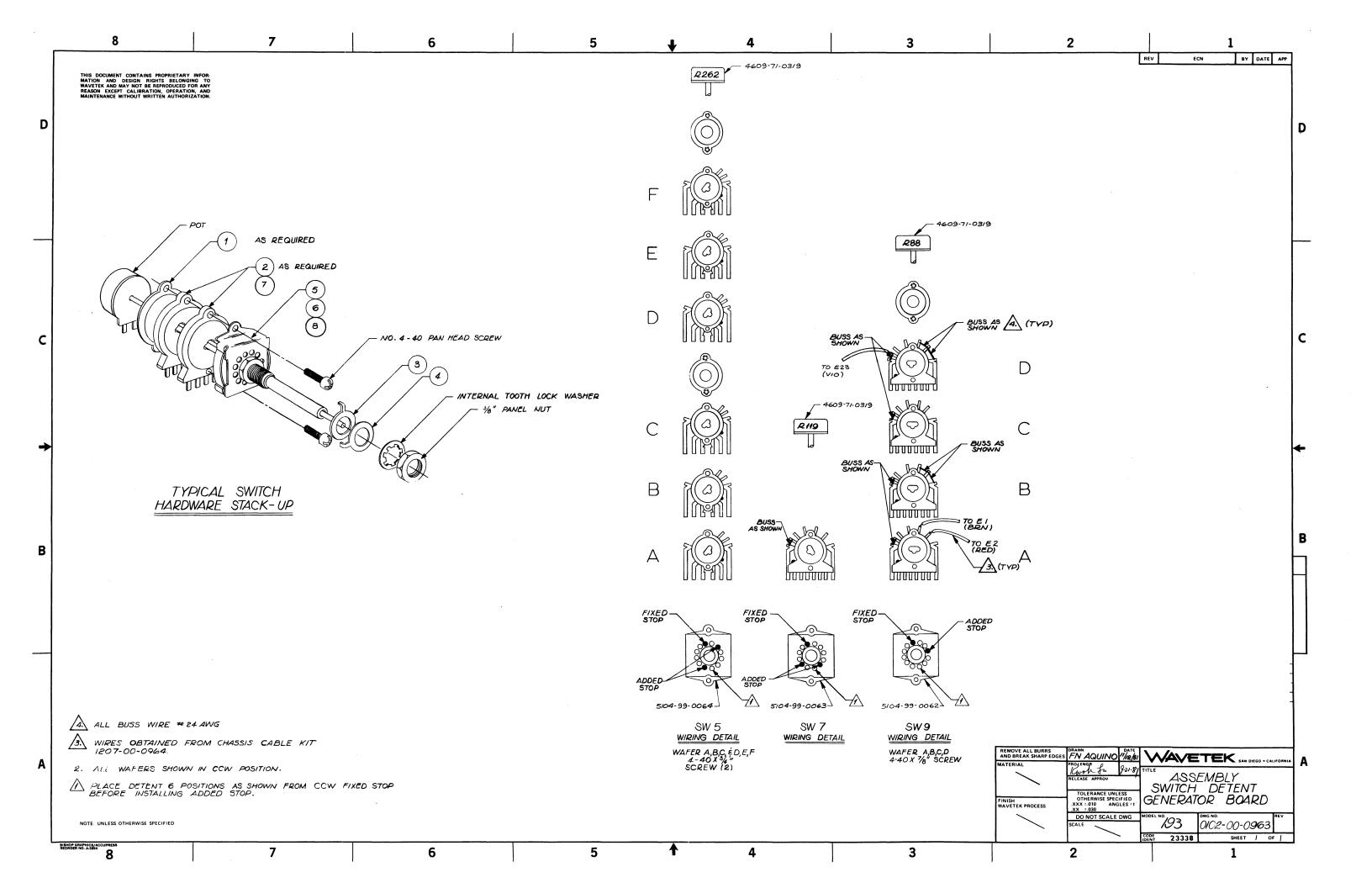


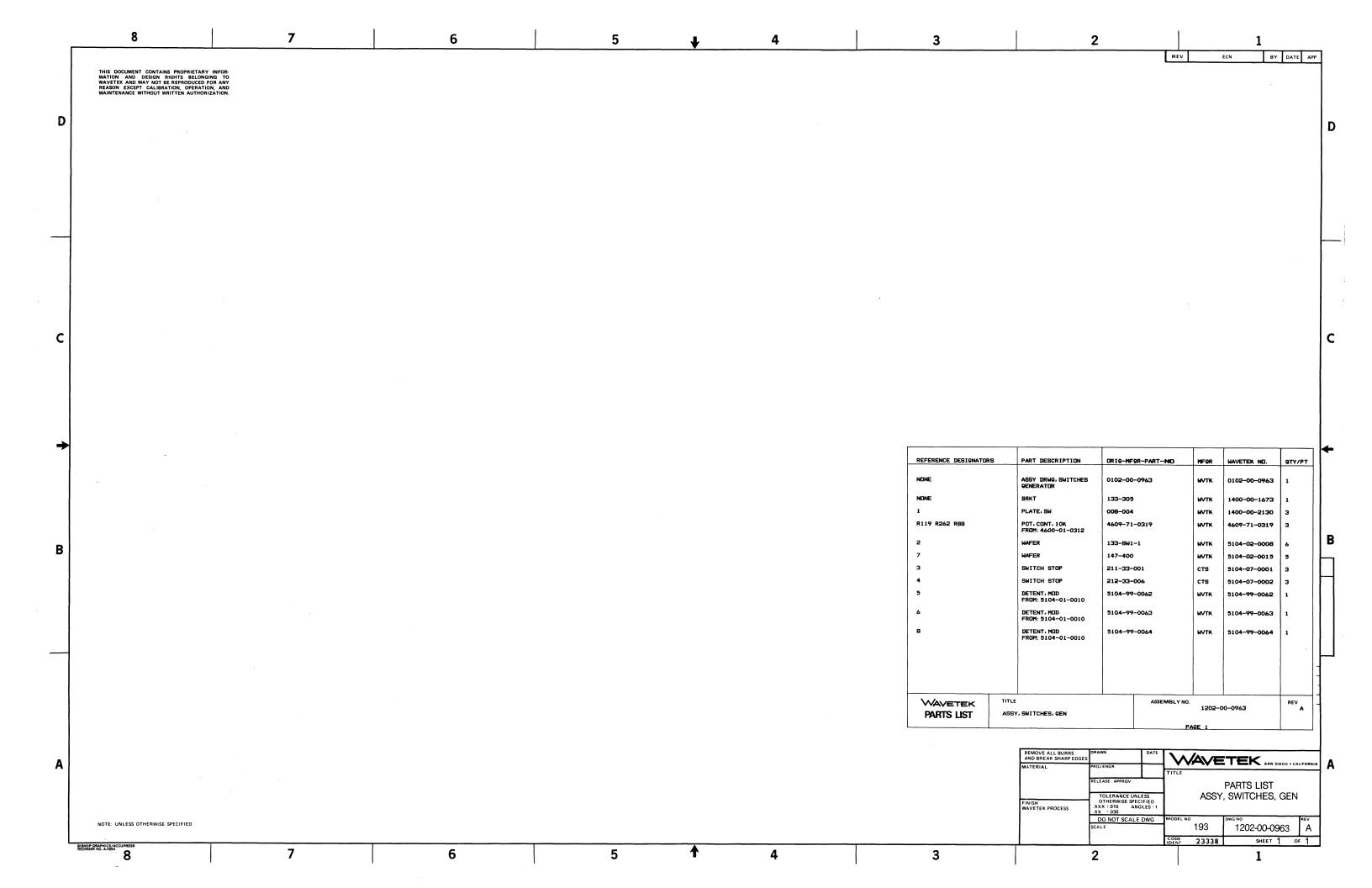
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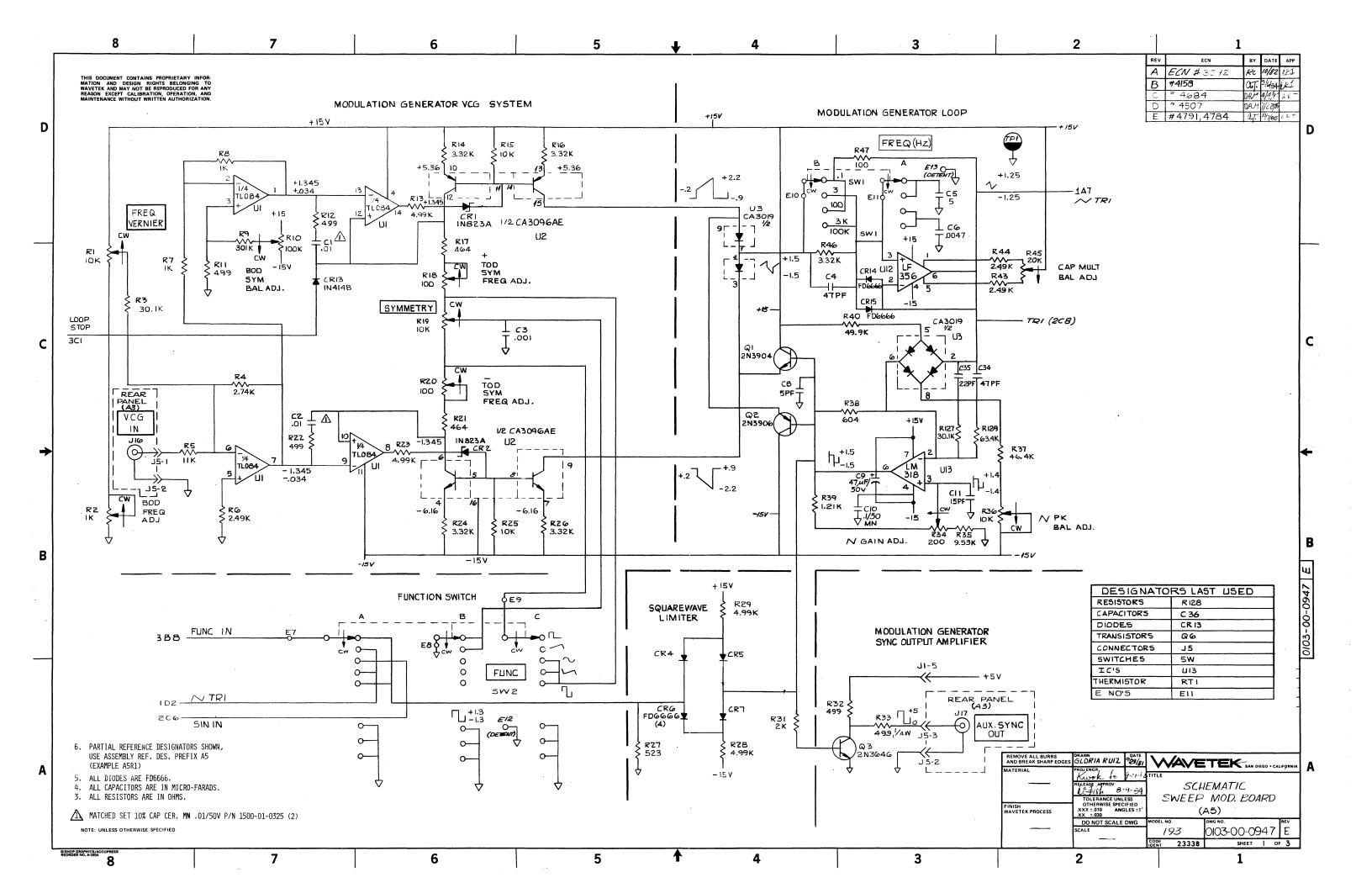
AND BREAK SHARP EDGES	DRAWN	DATE		ΔVE	TEK SAN DIEGO.	CALIEDBAIA
MATERIAL	PROJ ENGR		TITLE		The same state of	CALIFORNIA
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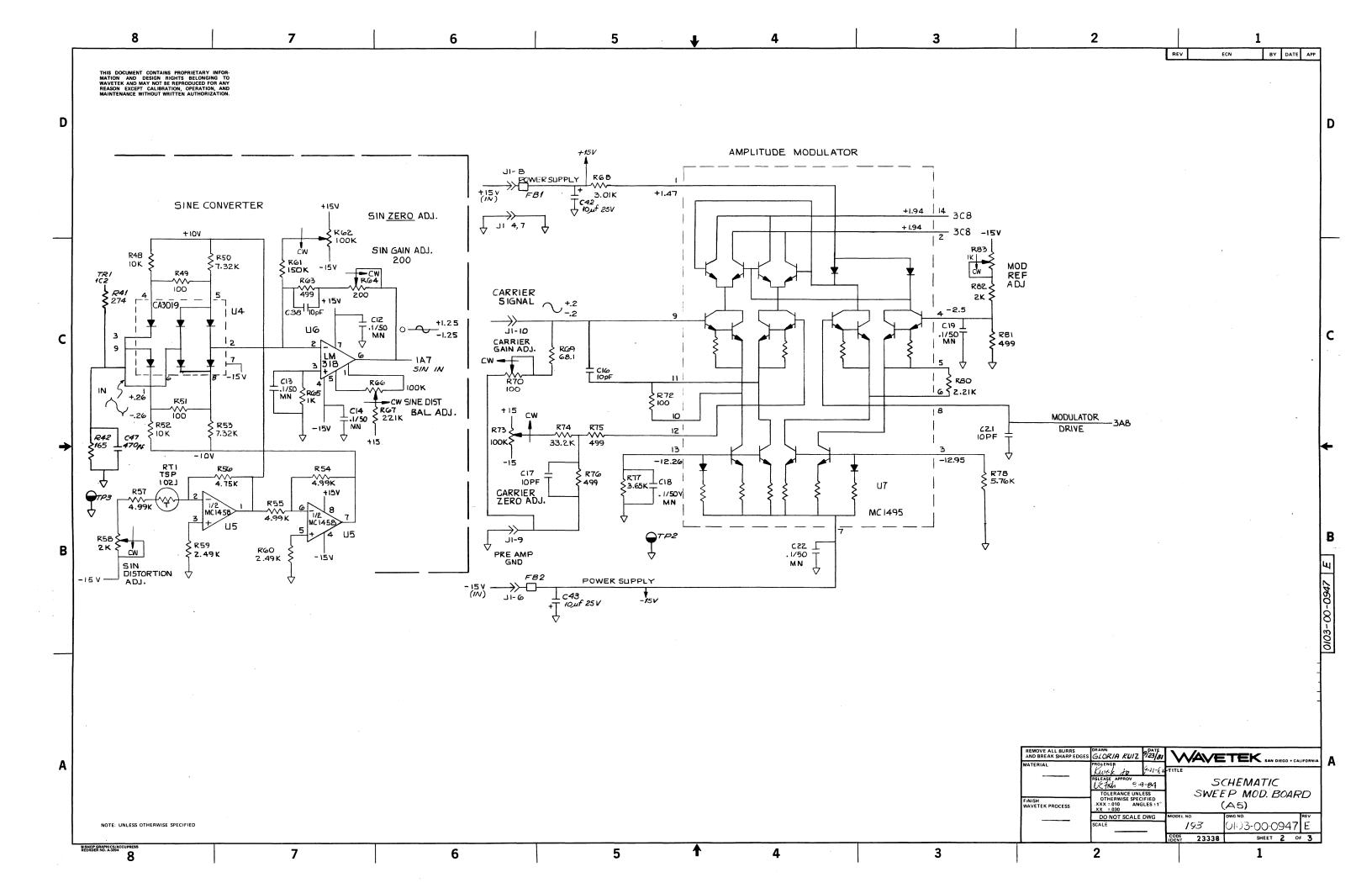
8 6 5 3 2 BY DATE APP ECN THIS DOCUMENT CONTAINS PROPRIETARY INFOR-MATION AND DESIGN RIGHTS BELONGING TO WAVETEK AND MAY NOT BE REPRODUCED FOR AMY REASON EXCEPT CALIBRATION, OPERATION, AND MAINTENANCE WITHOUT WRITTEN BUTHORIZATION. D D REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MEGR-PART-NO MEGR WAVETEK NO QTY/PT FERENCE DESIGNATORS PART DESCRIPTION ORIG-MFGR-PART-NO WAVETEK NO. QTY/PT REFERENCE DESIGNATORS QTY/PT NONE ASSY DRWG, GENERATOR 0101-00-0960 RADIAL LEAD, SP . 50 R134 R14 R140 R15 R200 R23 R46 R48 R53 R6 R74 R82 R83 R84 R85 R86 R87 R93 R97 NONE 193-0963 WVTK 1202-00-0963 C55 CAP, MYLAR, . 1MF, 100V 225P10491WD3 SPRAG 1500-41-0444 NONE SPK ASSY, XSISTER MNTG BRACKET R151 R190 R256 R257 R260 R36 R67 R69 190-0996 WUTK 1204-00-0994 C34 C5R472F ELPAC 1500-44-7203 CAP, MYLR, . 0047MF, 50V RES. MF, 1/8W, 1%, 10K RN55D-1002F 4701-03-1002 C35 C5R473F ELPAC 1500-44-7303 CAP, MYLR, . 047MF, 50V NONE SPK ASSY, BOARD MNTG ANGLE 190-1024 1206-00-1024 R110 RES, MF, 1/8W, 1%, 100K RN55D-1003F TRW 4701-03-1003 C36 C57 CAP, MYLR, . 47MF, 50V C5R474F ELPAC 1500-44-7403 R111 R52T R64 RES, MF, 1/8W, 1%, 1M RN55D-1004F TRW 4701-03-1004 C153 C66T C85 CAP, CER, 5PF, 1KV DD-050 CRL 1500-00-5011 C37 C81 VARI, 3. 