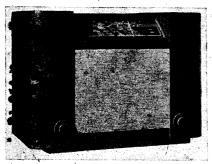
"TRADER" SERVICE SHEET

REVISED ISSUE OF SERVICE SHEET No. 268



The appearance of the table models 747A and 747AX.

IILIPS 747A

747AX AND 698A, 699A RADIOGRAMS

N interesting point in the design of the latest Philips 747A receiver is the use of two separate diodes for AVC purposes in order to overcome a form of distortion sometimes found when only one diode is employed. The receiver is a 4-valve (plus rectifier) AC 3-band superhet suitable for mains of 100-250 V, 50-100 C/S.

A very similar chassis is fitted in the 747AX receiver, but this is arranged so that it can be operated on DC mains via a vibrator unit, and has a different mains input plug.

In addition, a similar chassis is employed in the 698A and 699A radiograms. This Service Sheet was prepared on a 747A of the latest type.

It should be noted that some of the first models which were produced had an ordinary output pentode in place of the double diode pentode and did not use two diodes for AVC, the circuit being considerably different. These models are not covered here.

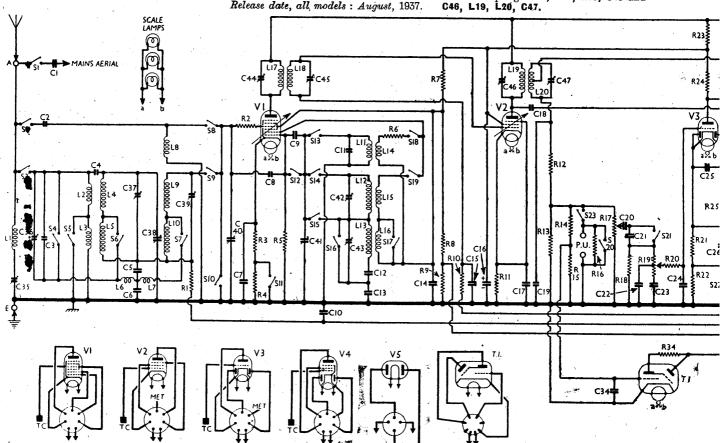
Release date, all models: August, 1937.

CIRCUIT DESCRIPTION

Aerial input on MW and LW via coupling coils L2, L3 and condenser C4 to mixed coupled band-pass filter. Primary coils L4, L5 are tuned by C38; secondaries L9, L10 by C40. Coupling is effected by condensers C6 (MW), C5 (LW) and coils L6, L7, which also form part of a selectivity circuit. On SW, input is via coupling condenser C2 to single-tuned circuit L8, C40. IF filtering by L1, C35 across aerial circuit.

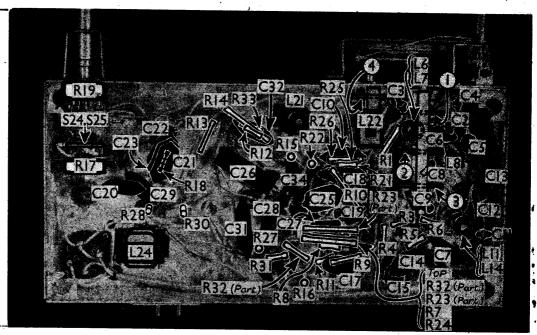
First valve (V1, Mullard metallised FC4) is an octode operating as frequency changer with electron coupling. Oscillator grid coils L11 (SW), L12 (MW) and L13 (LW) are tuned by C41; parallel trimming by C11 (SW), C42 (MW) and C43 (LW); series tracking by C13 (MW), and C12 (LW). Anode reaction by L14 (SW), L15 (MW) and L16 (LW). Second valve (V2, Mullard metallised VP4B) is a variable-mu RF pentode operating as intermediate frequency amplifies with tuned-primary tuned-secondary trans-

with tuned-primary tuned-secondary transformer couplings C44, L17, L18, C45 and C46, L19, L20, C47.



