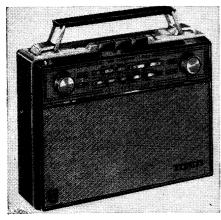
"TRADER" SERVICE SHEET



MPLOYING six p-n-p transistors, a press-button waveband switch and a printed circuit the Partition and a printed circuit, the Philips L3G91T "Philette" is a 2-band portable receiver "Philette" is a 2-band portable receiver housed in a plastics case. The wavebands covered are 188-555m (M.W.) and 1,177-2,000m (L.W.). The receiver is fitted with a ferrite rod aerial and a co-axial socket is provided to enable it to be operated with a oar aerial.

Release date and original price: May, 1959, £14 6s 2d. Purchase tax extra.

CIRCUIT DESCRIPTION

M.W. and L.W. aerial coils L2, L3 and their associated low impedance coupling coils L4, L5 are mounted on a ferrite rod to form an internal aerial. For L.W. reception, the switched coils L3 and L5 are connected in series with M.W. coils L2 and L4, and the tuning circuit is then shunted by C5. On both bands tuning is by variable capacitor C3, with parallel trimming by C4. Provision is made for the connection of a car aerial via C1, C2 and aerial loading coil L1.

TR1 is employed as self-oscillating frequency changer, the local oscillator circuit being formed by L6, L7 and L8. L7 and L8 comprise the oscillator coil, L8 providing a low impedance coupling for TR1 emitter. Oscillator tuning by variable capacitor C10 and trimmer capacitor C9 on M.W., and ad-

PHILIPS G91T

"Philette" Portable

ditionally by C8 on L.W. Tracking by the shaped vanes of C10.

TR2 and TR3 operate as a two-stage earthed-emitter I.F. amplifier with transformer couplings L9, L10; L11, L12; and L13, L14. Positive feedback due to the inherent internal coupling in TR2 and TR3 is neutralized by C13, R8 and C17, R12. Base bias for TR2 and TR3 is provided by the potential dividers R5, R6, R13 and R9, R10. Collector currents are stabilized by emitter resistors R7, R11.

Intermediate frequency 470 kc/s

The audio frequency component in the rectified output of germanium diode detector XI is developed across load resistor R13 and volume control R14, and is passed via R15, R16 and electrolytic coupling capacitor C22 to the base of driver stage TR4.

Under weak signal conditions the detector difference is improved by the forward bias

efficiency is improved by the forward bias applied to it from the potential divider R5, As the signal strength rises the positivegoing D.C. potential developed across R13, R14 is fed back as bias to the base of TR2,

giving automatic gain control.

The amplified output of TR4 is coupled to the output stage comprising the commonemitter transistors TR5, TR6 in a single-ended class B push-pull circuit. TR5 and TR6 are biased from the potential divider R21-R24. Coupling to high impedance speaker L15 via electrolytic capacitor C25. Tone compensation by negative feedback between L15 and the emitter of TR4 via R25, R26 and R20.

CIRCUIT ALIGNMENT

Equipment Required.—A signal generator Equipment Required.—A signal generator modulated 30 per cent at 400c/s; an output nieter with a 30Ω resistor for use as a dummy load; a 0.05μ F capacitor and a $2.2M\Omega$ resistor connected in parallel; a non-metallic slotted trimming tool, which can easily be made by cutting a slot in an insulated No.

10. knitting needle.

With the tuning gang at maximum capacitance, check that the cursor coincides with the right-hand end of the horizontal line between the M.W. and L.W. tuning scale aper-

Adjust the signal generator attenuator to maintain a 50mW reading on the output meter at all times during the alignment procedure

 Disconnect the speaker and connect the 30Ω dummy load resistor in its place; connect the output meter in parallel with the load resistor. Connect the signal generator, via the 0.05 µF capacitor connected in parallel with the 2.2M\Omega resistor, to the base of TR1.