5-13PF, 250V 75-TRIKO-02 3, 5/13Pf TRIKO 500-51-3000 R104 R113 R162 R182 R196 RES, MF, 1/8W, 1%, 10 RN55D-10R0 4701-03-1009 C156 C30 C73 C86 CAP, CER, 10PF, 1KV DD-100 CRL 1500-01-0011 GENERATOR BOARD 190-0834 700-00-0834 R204 R205 R206 R248 R251 R252 R255 R45 R54 R55 CAP, CER, 100PF, 1KV DD-101 1500-01-0111 J5 HEADER 1-640386-0 2100-02-0079 R21 RES, MF, 1/8W, 1%, 1. 1K RN55D-1101F TRM 4701-03-1101 C104 C105 C91 CAP, CER, . 001MF, 1KV DD-102 1500-01-0211 J1 J4 CONN. BOTTOM ENTRY, PC 09-52-3102 MOLEX 2100-02-0128 R1 R143 R218 R5 RES, MF, 1/8W, 1%, 121 RN55D-1210F TRW 4701-03-1210 C1 C10 C100 C101 C103 C108 CAP, CER, MON. 1MF, 50V CACO3Z5U104Z050A CORNG 1500-01-0405 SOCKET, PIN NS-430-25 ROBNU 2100-03-0064 C109 C111 C112 C113 C114 C116 C117 C118 C119 C120 R16 R192 R2 R216 R229 R253 RES, MF, 1/8W, 1%, 1. 21K RN55D-1211F 4701-03-1211 2000B1 USECO 2100-05-0009 C121 C122 C123 C152 C157 C121 C122 C123 C152 C157 C158 C18 C20 C21 C22 C23 C26 C27 C28 C29 C38 C39 C40 C41 C42 C43 C44 C46 C47 C48 C49 C5 C50 C52 C53 C54 C58 C61 C63 C64 C65 C67 C70 C71 TP1 TP2 TP3 TP4 TP5 TP4 TP7 BUSS BAR STANDOFF 2110-001 ARTWR 2100-05-0024 R170 R175 RES, MF, 1/8W, 1%, 12, 1K RN55D-1212F TRW 4701-03-1212 STANDOFF, SWAGE . 187L BR6911SPB-0. 187-34 I VNTR 2800-04-0018 R163 R31 RES, MF, 1/8W, 1%, 124 RN55D-1240F TRW 4701-03-1240 RES, MF, 1/8W, 1%, 12. 4K RN55D-1242F TRW 4701-03-1242 C74 C75 C76 C77 C79 C8 C80 TRANSIPAD 10123N 2800-11-0003 C83 C87 C88 C92 C96 C97 C9 R231 R264 RES, MF, 1/8W, 1%, 140 RN55D-1400F TRM 4701-03-1400 TRANSIPAD NONE 10160 METRS 2800-11-0004 R152 RES, MF, 1/8W, 1%, 150 RN55D-1500F TRU 4701-03-1500 C102T CAP, CER, 150PF, 1KV DD-151 1500-01-5111 NONE HEATSTNK 204 HAKE 2800-11-000B R114 R117 RES, MF, 1/8W, 1%, 1. 