Circuit diagram of the Philips 747A AC superhet. V3 and V4 are both multiple valves with double diodes. One diode of V3 is strapped to the car directly to V2; the first diode of V4 is the signal rectifier, while the second, which is biased from the junction of R8, R9, provides a delay ac occur in some chassis, and in early models V4 was a pentode without diodes. Except for the mains input circuit, the circuit

Under chassis yiew. In several cases, where components appear to be bunched, they are actually on small racks forming vertical assemblies. The waveband switch units are numbered I to 4 and indicated here b**∦** arrows. Thev are shown in detail in the diagrams overleaf, where they are viewed in the direction of these arrows. Resistances R23 and R32 each consist of two resistors.



Intermediate frequency 128 KC/S.
Diode second detector is not part of V3 but is part of double diode pentode output valve (V4, Mullard Fen4DD).
Audio frequency component in rectified

the cathode, while the other provides AVC voltage lay action for AVC to VI. Minor modifications circuit of the 747AX is identical.

output is developed across load resistance R13 and passed via manual volume control R17, coupling condenser C20 (and on SW C21), variable RC tone filter R18, R19, C22, C23 and IF filter circuit R20, C24 to CG of triode section of double diode triode valve (V3, Mullard metallised TDD4), which operates as AF amplifier.

Control voltage for cathode-ray tuning indicator (T.I., Mullard TV4) is obtained from potential divided R14, R15, across R13, and decoupled by condenser C34.

One diode of V3 is strapped to the cathode; the second, fed from V2 anode via C18, provides DC potential which is developed across load resistance R25 and fed back through decoupling circuit as GB to IF valve, to provide undelayed automatic volume control for this stage. AVC to V1 is delayed, however, by a rather complicated system. The controlling bias potential is that at the second (delay) diode of V4 which is positively biased, via its load resistance R32, from potential divided R23, R7, R8, R9. While no signal is being received a very small voltage drop will occur, due to the diode current, as the internal resistance of the diode is low compared with R32. This condition is maintained until the negative potential of V3 AVC diode rises, due to the strength of an incoming signal, to a value sufficient to neutralise, via R26, the opposite polarity of V4 delay diode, after which normal AVC action will occur in proportion to the strength of the signal. On SW, V1 is not AVC controlled.

Resistance-capacity coupling by **£24**, **C28** and **£27**, via grid stopper **£29**, between **V3** triode and pentode section of **V4**. GB for **V4** is obtained from drop along **£30** in HT negative lead to chassis. Fixed tone correction in anode circuit by **C31**. Provision for connection of low impedance external speaker across secondary of internal speaker input transformer **T1**.

Negative feed-back circuit C32, L21, R33 between secondary of T1 and cathode circuit of V3 is coupled (except on SW

and Gram.) by R22, L22. S22 short-circuits these two latter on SW and Gram.

HT current is supplied by full-wave rectifying valve (V5, Philips 1321). Smoothing by iron-cored choke L24 and large capacity electrolytic condensers C30, C33. Provision for mains aerial coupling via S1, C1, the lever of S1 covering the aerial socket when the switch is closed.