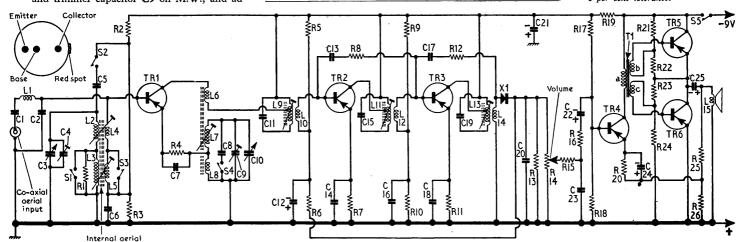
—Switch the receiver to M.W. Turn the

tuning gang to minimum and the volume (Continued overleaf, col. 1)

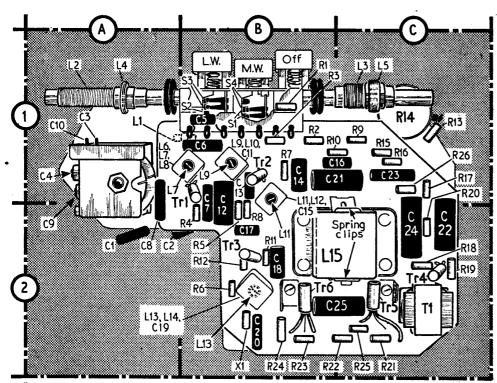
Coils*			R11	1kΩ	В2	
Ll	·	A1	R12	3·3kΩ	B2	
L2	1.6	Αl	R13	18kΩ	Čĩ	
L3	3.8	Ĉi	R14	50kΩ	Čî	
L4		Ăi	R15	ikΩ	či	
Ľ5		Ĉi	R16	1.5kΩ	či	
L6		Βi	R17	47kΩ	ČÎ	
Ľ7		Βi	R18	12kΩ	\tilde{C}_2	
Ĺ8	_	Βĩ	R19	680Ω	C2	
Ĺ9	5.6 (total)	Βī	R20	470Ω	$\tilde{C}2$	
Ĩ10	J 0 (101m2)	Βī	R21	2·7kΩ¹	C2	
Lii	5.5 (total)	Bī	R22	$100\Omega^1$	C2	
L12	J 5 (101)	Bi	R23	$2.7k\Omega^1$	B2	
L13	5.3 (total)	B2	R24	$100\Omega^{1}$	B2	
L14		B2	R25	2·2kΩ	C2	
L15	24.0	C2	R26	15Ω	C1	
Resist	Resistors			Capacitors		
R1	$270 \mathrm{k}\Omega$	B1	C1 C2	$0.01 \mu F$	A2	
R2	39kΩ	B1	C2	18pF ¹	A2	
R3	$10k\Omega$	B1	C3 C4 C5 C6 C7	196·1pF	A1	
R4	3·9kΩ	B2	C4	30pF	A1	
R5	$68k\Omega$	B2	C5	$68pF^2$	B1	
R6	$12k\Omega$	B2	C6	$0.01 \mu \mathrm{F}$	B1	
R7	680Ω	B1	C7	6,800pF	B2	
R8	1·2kΩ	B2	C8 C9	250pF ²	A2	
R9	22kΩ	C1	[] C9	30pF	A1	
R10	4·7kΩ	C1	[] C10	110pF	A1	

C11 C12 C13 C14 C15 C16 C16 C17 C18 C20 C21 C21 C22 C23 C24 C25	91pF 10µF 56pF 0-082µF 91pF 0-047µF 18pF 0-01µF 100µF 100µF 100µF 100µF	B1 B2 B1 B1 C1 B2 B2 B2 B2 C1 C2 C2
T1 {	### Add to the control of the contro	C2 B1

- *Approximate D.C. resistance in ohms.
- 15 per cent tolerance.
- 21 per cent tolerance.



Circuit diagram of the Philips G91T, or more fully L3G91T. The co-axial aerial socket permits the use of a car aerial.



View of the chassis with the back cover removed. The aerial rod is cut away to show the switch. L1 is hidden by the printed circuit panel.

Circuit Alignment—continued

control to maximum. Feed in a modulated 470kc's signal and adjust the cores of L11 (B1), L13 (B2), L9 (B1), and then L11 and L9 again, in that order, for maximum output. To adjust L13 the screening cover must first be removed (pull off), but it should be replaced immediately after L13 has been adjusted.

3.—Turn the tuning gang to maximum capacitance. Feed in a modulated capacitance. modulated 537kc/s signal and adjust the core of L7 (B1) for maximum output.

Turn the tuning gang to minimum capacitance. Feed in a modulated capacitance. 1,610kc/s signal and adjust C9 (A1) for maximum output.