5K RN55D-1501F 4701-03-1501 WAVETEK WAVETEK WAVETEK ASSEMBLY NO. REV M 1100-00-0960 1100-00-0960 1100-00-0960 PARTS LIST PCA, GENERATOR PCA, GENERATOR PARTS LIST PARTS LIST PCA. GENERATOR PACE 1 REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFGR-PART-NO WAVETEK NO. QTY/PT MFCR REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFGR-PART-NO WAVETEK NO. QTY/PT REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFCR-PART-NO MFOR AVETEK NO. QTY/PT C59 CAP, CER, 1. 5PF, 1KV 10TCC-V15 SPRAC 1500-01-5507 FERRITE BEAD FB3 FB4 56-590-65/3E FERRX 3100-00-0001 R168 R173 RES, MF, 1/8W, 1%, 15K RN55D-1502F TRW 4701-03-1502 C17 CAP, CER, 22PF, 1KV DD-220 1500-02-2011 FB1 FB2 BALUN CORE 2873000902 FARIT 3100-00-0002 R202 R34 RES, MF, 1/8W, 1%, 15 RN55D-15R0F TRW 4701-03-1509 C107 C94 DD-221 CRL 1500-02-2111 CAP, CER, 220PF, 1KV R17 R203 R211 R245 R96 POT, TRIM, 100 91AR100 BECK 4600-01-0103 R150 RES, MF, 1/8W, 1%, 1. 62K RN55D-1621F TRW 4701-03-1621 C16 C24 C25 C60 C68 C69 CAP, CER, 30PF, 1KV DD-300 CRL 1500-03-0001 R106 R185 R94 POT, TRIM, 10K 91AR10K BECK 1600-01-0315 R267 RES. MF. 1/8W. 1%. 165 RN55D-1450F TRM 4701-03-1650 C45 C51 C62 C82 CAP, CER, 33PF, 1KV DD-330 CRL 500-03-3011 R165 R63 R65 POT, TRIM, 100K 91AR100K BECK 4600-01-0402 RES, MF, 1/8W, 1%, 182 RN55D-1820F TRW 4701-03-1820 CK-502 1500-05-0210 CAP, CER, . 005MF, 50V R159 R254 R3 POT, TRIM, 200 91AR200 BECK 4600-02-0101 R61T RES, MF, 1/8W, 1%, 19. 