COMPONENTS AND VALUES

	RESISTANCES	Values (ohms)
R1	V1 pent, CG decoupling	100,000
$\tilde{R}\tilde{2}$	V1 pent. CG stopper	32
R3	V1 fixed GB resistance	250
		10.000
R4	Radio muting on gram.	
R5	V1 osc. CG resistance	50,000
$\mathbf{R6}$	V1 osc. anode SW stabil-	
	iser	▲ 50
	(Parts V1, V2, SG's, V1)	· . J
$\mathbf{R7}$	osc. anode, and V3 tri-	10,000
$\mathbf{R8}$	de anode HT feed; V4 }	32,000
$\mathbf{R9}$	AVC delay diode pos.	64,000
	bias potentiometer)	
R10	V2 CG decoupling	2.006.000
R11	V2 fixed GB resistance	2,00 4 000
R12	IF stopper	250,000
R13	V4 signal diode load	250,000 1,600,000
R14	T.I. control potential	5,000,000
R15	}	1.60
R16		5,000,000 1,600,000 100,000
R17		100,000
	Manual volume control	500,000
R18	Part of variable TC filter	800,000
R19	Variable tone control, total	600,000
R20	IF stopper	160,000
R21	V3 triode GB resistance	3,200
R22	Part negative feed-back	
	circuit	20
R23	Part of pot, with R7, R8	
	and R9	10,000*
R24	V3 triode load resistance	100,000
R25	V3 AVC diode load	500,000
R26	V3 AVC diode and V4 de-	223,000
	lay diode coupling	1,000,000
R27	V4 pent. CG resistance	500,000
R28	V4 pent. CG decoupling	320,000
R29	V4 pent. CG decoupling V4 pent. CG stopper	
R30	V4 GB resistance	1,000 125
R31		
	V4 SG HT feed	50
R32	V4 delay diode load	9,000,000†
	Part negative feed-back	
R33		
R34	circuit T.I. anode HT feed	2,000,000

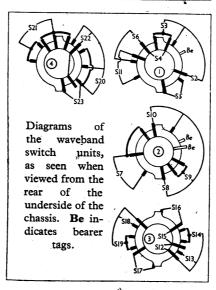
* Two 20,000 O in parallel. † One 5 MO and one 4 MO in series

		CONDENSERS	Values (μF)
1	C1	Mains aerial condenser	0.0005
-	Č2	Aerial SW series cond.	0.000016
ı	Č3	Image suppressor (fixed)	0.00002
١	C4	MW and LW aerial	0 00002
١	· .	coupling	0.00001
1	C5		0.016
١	Č6	Parts band-pass coupling {	0.025
1	Č7	V1 cathode by-pass	0.025
ı	čs	Small coupling	0.000002
	Č9	V1 osc. CG condenser	0.000002
ı	C10	V1 AVC line decoupling	0.0001
١	Cii	Osc. circuit SW trimmer	0.0000064
1	C12	Osc. circuit SW training	
Į		Osc. circuit LW tracker	0.0007
	C13 C14	Vi CC and accept	0.00149
1	C14	V1 SG and osc. anode	0.1
1		RF by-pass	0.1
ı	C15	V2 CG decoupling V2 SG and V3 triode	0.001
١	C16*	V2 SG and V3 triode	00.0
١	015	anode decoupling	32.0
١	C17	V2 cathode by-pass Coupling to V3 AVC diode	0.1
١	C18	Coupling to V3 A VC diode	0.00002
1	C19 .	IF by-pass	0.00005
İ	C20	AF coupling to V3 triode	0.005
	C21	SW AF coupling to V3	0 0000t
ı	C22	triode	0.00025
ı	C23	Parts variable TC filter	0.00025
ı	C24	IF by-pass	0.00064
	C25		0.0001 0.0016
ı	C26*	V3 cathode by-pass	50.0
1	C27	IF by-pass	0.0004
1	C28	V3 triode to V4 AF coup-	0.0004
1	020	ling	0.02
	C29	V4 CG decoupling	0.125
1	C30*	Part HT smoothing	32.0
	C3T	Fixed tone corrector	0.002
	C32	Part negative feed-back	0.05
	C33*	Part HT smoothing	32.0
	C34	T.I. feed decoupling	0.05
	C35‡	Aerial IF filter tuning	0.00017
	C36‡	Image suppression adjuster	.0.00003
	C371	Band-pass Pri. MW trimmer	0.00003
	C38† `	Band-pass primary tuning	0.00049
-	C39‡	Band-pass sec. MW	ŀ
1		trimmer	0.00003
	C40†	Band-pass secondary and	l
1	~441	SW aerial tuning	0.00049
1	C41†	Oscillator circuit tuning	0.00049
	C42; *	Osc. circuit MW trimmer	0.00003
-	C43‡	Osc. circuit LW trimmer	0.00003
	C44‡	1st IF trans. pri. tuning	0.00017
	C45‡	1st IF trans. sec. tuning	0.00017
	C46‡	2nd IF trans. pri. tuning	0.00017
-	C47‡	2nd IF trans. pri. tuning	0.00017