5.—Repeat operations 3 and 4.

-Feed in a modulated 640kc/s signal and tune it in on the receiver, then without disturbing the setting of the tuning gang, disconnect the signal generator from TR1 and loosely couple its output to the aerial circuit. This can be done conveniently by clipping the signal generator output lead onto the insulated covering of the connecting leads between L2 and L3 (B1). Still feeding in a modulated 640kc/s signal, slide the former of L2, L4 (A1) along the ferrite rod for maximum output.

7.—Feed in a 1,425kc/s signal, tune it in on the receiver, and then adjust C4 (A1) for maximum output.

-Switch the receiver to L.W. Transfer the signal generator to the base of TR1 via the 0.05μ F capacitor connected in parallel with the $2.2M\Omega$ resistor. Feed in a modulated 180 kc/s signal and tune it in on the receiver. Without disturbing the setting of the tuning gang, transfer the signal generator to the aerial circuit as described in operation 6. Still feeding in a 180kc/s signal, slide the former of L3, L5 (C1) along the ferrite rod for maximum output.

 Switch receiver to M.W. and repeat operations 6 and 7. Seal the formers of L2, L4 and L3, L5 to the ferrite rod with wax to prevent them from moving.

GENERAL NOTES

Switches.—S1-S5 are the waveband and battery on/off switches ganged in a slide-type unit and actuated by press-buttons. The unit is

shown in our rear view illustration of the chassis in location reference B1, where the switch contacts associated with S1-S4 are also shown. The battery switch contacts S5 are shown in the sketch of a section of the reverse side of the chassis in column 1. S1 and S3 close on M.W., S2 and S4 close on L.W.

Battery.—The battery recommended by the manufacturers is an Ever Ready PP9 or a Vidor T6009, rated at 9V.

Transistors.—In the event of replacement of either TR5 or TR6 being necessary, both transistors must be replaced with a matched pair.

Drive Cord Replacement.—About 25in of nylon-braided glass yarn is required for a new drive cord. As shown in the sketch of the tuning drive system (column 1), two metal collars are pinched on the cord to form a loop at each end, and the overall length when made up should be 610mm (24.2in).

Modifications.—The following variations in component values may be found.

R1, 220kΩ or 270kΩ; R12, 3.3kΩ; C17, 22pF; C20, 0.01 μF. In later receivers the battery switch S5 is connected in the positive battery lead.

DISMANTLING

Removing Chassis.—Unscrew two screws from the rear of the receiver, then remove the rear section of the case, which is held by a hidden press-stud, by pulling its lower edge away from the body of the case; unsolder two leads from the external aerial socket;

remove control knobs (pull off), springs and

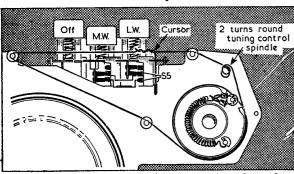
remove control knows (pair only), remove the three brass fixing pillars (unscrew). The bottom pillar carries the press-stud. The central section of the case and the chassis may then be separated from the front section of the case.

may then be separated from the front section of the case.
The printed circuit panel is mounted on the rear of the speaker unit, whose magnet passes through it. Two spring clips, one on each side of the magnet, clamp the panel to the speaker chassis, being sprung between the panel and the arch of the magnet. It is important to replace the insulated washer between the speaker and the printed circuit panel when re-assembling.

TRANSISTOR ANALYSIS

Transistor voltages and currents given in the table below are those derived from the manufacturers' information. Voltages were measured with a $20,000\Omega/V$ meter, chassis being the positive connection in every case. The receiver was switched to M.W., but there was no signal input. The total battery current was 6-9mA.

Transistor	Emitter (V)	Base (V)	Collector	
Tansistor	(*)	//	(V)	(mA)
TR1 OC44 TR2 OC45 TR3 OC45 TR4 OC78D TR5 OC78 TR6 OC78	1·6 0·56 0·9 1·13 4·5	1·3 0·66 1·0 1·12 4·66 0·16	7·0 7·0 7·42 8·42 9·0 4·5	0·4 0·8 0·95 2·2 0·8 0·8



Top: Diagram of the tuning drive system including a front view of the switches.

Right: View of the front side of the printed circuit panel, showing component connections.