6K RN55D-1962 4701-03-1962 C115 CAP, CER, 560PF, 1KV DD-561SLL CRL 1500-05-6101 R102 POT, TRIM, 2K 91AR2K BECK 1600-02-0201 R142 R144 R146 R158 R210 R214 R243 R249 R30 R32 R73 RN55D-2000 4701-03-2000 RES, MF, 1/8W, 1%, 200 TRW C15 CAP, MICA, 150PF, 500V DM15-151J ARCO 1500-11-5100 POT, TRIM, 20K 91AR20K BECK 1600-02-0301 СЗЭТ CAP, MICA, 56PF, 500V 1500-15-6000 R81 POT, TRIM, 500 91AR500 BECK 4600-05-0104 R223 R35T R40 RES, MF, 1/8W, 1%, 2K RN550-2001E TRW 4701-03-2001 CAP, MICA, 560PF, 300V CD15FC561F03 CDE 1500-15-6102 R201 POT, CONT, 500 FROM: 4600-05-0105 4609-75-0106 MUTK 4609-75-0106 R59 RES, MF, 1/8W, 1%, 20K RN55D-2002F TRW 4701-03-2002 C106 C19 C93 CAP, ELECT, 10MF/25V CRE SERIES 10/25 CAPAR 1500-31-0002 R213 R220 R221 R228 R230 R247 R188 TRW RES, MF, 1/8W, 1%, 215 RN55D-2150F 4701-03-2150 RES, C, 1/2W, 5%, 10 RC-1/2-100J STKPL 4700-25-0100 R191 R193 RES. MF, 1/8W, 1%, 21. 5K RN55D-2152F TRU 4701-03-2152 C11 C110 C12 C2 C6 C95 CAP, ELECT, 100MF, 25V RADIAL LEAD, SP . 20 ULB1V101M 1500-31-0102 NICH R68 RES, C, 1/2W, 5%, 2, 7M RC-1/2-275J STKPL 4700-25-2704 R124 R153 RES, MF, 1/8W, 1%, 221 RN55D-2210F TRW 4701-03-2210 C13 C14 C3 C4 CAP, ELECT, 1000MF/50V RADIAL LEAD, SP . 30 CRE SERIES 1000/50 CAPAR 1500-31-0203 RAD RES, C, 1/2W, 5%, 3, 9M RC-1/2-395J STKPI 4700-25-3904 R11 R37T RES, MF, 1/8W, 1%, 2. 21K RN55D-2211F 4701-03-2211 4701-03-1000 RES, MF, 1/8W, 1%, 100 RN55D-1000F R39T RES, MF, 1/8W, 1%, 2. 37K TRW RN55D-2371F 4701-03-2371 C7 CAP, ELECT, 2200MF, 16V RADIAL LEAD, SP . 30 CRE SERIES 2200/1A CAPAR 1500-32-2201 R18 RES. MF. 1/8W. 1%, 240 RN55D-2400F MEPCO 4701-03-2400 CAP, ELECT, 6800MF, 16V CAPAR 1500-36-8201 R10 R121 R125 R130 R133 RES. MF. 1/8W. 1%. 