* Electrolytic. † Variable. ‡ P.	re-set.
OTHER COMPONENTS	Approx. Values (ohms)
L1 Aerial IF filter coil L2 Aerial band-pass coupling coils	140·0 40·0 100·0
L4 Band-pass primary coils L6	4.5 48.0 1.0
$\left\{\begin{array}{c} L7 \\ L8 \end{array}\right\}$ Band-pass coupling coils $\left\{\begin{array}{c} A \\ A \end{array}\right\}$ Aerial SW tuning coil	1.0 0.05
$\left\{ egin{array}{c} \mathbf{L9} \\ \mathbf{L10} \\ \mathbf{L11} \end{array} \right\}$ Band-pass secondary coils $\left\{ \mathbf{C} \right\}$	4.5 48.0 0.05
L12 Osc. circuit MW tuning	12.0
L13 Osc. circuit LW tuning coil Cosc. anode SW reaction	35.0
L15 Coil	40.0
L16 Osc. anode LW reaction	9.0
L17 L18	135·0 135·0
L20 2nd 1F trans. { Sec., total L21 Parts negative feed-back	135·0 135·0 175·0
L22 Speaker speech coil \ L24 HT smoothing choke	3·0 3·5 375·0
T1 Speaker input trans. $\begin{cases} Pri. \\ Sec. \end{cases}$	800·0 0·4
T2 Mains trans. { Pri., total Heater sec. Rect. ht. sc.	43·0 0·05 0·15
S1 Mains aerial switch Waveband switches	245 0
S23 Gram. PU switch S24 Mains switches, ganged (-
825 R17	

Switch Table and Diagrams

	· · · · · · · · · · · · · · · · · · ·	1		
Switch	sw	MW	LW	Gram.
82	0	0 000 0 00 0 00 000	 	_
, 83 , 84 85 86 87 88 89 810			U	000000000000000000000000000000000000000
Q.F			_	
98				_
87	U			Č
80	_	C	_	C
80	C		• -	
210		U	C	_
S10 S11	_			C.
S11 S12	G.	Č	Č	i —
012	_	· ·	G	_
S13	C	_		
S14		C	C.	
S15				C
816		C		-
S17		C		
S18	C		_	l
\$19 \$20 \$21		C	C	
820	·C	C	C	
821		С	С	С
822	C·	`		C
S23	-	_		C



DISMANTLING THE SET

The cabinet is fitted with a detachable bottom, upon removal of which access may be gained to most of the components in the under-chassis compartment.

Removing Chassis.—Remove the four

control knobs (recessed grub screws, two in each of the large knobs);

remove the four bolts (with washers) holding the chassis to the bottom of the cabinet;

unsolder the earthing lead to the screen on the bottom of the cabinet;

unsolder from the tuning indicator holder the leads going to the scale lamps; from the speaker and its transformer T1

unsolder the leads connecting them to chassis; free the scale lamp leads from the two

cleats holding them to the sub-baffle, and remove the entire scale assembly (four round-head wood-screws with washers).

Chassis and scale assembly may now be withdrawn from the cabinet.

When replacing chassis, take the speaker lead with green paint on it to the secondary (wide-spaced contacts) righthand or rear contact on the transformer and to tags 1 and 2 (numbering from left to right when viewed from rear) on the speaker connecting panel;

take the red-painted lead of the primary pair to the primary (close-spaced contacts) right-hand or rear contact.

It is important that the two painted leads

mentioned should be connected respectively to primary and secondary contacts at the same end of the transformer. If one pair is reversed, negative feed-back will become positive, and the receiver will become unstable.