1K RN55D-1001F 4701-03-1001 R20 R232 R233 R234 R265 RES, MF, 1/8W, 1%, 249 RN55D-2490F 4701-03-2490 WAVETEK TITLE ASSEMBLY NO. TITLE ASSEMBLY NO. REV M WAVETEK WAVETEK 1100-00-0960 1100-00-0960 1100-00-0960 PCA, GENERATOR PARTS LIST PCA, GENERATOR PARTS LIST **PARTS LIST** PCA, GENERATOR PAGE 2 REMOVE ALL BURRS WAVETEK SAN DIEGO + CALIFORI PARTS LIST PCA, GENERATOR INISH NAVETEK PROCESS DO NOT SCALE DWG NOTE: UNLESS OTHERWISE SPECIFIED 193 1100-00-0960 М or 2 23338 SHEET 1 7 5 1 6 4 8 3 2 1

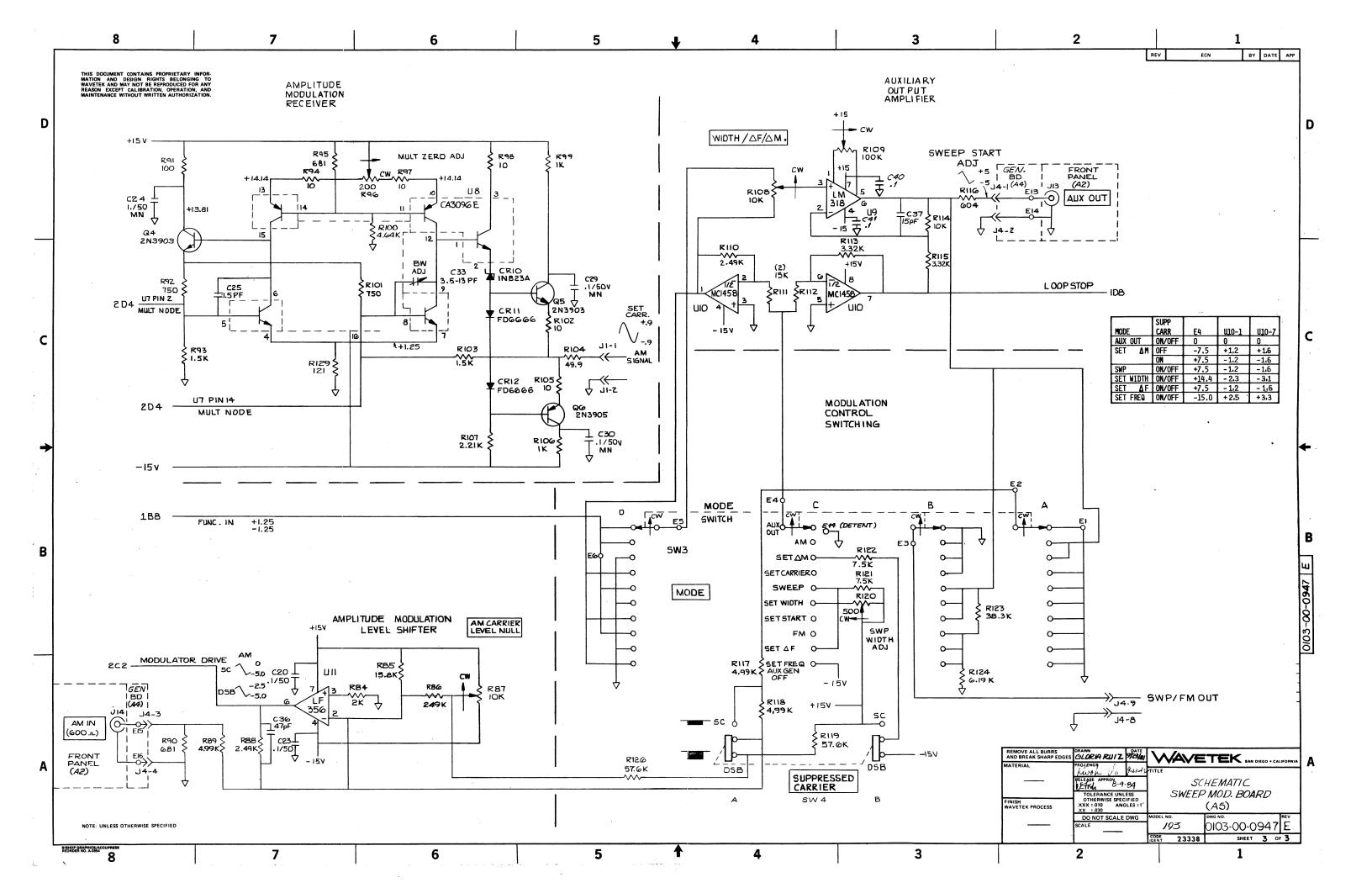
8 7 2 5 3 6 REV ECN BY DATE APP THIS DOCUMENT CONTAINS PROPRIETARY INFORMATION AND DESIGN RIGHTS BELONGING TO MAYETER AND MAY NOT BE REPRODUCED FOR ANY REASON EXCEPT CALIBRATION, OPERATION, AND MAINTENANCE WITHOUT WRITTEN AUTHORIZATION. D D REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFGR-PART-NO WAVETEK NO. QTY/PT REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFGR-PART-NO WAVETEK NO. QTY/PT MFGR MFGR REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MEGR-PART-NO MEGR WAVETEK NO. QTY/PT RES, MF, 1/8W, 1%, 909 RN55D-9090F 4701-03<del>-9</del>090 2N5160-18 4901-05-1600 R147 R157 R215 R225 R66 R78 RES, MF, 1/8W, 1%, 2. 49K RN55D-2491F 4701-03-2491 R108 RES, MF, 1/8W, 1%, 9. 09K RN55D-9091F 4701-03-9091 2N5486 Q11 MOT 4901-05-4860 TRANS RES, MF, 1/8W, 1%, 24. 9K RN55D-2492F 4701-03-2492 4701-03-9099 TRW R154 RES, MF, 1/8W, 1%, 90, 9 RN55D-90R9F TRW Q1 Q2 Q24 Q27 Q31 Q33 TRANS 2N5771 NSC 4901-05-7710 R177 RES, MF, 1/8W, 1%, 2, 74K RN55D-2741F 4701-03-2741 TRW R217 R219 R227 RES, MF, 1/4W, 1%, 1K RN60D1001F TRW 4701-13-1001 U7 TRANS IT 139 INTSL 4902-00-1390 R160 RES, MF, 1/8W, 1%, 27, 4 RN55D-27R4F TRW 4701-03-2749 R25 R26 RES, MF, 1/4W, 1%, 49. 