When replacing the control knobs, the large knob engraved with a white spot should go on the left-hand (tone-control) spindle.

Removing Speaker.—Remove the chassis as already described so far as in necessary to give access to the speaker and free its leads;

slacken the four clamps (nuts, lock-nuts, and washers) holding it to the subbaffle, and swivel the clamps.

When replacing, the connecting should be at the top, and the leads should be connected as previously described.

Model 747AX

Removing Converter Unit .-- After removing the back cover from the receiver, the unit will be seen suspended from the top of the cabinet. At about the centre of its length is a join where the solid cap overlaps the perforated main

body; slacken the two milled head screws holding the cap to the body at the join; grasp the cap with the left hand, and the

body with the right hand, and ease the two sections slightly apart, when they will slip out of the two slotted brackets holding them to the top of the cabinet.

The left-hand (cap) portion contains the vibrator and its six connecting prongs, and the right-hand (body) portion the

input and smoothing components.

To free the body (which is attached to the mains input panel of the receiver

by a 5-way cable) entirely, remove the two set-screws from the two bottom corners of the upper section of the panel.

VALVE ANALYSIS

Valve voltages and currents given in the table below are those measured in our receiver when it was operating on mains of 231 V, using the 220 V tapping on the mains transformer. The receiver was tuned to the lowest wavelength on the medium band and the volume control was at maximum, but there was no signal in-

put.
Voltages were measured on the 400 V scale of a model 7 Universal Avometer, dechassis being negative.

Valve	Anode	Anode	Screen	Screen
	Voltage	Current	Voltage	Current
	(V)	(mA)	(V)	(mA)
V1 FC4 V2 VP4B V3 TDD4 V4 Pen4DD V5 1821	$\begin{cases} 265 \\ {}_{13}\text{Oscil} \\ 76 \\ 265 \\ 62 \\ 245 \\ 275 \dagger \end{cases}$	1·4 } ator } 7·4 0·8 34·0	82 — 155 265	4·7 — 2·4 6·0

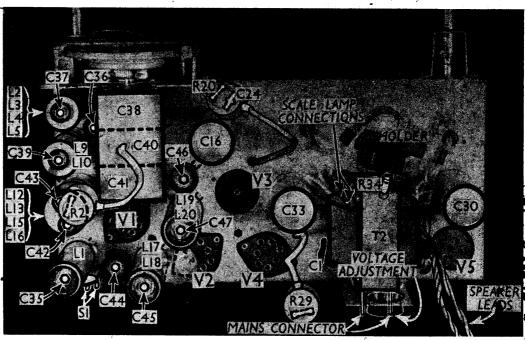
† Each anode, AC.

GENERAL NOTES

Switches.—S1 is the mains aerial switch, mounted on a bracket at the rear of the chassis, just above the A and E

PHILIPS' 747'A

Plan view of the chassis. Note the components mounted in V1, V3 and V4 top cap The connectors. speaker input transformer TI is not shown, but it may be mounted on the speaker or baffle or, in some cases, on the front right - hand corner of the chassis below T.I. holder. R34 is mounted directly on the tags of the T.I. holder. In the 747AX model an additional panel is mounted above the mains connecttion panel shown in illustration.



sockets. Two contacts of the switch unit are blank, while the other two forming the switch are closed when the lever control is moved so as to cover the aerial socket

\$2-\$23 are the waveband and pick-up switches, ganged in four rotary units beneath the chassis. These are indicated in our under-chassis view, and are shown in detail in the diagrams in col. 2, where they are drawn as seen when viewed from the rear of the underside of the chassis. There is a good deal of interconnection between the switches, particularly on the fourth unit, which we have indicated as clearly as possible.

The table (col. 2) gives the switch positions for the four control settings, starting from fully anti-clockwise. A dash ing from fully anti-clockwise.

indicates open, and C, closed.

824 and S25 are the QMB mains switches, ganged with the volume control

Scale Lamps.—These are three Philips MES types, with frosted tubular bulbs, type 8042-07.