9 TRW 4701-13-4999 937 938 939 4998-00-0051 WVTK 4998-00-0051 RES, MF, 1/8W, 1%, 301 4701-03-3010 RES, MF, 1/4W, 1%, 619 TRW 4701-13-6190 R184 RN60D-6190F SWITCH SU1 SU2 SU3 SU4 5102-00-0009 HUTK 5102-00-0009 R118 R183 R49 R50 R7 R79 RES, MF, 1/8W, 1%, 3. 01K 4701-03-3011 RN55D-3011F TRW TRM 4701-23-9539 P235 P23A RES, MF, 1/2W, 1%, 95, 3 PN450-9583E SWITCH, 2PDT, MOM. 5102-00-0010 5102-00-0010 R222 R224 RES, MF, 1W, 1%, 100 4701-33-1000 RN70D-1000F R186 RES, MF, 1/8W, 1%, 301K RN55D-3013F TRM 4701-03-3013 SWITCH, 4PDT, P-P 5102-00-0011 WVTK 5102-00-0011 4701-33-6199 R237 R238 R239 R241 RES, MF, 1W, 1%, 61. 9 RN70D-61R9F R212 RES, MF, 1/8W, 1%, 30. 1 RN55D-30R1F 4701-03-3019 NONE BUTTON, CONICAL F01-01 (BLACK) 5103-04-0006 SHADW 4770-00-0009 BOURN R12 RES MODULE 4310R-101-471 R259 RES, MF, 1/8W, 1%, 332 RN55D-3320F 4701-03-3320 TRW THERMISTER R161 155-180FAK-801 FENUL 5300-00-0002 R124 R24 RES MODULE 4308R-101-681 BOURN 4770-00-0031 R19 RES. MF. 1/8W. 17. 3 32K RN55D-3321F TRM 4701-03-3321 119 IC. OP-AMP TLOBOCP 7000-00-8001 CADDO 4799-00-0003 R112 RES, MF, . 6W, 1%, 10M ML-181 959 RES, MF, 1/8W, 1%, 33. 2K RN55D-3322F TRW 4701-03-3322 U11 U13 IC TLOBOCN 7000-00-B300 TI VARISTOR 1899-00-0045 WYTK 4799--00--0048 RV1 R77 R90 RES, MF, 1/8W, 1%, 3, 65K 4701-03-3651 IC U14 LF353N NSC 7000-03-5300 1N751A FAIR 4801-01-0751 CR29 DIODE R132 R149 4701-03-3920 RES, MF, 1/8W, 1%, 392 RN55D-3920F TRW UD IC, OP-AME LF356N NSC 7000-03-5600 CR27 CR28 DIODE, ZENER 6, 2V LN823A MOT 4801-01-0823 R176 R33 RES, MF, 1/8W, 1%, 464 RN55D-4A40F TRU 4701-03-4640 IC CA-3019 7000-30-1900 CR32 CR33 LN959 SIEM 4801-01-0959 DIODE R131 R141 R22 RES, MF, 1/8W, 1%, 4. 64K RN55D-4641F 4701-03-4641 CA3083 7000-30-8300 IC FAIR 1N5282 FAIR 4801-01-5282 CR7 DIODE ASSEMBLY NO. TITLE ASSEMBLY NO. REV M REV M TITLE WAVETEK WAVETEK WAVETEK 1100~00~0960 1100-00-0960 1100-00-0960 PCA, GENERATOR PCA, GENERATOR PARTS LIST PCA, GENERATOR **PARTS LIST** PARTS LIST PAGE 11 REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFGR-PART-NO WAVETEK NO. QTY/PT REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MFCR-PART-NO QTY/PT MFCR MFGR WAVETEK NO. REFERENCE DESIGNATORS PART DESCRIPTION ORIG-MEGR-PART-NO MFGR WAVETEK NO. GTY/PT R44 R57 RES, MF, 1. BW, 1%, 46. 4 RN55D-46R4F TRW 4701-03-4649 CA3127E 000-31-2700 4801-02-0254 R120 R129 R155 R181 R207 R250 R41 RES, MF, 1/8, 1%, 499 RN55D-4990F 4701-03-4990 U4 U6 ECL, NOR, QUAD 21NP MC10102 MOT 8001-01-0200 CR22 CR23 CR24 DIODE, ULTRA FAST FD777 FAIR 4807-02-0777 U10 U12 ECL, RCVR, TRI LINE MC1011AF MOT 8001-01-1600 R100 R105 R178 R189 R71 R72 RES, MF, 1/8W, 1%, 4. 99K 4701-03-4991 FAIR 4807-02-6666 RN55D-4991F CR10 CR11 CR12 CR13 CR14 1N4148 DIODE ECL. D-F/F, DUAL MC10131 8001-01-3100 R164 R98 RFS. MF. 1/84. 17. 49 9K RN55D-4992F TRM 4701-03-4992 CR30 CR31 CRB CR9 DIODE 5082-281 4809-02-2811 R128 R136 R197 RES, MF, 1/8W, 1%, 51. 1 RN55D-51R1F TRW 4701-03-5119 CR18 CR19 DIODE, SET, 2-FD-777 GTY: 2: 4807-02-0777 142-501-5E WVTK 4898-00-0003 RES, MF, 1/8W, 1%, 523 4701-03-5230 R208 R244 RES, MF, 1/8W, 1%, 59 RN55D-59RO 4701-03-5909 TRW BRIDGE ASSY, 4 AME DIODE 4<del>89</del>9-00-0037 CR2 RSA02 R70 RES, MF, 1/8W, 1%, 604 RN55D-A040E TRW 4701-03-6040 Q34 2N22194 NSC 4901-02-2191 R145 R209 R242 RES, MF, 1/8W, 1%, 619 RN55D-6190F TRW 4701-03-6190 MOT Q5 Q6 2N2369/ R240 R266 RES, MF, 1/8W, 1%, 61. 