External Speaker.—Two sockets are pro-

vided at the rear of the chassis for a low

mpedance (5.5 O) external speaker.

Resistances R23, R32.—R23 consists of two 20,000 O resistors in parallel, and R32 of a 4 MO and 5 MO in series.

Components R2, R20, C24 and R29.— R2 is inside the top cap connector of V1, R20 and C24 are inside the top cap connector of V3, and R29 is inside the top cap connector of V4. V4 Connections.—The Pen4DD has a 7-

pin base with connections differing from usual. The anode and cathode are interchanged, so that pin 2 is cathode and pin

6 is anode.

Chassis Divergencies.—We have examined circuit diagrams of the 747A and 747AX, and neither agrees entirely with our own chassis, which resembles most nearly the 747AX. The early 747A chassis had an ordinary pentode for V4, and not a double diode pentode, and this, of course, affects the AF end of the set considerably.

Some of the values for components do not agree with the makers' figures for the 747AX, but the differences are not important.

R33 may be 320 O, L22 may not be present, and C25 may not be present.

The radiogram models are similar in

the main, but again certain component values in the AF end may be different.

To cover all the divergencies here would be very confusing, and, working from the Service Sheet, the competent engineer should not find serious difficulty in following any of the divergent chassis.

The main difference between the chassis of the 747A and that of the 747AX is in the arrangement of the mains input panel. The panel of the 747A has two prongs mounted on it, to accept the mains connector, and a circular voltage adjustment

device just below the prongs.
On the 747AX there are two panels, one mounted directly above the other. The lower one carries several connecting prongs, each solid and much smaller than the two on the 747A, and the voltage adjustment device. The upper panel carries five pins in a row at the bottom and two filter condensers at the top. mains connector is reversible, for AC or DC, and the row of prongs on the upper panel are engaged only for DC operation. The cable from the converter unit is connected only to the upper panel.

CIRCUIT ALIGNMENT

It is not necessary to remove the chassis when aligning; merely place the receiver on its left side, and remove the bottom plate and the back of the cabinet.

IF Stages.—Connect up an earth wire, switch set to LW, turn volume and tuning controls to maximum. Connect signal

generator via a 0.032 µF condenser to control grid (top cap) of V1 (leaving existing top cap connection) and chassis. Feel in a 128 KC/S signal.

Connect a 25,000 O resistor and 0.1 µF condenser in-series, between top cap of V2 and chassis. Adjust C44 for maximum output. Connect the resistor and condenser between anode of V1 and chassis, and adjust C45 for maximum output. Connect the resistor and condenser across L20, and adjust C46 for maximum output. Connect resistor and condenser between anode of V2 and chassis, and adjust C47

for maximum output.

Seal trimmers with wax, remove resistor and condenser, and signal generator coupling condenser.

RF and Oscillator Stages.—If chassis has not been removed from cabinet, the usual method employing a 15 degree jig cannot be used. In this case, ordinary scale settings may be used, providing malibration is not suspected of being inacu-

MW.—Switch set to MW, turn volume control to maximum, and tune to 200 m control to maximum, and tune to 255 m on scale. Connect signal generator $^{\circ}$ A and $^{\circ}$ sockets via a dummy aerial, and feed in a 208 m (1,442 KC/S) signal. Connect 25,000 O resistor and 0.1 μ F condenser in series between top cap of V2 and chassis. Adjust C42, C39 and C37 for maximum output. Readjust C39, then C42, and then seal all three trimmers.

LW.—Switch set to LW. Feed in a 395 KC/S (760 m) signal, tune it in, and adjust C43 for maximum output, then re-

Image Suppressor.—Switch set to MW. Feed in a strong 1,000 KC/S (300 m) signal, tune set to 403 m on scale, adjust C36 for minimum output, and re-seal.

IF Filter.—Switch set to LW, feed in a strong 128 KC/S signal, turn tuning conto maximum, and adjust C35 for minimum output, then re-seal.