9 4701-03-6199 4901-02-9051 **Q35** NSC TRANS 2N29054 R115 R116 R122 R123 R199 RES, MF, 1/8W, 1%, 681 TRW 4701-03-6810 RN55D-6810F FAIR 4901-03-5630 912 913 917 918 919 922 923 TRANS 2N3563 R13 RES, MF, 1/8W, 1%, 6, 81K RN55D-A811E TRM 4701-03-6811 R263 R268 RES, MF, 1/8W, 1%, 68. 1 4701-03-6819 Q10 2N3646 NSC 4901-03-6460 TRANS R179 R195 R261 R42 R43 RES, MF, 1/8W, 1%, 750 RN55D-7500F 4701-03-7500 4901-03-9030 929 TRANS 2N3903 NSC 4701-03-7879 TRW R246 R47 RES, MF, 1/8W, 1%, 78, 7 RN55D-78R7F Q14 TRANS 2N3904 FAIR 4901-03-9040 RES, MF, 1/8W, 1%, 8. 25K RN55D-8251F 4701-03-8251 Q15 Q16 Q20 Q21 2N3906 FAIR 4901-03-9060 TITLE ASSEMBLY NO. ASSEMBLY NO. ASSEMBLY NO. REV M WAVETEK WAVETEK WAVETEK 1100-00-0960 1100-00-0960 1100--00--0960 PARTS LIST PCA, GENERATOR PARTS LIST PCA, GENERATOR PCA, GENERATOR PARTS LIST PAGE 12 REMOVE ALL BURRS AND BREAK SHARP EDGES WAVETEK SAN DIEGO • CALIFORNI RELEASE APPROV PARTS LIST TOLERANCE UNLESS OTHERWISE SPECIFIED .XXX +.010 ANGLES + .XX +.030 PCA. GENERATOR FINISH WAVETEK PROCESS DO NOT SCALE DWG М NOTE: UNLESS OTHERWISE SPECIFIED 193 1100-00-0960 23338 SHEET 2 2 5 4 3 7 6 1

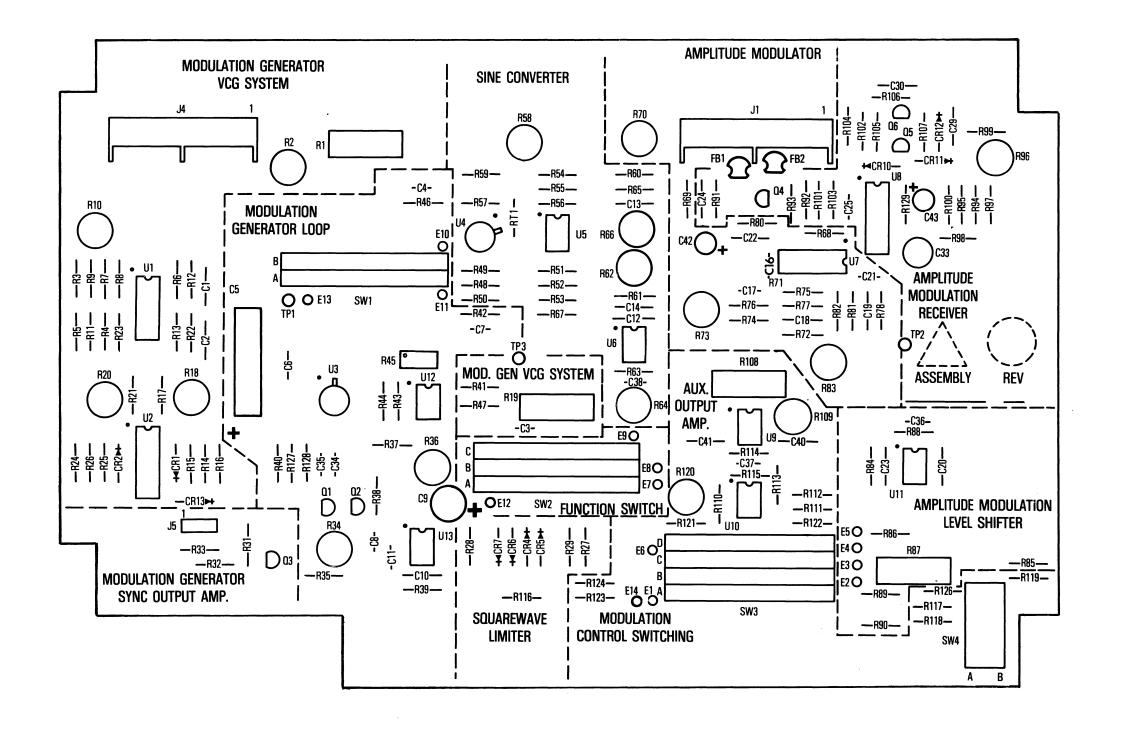












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REMOVE ALL BURRS AND BREAK SHARP EDGES	DRAWN	DATE	V/AV	ETEK SAN DIEGO • CALIFORNIA
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FINISH WAVETEK PROCESS	TOLERANCE UNL OTHERWISE SPEC XXX · 010 ANG XX · 030		SWEEP/MOD BOARD	
	DO NOT SCALE	DWG	MODEL NO 193	1100-00-0947 F
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